







ISO6760L SLLSFO3A - DECEMBER 2021 - REVISED FEBRUARY 2023

# ISO6760L Six-Channel Reinforced Digital Isolators with Integrated Interlock and **Robust EMC**

#### 1 Features

- ISO6760 with integrated Interlock function
  - Designed to support opposite polarity of adjacent channels
  - Three sets of paired interlock channels
- Robust isolation barrier:
  - High lifetime at 1500 V<sub>RMS</sub> working voltage
  - Up to 5000 V<sub>RMS</sub> isolation rating
  - Up to 10 kV surge capability
  - ±130 kV/µs typical CMTI
- Wide supply range: 1.71 V to 1.89 V and 2.25 V to
- Channel output non-inverting (ISO6760L) and inverting (ISO6760LN) options
- 50 Mbps data rate
- 1.71 V to 5.5 V level translation
- Wide temperature range: -40°C to 125°C
- 1.4 mA per channel typical at 1 Mbps
- Robust electromagnetic compatibility (EMC)
  - System-level ESD, EFT, and surge immunity
  - Low emissions
- Wide-SOIC (DW-16) Package
- Safety-Related Certifications:
  - DIN EN IEC 60747-17 (VDE 0884-17)
  - UL 1577 component recognition program
  - IEC 62368-1, IEC 61010-1, IEC 60601-1 and GB 4943.1 certifications

## 2 Applications

- Motor drives
- **Appliances**
- Grid
- **Building Automation**

## 3 Description

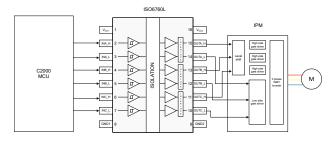
The ISO6760L and ISO6760LN devices are highperformance, six-channel digital isolators integrated interlock function for applications requiring up to 5000  $V_{RMS}$  isolation ratings per UL 1577. These devices are also certified by VDE, TUV, CSA, and CQC.

The ISO6760L family of devices integrate a series of logic gates to provide hardware interlock functionality for adjacent channels. The interlock feature ensures that each channel, in a channel pairing, will not be enabled at the same time. If both channels in the pairing share the same input logic, the output logic will always be low. The ISO6760L family of devices have all six channels in the same direction and provide high electromagnetic immunity and low emissions at low power consumption, while isolating CMOS or LVCMOS digital I/Os. Each isolation channel has a logic input and output buffer separated by TI's double capacitive silicon dioxide (SiO<sub>2</sub>) insulation barrier.

Used in conjunction with intelligent power modules (IPMs), the interlock feature in these devices help prevent shoot through current between the high side and low side gate driver during turn on and turn off events. Six channels, including three pairings of interlock circuitry, are integrated in a 16-pin SOIC wide-body (DW) package with space savings greater than 50% compared to optocoupler solutions. Through innovative chip design and layout techniques, the electromagnetic compatibility of the ISO6760L devices has been significantly enhanced to ease system-level ESD, EFT, surge, and emissions compliance.

### **Device Description**

Part Number	Package	Body Size
ISO6760L, ISO6760LN	SOIC (DW)	10.30 mm × 7.50 mm



Simplified Schematic



#### Table of Contents

1 Features1	6.21 Insulation Characteristics Curves	. 18
2 Applications 1	6.22 Typical Characteristics	. 19
3 Description1	7 Parameter Measurement Information	. 20
4 Revision History2	8 Detailed Description	21
5 Pin Configuration and Functions3	8.1 Overview	. 21
Specifications4	8.2 Functional Block Diagram	. 21
6.1 Absolute Maximum Ratings4	8.3 Feature Description	22
6.2 ESD Ratings 4	8.4 Device Functional Modes	24
6.3 Recommended Operating Conditions5	9 Application and Implementation	. 25
6.4 Thermal Information6	9.1 Application Information	. 25
6.5 Power Ratings6	9.2 Typical Application	. 26
6.6 Insulation Specifications7	10 Insulation Lifetime	.29
6.7 Safety-Related Certifications9	11 Power Supply Recommendations	.30
6.8 Safety Limiting Values9	12 Layout	31
6.9 Electrical Characteristics—5-V Supply10	12.1 Layout Guidelines	. 31
6.10 Supply Current Characteristics—5-V Supply10	12.2 Layout Example	. 32
6.11 Electrical Characteristics—3.3-V Supply11	13 Device and Documentation Support	33
6.12 Supply Current Characteristics—3.3-V Supply 11	13.1 Documentation Support	. 33
6.13 Electrical Characteristics—2.5-V Supply12	13.2 Receiving Notification of Documentation Updates.	
6.14 Supply Current Characteristics—2.5-V Supply12	13.3 Support Resources	
6.15 Electrical Characteristics—1.8-V Supply13	13.4 Trademarks	
6.16 Supply Current Characteristics—1.8-V Supply13	13.5 Electrostatic Discharge Caution	.33
6.17 Switching Characteristics—5-V Supply14	13.6 Glossary	.33
6.18 Switching Characteristics—3.3-V Supply15	14 Mechanical, Packaging, and Orderable	
6.19 Switching Characteristics—2.5-V Supply16	Information	. 33
6.20 Switching Characteristics—1.8-V Supply17		

### 4 Revision History

NOTE: Page numbers for previous revisions may differ from page numbers in the current version.

Changes from Revision \* (December 2021) to Revision A (February 2023)

С	hanges from Revision * (December 2021) to Revision A (February 2023)	Page
•	Changed standard name From: "DIN V VDE V 0884-11:2017-01" To: "DIN EN IEC 60747-17 (VDE 088	34-17)"
	throughout the document	1
•	Removed references to standard IEC/EN/CSA 60950-1 throughout the document	1

- Updated standards marked as "planned" to include certificate numbers thoughout the document......1 Removed standard revision and year references from all standard names thoughout the document......1 Added Maximum impulse voltage (V<sub>IMP</sub>) specification per DIN EN IEC 60747-17 (VDE 0884-17)......7 Changed test conditions and values of Maximum surge isolation voltage (V<sub>IOSM</sub>) specification per DIN EN IEC
- 60747-17 (VDE 0884-17).......7 Clarified method b test conditions of Apparent charge (q<sub>PD</sub>)......7 Changed Maximum surge isolation voltage (V<sub>IOSM</sub>) from 6250 V<sub>PK</sub> to 10000 V<sub>PK</sub> ......9
- Changed working voltage lifetime margin From: 87.5% To: 50%, minimum required insulation lifetime From: 37.5 years To: 30 years and insulation lifetime per TDDB From: 220 years To: 36 years per DIN EN IEC
- Updated to DW0016B mechanical drawing......33

Product Folder Links: ISO6760L

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# **5 Pin Configuration and Functions**

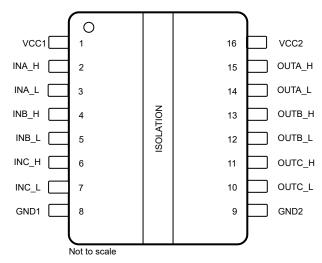


Figure 5-1. ISO6760L DW Package 16-Pin SOIC-WB Top View

**Table 5-1. Pin Functions** 

	PIN	1/0	DECODINE
NAME	ISO6760L	I/O	DESCRIPTION
GND1	8	_	Ground connection for V <sub>CC1</sub>
GND2	9	_	Ground connection for V <sub>CC2</sub>
INA_H	2	I	Input, channel A_H (Interlock paired with channel A_L)
INA_L	3	I	Input, channel A_L (Interlock paired with channel A_H)
INB_H	4	I	Input, channel B_H (Interlock paired with channel B_L)
INB_L	5	I	Input, channel B_L (Interlock paired with channel B_H)
INC_H	6	I	Input, channel C_H (Interlock paired with channel C_L)
INC_L	7	I	Input, channel C_L (Interlock paired with channel C_H)
OUTA_H	15	0	Output, channel A_H (Interlock paired with channel A_L)
OUTA_L	14	0	Output, channel A_L (Interlock paired with channel A_H)
OUTB_H	13	0	Output, channel B_H (Interlock paired with channel B_L)
OUTB_L	12	0	Output, channel B_L (Interlock paired with channel B_H)
OUTC_H	11	0	Output, channel C_H (Interlock paired with channel C_L)
OUTC_L	10	0	Output, channel C_L (Interlock paired with channel C_H)
V <sub>CC1</sub>	1	_	Power supply, side 1
V <sub>CC2</sub>	16	_	Power supply, side 2



## **6 Specifications**

# **6.1 Absolute Maximum Ratings**

See(1)

		MIN	MAX	UNIT
Supply Voltage (2)	V <sub>CC1</sub> to GND1	-0.5	6	V
Supply Voltage V	V <sub>CC2</sub> to GND2	-0.5	6	V
Input/Output	INx to GNDx	-0.5	V <sub>CCX</sub> + 0.5 <sup>(3)</sup>	V
Voltage	OUTx to GNDx	-0.5	V <sub>CCX</sub> + 0.5 <sup>(3)</sup>	V
Output Current	lo	-15	15	mA
Temperature	Operating junction temperature, T <sub>J</sub>		150	°C
Temperature	Storage temperature, T <sub>stg</sub>	-65	150	°C

- (1) Stresses beyond those listed under Absolute Maximum Ratings may cause permanent damage to the device. These are stress ratings only, which do not imply functional operation of the device at these or any other conditions beyond those indicated under Recommended Operating Conditions. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.
- (2) All voltage values except differential I/O bus voltages are with respect to the local ground terminal (GND1 or GND2) and are peak voltage values
- (3) Maximum voltage must not exceed 6 V.

### 6.2 ESD Ratings

(1) (2)

			VALUE	UNIT
	Electrostatic discharge	Human body model (HBM), per ANSI/ ESDA/JEDEC JS-001, all pins <sup>(1)</sup>	±2000	
V <sub>(ESD)</sub>	Electrostatic discharge	Charged device model (CDM), per JEDEC specification JESD22-C101, all pins <sup>(2)</sup>	±1500	V

Product Folder Links: ISO6760L

- (1) JEDEC document JEP155 states that 500-V HBM allows safe manufacturing with a standard ESD control process.
- (2) JEDEC document JEP157 states that 250-V CDM allows safe manufacturing with a standard ESD control process.

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# **6.3 Recommended Operating Conditions**

			MIN	NOM	MAX	UNIT
V <sub>CC1</sub> (1)	Supply Voltage Side 1 <sup>(3)</sup>		1.71		1.89	V
V <sub>CC1</sub> (1)	Supply Voltage Side 1 <sup>(3)</sup>		2.25		5.5	V
V <sub>CC2</sub> (1)	Supply Voltage Side 2 <sup>(3)</sup>		1.71		1.89	V
V <sub>CC2</sub> (1)	Supply Voltage Side 2 <sup>(3)</sup>		2.25		5.5	V
Vcc (UVLO+)	UVLO threshold when supply	voltage is rising		1.53	1.71	V
Vcc (UVLO-)	UVLO threshold when supply	voltage is falling	1.1	1.41		V
Vhys (UVLO)	Supply voltage UVLO hystere	sis	0.08	0.13		V
V <sub>IH</sub>	High level Input voltage		0.7 x V <sub>CCI</sub>		V <sub>CCI</sub>	V
V <sub>IL</sub>	Low level Input voltage		0		0.3 x V <sub>CCI</sub>	V
- 112		$V_{CCO}^{(2)} = 5 V$	-4			mA
	High level output current	V <sub>CCO</sub> = 3.3 V	-2			v V
'ОН	High level output current	V <sub>CCO</sub> = 2.5 V	-1			mA
		V <sub>CCO</sub> = 1.8 V	-1			mA
		V <sub>CCO</sub> = 5 V			4	mA
	Low level output current	V <sub>CCO</sub> = 3.3 V			2	mA
l <sub>OL</sub>		V <sub>CCO</sub> = 2.5 V			1	mA
		V <sub>CCO</sub> = 1.8 V			1	mA
DR	Data Rate	V <sub>CC</sub> = 2.25 V to 5.5 V	0		50	Mbps
טוז	Data Nate	V <sub>CC</sub> = 1.71 V to 1.89 V	0		25	Mbps
T <sub>A</sub>	Ambient temperature		-40	25	125	°C

<sup>(2)</sup> (3)

 $V_{CC1}$  and  $V_{CC2}$  can be set independent of one another  $V_{CCI}$  = Input-side  $V_{CC}$ ;  $V_{CCO}$  = Output-side  $V_{CC}$  The channel outputs are in undetermined state when 1.89 V <  $V_{CC1}$ ,  $V_{CC2}$  < 2.25 V and 1.05 V <  $V_{CC1}$ ,  $V_{CC2}$  < 1.71 V



# **6.4 Thermal Information**

		ISO6760L	
	THERMAL METRIC (1)	DW (SOIC)	UNIT
		16 PINS	
$R_{\theta JA}$	Junction-to-ambient thermal resistance	68.8	°C/W
R <sub>0JC(top)</sub>	Junction-to-case (top) thermal resistance	31.8	°C/W
$R_{\theta JB}$	Junction-to-board thermal resistance	32.7	°C/W
$\Psi_{JT}$	Junction-to-top characterization parameter	13.5	°C/W
$\Psi_{JB}$	Junction-to-board characterization parameter	32.1	°C/W
R <sub>0JC(bot)</sub>	Junction-to-case (bottom) thermal resistance	n/a	°C/W

# **6.5 Power Ratings**

	PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
ISO6760	L					
P <sub>D</sub>	Maximum power dissipation (both sides)	V <sub>CC1</sub> = V <sub>CC2</sub> = 5.5 V, T <sub>J</sub> = 150°C, C <sub>L</sub> = 15 pF, Input a 25-MHz 50% duty cycle			200	mW
P <sub>D1</sub>	Maximum power dissipation (side-1)				45	mW
P <sub>D2</sub>	Maximum power dissipation (side-2)	square wave			155	mW

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### 6.6 Insulation Specifications

DADAMETED		VALUE		
	PARAMETER	TEST CONDITIONS	DW-16	UNIT
CLR	External clearance <sup>(1)</sup>	Shortest terminal-to-terminal distance through air	>8	mm
CPG	External creepage <sup>(1)</sup>	Shortest terminal-to-terminal distance across the package surface	>8	mm
DTI	Distance through the insulation	Minimum internal gap (internal clearance)	>17	um
CTI	Comparative tracking index	DIN EN 60112 (VDE 0303-11); IEC 60112	>600	V
	Material group	According to IEC 60664-1	I	
	Over-18-18-18-18-18-18-18-18-18-18-18-18-18-	Rated mains voltage ≤ 600 V <sub>RMS</sub>	I-IV	
	Overvoltage category per IEC 60664-1	Rated mains voltage ≤ 1000 V <sub>RMS</sub>	1-111	
DIN EN I	IEC 60747-17 (VDE 0884-17) (2)			
V <sub>IORM</sub>	Maximum repetitive peak isolation voltage	AC voltage (bipolar)	2121	V <sub>PK</sub>
V <sub>IOWM</sub>	Maximum working isolation voltage	AC voltage; Time dependent dielectric breakdown (TDDB) Test; See Insulation Lifetime Projection Data	1500	V <sub>RMS</sub>
		DC voltage	2121	V <sub>DC</sub>
$V_{IOTM}$	Maximum transient isolation voltage	$V_{TEST} = V_{IOTM}$ , t = 60 s (qualification); $V_{TEST} = 1.2 \text{ x } V_{IOTM}$ , t= 1 s (100% production)	7071	V <sub>PK</sub>
V <sub>IMP</sub>	Maximum impulse voltage <sup>(3)</sup>	Tested in air, 1.2/50-us waveform per IEC 62368-1	7692	V <sub>PK</sub>
V <sub>IOSM</sub>	Maximum surge isolation voltage <sup>(4)</sup>	V <sub>IOSM</sub> ≥ 1.3 x V <sub>IMP</sub> ; Tested in oil (qualification test), 1.2/50-µs waveform per IEC 62368-1	10000	V <sub>PK</sub>
		Method a, After Input-output safety test subgroup 2/3, $V_{ini} = V_{IOTM}$ , $t_{ini} = 60 \text{ s}$ ; $V_{pd(m)} = 1.2 \text{ x } V_{IORM}$ , $t_m = 10 \text{ s}$	≤5	
$q_{pd}$	Apparent charge <sup>(5)</sup>	Method a, After environmental tests subgroup 1, $V_{ini} = V_{IOTM}$ , $t_{ini} = 60 \text{ s}$ ; $V_{pd(m)} = 1.6 \text{ x } V_{IORM}$ , $t_m = 10 \text{ s}$	≤5	pC
		Method b: At routine test (100% production) and preconditioning (type test); $V_{ini} = 1.2 \times V_{IOTM}, t_{ini} = 1 \text{ s}; \\ V_{pd(m)} = 1.875 \times V_{IORM}, t_m = 1 \text{ s (method b1) or } \\ V_{pd(m)} = V_{ini}, t_m = t_{ini} \text{ (method b2)}$	≤5	
C <sub>IO</sub>	Barrier capacitance, input to output <sup>(6)</sup>	$V_{IO} = 0.4 \text{ x sin } (2\pi \text{ft}), f = 1 \text{ MHz}$	~1	pF
		V <sub>IO</sub> = 500 V, T <sub>A</sub> = 25°C	>10 <sup>12</sup>	
$R_{IO}$	Isolation resistance <sup>(6)</sup>	V <sub>IO</sub> = 500 V, 100°C ≤ T <sub>A</sub> ≤ 125°C	>10 <sup>11</sup>	Ω
		V <sub>IO</sub> = 500 V at T <sub>S</sub> = 150°C	>10 <sup>9</sup>	
	Pollution degree		2	
	Climatic category		40/125/21	
UL 1577				
V <sub>ISO</sub>	Maximum withstanding isolation voltage	$V_{TEST} = V_{ISO}$ , t = 60 s (qualification), $V_{TEST} = 1.2 \times V_{ISO}$ , t = 1 s (100% production)	5000	V <sub>RMS</sub>

- 1) Creepage and clearance requirements should be applied according to the specific equipment isolation standards of an application. Care should be taken to maintain the creepage and clearance distance of a board design to ensure that the mounting pads of the isolator on the printed-circuit board do not reduce this distance. Creepage and clearance on a printed-circuit board become equal in certain cases. Techniques such as inserting grooves and/or ribs on a printed-circuit board are used to help increase these specifications.
- (2) This coupler is suitable for *safe electrical insulation* only within the safety ratings. Compliance with the safety ratings shall be ensured by means of suitable protective circuits.

- (3) Testing is carried out in air to determine the surge immunity of the package.
- (4) Testing is carried out in oil to determine the intrinsic surge immunity of the isolation barrier.
- (5) Apparent charge is electrical discharge caused by a partial discharge (pd).



(6) All pins on each side of the barrier tied together creating a two-terminal device.

### 6.7 Safety-Related Certifications

VDE	CSA	UL	CQC	TUV
Certified according to DIN EN IEC 60747-17 (VDE 0884-17)	Certified according to IEC 62368-1, IEC 61010-1 and IEC 60601	Certified according to UL 1577 Component Recognition Program	Certified according to GB 4943.1	Certified according to EN 61010-1 and EN 62368-1
Maximum transient isolation voltage, 7071 V <sub>PK</sub> ; Maximum repetitive peak isolation voltage, 2121 V <sub>PK</sub> ; Maximum surge isolation voltage, 10000 V <sub>PK</sub>	600 V <sub>RMS</sub> reinforced insulation per CSA 62368-1 and IEC 62368-1; 600 V <sub>RMS</sub> reinforced insulation per CSA 61010-1and IEC 61010-1 (pollution degree 2, material group I); 2 MOPP (Means of Patient Protection) per CSA 60601-1 and IEC 60601-1, 250 V <sub>RMS</sub> max working voltage	Single protection, 5000 V <sub>RMS</sub>	Reinforced insulation, Altitude ≤ 5000 m, Tropical Climate, 700 V <sub>RMS</sub> maximum working voltage	5000 V <sub>RMS</sub> reinforced insulation per EN 61010-1 and EN 62368-1 up to working voltage of 600 V <sub>RMS</sub>
Certificate number: 40040142	Master contract number: 220991	File number: E181974	Certificate number: CQC21001304083	Client ID number: 077311

### 6.8 Safety Limiting Values

Safety limiting<sup>(1)</sup> intends to minimize potential damage to the isolation barrier upon failure of input or output circuitry

	PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
DW-16 PA	ACKAGE					
		$R_{\theta JA} = 68.8$ °C/W, $V_I = 5.5$ V, $T_J = 150$ °C, $T_A = 25$ °C			330	mA
	Safety input, output, or supply current (1)	$R_{\theta JA} = 68.8^{\circ}C/W, V_{I} = 3.6 \text{ V}, T_{J} = 150^{\circ}C, T_{A} = 25^{\circ}C$			504	mA
I <sub>S</sub>	F	$R_{\theta,JA} = 68.8^{\circ}C/W, V_1 = 2.75 V, T_J = 150^{\circ}C, T_A = 25^{\circ}C$			660	mA
		$R_{\theta JA} = 68.8^{\circ}C/W, V_{I} = 1.89 V, T_{J} = 150^{\circ}C, T_{A} = 25^{\circ}C$			956	mA
Ps	Safety input, output, or total power (1)	R <sub>0JA</sub> = 68.8°C/W, T <sub>J</sub> = 150°C, T <sub>A</sub> = 25°C			1820	mW
T <sub>S</sub>	Maximum safety temperature (1)				150	°C

The maximum safety temperature, T<sub>S</sub>, has the same value as the maximum junction temperature, T<sub>J</sub>, specified for the device. The  $I_S$  and  $P_S$  parameters represent the safety current and safety power respectively. The maximum limits of  $I_S$  and  $P_S$  should not be exceeded. These limits vary with the ambient temperature, T<sub>A</sub>.

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The junction-to-air thermal resistance, R<sub>0,JA</sub>, in the table is that of a device installed on a high-K test board for leaded surface-mount packages. Use these equations to calculate the value for each parameter:

 $T_J = T_A + R_{\theta JA} \times P$ , where P is the power dissipated in the device.  $T_{J(max)} = T_S = T_A + R_{\theta JA} \times P_S$ , where  $T_{J(max)}$  is the maximum allowed junction temperature.  $P_S = I_S \times V_I$ , where  $V_I$  is the maximum input voltage.



## 6.9 Electrical Characteristics—5-V Supply

over recommended operating conditions (unless otherwise noted)

	PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
V <sub>OH</sub>	High-level output voltage	I <sub>OH</sub> = -4 mA; See Switching Characteristics Test Circuit and Voltage Waveforms	V <sub>CCO</sub> - 0.4			V
V <sub>OL</sub>	Low-level output voltage	I <sub>OL</sub> = 4 mA; See Switching Characteristics Test Circuit and Voltage Waveforms			0.4	V
V <sub>IT+(IN)</sub>	Rising input switching threshold			0.7	x V <sub>CCI</sub> <sup>(1)</sup>	V
V <sub>IT-(IN)</sub>	Falling input switching threshold		0.3 x V <sub>CCI</sub>			V
V <sub>I(HYS)</sub>	Input threshold voltage hysteresis		0.1 x V <sub>CCI</sub>			V
I <sub>IH</sub>	High-level input current	V <sub>IH</sub> = V <sub>CCI</sub> <sup>(1)</sup> at INx			10	μA
I <sub>IL</sub>	Low-level input current	V <sub>IL</sub> = 0 V at INx	-10			μA
СМТІ	Common mode transient immunity	V <sub>I</sub> = V <sub>CC</sub> or 0 V, V <sub>CM</sub> = 1200 V; see Common-Mode Transient Immunity Test Circuit	50	130		kV/us
C <sub>i</sub>	Input Capacitance (2)	$V_I = V_{CC}/2 + 0.4 \times \sin(2\pi ft), f = 2$ MHz, $V_{CC} = 5 \text{ V}$		2.8		pF

 $V_{\rm CCI}$  = Input-side  $V_{\rm CC}$ ;  $V_{\rm CCO}$  = Output-side  $V_{\rm CC}$  Measured from input pin to same side ground.

# 6.10 Supply Current Characteristics—5-V Supply

 $V_{CC1} = V_{CC2} = 5 \text{ V} \pm 10\%$  (over recommended operating conditions unless otherwise noted).

PARAMETER	TEST CONDITION	s	SUPPLY CURRENT	MIN TYP	MAX	UNIT
ISO6760L						
Supply current - DC signal	Output A: GND for ISO6760L and Vcc for ISO6760LN Output B: Vcc for ISO6760L and GND for ISO6760LN		I <sub>CC1</sub>	5.11	6.97	
			I <sub>CC2</sub>	3.3	5.38	
	All channels switching with square	1 Mbps	I <sub>CC1</sub>	5.13	6.99	
		1 Mbp3	I <sub>CC2</sub>	3.7	5.83	mA
Supply current - AC signal		10 Mbps	I <sub>CC1</sub>	5.29	7.19	ША
Supply current - AC signal	wave clock input; CL = 15 pF	TO Mibbs	I <sub>CC2</sub>	7.27	9.9	
		50 Mbps	I <sub>CC1</sub>	6.12	8.16	
			I <sub>CC2</sub>	23.62	27.74	

Product Folder Links: ISO6760L

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# 6.11 Electrical Characteristics—3.3-V Supply

 $V_{CC1} = V_{CC2} = 3.3 \text{ V} \pm 10\%$  (over recommended operating conditions unless otherwise noted).

	PARAMETER	TEST CONDITIONS	MIN	TYP MAX	UNIT
V <sub>OH</sub>	High-level output voltage	I <sub>OH</sub> = -2 mA; See Switching Characteristics Test Circuit and Voltage Waveforms	V <sub>CCO</sub> - 0.2		V
V <sub>OL</sub>	Low-level output voltage	I <sub>OL</sub> = 2 mA; See Switching Characteristics Test Circuit and Voltage Waveforms		0.2	V
V <sub>IT+(IN)</sub>	Rising input switching threshold			0.7 x V <sub>CCI</sub> <sup>(1)</sup>	V
V <sub>IT-(IN)</sub>	Falling input switching threshold		0.3 x V <sub>CCI</sub>		V
V <sub>I(HYS)</sub>	Input threshold voltage hysteresis		0.1 x V <sub>CCI</sub>		V
I <sub>IH</sub>	High-level input current	V <sub>IH</sub> = V <sub>CCI</sub> <sup>(1)</sup> at INx		10	μA
I <sub>IL</sub>	Low-level input current	V <sub>IL</sub> = 0 V at INx	-10		μA
CMTI	Common mode transient immunity	V <sub>I</sub> = V <sub>CC</sub> or 0 V, V <sub>CM</sub> = 1200 V; see Common-Mode Transient Immunity Test Circuit	50	130	kV/us
Ci	Input Capacitance <sup>(2)</sup>	$V_{I} = V_{CC}/2 + 0.4 \times \sin(2\pi f t), f = 2$ MHz, $V_{CC} = 3.3 \text{ V}$		2.8	pF

 <sup>(1)</sup> V<sub>CCI</sub> = Input-side V<sub>CC</sub>; V<sub>CCO</sub> = Output-side V<sub>CC</sub>
 (2) Measured from input pin to same side ground.

# 6.12 Supply Current Characteristics—3.3-V Supply

 $V_{CC1} = V_{CC2} = 3.3 \text{ V} \pm 10\%$  (over recommended operating conditions unless otherwise noted).

PARAMETER	TEST CONDITION	s	SUPPLY CURRENT	MIN	TYP	MAX	UNIT
ISO6760L							
Supply current - DC signal	Output A: GND for ISO6760L and Vo	cc for ISO6760LN	I <sub>CC1</sub>		5.08	6.89	
Supply current - DC signal	Output B: Vcc for ISO6760L and GND for ISO6760LN		I <sub>CC2</sub>		3.28	5.36	
	All channels switching with square wave clock input; CL = 15 pF	1 Mbps	I <sub>CC1</sub>		5.1	6.9	
		1 Mbps	I <sub>CC2</sub>		3.57	5.68	mA
Supply current - AC signal		10 Mbps	I <sub>CC1</sub>		5.18	7.04	ША
Supply current - AC signal		TO MIDPS	I <sub>CC2</sub>		6.07	8.62	-
		50 Mbps	I <sub>CC1</sub>		5.74	7.68	
			I <sub>CC2</sub>		17.54	21.5	



## 6.13 Electrical Characteristics—2.5-V Supply

 $V_{CC1} = V_{CC2} = 2.5 \text{ V} \pm 10\%$  (over recommended operating conditions unless otherwise noted).

	PARAMETER	TEST CONDITIONS	MIN	TYP MAX	UNIT
V <sub>OH</sub>	High-level output voltage	I <sub>OH</sub> = -1 mA; See Switching Characteristics Test Circuit and Voltage Waveforms	V <sub>CCO</sub> - 0.1		V
V <sub>OL</sub>	Low-level output voltage	I <sub>OL</sub> = 1 mA; See Switching Characteristics Test Circuit and Voltage Waveforms		0.1	V
V <sub>IT+(IN)</sub>	Rising input switching threshold			0.7 x V <sub>CCI</sub> <sup>(1)</sup>	V
V <sub>IT-(IN)</sub>	Falling input switching threshold		0.3 x V <sub>CCI</sub>		V
V <sub>I(HYS)</sub>	Input threshold voltage hysteresis		0.1 x V <sub>CCI</sub>		V
I <sub>IH</sub>	High-level input current	V <sub>IH</sub> = V <sub>CCI</sub> <sup>(1)</sup> at INx		10	μA
I <sub>IL</sub>	Low-level input current	V <sub>IL</sub> = 0 V at INx	-10		μA
СМТІ	Common mode transient immunity	V <sub>I</sub> = V <sub>CC</sub> or 0 V, V <sub>CM</sub> = 1200 V; see Common-Mode Transient Immunity Test Circuit	50	130	kV/us
Ci	Input Capacitance <sup>(2)</sup>	$V_I = V_{CC}/2 + 0.4 \times \sin(2\pi ft), f = 2$ MHz, $V_{CC} = 2.5 \text{ V}$		2.8	pF

 <sup>(1)</sup> V<sub>CCI</sub> = Input-side V<sub>CC</sub>; V<sub>CCO</sub> = Output-side V<sub>CC</sub>
 (2) Measured from input pin to same side ground.

# 6.14 Supply Current Characteristics—2.5-V Supply

 $V_{CC1} = V_{CC2} = 2.5 \text{ V} \pm 10\%$  (over recommended operating conditions unless otherwise noted).

PARAMETER	TEST CONDITION	S	SUPPLY CURRENT	MIN	TYP	MAX	UNIT
ISO6760L							
Supply current - DC signal	Output A: GND for ISO6760L and Vo	cc for ISO6760LN	I <sub>CC1</sub>		5.07	6.85	
Supply current - DC signal	Output B: Vcc for ISO6760L and GND for ISO6760LN		I <sub>CC2</sub>		3.28	5.35	
	All channels switching with square wave clock input; CL = 15 pF	1 Mbps	I <sub>CC1</sub>		5.08	6.87	
		i ivibps	I <sub>CC2</sub>		3.49	5.59	mA
Supply current - AC signal		10 Mbps	I <sub>CC1</sub>		5.14	6.97	ША
Supply current - AC signal		TO MDPS	I <sub>CC2</sub>		5.34	7.8	
		25 Mbps	I <sub>CC1</sub>		5.59	7.49	
			I <sub>CC2</sub>		13.83	17.47	

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# 6.15 Electrical Characteristics—1.8-V Supply

 $V_{CC1} = V_{CC2} = 1.8 \text{ V} \pm 10\%$  (over recommended operating conditions unless otherwise noted).

	PARAMETER	TEST CONDITIONS	MIN	TYP MA	X UNIT
V <sub>OH</sub>	High-level output voltage	I <sub>OH</sub> = -1 mA; See Switching Characteristics Test Circuit and Voltage Waveforms	V <sub>CCO</sub> - 0.1		V
V <sub>OL</sub>	Low-level output voltage	I <sub>OL</sub> = 1 mA; See Switching Characteristics Test Circuit and Voltage Waveforms		C	.1 V
V <sub>IT+(IN)</sub>	Rising input switching threshold			0.7 x V <sub>CCI</sub>	(1) V
V <sub>IT-(IN)</sub>	Falling input switching threshold		0.3 x V <sub>CCI</sub>	-	V
V <sub>I(HYS)</sub>	Input threshold voltage hysteresis		0.1 x V <sub>CCI</sub>		V
I <sub>IH</sub>	High-level input current	V <sub>IH</sub> = V <sub>CCI</sub> <sup>(1)</sup> at INx			10 μA
I <sub>IL</sub>	Low-level input current	V <sub>IL</sub> = 0 V at INx	-10		μA
СМТІ	Common mode transient immunity	V <sub>I</sub> = V <sub>CC</sub> or 0 V, V <sub>CM</sub> = 1200 V; see Common-Mode Transient Immunity Test Circuit	50	75	kV/us
C <sub>i</sub>	Input Capacitance <sup>(2)</sup>	$V_1 = V_{CC}/2 + 0.4 \times \sin(2\pi ft), f = 2$ MHz, $V_{CC} = 1.8 \text{ V}$		2.8	pF

 <sup>(1)</sup> V<sub>CCI</sub> = Input-side V<sub>CC</sub>; V<sub>CCO</sub> = Output-side V<sub>CC</sub>
 (2) Measured from input pin to same side ground.

# 6.16 Supply Current Characteristics—1.8-V Supply

 $V_{CC1} = V_{CC2} = 1.8 \text{ V} \pm 10\%$  (over recommended operating conditions unless otherwise noted).

PARAMETER	TEST CONDITION	s	SUPPLY CURRENT	MIN	TYP	MAX	UNIT
ISO6760L				-			
Supply current - DC signal	Output A: GND for ISO6760L and Vcc for ISO6760LN Output B: Vcc for ISO6760L and GND for ISO6760LN		I <sub>CC1</sub>		4.27	6.24	
			I <sub>CC2</sub>		3.15	5.39	
	All channels switching with square	1 Mbps	I <sub>CC1</sub>		4.28	6.25	
		1 Mbps	I <sub>CC2</sub>		3.3	5.55	mA
		10 Mbps	I <sub>CC1</sub>	-	4.37	6.37	
Supply current - AC signal	wave clock input; CL = 15 pF		I <sub>CC2</sub>		4.6	7.04	
		50 Mbps	I <sub>CC1</sub>		4.5	6.5	
		(25Mbps)	I <sub>CC2</sub>		6.84	9.47	



## 6.17 Switching Characteristics—5-V Supply

V<sub>CC1</sub> = V<sub>CC2</sub> = 5 V ±10% (over recommended operating conditions unless otherwise noted)

	PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT
ISO6760L							
t <sub>PLH</sub> , t <sub>PHL</sub>	Propagation delay time		One input in static state and other input		13	20.5	ns
PWD	Pulse width distortion <sup>(1)</sup>  t <sub>PHL</sub> - t <sub>PLH</sub>		is toggled at 100kbps. See Switching Characteristics Test Circuit and Voltage Waveforms		1	7	ns
t <sub>sk(o)</sub>	Channel-to-channel output skew time <sup>(2)</sup>		Same-direction channels			6	ns
t <sub>sk(pp)</sub>	Part-to-part skew time <sup>(3)</sup>					6	ns
t <sub>r</sub>	Output signal rise time		See Switching Characteristics Test		2.6	4.5	ns
t <sub>f</sub>	Output signal fall time		Circuit and Voltage Waveforms		2.6	4.5	ns
tPU	Time from UVLO to valid output data	Time from UVLO to valid output data				300	us
t <sub>DO</sub>	Default output delay time from input power loss		Measured from the time VCC goes below 1.2V. See Default Output Delay Time Test Circuit and Voltage Waveforms		0.1	0.3	us
t <sub>ie</sub>	Time interval error		2 <sup>16</sup> – 1 PRBS data at 50 Mbps		1		ns

<sup>(1)</sup> Also known as pulse skew.

<sup>(2)</sup>  $t_{sk(o)}$  is the skew between outputs of a single device with all driving inputs connected together and the outputs switching in the same direction while driving identical loads.

<sup>(3)</sup> t<sub>sk(pp)</sub> is the magnitude of the difference in propagation delay times between any terminals of different devices switching in the same direction while operating at identical supply voltages, temperature, input signals and loads.

# 6.18 Switching Characteristics—3.3-V Supply

V<sub>CC1</sub> = V<sub>CC2</sub> = 3.3 V ±10% (over recommended operating conditions unless otherwise noted)

	PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT
ISO6760L							
t <sub>PLH</sub> , t <sub>PHL</sub>	Propagation delay time		One input in static state and other input		13	21	ns
PWD	Pulse width distortion <sup>(1)</sup>  t <sub>PHL</sub> - t <sub>PLH</sub>		is toggled at 100kbps. See Switching Characteristics Test Circuit and Voltage Waveforms		1	7	ns
t <sub>sk(o)</sub>	Channel-to-channel output skew time(2)		Same-direction channels			6	ns
t <sub>sk(pp)</sub>	Part-to-part skew time <sup>(3)</sup>					7	ns
t <sub>r</sub>	Output signal rise time				1.6	2.8	ns
t <sub>f</sub>	Output signal fall time		Circuit and Voltage Waveforms		1.6	2.8	ns
tPU	Time from UVLO to valid output data	Time from UVLO to valid output data				300	us
t <sub>DO</sub>	Default output delay time from input pov	efault output delay time from input power loss			0.1	0.3	us
t <sub>ie</sub>	Time interval error		2 <sup>16</sup> – 1 PRBS data at 50 Mbps		1		ns

<sup>(1)</sup> Also known as pulse skew.

<sup>(2)</sup>  $t_{sk(o)}$  is the skew between outputs of a single device with all driving inputs connected together and the outputs switching in the same direction while driving identical loads.

<sup>(3)</sup>  $t_{sk(pp)}$  is the magnitude of the difference in propagation delay times between any terminals of different devices switching in the same direction while operating at identical supply voltages, temperature, input signals and loads.



## 6.19 Switching Characteristics—2.5-V Supply

V<sub>CC1</sub> = V<sub>CC2</sub> = 2.5 V ±10% (over recommended operating conditions unless otherwise noted)

	PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT
ISO6760L							
t <sub>PLH</sub> , t <sub>PHL</sub>	Propagation delay time		One input in static state and other input		14.5	23.5	ns
PWD	Pulse width distortion <sup>(1)</sup>  t <sub>PHL</sub> - t <sub>PLH</sub>		is toggled at 100kbps. See Switching Characteristics Test Circuit and Voltage Waveforms		1	7.1	ns
t <sub>sk(o)</sub>	Channel-to-channel output skew time(2)		Same-direction channels			6	ns
t <sub>sk(pp)</sub>	Part-to-part skew time <sup>(3)</sup>					7.9	ns
t <sub>r</sub>	Output signal rise time				2	4	ns
t <sub>f</sub>	Output signal fall time		Circuit and Voltage Waveforms		2	4	ns
tPU	Time from UVLO to valid output data	Time from UVLO to valid output data				300	us
t <sub>DO</sub>	Default output delay time from input pov	Default output delay time from input power loss			0.1	0.3	us
t <sub>ie</sub>	Time interval error		2 <sup>16</sup> – 1 PRBS data at 50 Mbps		1		ns

<sup>(1)</sup> Also known as pulse skew.

<sup>(2)</sup>  $t_{sk(o)}$  is the skew between outputs of a single device with all driving inputs connected together and the outputs switching in the same direction while driving identical loads.

<sup>(3)</sup> t<sub>sk(pp)</sub> is the magnitude of the difference in propagation delay times between any terminals of different devices switching in the same direction while operating at identical supply voltages, temperature, input signals and loads.

# 6.20 Switching Characteristics—1.8-V Supply

V<sub>CC1</sub> = V<sub>CC2</sub> = 1.8 V ±5% (over recommended operating conditions unless otherwise noted)

	PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT
ISO6760L							
t <sub>PLH</sub> , t <sub>PHL</sub>	Propagation delay time		One input in static state and other input		18	31	ns
PWD	Pulse width distortion  t <sub>PHL</sub> - t <sub>PLH</sub>		is toggled at 100kbps. See Switching Characteristics Test Circuit and Voltage Waveforms		1	8.2	ns
t <sub>sk(o)</sub>	Channel-to-channel output skew time <sup>(1)</sup>		Same-direction channels			6	ns
t <sub>sk(pp)</sub>	Part-to-part skew time <sup>(2)</sup>					11.7	ns
t <sub>r</sub>	Output signal rise time		See Switching Characteristics Test		2.7	5.3	ns
t <sub>f</sub>	Output signal fall time		Circuit and Voltage Waveforms		2.7	5.3	ns
tPU	Time from UVLO to valid output data	Time from UVLO to valid output data				300	us
t <sub>DO</sub>	Default output delay time from input power loss		Measured from the time VCC goes below 1.2V. See Default Output Delay Time Test Circuit and Voltage Waveforms		0.1	0.3	us
t <sub>ie</sub>	Time interval error		2 <sup>16</sup> – 1 PRBS data at 50 Mbps		1		ns

 $t_{sk(p)}$  is the skew between outputs of a single device with all driving inputs connected together and the outputs switching in the same direction while driving identical loads.  $t_{sk(pp)}$  is the magnitude of the difference in propagation delay times between any terminals of different devices switching in the same direction while operating at identical supply voltages, temperature, input signals and loads.



# **6.21 Insulation Characteristics Curves**

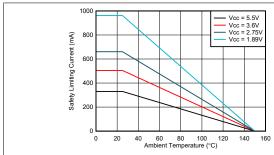


Figure 6-1. Thermal Derating Curve for Safety Limiting Current for DW-16 Package

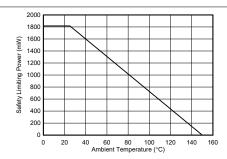
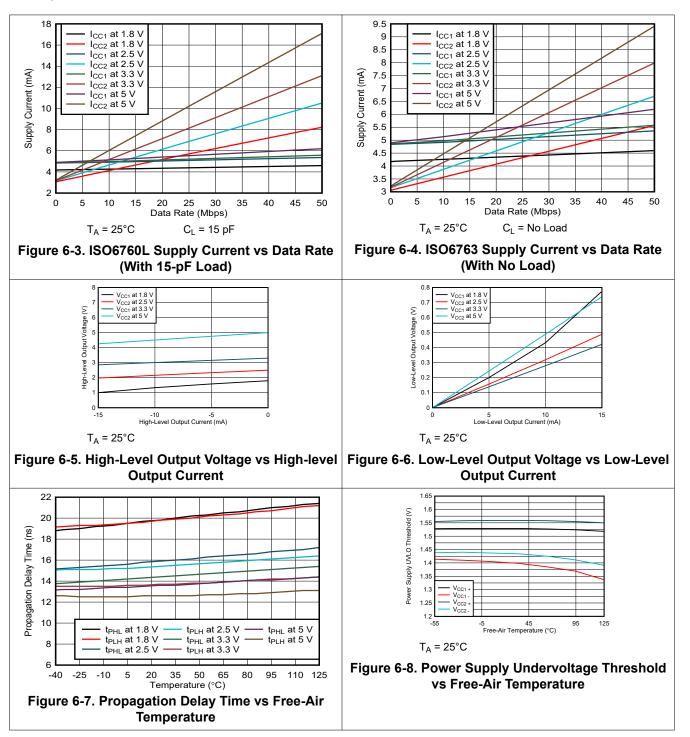


Figure 6-2. Thermal Derating Curve for Safety Limiting Power for DW-16 Package

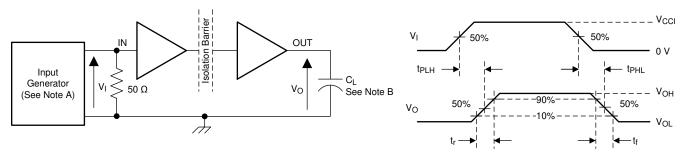


## **6.22 Typical Characteristics**





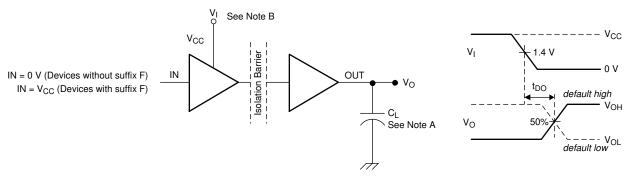
### 7 Parameter Measurement Information



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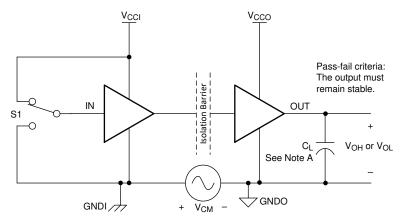
- A. The input pulse is supplied by a generator having the following characteristics: PRR  $\leq$  50 kHz, 50% duty cycle,  $t_r \leq$  3 ns,  $t_f \leq$  3ns,  $Z_O =$  50 Ω. At the input, 50 Ω resistor is required to terminate Input Generator signal. It is not needed in actual application.
- B. C<sub>L</sub> = 15 pF and includes instrumentation and fixture capacitance within ±20%.

Figure 7-1. Switching Characteristics Test Circuit and Voltage Waveforms



- A. C<sub>L</sub> = 15 pF and includes instrumentation and fixture capacitance within ±20%.
- B. Power Supply Ramp Rate = 10 mV/ns

Figure 7-2. Default Output Delay Time Test Circuit and Voltage Waveforms



- A. C<sub>L</sub> = 15 pF and includes instrumentation and fixture capacitance within ±20%.
- B. For optimized CMTI performance, a  $0.1~\mu\text{F} + 1~\mu\text{F}$  decoupling capacitor should be placed close to V<sub>CC1</sub> and V<sub>CC2</sub>. Please see Section 12.2 for capacitor placement details. A recommended  $0.1\mu\text{F}$  capacitor is LLL185R71A104MA11L (CAP CER 0.1UF 10V X7R 0306 LW Reversed Low ESL Chip Ceramic Capacitors) or equivalent.

Figure 7-3. Common-Mode Transient Immunity Test Circuit

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# 8 Detailed Description

### 8.1 Overview

The ISO6760L family of devices have an ON-OFF keying (OOK) modulation scheme to transmit the digital data across a silicon dioxide based isolation barrier. The transmitter sends a high frequency carrier across the barrier to represent one digital state and sends no signal to represent the other digital state. The receiver demodulates the signal after advanced signal conditioning and produces the output which goes through an interlock stage before an output buffer. The ISO6760L family offers two options, a standard non-inverting channel, ISO6760L, and a channel inverting ISO6760LN. The two offerings make the ISO6760L family compatable with historical optocoupler based solutions. The ISO6760L devices also incorporate advanced circuit techniques to maximize the CMTI performance and minimize the radiated emissions due to the high frequency carrier and IO buffer switching. The conceptual block diagram of a digital capacitive isolator, Figure 8-1, shows a functional block diagram of a typical channel.

### 8.2 Functional Block Diagram

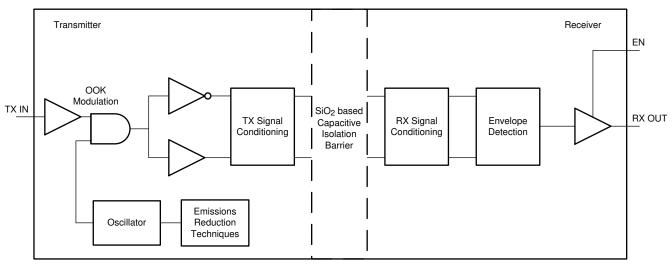


Figure 8-1. Conceptual Block Diagram of a Digital Capacitive Isolator

Figure 8-2 shows a conceptual detail of how the ON-OFF keying scheme works.

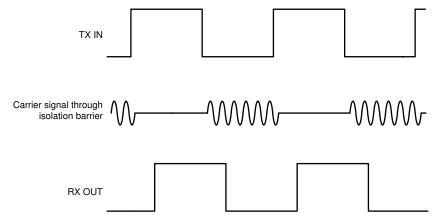


Figure 8-2. On-Off Keying (OOK) Based Modulation Scheme



### 8.3 Feature Description

Table 8-1 provides an overview of the device features.

Table 8-1. Device Features

PART NUMBER	CHANNEL DIRECTION	MAXIMUM DATA RATE	OUTPUT	PACKAGE	RATED ISOLATION(1)
ISO6760L	6 Forward, 3 Interlock Pairs	50 Mbps	Non-Inverted	DW-16	5000 V <sub>RMS</sub> / 7000 V <sub>PK</sub>
ISO6760LN	6 Forward, 3 Interlock Pairs	50 Mbps	Inverted	DW-16	5000 V <sub>RMS</sub> / 7000 V <sub>PK</sub>

(1) See for detailed isolation ratings.

## 8.3.1 Electromagnetic Compatibility (EMC) Considerations

Many applications in harsh industrial environment are sensitive to disturbances such as electrostatic discharge (ESD), electrical fast transient (EFT), surge and electromagnetic emissions. These electromagnetic disturbances are regulated by international standards such as IEC 61000-4-x and CISPR 32. Although system-level performance and reliability depends, to a large extent, on the application board design and layout, the ISO676x family of devices incorporates many chip-level design improvements for overall system robustness. Some of these improvements include:

- Robust ESD protection cells for input and output signal pins and inter-chip bond pads.
- · Low-resistance connectivity of ESD cells to supply and ground pins.
- Enhanced performance of high voltage isolation capacitor for better tolerance of ESD, EFT and surge events.
- Bigger on-chip decoupling capacitors to bypass undesirable high energy signals through a low impedance path.
- PMOS and NMOS devices isolated from each other by using guard rings to avoid triggering of parasitic SCRs.
- Reduced common mode currents across the isolation barrier by ensuring purely differential internal operation.

### 8.3.2 Interlock Capability

The ISO6760L family incorporates a series of logic gates to protect adjacent channel pairings from both registering high simultaneously. When paired with an IPM, this interlock circuitry provides protection proventing shoot through current to both the high-side and low-side switch of the module. This design, shown in ISO6760L Channel Pairing Block Diagram of Interlock, is used to make sure that when one of the channel pairings is logic high, the other channel will output logic low. ISO6760L Device Truth Table provides the logic output state to the corresponding input state for ISO6760L and ISO6760LN (Inverted) Device Truth Table provides the logic output state to the corresponding input state for ISO6760LN (inverted output version).

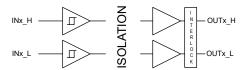


Figure 8-3. ISO6760L Channel Pairing Block Diagram of Interlock

### **ISO6760L Device Truth Table**

INx_H	INx_L	OUTx_H	OUTx_L	
High	Low	High	Low	
Low	High	Low	High	
High	High	High Low		
Low	Low	Low	Low	
Floating	Floating	Low	Low	



# Table 8-2. ISO6760LN (Inverted) Device Truth Table

INx_H	INx_L	OUTx_H	OUTx_L	
High	Low	Low	High	
Low	High	High	Low	
High	High	Low	Low	
Low	Low	Low	Low	
Floating	Floating	Floating Low		



#### 8.4 Device Functional Modes

Function Table lists the functional modes for the ISO6760L devices.

#### **Function Table**

V <sub>CCI</sub> <sup>(1)</sup>	V <sub>cco</sub>	INPUT (INx_H and INx_L) <sup>(3)</sup>	OUTPUT (OUTx_H and OUTx_L)	COMMENTS			
		Н		Normal Operation: A channel output assumes the logic state of its input			
PU	PU	L	Normal	noted in ISO6760L Device Truth Table and ISO6760LN (Inverted) Device Tru			
		Open		Table .			
PD	PU	X	Low	Output Low: When $V_{CCI}$ is unpowered and $V_{CCO}$ is powered up, the output interlock circuit will set the output to logic low. When $V_{CCI}$ transitions from unpowered to powered-up, a channel output assumes the logic state in ISO6760L Device Truth Table and ISO6760LN (Inverted) Device Truth Table . When $V_{CCI}$ transitions from powered-up to unpowered, channel output will be the output low state.			
х	PD	Х	Undetermined	When $V_{CCO}$ is unpowered, a channel output is undetermined <sup>(2)</sup> . When $V_{CCO}$ transitions from unpowered to powered-up, a channel output assumes the logic state of its input noted in ISO6760L Device Truth Table and ISO6760LN (Inverted) Device Truth Table .			

- $V_{CCI}$  = Input-side  $V_{CC}$ ;  $V_{CCO}$  = Output-side  $V_{CC}$ ; PU = Powered up ( $V_{CC} \ge 1.71 \text{ V}$ ); PD = Powered down ( $V_{CC} \le 1.05 \text{ V}$ ); X = Irrelevant; H = High level; L = Low level; Z = High Impedance
- The outputs are in undetermined state when 1.7 V <  $V_{CCI}$ ,  $V_{CCO}$  < 2.25 V and 1.05 V <  $V_{CCI}$ ,  $V_{CCO}$  < 1.71 V A strongly driven input signal can weakly power the floating  $V_{CC}$  through an internal protection diode and cause undetermined output

### 8.4.1 Device I/O Schematics

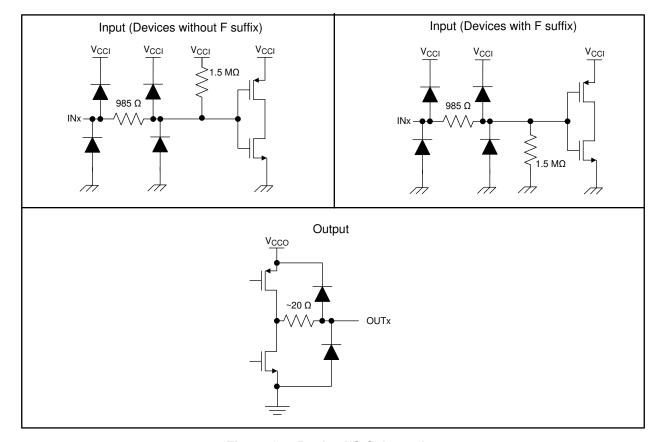


Figure 8-4. Device I/O Schematics



# 9 Application and Implementation

#### Note

Information in the following applications sections is not part of the TI component specification, and TI does not warrant its accuracy or completeness. TI's customers are responsible for determining suitability of components for their purposes. Customers should validate and test their design implementation to confirm system functionality.

### 9.1 Application Information

The ISO6760L devices are high-performance, six-channel digital isolators. The ISO6760L devices use single-ended CMOS-logic switching technology with built in hardware interlock logic. The supply voltage range is from 1.71 V to 5.5 V for both supplies,  $V_{CC1}$  and  $V_{CC2}$ . Since an isolation barrier separates the two sides, each side can be sourced independently with any voltage within recommended operating conditions. As an example, it is possible to supply ISO6760L  $V_{CC1}$  with 3.3 V (which is within 1.71 V to 5.5 V) and  $V_{CC2}$  with 5V (which is also within 1.71 V to 5.5 V). You can use the digital isolator as a logic-level translator in addition to providing isolation. When designing with digital isolators, keep in mind that because of the single-ended design structure, digital isolators do not conform to any specific interface standard and are only intended for isolating single-ended CMOS or TTL digital signal lines. The isolator is typically placed between the data controller (that is, MCU or FPGA), and a data converter or a line transceiver, regardless of the interface type or standard.



# 9.2 Typical Application

Figure 9-1 shows the isolated connections between a processor and Intelligent Power Module (IPM) interface implementation.

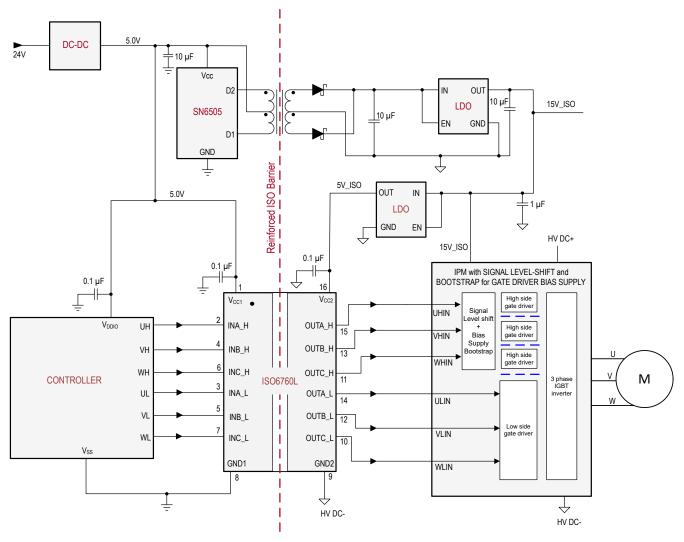


Figure 9-1. Isolation for Intelligent Power Module (IPM) Interface

### 9.2.1 Design Requirements

To design with these devices, use the parameters listed in Table 9-1.

Table 9-1. Design Parameters

PARAMETER	VALUE		
Supply voltage, V <sub>CC1</sub> and V <sub>CC2</sub>	1.71 V to 1.89 V and 2.25 V to 5.5 V		
Decoupling capacitor between V <sub>CC1</sub> and GND1	0.1 μF		
Decoupling capacitor from V <sub>CC2</sub> and GND2	0.1 μF		

### 9.2.2 Detailed Design Procedure

Unlike optocouplers, which require external components to improve performance, provide bias, or limit current, the ISO6760L family of devices only require two external bypass capacitors to operate.

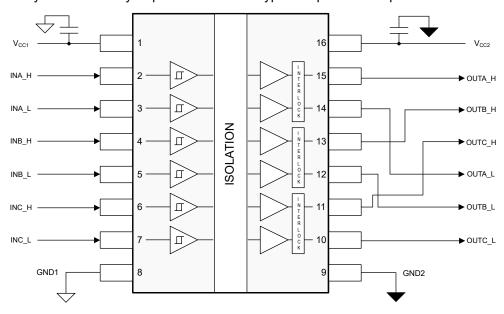


Figure 9-2. Typical ISO6760L Circuit Hook-up



### 9.2.3 Application Curve

The following diagrams of the ISO6760L family of devices show how the hardware interlock circuitry protects against shoot through current. Within a channel pairing, both outputs cannot be high simultaneously. ISO6760L Interlock Diagram shows the ISO6760L demonstrating the hardware interlock on a 200 Hz input signal out of phase between two adjacent channels. ISO6760L Interlock Diagram shows the normal ISO6760 (device offered without interlock circuitry) with a 200 Hz input signal out of phase between two adjacent channels for comparison.



Figure 9-3. ISO6760L Interlock Diagram

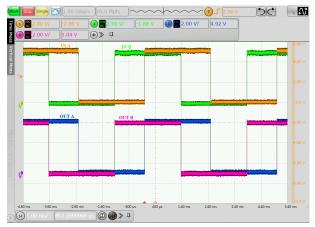


Figure 9-4. ISO6760 (Device without Interlock)

Diagram



#### 10 Insulation Lifetime

Insulation lifetime projection data is collected by using industry-standard Time Dependent Dielectric Breakdown (TDDB) test method. In this test, all pins on each side of the barrier are tied together creating a two-terminal device and high voltage applied between the two sides; See Figure 10-1 for TDDB test setup. The insulation breakdown data is collected at various high voltages switching at 60 Hz over temperature. For reinforced insulation, VDE standard requires the use of TDDB projection line with failure rate of less than 1 part per million (ppm). Even though the expected minimum insulation lifetime is 20 years at the specified working isolation voltage, VDE reinforced certification requires additional safety margin of 20% for working voltage and 50% for lifetime which translates into minimum required insulation lifetime of 30 years at a working voltage that's 20% higher than the specified value.

Figure 10-2 shows the intrinsic capability of the isolation barrier to withstand high voltage stress over its lifetime. Based on the TDDB data, the intrinsic capability of the insulation is  $1500 \, V_{RMS}$  with a lifetime of 36 years. Other factors, such as package size, pollution degree, material group, etc. can further limit the working voltage of the component. The working voltage of DW-16 package is specified upto  $1500 \, V_{RMS}$ . At the lower working voltages, the corresponding insulation lifetime is much longer than 36 years.

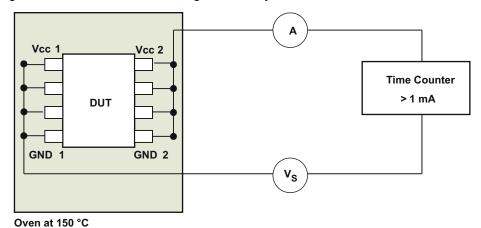


Figure 10-1. Test Setup for Insulation Lifetime Measurement



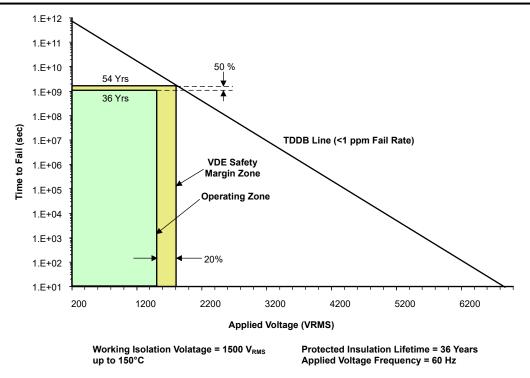


Figure 10-2. Insulation Lifetime Projection Data

## 11 Power Supply Recommendations

To help ensure reliable operation at data rates and supply voltages, a  $0.1-\mu F$  bypass capacitor is recommended at the input and output supply pins ( $V_{CC1}$  and  $V_{CC2}$ ). The capacitors should be placed as close to the supply pins as possible. If only a single primary-side power supply is available in an application, isolated power can be generated for the secondary-side with the help of a transformer driver. For industrial applications, please use Texas Instruments' SN6501 or SN6505B. For such applications, detailed power supply design and transformer selection recommendations are available in SN6501 Transformer Driver for Isolated Power Supplies or SN6505B-Q1 Low-noise, 1-A Transformer Drivers for Isolated Power Supplies.

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# 12 Layout

### 12.1 Layout Guidelines

A minimum of two layers is required to accomplish a low EMI PCB design. To further improve EMI, a four layer board can be used (see Figure 12-2). Layer stacking for a four layer board should be in the following order (top-to-bottom): high-speed signal layer, ground plane, power plane and low-frequency signal layer.

- Routing the high-speed traces on the top layer avoids the use of vias (and the introduction of their inductances) and allows for clean interconnects between the isolator and the transmitter and receiver circuits of the data link.
- Placing a solid ground plane next to the high-speed signal layer establishes controlled impedance for transmission line interconnects and provides an excellent low-inductance path for the return current flow.
- Placing the power plane next to the ground plane creates additional high-frequency bypass capacitance of approximately 100 pF/inch<sup>2</sup>.
- Routing the slower speed control signals on the bottom layer allows for greater flexibility as these signal links usually have margin to tolerate discontinuities such as vias.

If an additional supply voltage plane or signal layer is needed, add a second power or ground plane system to the stack to keep it symmetrical. This makes the stack mechanically stable and prevents it from warping. Also the power and ground plane of each power system can be placed closer together, thus increasing the high-frequency bypass capacitance significantly.

For detailed layout recommendations, refer to the Digital Isolator Design Guide.

#### 12.1.1 PCB Material

For digital circuit boards operating below 150 Mbps, (or rise and fall times higher than 1 ns), and trace lengths of up to 10 inches, use standard FR-4 UL94V-0 printed circuit boards. This PCB is preferred over cheaper alternatives due to its lower dielectric losses at high frequencies, less moisture absorption, greater strength and stiffness, and self-extinguishing flammability-characteristics.

Product Folder Links: ISO67601



# 12.2 Layout Example

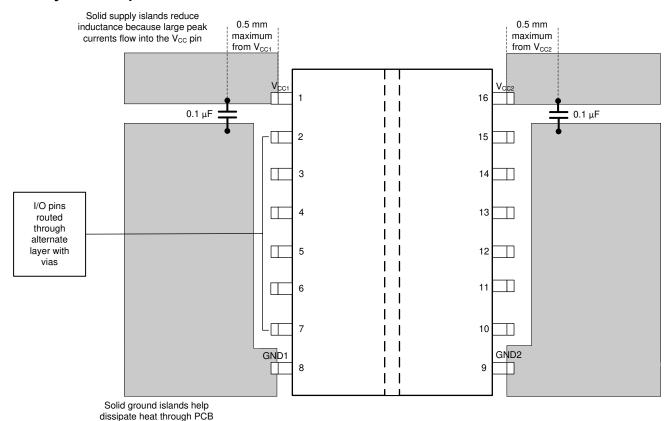


Figure 12-1. Layout Example

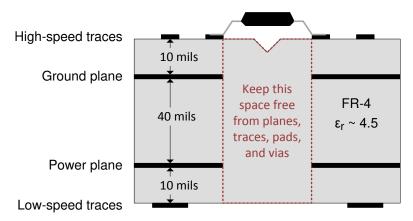


Figure 12-2. Four Layer Board Layout Example

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# 13 Device and Documentation Support

## 13.1 Documentation Support

#### 13.1.1 Related Documentation

For related documentation, see the following:

- Texas Instruments, Digital Isolator Design Guide
- Texas Instruments, Digital Isolator Design Guide
- · Texas Instruments, Isolation Glossary
- Texas Instruments, How to use isolation to improve ESD, EFT, and Surge immunity in industrial systems application report
- Texas Instruments, ADS79xx 12/10/8-Bit, 1 MSPS, 16/12/8/4-Channel, Single-Ended, MicroPower, Serial Interface ADCs data sheet
- Texas Instruments, DAC161P997 Single-Wire 16-bit DAC for 4- to 20-mA Loops data sheet
- Texas Instruments, MSP430G2132Mixed Signal Microcontroller data sheet
- Texas Instruments, SN6501 Transformer Driver for Isolated Power Supplies data sheet
- Texas Instruments, TPS76333Low-Power 150-mA Low-Dropout Linear Regulators data sheet

### 13.2 Receiving Notification of Documentation Updates

To receive notification of documentation updates, navigate to the device product folder on ti.com. In the upper right corner, click on *Alert me* to register and receive a weekly digest of any product information that has changed. For change details, review the revision history included in any revised document.

### 13.3 Support Resources

TI E2E™ support forums are an engineer's go-to source for fast, verified answers and design help — straight from the experts. Search existing answers or ask your own question to get the quick design help you need.

Linked content is provided "AS IS" by the respective contributors. They do not constitute TI specifications and do not necessarily reflect TI's views; see TI's Terms of Use.

#### 13.4 Trademarks

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### 13.5 Electrostatic Discharge Caution



This integrated circuit can be damaged by ESD. Texas Instruments recommends that all integrated circuits be handled with appropriate precautions. Failure to observe proper handling and installation procedures can cause damage.

ESD damage can range from subtle performance degradation to complete device failure. Precision integrated circuits may be more susceptible to damage because very small parametric changes could cause the device not to meet its published specifications.

#### 13.6 Glossary

TI Glossary This glossary lists and explains terms, acronyms, and definitions.

### 14 Mechanical, Packaging, and Orderable Information

The following pages include mechanical packaging and orderable information. This information is the most current data available for the designated devices. This data is subject to change without notice and revision of this document. For browser-based versions of this data sheet, refer to the left-hand navigation.

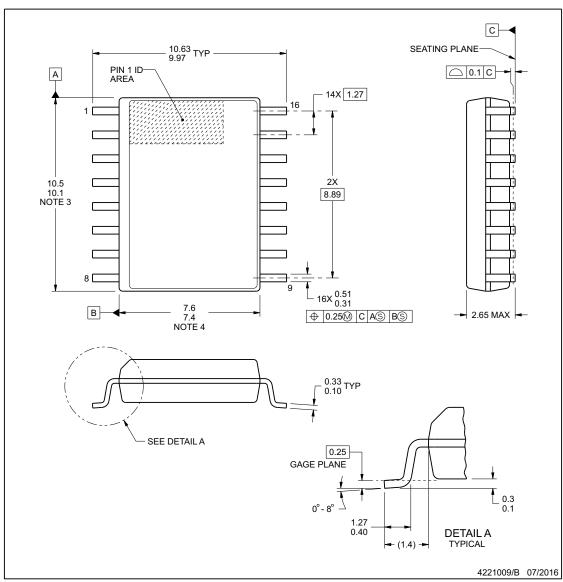
**DW0016B** 





### **PACKAGE OUTLINE**

# SOIC - 2.65 mm max height



### NOTES:

- 1. All linear dimensions are in millimeters. Dimensions in parenthesis are for reference only. Dimensioning and tolerancing per ASME Y14.5M.

  2. This drawing is subject to change without notice.
- 3. This dimension does not include mold flash, protrusions, or gate burrs. Mold flash, protrusions, or gate burrs shall not exceed 0.15 mm, per side.

  4. This dimension does not include interlead flash. Interlead flash shall not exceed 0.25 mm, per side.
- 5. Reference JEDEC registration MS-013.

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Product Folder Links: ISO6760L

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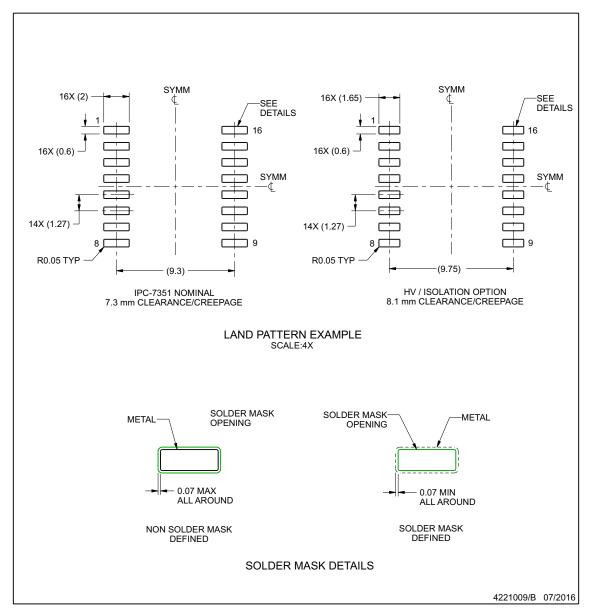


### **EXAMPLE BOARD LAYOUT**

# DW0016B

SOIC - 2.65 mm max height

SOIC



NOTES: (continued)

6. Publication IPC-7351 may have alternate designs.

7. Solder mask tolerances between and around signal pads can vary based on board fabrication site.

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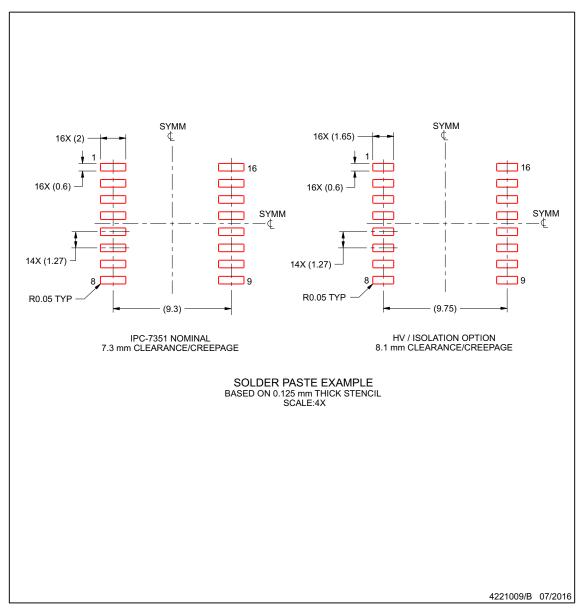


# **EXAMPLE STENCIL DESIGN**

# **DW0016B**

SOIC - 2.65 mm max height

SOIC



NOTES: (continued)

- 8. Laser cutting apertures with trapezoidal walls and rounded corners may offer better paste release. IPC-7525 may have alternate design recommendations
- design recommendations.

  9. Board assembly site may have different recommendations for stencil design.

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www.ti.com 2-May-2025

#### PACKAGING INFORMATION

Orderable part number	Status	Material type	Package   Pins	Package qty   Carrier	RoHS	Lead finish/ Ball material	MSL rating/ Peak reflow	Op temp (°C)	Part marking (6)
						(4)	(5)		
ISO6760LDWR	Active	Production	SOIC (DW)   16	2000   LARGE T&R	Yes	NIPDAU	Level-2-260C-1 YEAR	-40 to 125	ISO6760L
ISO6760LNDWR	Active	Production	SOIC (DW)   16	2000   LARGE T&R	Yes	NIPDAU	Level-2-260C-1 YEAR	-40 to 125	ISO6760LN

<sup>(1)</sup> Status: For more details on status, see our product life cycle.

Multiple part markings will be inside parentheses. Only one part marking contained in parentheses and separated by a "~" will appear on a part. If a line is indented then it is a continuation of the previous line and the two combined represent the entire part marking for that device.

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<sup>(2)</sup> Material type: When designated, preproduction parts are prototypes/experimental devices, and are not yet approved or released for full production. Testing and final process, including without limitation quality assurance, reliability performance testing, and/or process qualification, may not yet be complete, and this item is subject to further changes or possible discontinuation. If available for ordering, purchases will be subject to an additional waiver at checkout, and are intended for early internal evaluation purposes only. These items are sold without warranties of any kind.

<sup>(3)</sup> RoHS values: Yes, No, RoHS Exempt. See the TI RoHS Statement for additional information and value definition.

<sup>(4)</sup> Lead finish/Ball material: Parts may have multiple material finish options. Finish options are separated by a vertical ruled line. Lead finish/Ball material values may wrap to two lines if the finish value exceeds the maximum column width.

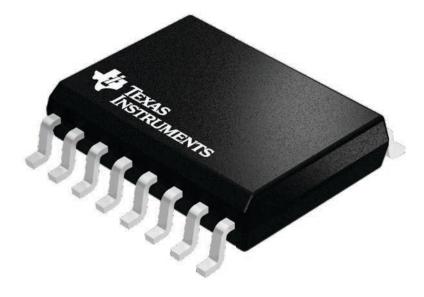
<sup>(5)</sup> MSL rating/Peak reflow: The moisture sensitivity level ratings and peak solder (reflow) temperatures. In the event that a part has multiple moisture sensitivity ratings, only the lowest level per JEDEC standards is shown. Refer to the shipping label for the actual reflow temperature that will be used to mount the part to the printed circuit board.

<sup>(6)</sup> Part marking: There may be an additional marking, which relates to the logo, the lot trace code information, or the environmental category of the part.

7.5 x 10.3, 1.27 mm pitch

SMALL OUTLINE INTEGRATED CIRCUIT

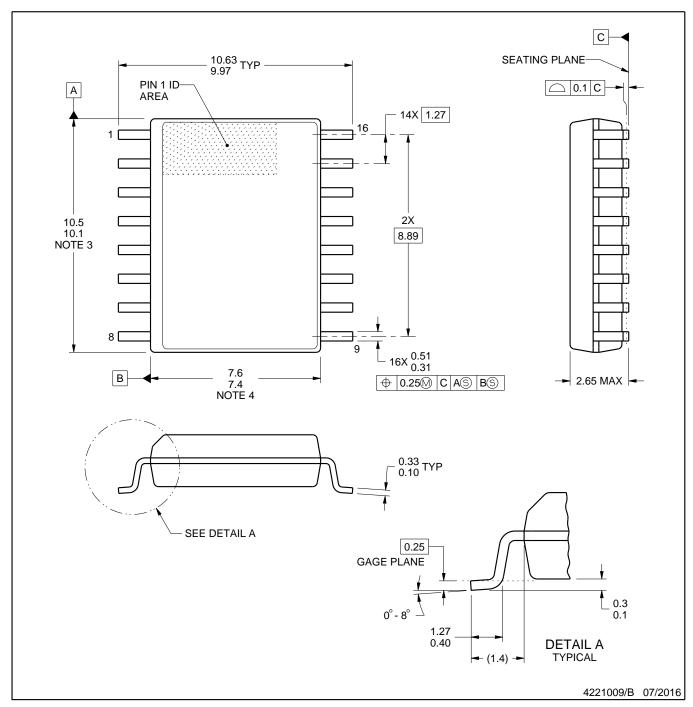
This image is a representation of the package family, actual package may vary. Refer to the product data sheet for package details.







SOIC



#### NOTES:

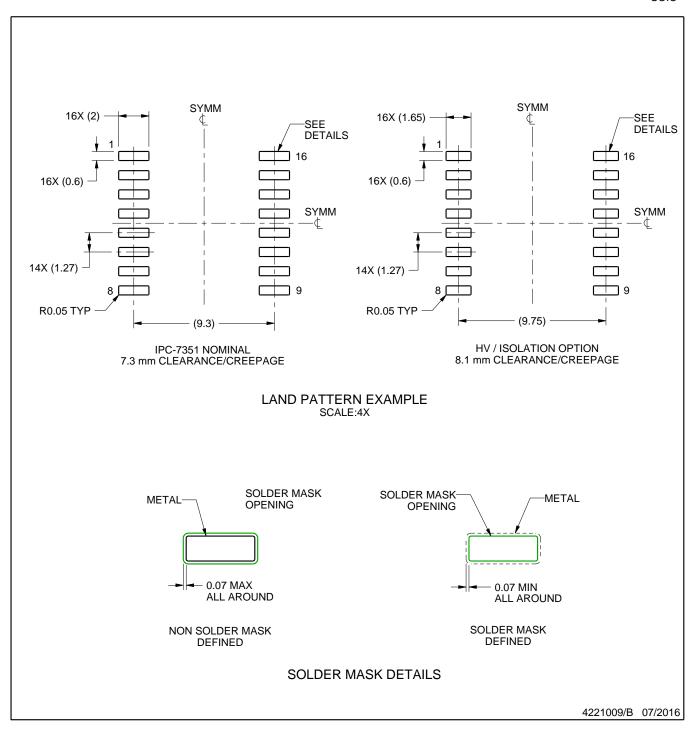
- 1. All linear dimensions are in millimeters. Dimensions in parenthesis are for reference only. Dimensioning and tolerancing
- per ASME Y14.5M.

  2. This drawing is subject to change without notice.

  3. This dimension does not include mold flash, protrusions, or gate burrs. Mold flash, protrusions, or gate burrs shall not exceed 0.15 mm, per side.
- 4. This dimension does not include interlead flash. Interlead flash shall not exceed 0.25 mm, per side.
- 5. Reference JEDEC registration MS-013.



SOIC

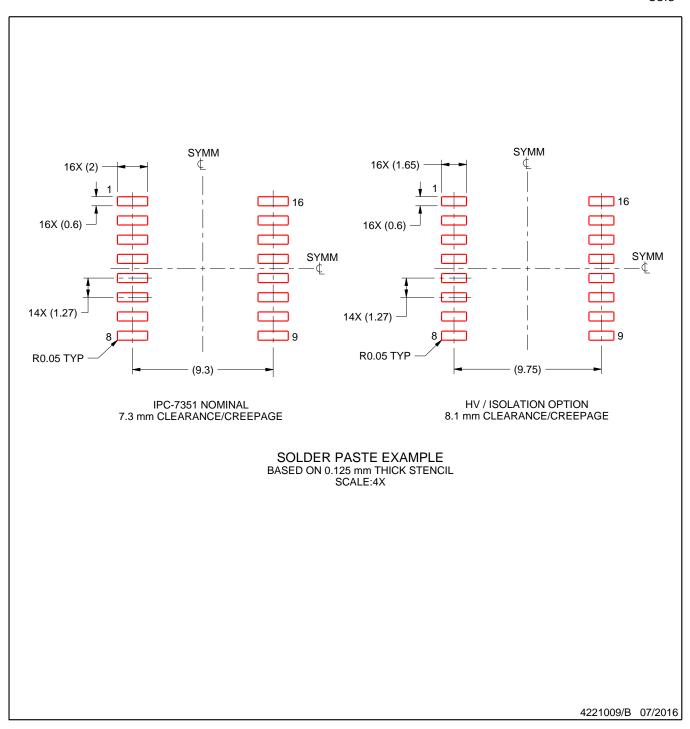


NOTES: (continued)

6. Publication IPC-7351 may have alternate designs.

7. Solder mask tolerances between and around signal pads can vary based on board fabrication site.

SOIC



#### NOTES: (continued)

- 8. Laser cutting apertures with trapezoidal walls and rounded corners may offer better paste release. IPC-7525 may have alternate design recommendations.
- 9. Board assembly site may have different recommendations for stencil design.



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