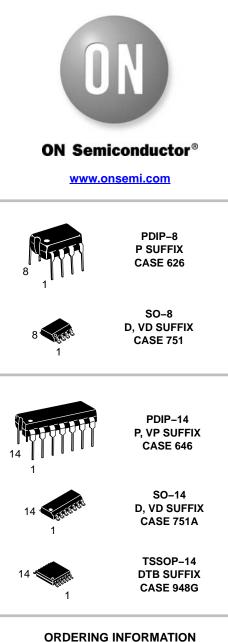
# Operational Amplifiers, Single Supply 3.0 V to 44 V, Low Power

Quality bipolar fabrication with innovative design concepts are employed for the MC33171/72/74, NCV33172/74 series of monolithic operational amplifiers. These devices operate at 180  $\mu$ A per amplifier and offer 1.8 MHz of gain bandwidth product and 2.1 V/ $\mu$ s slew rate without the use of JFET device technology. Although this series can be operated from split supplies, it is particularly suited for single supply operation, since the common mode input voltage includes ground potential (V<sub>EE</sub>). With a Darlington input stage, these devices exhibit high input resistance, low input offset voltage and high gain. The all NPN output stage, characterized by no deadband crossover distortion and large output voltage swing, provides high capacitance drive capability, excellent phase and gain margins, low open loop high frequency output impedance and symmetrical source/sink AC frequency response.

The MC33171/72/74, NCV33172/74 are specified over the industrial/automotive temperature ranges. The complete series of single, dual and quad operational amplifiers are available in plastic as well as the surface mount packages.

# Features

- Low Supply Current: 180 µA (Per Amplifier)
- Wide Supply Operating Range: 3.0 V to 44 V or  $\pm 1.5$  V to  $\pm 22$  V
- Wide Input Common Mode Range, Including Ground (V<sub>EE</sub>)
- Wide Bandwidth: 1.8 MHz
- High Slew Rate: 2.1 V/µs
- Low Input Offset Voltage: 2.0 mV
- Large Output Voltage Swing: -14.2 V to +14.2 V (with ±15 V Supplies)
- Large Capacitance Drive Capability: 0 pF to 500 pF
- Low Total Harmonic Distortion: 0.03%
- Excellent Phase Margin: 60°
- Excellent Gain Margin: 15 dB
- Output Short Circuit Protection
- ESD Diodes Provide Input Protection for Dual and Quad
- NCV Prefix for Automotive and Other Applications Requiring Unique Site and Control Change Requirements; AEC–Q100 Qualified and PPAP Capable
- These Devices are Pb–Free, Halogen Free/BFR Free and are RoHS Compliant

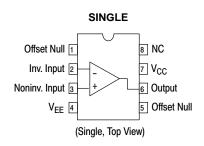


See detailed ordering and shipping information in the package dimensions section on page 9 of this data sheet.

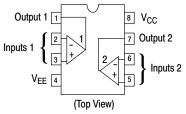
# **DEVICE MARKING INFORMATION**

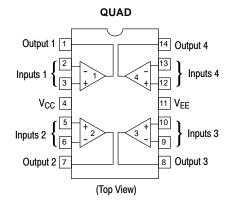
See general marking information in the device marking section on page 10 of this data sheet.

# **PIN CONNECTIONS**









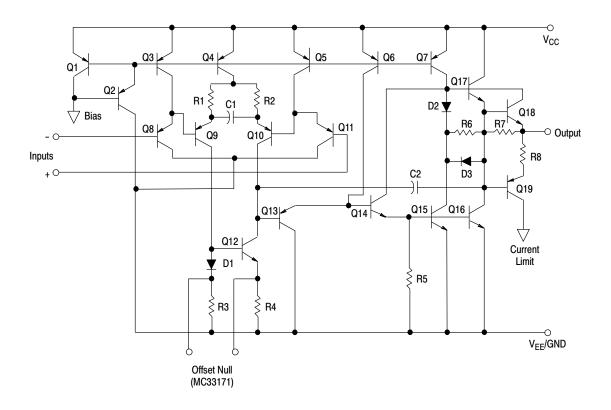


Figure 1. Representative Schematic Diagram (Each Amplifier)

## MAXIMUM RATINGS

Rating	Symbol	Value	Unit
Supply Voltage	V <sub>CC</sub> /V <sub>EE</sub>	±22	V
Input Differential Voltage Range	V <sub>IDR</sub>	(Note 1)	V
Input Voltage Range	V <sub>IR</sub>	(Note 1)	V
Output Short Circuit Duration (Note 2)	t <sub>SC</sub>	Indefinite	sec
Operating Ambient Temperature Range	T <sub>A</sub>	(Note 3)	°C
Operating Junction Temperature	TJ	+150	°C
Storage Temperature Range	T <sub>stg</sub>	-65 to +150	°C

Stresses exceeding those listed in the Maximum Ratings table may damage the device. If any of these limits are exceeded, device functionality should not be assumed, damage may occur and reliability may be affected.

Characteristics	Symbol	Min	Тур	Max	Unit
Input Offset Voltage ( $V_{CM} = 0 V$ ) $V_{CC} = +15 V$ , $V_{EE} = -15 V$ , $T_A = +25^{\circ}C$ $V_{CC} = +5.0 V$ , $V_{EE} = 0 V$ , $T_A = +25^{\circ}C$ $V_{CC} = +15 V$ , $V_{EE} = -15 V$ , $T_A = T_{low}$ to $T_{high}$ (Note 3)	V <sub>IO</sub>		2.0 2.5 -	4.5 5.0 6.5	mV
Average Temperature Coefficient of Offset Voltage	$\Delta V_{IO} / \Delta T$	-	10	-	μV/°C
Input Bias Current ( $V_{CM} = 0 V$ ) $T_A = +25^{\circ}C$ $T_A = T_{low}$ to $T_{high}$ (Note 3)	Ι <sub>ΙΒ</sub>		20 -	100 200	nA
Input Offset Current (V <sub>CM</sub> = 0 V) $T_A = +25^{\circ}C$ $T_A = T_{low}$ to $T_{high}$ (Note 3)	Ι <sub>ΙΟ</sub>		5.0 -	20 40	nA
Large Signal Voltage Gain (V <sub>O</sub> = $\pm$ 10 V, R <sub>L</sub> = 10 k) T <sub>A</sub> = +25°C T <sub>A</sub> = T <sub>low</sub> to T <sub>high</sub> (Note 3)	A <sub>VOL</sub>	50 25	500 -		V/mV
Output Voltage Swing $V_{CC} = +5.0 \text{ V}, V_{EE} = 0 \text{ V}, R_L = 10 \text{ k}, T_A = +25^{\circ}\text{C}$ $V_{CC} = +15 \text{ V}, V_{EE} = -15 \text{ V}, R_L = 10 \text{ k}, T_A = +25^{\circ}\text{C}$ $V_{CC} = +15 \text{ V}, V_{EE} = -15 \text{ V}, R_L = 10 \text{ k}, T_A = T_{low} \text{ to } T_{high} \text{ (Note 3)}$	V <sub>OH</sub>	3.5 13.6 13.3	4.3 14.2 -	- - -	V
	V <sub>OL</sub>	- - -	0.05 -14.2 -	0.15 -13.6 -13.3	
Output Short Circuit (T <sub>A</sub> = +25°C) Input Overdrive = 1.0 V, Output to Ground Source Sink	I <sub>SC</sub>	3.0 15	5.0 27		mA
Input Common Mode Voltage Range $T_A = +25^{\circ}C$ $T_A = T_{low}$ to $T_{high}$ (Note 3)	V <sub>ICR</sub>		to (V <sub>CC</sub> – to (V <sub>CC</sub> –		V
Common Mode Rejection Ratio (R <sub>S</sub> $\leq$ 10 k), T <sub>A</sub> = +25°C	CMRR	80	90	-	dB
Power Supply Rejection Ratio (R <sub>S</sub> = 100 $\Omega$ ), T <sub>A</sub> = +25°C	PSRR	80	100	-	dB
Power Supply Current (Per Amplifier) $V_{CC} = +5.0 \text{ V}, V_{EE} = 0 \text{ V}, T_A = +25^{\circ}\text{C}$ $V_{CC} = +15 \text{ V}, V_{EE} = -15 \text{ V}, T_A = +25^{\circ}\text{C}$ $V_{CC} = +15 \text{ V}, V_{EE} = -15 \text{ V}, T_A = T_{Iow} \text{ to } T_{high} \text{ (Note 3)}$	ID		180 220 -	250 250 300	μΑ

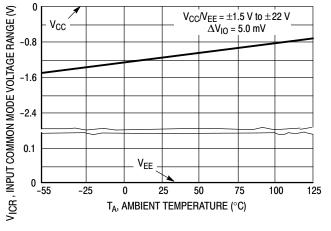
# DC ELECTRICAL CHARACTERISTICS (V<sub>CC</sub> = +15 V, V<sub>EE</sub> = -15 V, R<sub>L</sub> connected to ground, T<sub>A</sub> = +25°C, unless otherwise noted.)

1. Either or both input voltages must not exceed the magnitude of V<sub>CC</sub> or V<sub>EE</sub>.2. Power dissipation must be considered to ensure maximum junction temperature (T<sub>J</sub>) is not exceeded.3. MC3317x $T_{low} = -40^{\circ}C$ MC3317xV, NCV3317x $T_{low} = -40^{\circ}C$ Thigh = +85^{\circ}CMC3317xV, NCV3317xThe endowed of the endowed o

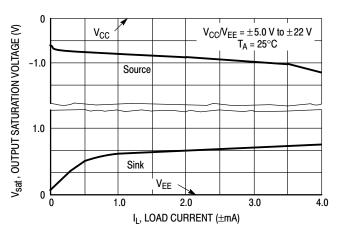
Characteristics	Symbol	Min	Тур	Max	Unit
Slew Rate (V <sub>in</sub> = -10 V to +10 V, R <sub>L</sub> = 10 k, C <sub>L</sub> = 100 pF) A <sub>V</sub> +1 A <sub>V</sub> -1	SR	1.6 _	2.1 2.1		V/µs
Gain Bandwidth Product (f = 100 kHz)	GBW	1.4	1.8	_	MHz
Power Bandwidth $A_V = +1.0 R_L = 10 k$ , $V_O = 20 V_{pp}$ , THD = 5%	BWp	_	35	_	kHz
Phase Margin $R_L = 10 k$ $R_L = 10 k$ , $C_L = 100 pF$	φ <sub>m</sub>		60 45		Deg
Gain Margin $R_L = 10 \text{ k}$ $R_L = 10 \text{ k}$ , $C_L = 100 \text{ pF}$	A <sub>m</sub>		15 5.0		dB
Equivalent Input Noise Voltage $R_S = 100 \ \Omega$ , f = 1.0 kHz	e <sub>n</sub>	-	32	-	nV/√ <u>H</u> z
Equivalent Input Noise Current (f = 1.0 kHz)	۱ <sub>n</sub>	_	0.2	_	pA/√Hz
Differential Input Resistance $V_{cm} = 0 V$	R <sub>in</sub>	_	300	_	MΩ
Input Capacitance	C <sub>in</sub>	-	0.8	-	pF
Total Harmonic Distortion AV = +10, RL = 10 k, 2.0 Vpp $\leq$ V0 $\leq$ 20 Vpp, f = 10 kHz	THD	_	0.03	_	%
Channel Separation (f = 10 kHz)	CS	-	120	-	dB
Open Loop Output Impedance (f = 1.0 MHz)	z <sub>o</sub>	_	100	_	Ω

# AC ELECTRICAL CHARACTERISTICS ( $V_{CC}$ = +15 V, $V_{EE}$ = -15 V, $R_L$ connected to ground, $T_A$ = +25°C, unless otherwise noted.)

Product parametric performance is indicated in the Electrical Characteristics for the listed test conditions, unless otherwise noted. Product performance may not be indicated by the Electrical Characteristics if operated under different conditions.









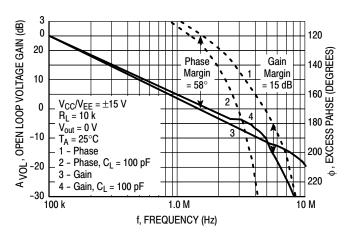


Figure 4. Open Loop Voltage Gain and Phase versus Frequency

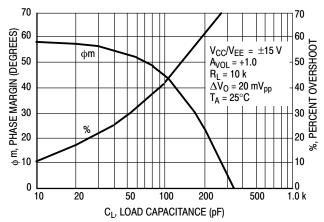


Figure 5. Phase Margin and Percent Overshoot versus Load Capacitance

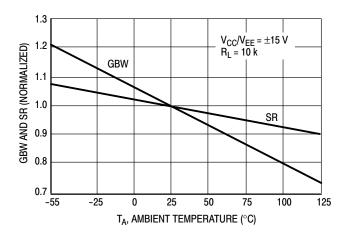
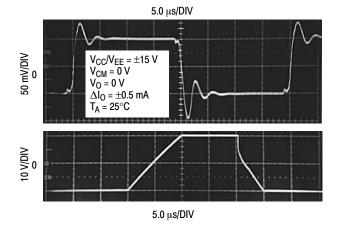
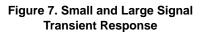


Figure 6. Normalized Gain Bandwidth Product and Slew Rate versus Temperature





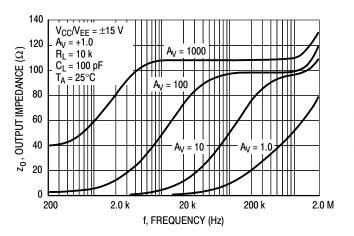


Figure 8. Output Impedance and Frequency

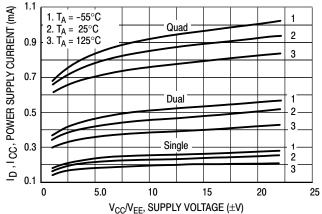


Figure 9. Supply Current versus Supply Voltage

## APPLICATIONS INFORMATION – CIRCUIT DESCRIPTION/PERFORMANCE FEATURES

Although the bandwidth, slew rate, and settling time of the MC33171/72/74 amplifier family is similar to low power op amp products utilizing JFET input devices, these amplifiers offer additional advantages as a result of the PNP transistor differential inputs and an all NPN transistor output stage.

Because the input common mode voltage range of this input stage includes the  $V_{EE}$  potential, single supply operation is feasible to as low as 3.0 V with the common mode input voltage at ground potential.

The input stage also allows differential input voltages up to  $\pm 44$  V, provided the maximum input voltage range is not exceeded. Specifically, the input voltages must range between  $V_{CC}$  and  $V_{EE}$  supply voltages as shown by the maximum rating table. In practice, although not recommended, the input voltages can exceed the V<sub>CC</sub> voltage by approximately 3.0 V and decrease below the VEE voltage by 0.3 V without causing product damage, although output phase reversal may occur. It is also possible to source up to 5.0 mA of current from  $V_{EE}$  through either inputs' clamping diode without damage or latching, but phase reversal may again occur. If at least one input is within the common mode input voltage range and the other input is within the maximum input voltage range, no phase reversal will occur. If both inputs exceed the upper common mode input voltage limit, the output will be forced to its lowest voltage state.

Since the input capacitance associated with the small geometry input device is substantially lower (0.8 pF) than that of a typical JFET (3.0 pF), the frequency response for a given input source resistance is greatly enhanced. This becomes evident in D-to-A current to voltage conversion applications where the feedback resistance can form a pole with the input capacitance of the op amp. This input pole creates a 2nd Order system with the single pole op amp and is therefore detrimental to its settling time. In this context, lower input capacitance is desirable especially for higher values of feedback resistances (lower current DACs). This input pole can be compensated for by creating a feedback zero with a capacitance across the feedback resistance, if necessary, to reduce overshoot. For 10 k $\Omega$  of feedback resistance, the MC33171/72/74 family can typically settle to within 1/2 LSB of 8 bits in 4.2 µs, and within 1/2 LSB of 12 bits in 4.8 µs for a 10 V step. In a standard inverting unity gain fast settling configuration, the symmetrical slew rate is typically  $\pm 2.1$  V/µs. In the classic noninverting unity gain configuration the typical output positive slew rate is also 2.1 V/µs, and the corresponding negative slew rate will usually exceed the positive slew rate as a function of the fall time of the input waveform.

The all NPN output stage, shown in its basic form on the equivalent circuit schematic, offers unique advantages over the more conventional NPN/PNP transistor Class AB output stage. A 10 k $\Omega$  load resistance can typically swing within

0.8 V of the positive rail (V<sub>CC</sub>) and negative rail (V<sub>EE</sub>), providing a 28.4 Vpp swing from ±15 V supplies. This large output swing becomes most noticeable at lower supply voltages.

The positive swing is limited by the saturation voltage of the current source transistor Q7, the V<sub>BE</sub> of the NPN pull–up transistor Q17, and the voltage drop associated with the short circuit resistance, R5. For sink currents less than 0.4 mA, the negative swing is limited by the saturation voltage of the pull–down transistor Q15, and the voltage drop across R4 and R5. For small valued sink currents, the above voltage drops are negligible, allowing the negative swing voltage to approach within millivolts of V<sub>EE</sub>. For sink currents (> 0.4 mA), diode D3 clamps the voltage across R4. Thus the negative swing is limited by the saturation voltage of Q15, plus the forward diode drop of D3 ( $\approx$ V<sub>EE</sub> +1.0 V). Therefore an unprecedented peak–to–peak output voltage swing is possible for a given supply voltage as indicated by the output swing specifications.

If the load resistance is referenced to  $V_{CC}$  instead of ground for single supply applications, the maximum possible output swing can be achieved for a given supply voltage. For light load currents, the load resistance will pull the output to  $V_{CC}$  during the positive swing and the output will pull the load resistance near ground during the negative swing. The load resistance value should be much less than that of the feedback resistance to maximize pull–up capability.

Because the PNP output emitter–follower transistor has been eliminated, the MC33171/72/74 family offers a 15 mA minimum current sink capability, typically to an output voltage of ( $V_{EE}$  +1.8 V). In single supply applications the output can directly source or sink base current from a common emitter NPN transistor for current switching applications.

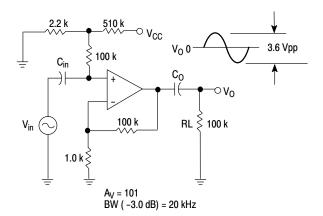
In addition, the all NPN transistor output stage is inherently faster than PNP types, contributing to the bipolar amplifier's improved gain bandwidth product. The associated high frequency low output impedance (200  $\Omega$  typ @ 1.0 MHz) allows capacitive drive capability from 0 pF to 400 pF without oscillation in the noninverting unity gain configuration. The 60° phase margin and 15 dB gain margin, as well as the general gain and phase characteristics, are virtually independent of the source/sink output swing conditions. This allows easier system phase compensation, since output swing will not be a phase consideration. The AC characteristics of the MC33171/72/74 family also allow excellent active filter capability, especially for low voltage single supply applications.

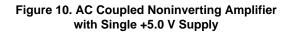
Although the single supply specification is defined at 5.0 V, these amplifiers are functional to at least 3.0 V @  $25^{\circ}$ C. However slight changes in parametrics such as bandwidth, slew rate, and DC gain may occur.

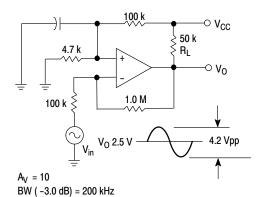
If power to this integrated circuit is applied in reverse polarity, or if the IC is installed backwards in a socket, large unlimited current surges will occur through the device that may result in device destruction.

As usual with most high frequency amplifiers, proper lead dress, component placement and PC board layout should be exercised for optimum frequency performance. For example, long unshielded input or output leads may result in unwanted input/output coupling. In order to preserve the relatively low input capacitance associated with these amplifiers, resistors connected to the inputs should be immediately adjacent to the input pin to minimize additional stray input capacitance. This not only minimizes the input pole for optimum frequency response, but also minimizes extraneous "pick up" at this node. Supply decoupling with adequate capacitance immediately adjacent to the supply pin is also important, particularly over temperature, since many types of decoupling capacitors exhibit great impedance changes over temperature.

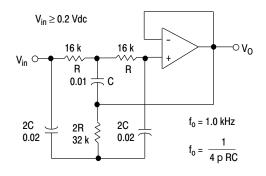
The output of any one amplifier is current limited and thus protected from a direct short to ground. However, under such conditions, it is important not to allow the device to exceed the maximum junction temperature rating. Typically for  $\pm 15$  V supplies, any one output can be shorted continuously to ground without exceeding the maximum temperature rating.













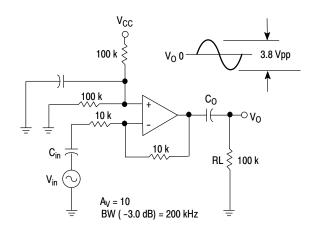
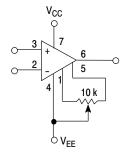


Figure 11. AC Coupled Inverting Amplifier with Single +5.0 V Supply



Offset Nulling range is approximately  $\pm 80 \text{ mV}$  with a 10 k potentiometer, MC33171 only.

Figure 13. Offset Nulling Circuit

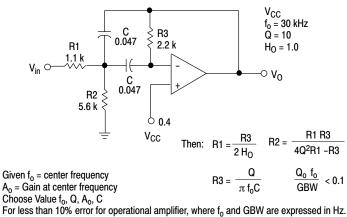


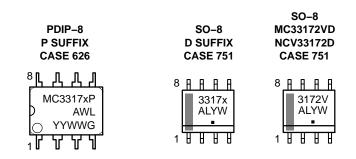
Figure 15. Active Bandpass Filter

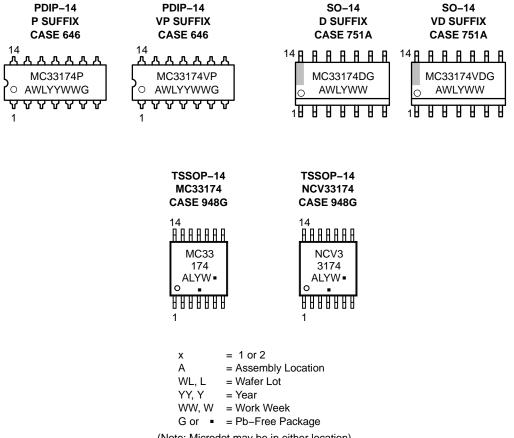
# **ORDERING INFORMATION**

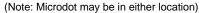
Op Amp Function	Device	Operating Temperature Range	Package	Shipping <sup>†</sup>
	MC33171DG		SO–8 (Pb–Free)	98 Units/Rail
Single	MC33171DR2G	C33171DR2G $T_{A} = -40^{\circ} \text{ to } +85^{\circ}\text{C}$	SO-8 (Pb-Free)	2500 / Tape & Reel
	MC33171PG		PDIP (Pb-Free)	50 Units/Rail
	MC33172DG		SO-8 (Pb-Free)	98 Units/Rail
	MC33172DR2G	$T_A = -40^\circ$ to +85°C	SO-8 (Pb-Free)	2500 / Tape & Reel
Dual	MC33172PG		PDIP (Pb-Free)	50 Units/Rail
Duai	MC33172VDG	T <sub>A</sub> = −40° to +125°C	SO-8 (Pb-Free)	98 Units/Rail
	MC33172VDR2G		SO-8 (Pb-Free)	2500 / Tape & Reel
	NCV33172DR2G*		SO-8 (Pb-Free)	2500 / Tape & Reel
	MC33174DG		SO-14 (Pb-Free)	55 Units/Rail
	MC33174DR2G		SO-14 (Pb-Free)	2500 / Tape & Reel
	MC33174DTBG	$T_A = -40^\circ$ to +85°C	TSSOP-14 (Pb-Free)	96 Units/Rail
	MC33174DTBR2G		TSSOP-14 (Pb-Free)	2500 / Tape & Reel
Quad	MC33174PG		PDIP (Pb-Free)	25 Units/Rail
Ν	MC33174VDG		SO-14 (Pb-Free)	55 Units/Rail
	MC33174VDR2G	T <sub>A</sub> = −40° to +125°C	SO-14 (Pb-Free)	2500 / Tape & Reel
	MC33174VPG	$r_{A} = -40 \ 10 \pm 125 \ C$	PDIP (Pb-Free)	25 Units/Rail
	NCV33174DTBR2G*	F	TSSOP-14 (Pb-Free)	2500 / Tape & Reel

†For information on tape and reel specifications, including part orientation and tape sizes, please refer to our Tape and Reel Packaging Specifications Brochure, BRD8011/D.
 \*NCV prefix for automotive and other applications requiring site and change controls.

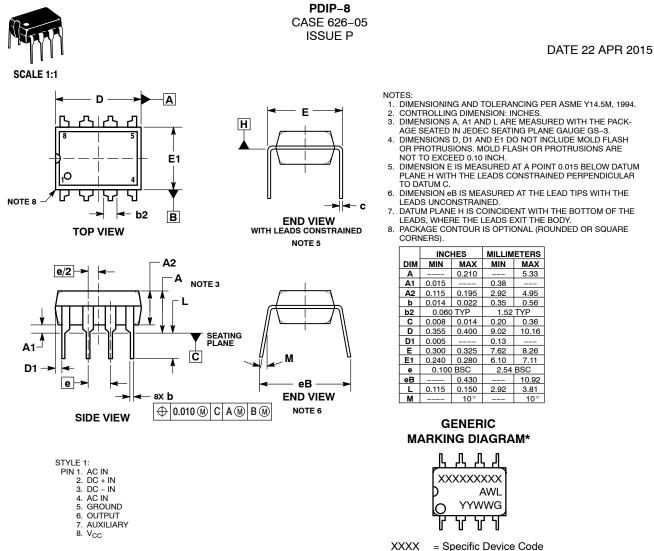
# **MARKING DIAGRAMS**







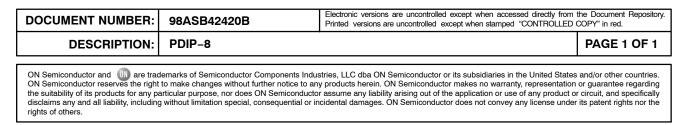


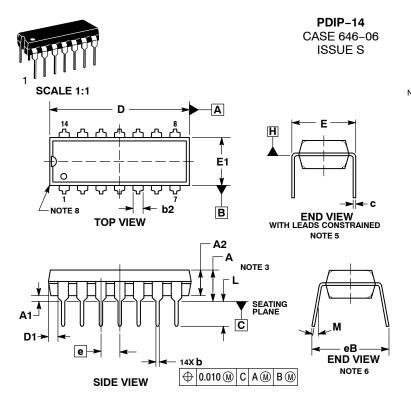


A = Assembly Location

- WL = Wafer Lot
- YY = Year
- WW = Work Week G = Pb-Free Package

\*This information is generic. Please refer to device data sheet for actual part marking. Pb–Free indicator, "G" or microdot " ■", may or may not be present.





**STYLES ON PAGE 2** 

#### **ON Semiconductor**

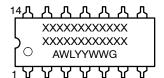


#### DATE 22 APR 2015

- NOTES:
  DIMENSIONING AND TOLERANCING PER ASME Y14.5M, 1994.
  CONTROLLING DIMENSION: INCHES.
  DIMENSIONS A, A1 AND L ARE MEASURED WITH THE PACK-AGE SEATED IN JEDEC SEATING PLANE GAUGE GS-3.
  DIMENSIONS D, D1 AND E1 DO NOT INCLUDE MOLD FLASH OR PROTRUSIONS. MOLD FLASH OR PROTRUSIONS ARE NOT DE VICE DA 10 INCH.
- NOT TO EXCEED 0.10 INCH. DIMENSION E IS MEASURED AT A POINT 0.015 BELOW DATUM 5. PLANE H WITH THE LEADS CONSTRAINED PERPENDICULAR TO DATUM C.
- 6.
- DIMENSION & BIS MEASURED AT THE LEAD TIPS WITH THE LEADS UNCONSTRAINED. DATUM PLANE H IS COINCIDENT WITH THE BOTTOM OF THE LEADS, WHERE THE LEADS EXIT THE BODY. PACKAGE CONTOUR IS OPTIONAL (ROUNDED OR SQUARE CORDINED) 7.
- 8. CORNERS).

	INCHES		MILLIM	ETERS
DIM	MIN	MAX	MIN	MAX
Α		0.210		5.33
A1	0.015		0.38	
A2	0.115	0.195	2.92	4.95
b	0.014	0.022	0.35	0.56
b2	0.060 TYP		1.52 TYP	
С	0.008	0.014	0.20	0.36
D	0.735	0.775	18.67	19.69
D1	0.005		0.13	
Е	0.300	0.325	7.62	8.26
E1	0.240	0.280	6.10	7.11
е	0.100 BSC		2.54	BSC
eB		0.430		10.92
L	0.115	0.150	2.92	3.81
М		10°		10°

## GENERIC **MARKING DIAGRAM\***



XXXXX = Specific Device Code

- = Assembly Location
- WL = Wafer Lot
- YY = Year

А

G

- ww = Work Week
  - = Pb-Free Package

\*This information is generic. Please refer to device data sheet for actual part marking. Pb-Free indicator, "G" or microdot " .", may or may not be present.

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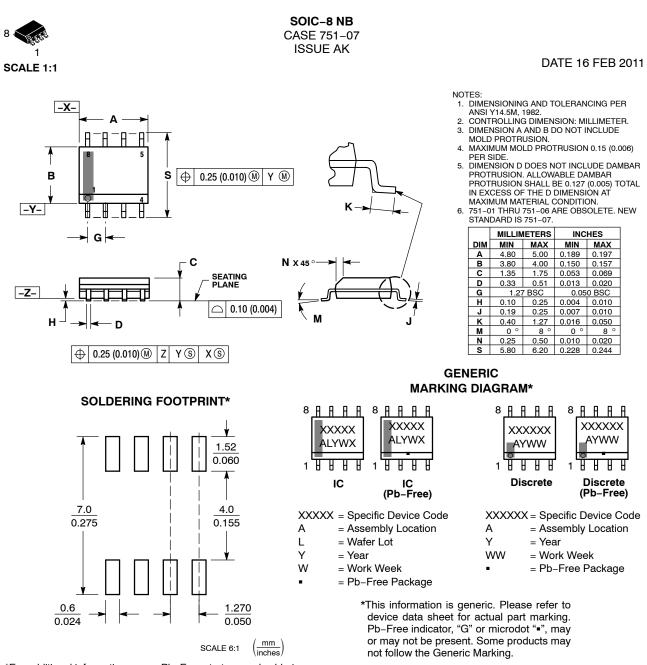
### PDIP-14 CASE 646-06 ISSUE S

# DATE 22 APR 2015

STYLE 1: PIN 1. COLLECTOR 2. BASE 3. EMITTER 4. NO CONNECTION 5. EMITTER 6. BASE 7. COLLECTOR 8. COLLECTOR 9. BASE 10. EMITTER 11. NO CONNECTION 12. EMITTER 13. BASE 14. COLLECTOR	STYLE 2: CANCELLED	STYLE 3: CANCELLED	STYLE 4: PIN 1. DRAIN 2. SOURCE 3. GATE 4. NO CONNECTION 5. GATE 6. SOURCE 7. DRAIN 8. DRAIN 9. SOURCE 10. GATE 11. NO CONNECTION 12. GATE 13. SOURCE 14. DRAIN
STYLE 5: PIN 1. GATE 2. DRAIN 3. SOURCE 4. NO CONNECTION 5. SOURCE 6. DRAIN 7. GATE 9. DRAIN 10. SOURCE 11. NO CONNECTION 12. SOURCE 13. DRAIN 14. GATE	STYLE 6: PIN 1. COMMON CATHODE 2. ANODE/CATHODE 3. ANODE/CATHODE 4. NO CONNECTION 5. ANODE/CATHODE 6. NO CONNECTION 7. ANODE/CATHODE 8. ANODE/CATHODE 9. ANODE/CATHODE 10. NO CONNECTION 11. ANODE/CATHODE 12. ANODE/CATHODE 13. NO CONNECTION 14. COMMON ANODE	STYLE 7: PIN 1. NO CONNECTION 2. ANODE 3. ANODE 4. NO CONNECTION 5. ANODE 6. NO CONNECTION 7. ANODE 8. ANODE 9. ANODE 10. NO CONNECTION 11. ANODE 12. ANODE 13. NO CONNECTION 14. COMMON CATHODE	STYLE 8: PIN 1. NO CONNECTION 2. CATHODE 3. CATHODE 4. NO CONNECTION 5. CATHODE 6. NO CONNECTION 7. CATHODE 9. CATHODE 10. NO CONNECTION 11. CATHODE 12. CATHODE 13. NO CONNECTION 14. COMMON ANODE
STYLE 9: PIN 1. COMMON CATHODE 2. ANODE/CATHODE 3. ANODE/CATHODE 4. NO CONNECTION 5. ANODE/CATHODE 6. ANODE/CATHODE 7. COMMON ANODE 8. COMMON ANODE 9. ANODE/CATHODE 10. ANODE/CATHODE 11. NO CONNECTION 12. ANODE/CATHODE 13. ANODE/CATHODE 14. COMMON CATHODE	STYLE 10: PIN 1. COMMON CATHODE 2. ANODE/CATHODE 3. ANODE/CATHODE 4. ANODE/CATHODE 5. ANODE/CATHODE 6. NO CONNECTION 7. COMMON ANODE 8. COMMON CATHODE 9. ANODE/CATHODE 11. ANODE/CATHODE 12. ANODE/CATHODE 13. NO CONNECTION 14. COMMON ANODE	STYLE 11: PIN 1. CATHODE 2. CATHODE 3. CATHODE 4. CATHODE 6. CATHODE 7. CATHODE 8. ANODE 9. ANODE 10. ANODE 11. ANODE 12. ANODE 13. ANODE 14. ANODE	STYLE 12: PIN 1. COMMON CATHODE 2. COMMON ANODE 3. ANODE/CATHODE 4. ANODE/CATHODE 5. ANODE/CATHODE 6. COMMON ANODE 7. COMMON CATHODE 8. ANODE/CATHODE 10. ANODE/CATHODE 11. ANODE/CATHODE 12. ANODE/CATHODE 13. ANODE/CATHODE 14. ANODE/CATHODE 14. ANODE/CATHODE

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#### SOIC-8 NB CASE 751-07 **ISSUE AK**

STYLE 1: PIN 1. EMITTER COLLECTOR 2. COLLECTOR 3. 4. EMITTER EMITTER 5. BASE 6. 7 BASE EMITTER 8. STYLE 5: PIN 1. DRAIN 2. DRAIN 3. DRAIN DRAIN 4. GATE 5. 6. GATE SOURCE 7. 8. SOURCE STYLE 9: PIN 1. EMITTER, COMMON COLLECTOR, DIE #1 COLLECTOR, DIE #2 2. З. EMITTER, COMMON 4. 5. EMITTER, COMMON 6 BASE. DIE #2 BASE, DIE #1 7. 8. EMITTER, COMMON STYLE 13: PIN 1. N.C. 2. SOURCE 3 GATE 4. 5. DRAIN 6. DRAIN DRAIN 7. DRAIN 8. STYLE 17: PIN 1. VCC 2. V2OUT V10UT З. TXE 4. 5. RXE 6. VFF 7. GND 8. ACC STYLE 21: PIN 1. CATHODE 1 2. CATHODE 2 3 CATHODE 3 CATHODE 4 4. 5. CATHODE 5 6. COMMON ANODE COMMON ANODE 7. 8. CATHODE 6 STYLE 25: PIN 1. VIN 2 N/C REXT З. 4. GND 5. IOUT 6. IOUT IOUT 7. 8. IOUT STYLE 29: BASE, DIE #1 PIN 1. 2 EMITTER, #1 BASE, #2 З. EMITTER, #2 4. 5 COLLECTOR, #2 COLLECTOR, #2 6.

STYLE 2: PIN 1. COLLECTOR, DIE, #1 2. COLLECTOR, #1 COLLECTOR, #2 3. 4 COLLECTOR, #2 BASE, #2 5. EMITTER, #2 6. 7 BASE #1 EMITTER, #1 8. STYLE 6: PIN 1. SOURCE 2. DRAIN 3. DRAIN SOURCE 4. SOURCE 5. 6. GATE GATE 7. 8. SOURCE STYLE 10: GROUND PIN 1. BIAS 1 OUTPUT 2. З. GROUND 4. 5. GROUND 6. BIAS 2 INPUT 7. 8. GROUND STYLE 14: PIN 1. N-SOURCE 2. N-GATE 3 P-SOURCE P-GATE 4. P-DRAIN 5 6. P-DRAIN N-DRAIN 7. N-DRAIN 8. STYLE 18: PIN 1. ANODE ANODE 2. SOURCE 3. GATE 4. 5. DRAIN 6 DRAIN CATHODE 7. CATHODE 8. STYLE 22: PIN 1. I/O LINE 1 2. COMMON CATHODE/VCC 3 COMMON CATHODE/VCC 4. I/O LINE 3 5. COMMON ANODE/GND 6. I/O LINE 4 7. I/O LINE 5 COMMON ANODE/GND 8. STYLE 26: PIN 1. GND 2 dv/dt З. ENABLE 4. ILIMIT 5. SOURCE SOURCE 6. SOURCE 7. 8. VCC STYLE 30: DRAIN 1 PIN 1. DRAIN 1 2 GATE 2 З. SOURCE 2 4. SOURCE 1/DRAIN 2 SOURCE 1/DRAIN 2 5.

6.

7.

8 GATE 1

SOURCE 1/DRAIN 2

STYLE 3: PIN 1. DRAIN, DIE #1 DRAIN, #1 2. DRAIN, #2 З. 4. DRAIN, #2 GATE, #2 5. SOURCE, #2 6. 7 GATE #1 8. SOURCE, #1 STYLE 7: PIN 1. INPUT 2. EXTERNAL BYPASS THIRD STAGE SOURCE GROUND З. 4. 5. DRAIN 6. GATE 3 SECOND STAGE Vd 7. FIRST STAGE Vd 8. STYLE 11: PIN 1. SOURCE 1 GATE 1 SOURCE 2 2. З. GATE 2 4. 5. DRAIN 2 6. DRAIN 2 DRAIN 1 7. 8. DRAIN 1 STYLE 15: PIN 1. ANODE 1 2. ANODE 1 ANODE 1 3 ANODE 1 4. 5. CATHODE, COMMON CATHODE, COMMON CATHODE, COMMON 6. 7. CATHODE, COMMON 8. STYLE 19: PIN 1. SOURCE 1 GATE 1 SOURCE 2 2. 3. GATE 2 4. 5. DRAIN 2 6. MIRROR 2 7. DRAIN 1 8. **MIRROR 1** STYLE 23: PIN 1. LINE 1 IN COMMON ANODE/GND COMMON ANODE/GND 2. 3 LINE 2 IN 4. LINE 2 OUT 5. COMMON ANODE/GND COMMON ANODE/GND 6. 7. LINE 1 OUT 8. STYLE 27: PIN 1. ILIMIT OVI O 2 З. UVLO 4. INPUT+ 5. 6. SOURCE SOURCE SOURCE 7. 8 DRAIN

#### DATE 16 FEB 2011

STYLE 4: PIN 1. 2. ANODE ANODE ANODE З. 4. ANODE ANODE 5. 6. ANODE 7 ANODE COMMON CATHODE 8. STYLE 8: PIN 1. COLLECTOR, DIE #1 2. BASE, #1 BASE #2 З. COLLECTOR, #2 4. COLLECTOR, #2 5. 6. EMITTER, #2 EMITTER, #1 7. 8. COLLECTOR, #1 STYLE 12: PIN 1. SOURCE SOURCE 2. 3. GATE 4. 5. DRAIN 6 DRAIN DRAIN 7. 8. DRAIN STYLE 16 EMITTER, DIE #1 PIN 1. 2. BASE, DIE #1 EMITTER, DIE #2 3 BASE, DIE #2 4. 5. COLLECTOR, DIE #2 6. COLLECTOR, DIE #2 COLLECTOR, DIE #1 7. COLLECTOR, DIE #1 8. STYLE 20: PIN 1. SOURCE (N) GATE (N) SOURCE (P) 2. 3. 4. GATE (P) 5. DRAIN 6. DRAIN DRAIN 7. 8. DRAIN STYLE 24: PIN 1. BASE EMITTER 2. 3 COLLECTOR/ANODE COLLECTOR/ANODE 4. 5. CATHODE 6. CATHODE COLLECTOR/ANODE 7. 8. COLLECTOR/ANODE STYLE 28: PIN 1. SW\_TO\_GND 2. DASIC OFF DASIC\_SW\_DET З. 4. GND 5. 6. V MON VBULK 7. VBULK 8 VIN

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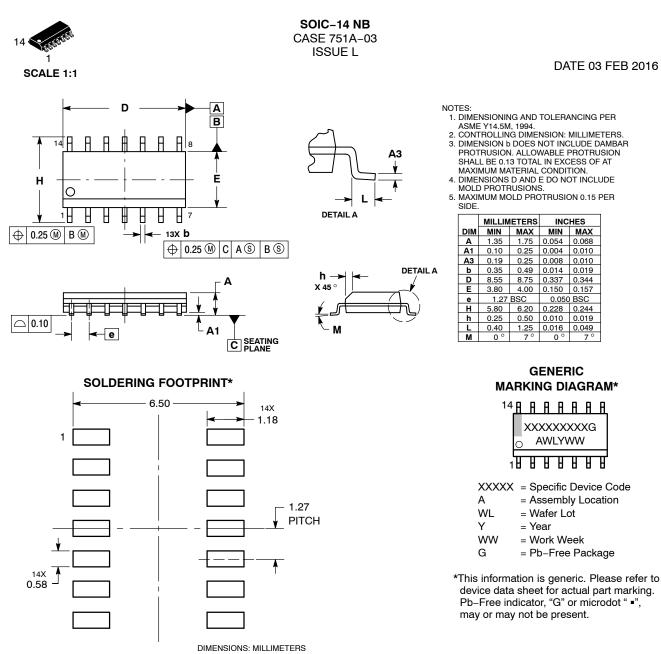
7.

8

COLLECTOR, #1

COLLECTOR, #1





\*For additional information on our Pb–Free strategy and soldering details, please download the ON Semiconductor Soldering and Mounting Techniques Reference Manual, SOLDERRM/D.

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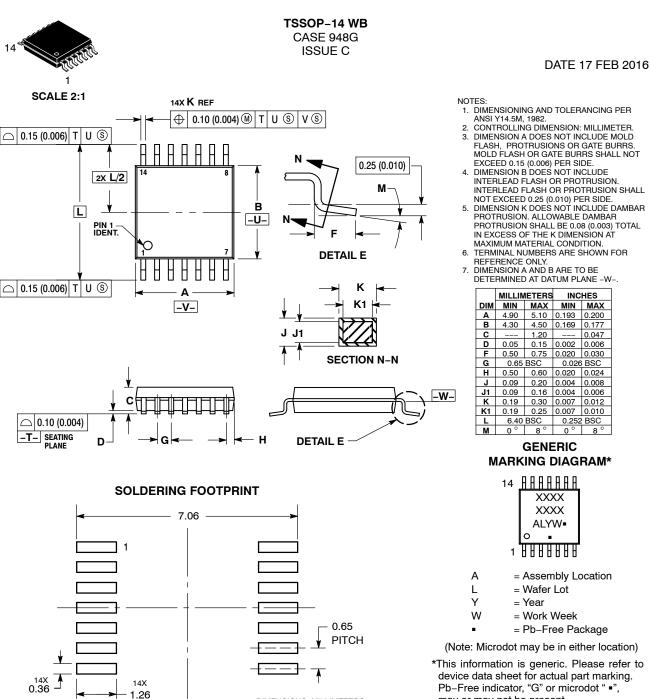
### SOIC-14 CASE 751A-03 ISSUE L

# DATE 03 FEB 2016

STYLE 1: PIN 1. COMMON CATHODE 2. ANODE/CATHODE 3. ANODE/CATHODE 4. NO CONNECTION 5. ANODE/CATHODE 6. NO CONNECTION 7. ANODE/CATHODE 8. ANODE/CATHODE 9. ANODE/CATHODE 10. NO CONNECTION 11. ANODE/CATHODE 12. ANODE/CATHODE 13. NO CONNECTION 14. COMMON ANODE	STYLE 2: CANCELLED	STYLE 3: PIN 1. NO CONNECTION 2. ANODE 3. ANODE 4. NO CONNECTION 5. ANODE 6. NO CONNECTION 7. ANODE 8. ANODE 9. ANODE 10. NO CONNECTION 11. ANODE 12. ANODE 13. NO CONNECTION 14. COMMON CATHODE	STYLE 4: PIN 1. NO CONNECTION 2. CATHODE 3. CATHODE 4. NO CONNECTION 5. CATHODE 6. NO CONNECTION 7. CATHODE 9. CATHODE 10. NO CONNECTION 11. CATHODE 12. CATHODE 13. NO CONNECTION 14. COMMON ANODE
STYLE 5: PIN 1. COMMON CATHODE 2. ANODE/CATHODE 3. ANODE/CATHODE 4. ANODE/CATHODE 5. ANODE/CATHODE 6. NO CONNECTION 7. COMMON ANODE 8. COMMON CATHODE 9. ANODE/CATHODE 10. ANODE/CATHODE 11. ANODE/CATHODE 12. ANODE/CATHODE 13. NO CONNECTION 14. COMMON ANODE	STYLE 6: PIN 1. CATHODE 2. CATHODE 3. CATHODE 4. CATHODE 5. CATHODE 6. CATHODE 7. CATHODE 8. ANODE 9. ANODE 10. ANODE 11. ANODE 12. ANODE 13. ANODE 14. ANODE	STYLE 7: PIN 1. ANODE/CATHODE 2. COMMON ANDDE 3. COMMON CATHODE 4. ANODE/CATHODE 5. ANODE/CATHODE 7. ANODE/CATHODE 8. ANODE/CATHODE 10. ANODE/CATHODE 11. COMMON CATHODE 12. COMMON ANODE 13. ANODE/CATHODE 14. ANODE/CATHODE	STYLE 8: PIN 1. COMMON CATHODE 2. ANODE/CATHODE 3. ANODE/CATHODE 4. NO CONNECTION 5. ANODE/CATHODE 6. ANODE/CATHODE 7. COMMON ANODE 8. COMMON ANODE 9. ANODE/CATHODE 10. ANODE/CATHODE 11. NO CONNECTION 12. ANODE/CATHODE 13. ANODE/CATHODE 14. COMMON CATHODE

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