

## 3.0V/3.3V Low Power Microprocessor Supervisory Circuits

April 2015 Rev. 3.0.0

#### **GENERAL DESCRIPTION**

The SP706R/S/T and SP708R/S/T series is a family of microprocessor ( $\mu$ P) supervisory circuits that integrate myriad components involved in discrete solutions which monitor power-supply and battery in  $\mu$ P and digital systems.

The SP706R/S/T and SP708R/S/T series will significantly improve system reliability and operational efficiency when compared to solutions obtained with discrete components. The features of the SP706R/S/T and SP708R/S/T series include a watchdog timer, a  $\mu P$  reset, a Power Fail Comparator, and a manual-reset input.

The SP706R/S/T and SP708R/S/T series is ideal for 3.0V or 3.3V applications in automotive systems, computers, controllers, and intelligent instruments. The SP706R/S/T and SP708R/S/T series is an ideal solution for systems in which critical monitoring of the power supply to the  $\mu P$  and related digital components is demanded.

#### **APPLICATIONS**

- Processors & DSPs Based Systems
- Industrial & Medical Instruments

#### **FEATURES**

- Precision Low Voltage Monitor
  - SP706R/SP708R: +2.63V
  - SP706S/SP708S: +2.93V
  - SP706T/SP708T: +3.08V
- 200ms RESET Pulse Width
  - SP706R/S/T: Active Low
  - SP708R/S/T: Active High and Active Low
- Independent Watchdog Timer
  - 1.6s Timeout (SP706R/S/T)
  - Enable/Disable Function
- 40µA Maximum Supply Current
- Debounced TTL/CMOS Manual Reset Input
- RESET Asserted Down to Vcc=1V
- Vcc Glitch Immunity
- Voltage Monitor for Power Failure or Low Battery Warning
- 8-Pin NSOIC and MSOP Packages
- Pin Compatible with Industry Standards 706R/S/T and 708R/S/T

Part Number	RESET Threshold	RESET Active	Manual RESET	Watchdog	PFI Accuracy
SP706R	2.63V	Low	Yes	Yes	4%
SP706S	2.93V	Low	Yes	Yes	4%
SP706T	3.08V	Low	Yes	Yes	4%
SP708R	2.63V	Low and High	Yes	No	4%
SP708S	2.93V	Low and High	Yes	No	4%
SP708T	3.08V	Low and High	Yes	No	4%

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#### **ABSOLUTE MAXIMUM RATINGS**

These are stress ratings only and functional operation of the device at these ratings or any other above those indicated in the operation sections of the specifications below is not implied. Exposure to absolute maximum rating conditions for extended periods of time may affect reliability.

ESD Rating (HBM - Human Body Model)	. 2kV
Continuous Power Dissipation	
SO (derate 5.88mW/°C above +70°C)473	1mW
Mini SO (derate 4.10mW/°C above +70°C)330	)mW
Storage Temperature65°C to 10	60°C
Lead Temperature (Soldering, 10 sec)3	00°C

#### **ELECTRICAL SPECIFICATIONS**

Unless otherwise indicated,  $V_{CC}$  = 2.7V to 5.5V (SP70xR),  $V_{CC}$  = 3.15V to 5.5V (SP70xS),  $V_{CC}$  = 3.0V to 5.5V (SP70xT),  $T_{A}$  =  $T_{MIN}$  to  $T_{MAX}$ , typical at 25°C.

Parameter	Min.	Тур.	Max.	Units	Conditions
Operating Voltage Range V <sub>CC</sub>	1.0		5.5	V	
Supply Current I <sub>SUPPLY</sub>		25	40	μА	MR= $V_{CC}$ or floating, WDI floating, $V_{CC} = 3.3V$
		40	80	μΑ	$MR=V_{CC}$ or floating, $V_{CC}=5.5V$
	2.55	2.63	2.70		SP706R. SP708R
Reset Threshold	2.85	2.93	3.00	V	SP706S. SP708S
	3.00	3.08	3.15		SP706T. SP708T
Reset Threshold Hysteresis		20		mV	Note 2
Reset Pulse Width t <sub>RS</sub>	140	200	280	ms	Note 2
RESET Output Voltage					
$V_OH$	0.8xV <sub>CC</sub>				V <sub>RST(MAX)</sub> <v<sub>CC&lt;3.6V, I<sub>SOURCE</sub>=500μA</v<sub>
$V_{OL}$			0.3	V	V <sub>RST(MAX)</sub> <v<sub>CC&lt;3.6V, I<sub>SINK</sub>=1.2mA</v<sub>
$V_OH$	V <sub>CC</sub> -1.5				4.5V <v<sub>CC&lt;5.5V, I<sub>SOURCE</sub>=800μA</v<sub>
VoL			0.4		4.5V <v<sub>CC&lt;5.5V, I<sub>SINK</sub>=3.2mA</v<sub>
RESET Output Voltage					
$V_OH$	V <sub>CC</sub> -0.6				V <sub>RST(MAX)</sub> < V <sub>CC</sub> < 3.6V, I <sub>SOURCE</sub> = 215µA
$V_{OL}$			0.3	V	V <sub>RST(MAX)</sub> < V <sub>CC</sub> < 3.6V, I <sub>SOURCE</sub> = 1.2mA
$V_OH$	V <sub>CC</sub> -1.5				4.5V <v<sub>CC&lt;5.5V, I<sub>SOURCE</sub>=800μA</v<sub>
VoL			0.4		4.5V <v<sub>CC&lt;5.5V, I<sub>SOURCE</sub>=3.2mA</v<sub>
Watchdog Timeout Period two	1.00	1.60	2.25	S	V <sub>CC</sub> <3.6V
WDI Pulse Width twp (note 1)	50			ns	$V_{IL}$ =0.4V, $V_{IH}$ =0.8x $V_{CC}$
WDI Input Threshold					
$V_{ ext{IL}}$			0.6		V <sub>RST(MAX)</sub> < V <sub>CC</sub> < 3.6 V
$V_{\mathrm{IH}}$	$0.7xV_{CC}$			V	$V_{RST(MAX)} < V_{CC} < 3.6V$
$V_{\mathtt{IL}}$			0.8		V <sub>cc</sub> =5V
$V_{\mathrm{IH}}$	3.5				V <sub>CC</sub> =5V
WDI Input Current	-1	0.02	1	μA	WDI=0V or WDI=V <sub>CC</sub>
WDO Output Voltage					
$V_{ ext{IL}}$	$0.8xV_{CC}$			V	$V_{RST(MAX)} < V_{CC} < 3.6V$ , $I_{SOURCE} = 500 \mu A$
$V_IH$			0.3	_ v	V <sub>RST(MAX)</sub> <v<sub>CC&lt;3.6V, I<sub>SINK</sub>=1.2mA</v<sub>
$V_{\mathrm{IL}}$	V <sub>CC</sub> -1.5				4.5V <v<sub>CC&lt;5.5V, I<sub>SOURCE</sub>=800μA</v<sub>



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Parameter	Min.	Тур.	Max.	Units	Conditions
V <sub>IH</sub>			0.4		4.5V <v<sub>CC&lt;5.5V, I<sub>SINK</sub>=3.2mA</v<sub>
MD Dull up Current	25	70	250		$MR = 0V$ , $V_{RST(MAX)} < V_{CC} < 3.6V$
MR Pull-up Current	100	250	600	μΑ	$MR = 0V, 4.5V < V_{CC} < 5.5V$
MR Pulse Width t <sub>MR</sub>	500			nc	V <sub>RST(MAX)</sub> <v<sub>CC&lt;3.6V</v<sub>
IVIR Puise Width t <sub>MR</sub>	150			ns	4.5V <v<sub>CC&lt;5.5V</v<sub>
MR Input Threshold					
$V_{\mathrm{IL}}$			0.6		V <sub>RST(MAX)</sub> <v<sub>CC&lt;3.6V</v<sub>
V <sub>IH</sub>	$0.7xV_{CC}$			V	$V_{RST(MAX)} < V_{CC} < 3.6V$
$V_{\mathrm{IL}}$			0.8		4.5V <v<sub>CC&lt;5.5V</v<sub>
V <sub>IH</sub>	2.0				4.5V <v<sub>CC&lt;5.5V</v<sub>
MD to Decet Out Delay t			750	nc	V <sub>RST(MAX)</sub> <v<sub>CC&lt;3.6V</v<sub>
MR to Reset Out Delay t <sub>MD</sub>			250	ns	4.5V <vcc<5.5v< td=""></vcc<5.5v<>
PFI Input Threshold	1.20	1.25	1.30	V	$V_{cc}$ =3.0V - SP70xR, PFI falling $V_{cc}$ =3.3V - SP70xS/T, PFI falling
PFI Input Current	-200.00	0.01	200.00	nA	$V_{PFI} = 1.36V$
PFO Output Voltage					
$V_{IL}$	0.8xV <sub>CC</sub>				V <sub>RST(MAX)</sub> <v<sub>CC&lt;3.6V, I<sub>SOURCE</sub>=500µA</v<sub>
V <sub>IH</sub>			0.3	V	V <sub>RST(MAX)</sub> <v<sub>CC&lt;3.6V, I<sub>SINK</sub>=1.2mA</v<sub>
$V_{\mathrm{IL}}$	V <sub>CC</sub> -1.5			Ī	4.5V <v<sub>CC&lt;5.5V, I<sub>SOURCE</sub>=800μA</v<sub>
V <sub>IH</sub>			0.4		4.5V <v<sub>CC&lt;5.5V, I<sub>SINK</sub>=3.2mA</v<sub>

Note 1:WDI minimum rise /fall time is  $1\mu s$ .

#### **BLOCK DIAGRAM**

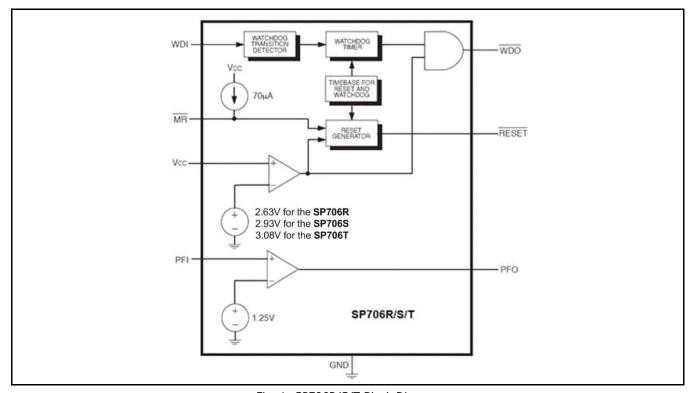


Fig. 1: SP706R/S/T Block Diagram



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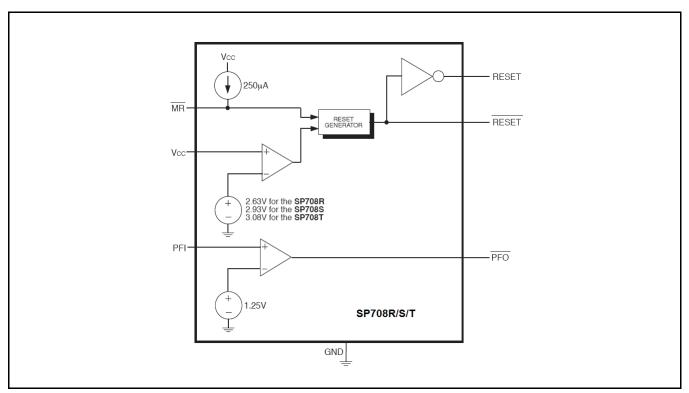


Fig. 2: SP708R/S/T Block Diagram

#### **PIN ASSIGNMENT**

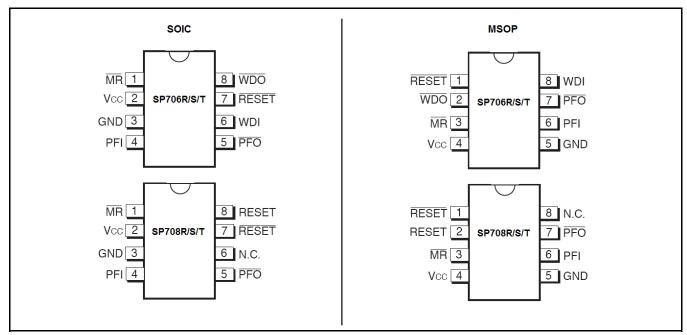


Fig. 3: Pin Assignment

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### **PIN DESCRIPTION**

N	Pin Number			Description		
Name	SP706			R/S/T	Description	
	SOIC	MSOP	SOIC	MSOP	M 10 1	
MR	1	3	1	3	Manual Reset This input triggers a reset pulse when pulled below 0.8V. This active LOW input has an internal 70µA pull-up current. It can be driven from a TTL or CMOS logic line or shorted to ground with a switch.	
Vcc	2	4	2	4	Voltage Input	
GND	3	5	3	5	Ground reference for all signals	
PFI	4	6	4	6	Power-Fail Input When this voltage monitor input is less than 1.25V, PFO goes LOW. Connect PFI to ground or Vcc when not in use.	
PFO	5	7	5	7	Power-Fail Output This output is LOW until PFI is less then 1.25V	
WDI	6	8	-	Watchdog Input  If this input remains HIGH or LOW for 1.6s, the internal watch timer times out and WDO goes LOW. Floating WDI or connecting to a high-impedance tri-state buffer disables the watchdog fea The internal watchdog timer clears whenever RESET is asserted is tri-stated, or whenever WDI sees a rising or falling edge.		
N.C.	-	-	6	8	No Connect	
RESET	7	1	7	1	Active-LOW RESET Output This output pulses LOW for 200ms when triggered and stays LOW whenever Vcc is below the reset threshold. It remains LOW for 200ms after Vcc rises above the reset threshold or MR goes from LOW to HIGH. A watchdog timeout will not trigger RESET unless WDO is connected to MR.	
WDO	8	2	-	Watchdog Output This output pulls LOW when the internal watchdog timer finished 1.6s count and does not go HIGH again until the watchdog is completed.  WDO also goes LOW during low-line conditions. Whenever Vcc is below the reset threshold, WDO stays LOW. However, unlike RIWDO does not have a minimum pulse width. As soon as Vcc is a the reset threshold, WDO goes HIGH with no delay.		
RESET	-	-	8	2	Active-HIGH RESET Output This output is the complement of RESET. Whenever RESET is HIGH, RESET is LOW and vice-versa. Note that the SP708R/S/T has a reset output only.	

### ORDERING INFORMATION

Part Number	Temperature Range	Marking	Package	Packing Quantity	Note 1	Note 2
SP706RCN-L	0°C≤T <sub>A</sub> ≤+70°C	SP706RC	8-pin NSOIC	Bulk	Lead Free	
SP706RCN-L/TR	0°C≤T <sub>A</sub> ≤+70°C	YYWWL X	8-pin NSOIC	2.5K/Tape & Reel	Lead Free	
SP706RCU-L	0°C≤T <sub>A</sub> ≤+70°C	706R	8-pin MSOP	Bulk	Lead Free	
SP706RCU-L/TR	0°C≤T <sub>A</sub> ≤+70°C	CXXX YWW	8-pin MSOP	2.5K/Tape & Reel	Lead Free	
SP706REN-L	-40°C≤T <sub>A</sub> ≤+85°C	SP706RE	8-pin NSOIC	Bulk	Lead Free	
SP706REN-L/TR	-40°C≤T <sub>A</sub> ≤+85°C	YYWWL X	8-pin NSOIC	2.5K/Tape & Reel	Lead Free	
SP706REU-L	-40°C≤T <sub>A</sub> ≤+85°C	706R	8-pin MSOP	Bulk	Lead Free	
SP706REU-L/TR	-40°C≤T <sub>A</sub> ≤+85°C	EXXX YWW	8-pin MSOP	2.5K/Tape & Reel	Lead Free	
SP706SCN-L	0°C≤T <sub>A</sub> ≤+70°C	SP706SC	8-pin NSOIC	Bulk	Lead Free	
SP706SCN-L/TR	0°C≤T <sub>A</sub> ≤+70°C	YYWWL X	8-pin NSOIC	2.5K/Tape & Reel	Lead Free	
SP706SCU-L	0°C≤T <sub>A</sub> ≤+70°C	706S	8-pin MSOP	Bulk	Lead Free	
SP706SCU-L/TR	0°C≤T <sub>A</sub> ≤+70°C	CXXX YWW	8-pin MSOP	2.5K/Tape & Reel	Lead Free	



# 3.0V/3.3V Low Power Microprocessor Supervisory Circuits

Part Number	Temperature Range	Marking	Package	Packing Quantity	Note 1	Note 2
SP706SEN-L	-40°C≤T <sub>A</sub> ≤+85°C	SP706SE YYWWL	8-pin NSOIC	Bulk	Lead Free	
SP706SEN-L/TR	-40°C≤T <sub>A</sub> ≤+85°C	X	8-pin NSOIC	2.5K/Tape & Reel	Lead Free	
SP706SEU-L	-40°C≤T <sub>A</sub> ≤+85°C	706S	8-pin MSOP	Bulk	Lead Free	
SP706SEU-L/TR	-40°C≤T <sub>A</sub> ≤+85°C	EXXX YWW	8-pin MSOP	2.5K/Tape & Reel	Lead Free	
SP706TCN-L	0°C≤T <sub>A</sub> ≤+70°C	SP706TC YYWWL	8-pin NSOIC	Bulk	Lead Free	
SP706TCN-L/TR	0°C≤T <sub>A</sub> ≤+70°C	X	8-pin NSOIC	2.5K/Tape & Reel	Lead Free	
SP706TCU-L	0°C≤T <sub>A</sub> ≤+70°C	706T	8-pin MSOP	Bulk	Lead Free	
SP706TCU-L/TR	0°C≤T <sub>A</sub> ≤+70°C	CXXX YWW	8-pin MSOP	2.5K/Tape & Reel	Lead Free	
SP706TEN-L	-40°C≤T <sub>A</sub> ≤+85°C	SP706TE YYWWL	8-pin NSOIC	Bulk	Lead Free	
SP706TEN-L/TR	-40°C≤T <sub>A</sub> ≤+85°C	X	8-pin NSOIC	2.5K/Tape & Reel	Lead Free	
SP706TEU-L	-40°C≤T <sub>A</sub> ≤+85°C	706T EXXX	8-pin MSOP	Bulk	Lead Free	
SP706TEU-L/TR	-40°C≤T <sub>A</sub> ≤+85°C	YWW	8-pin MSOP	2.5K/Tape & Reel	Lead Free	
SP708REN-L	-40°C≤T <sub>A</sub> ≤+85°C	SP708RE YYWWL	8-pin NSOIC	Bulk	Lead Free	
SP708REN-L/TR	-40°C≤T <sub>A</sub> ≤+85°C	X	8-pin NSOIC	2.5K/Tape & Reel	Lead Free	
SP708SCN-L	0°C≤T <sub>A</sub> ≤+70°C	SP708SC YYWWL	8-pin NSOIC	Bulk	Lead Free	
SP708SCN-L/TR	0°C≤T <sub>A</sub> ≤+70°C	X	8-pin NSOIC	2.5K/Tape & Reel	Lead Free	
SP708SCU-L	0°C≤T <sub>A</sub> ≤+70°C	708S UXXX	8-pin MSOP	Bulk	Lead Free	
SP708SCU-L/TR	0°C≤T <sub>A</sub> ≤+70°C	YWW	8-pin MSOP	2.5K/Tape & Reel	Lead Free	
SP708SEN-L	-40°C≤T <sub>A</sub> ≤+85°C	SP708SE	8-pin NSOIC	Bulk	Lead Free	
SP708SEN-L/TR	-40°C≤T <sub>A</sub> ≤+85°C	YYWWL X	8-pin NSOIC	2.5K/Tape & Reel	Lead Free	
SP708TCN-L	0°C≤T <sub>A</sub> ≤+70°C	SP708TC	8-pin NSOIC	Bulk	Lead Free	
SP708TCN-L/TR	0°C≤T <sub>A</sub> ≤+70°C	YYWWL X	8-pin NSOIC	2.5K/Tape & Reel	Lead Free	
SP708TEN-L	-40°C≤T <sub>A</sub> ≤+85°C	SP706TE YYWWL	8-pin NSOIC	Bulk	Lead Free	
SP708TEN-L/TR	-40°C≤T <sub>A</sub> ≤+85°C	X	8-pin NSOIC	2.5K/Tape & Reel	Lead Free	

<sup>&</sup>quot;YY" = Year - "WW" = Work Week - "X" = Lot Number

# 3.0V/3.3V Low Power Microprocessor Supervisory Circuits

#### TYPICAL PERFORMANCE CHARACTERISTICS

All data taken at  $V_{CC} = 2.7V$  to 5.5V (SP70xR),  $V_{CC} = 3.15V$  to 5.5V (SP70xS),  $V_{CC} = 3.0V$  to 5.5V (SP70xT),  $T_{A} = 25^{\circ}$ C, unless otherwise indicated.

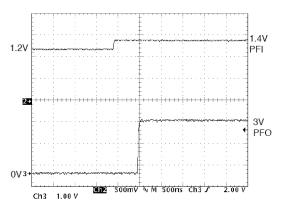


Fig. 4: Power-Fail Comparator De-Assertion Response Time

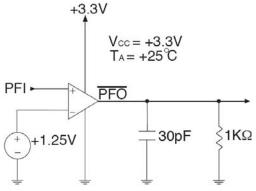


Fig. 5: Power-Fail Comparator De-Assertion Response Time Circuit

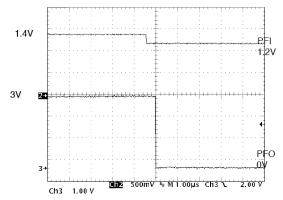


Fig. 6: Power-Fail Comparator Assertion Response Time

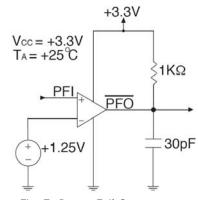


Fig. 7: Power-Fail Comparator Assertion Response Time Circuit

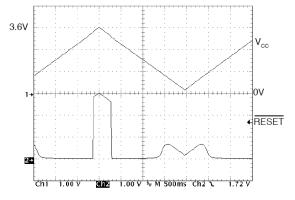


Fig. 8: SP706 RESET Output Voltage vs. Supply Voltage

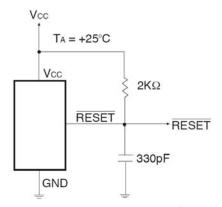


Fig. 9: SP706 RESET Output Voltage vs. Supply Voltage Circuit

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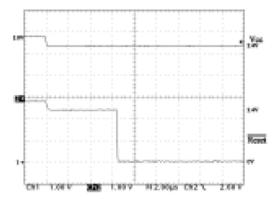


Fig. 10: SP706 RESET Response Time

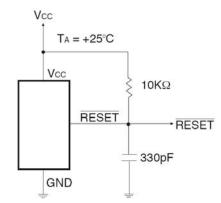


Fig. 11: SP706 RESET Response Time Circuit

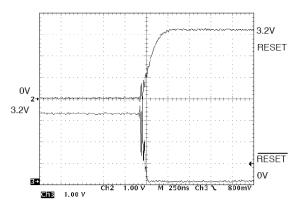


Fig. 12: SP708 RESET and RESET Assertion

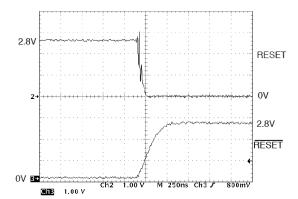


Fig. 13: SP708 RESET and RESET De-Assertion

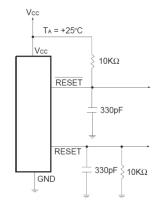


Fig. 14: SP708 RESET and RESET Assertion and De-Assertion Circuit



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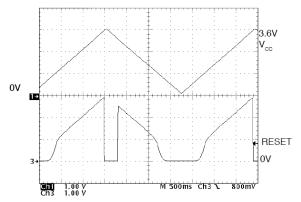


Fig. 15: SP708 RESET Output Voltage vs. Supply Voltage

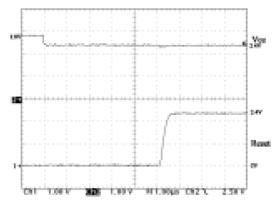


Fig. 16: SP708 RESET Response Time

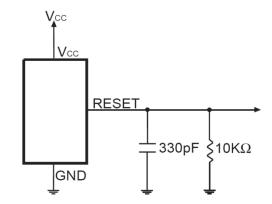


Fig. 17: SP708 RESET Output Voltage vs. Supply Voltage and RESET Response Time Circuit



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## 3.0V/3.3V Low Power Microprocessor Supervisory Circuits

#### **FEATURES**

The SP706R/S/T and SP708R/S/T series provides four key functions:

- 1. A reset output during power-up, power-down and brownout conditions.
- 2. An independent watchdog output that goes LOW if the watchdog input has not been toggled within 1.6 seconds.
- 3. A 1.25V threshold detector for power-fail warning, low battery detection, or monitoring a power supply other than +3.3V/+3.0V.
- 4. An active-LOW manual-reset that allows RESET to be triggered by a pushbutton switch.

The SP706R/S/T devices are the same as the SP708R/S/T devices except for the active-HIGH RESET substitution of the watchdog timer.

#### THEORY OF OPERATION

The SP706R/S/T-SP708R/S/T series is a microprocessor (µP) supervisory circuit that monitors the power supplied to digital circuits such as microprocessors, microcontrollers, or memory. The series is an ideal solution for portable, battery-powered equipment that requires power supply monitoring. Implementing this series will reduce the number of components and overall complexity. The watchdog functions of this product family will continuously oversee the operational status of a system. The operational features and benefits of the SP706R/S/T-SP708R/S/T series are described in more detail below.

#### **RESET OUTPUT**

A microprocessor's reset input starts the  $\mu P$  in a known state. The SP706R/S/T-SP708R/S/T series asserts reset during power-up and prevents code execution errors during power down or brownout conditions.

On power-up, once  $V_{\text{CC}}$  reaches 1V, RESET is a guaranteed logic LOW of 0.4V or less. As  $V_{\text{CC}}$  rises, RESET stays LOW. When  $V_{\text{CC}}$  rises above the reset threshold, an internal timer releases RESET after 200ms. RESET pulses LOW whenever  $V_{\text{CC}}$  dips below the reset threshold, such as in a brownout condition. When a brownout condition occurs in the middle of a

previously initiated reset pulse, the pulse continues for at least another 140ms. On power down, once  $V_{CC}$  falls below the reset threshold, RESET stays LOW and is guaranteed to be 0.4V or less until  $V_{CC}$  drops below 1V.

The active-HIGH RESET output is simply the complement of the RESET output and is guaranteed to be valid with Vcc down to 1.1V. Some  $\mu Ps$ , such as Intel's 80C51, require an active-HIGH reset pulse.

#### **WATCHDOG TIMER**

The SP706R/S/T-SP708R/S/T watchdog circuit monitors the  $\mu$ P's activity. If the  $\mu$ P does not toggle the watchdog input (WDI) within 1.6 seconds and WDI is not tri-stated, WDO goes LOW. As long as RESET is asserted or the WDI input is tri-stated, the watchdog timer will stay cleared and will not count. As soon as RESET is released and WDI is driven HIGH or LOW, the timer will start counting. Pulses as short as 50ns can be detected.

Typically, WDO will be connected to the non-maskable interrupt input (NMI) of a  $\mu P$ . When  $V_{CC}$  drops below the reset threshold, WDO will go LOW whether or not the watchdog timer had timed out. Normally this would trigger an NMI but RESET goes LOW simultaneously and thus overrides the NMI.

If WDI is left unconnected, WDO can be used as a low-line output. Since floating WDI disables the internal timer, WDO goes LOW only when  $V_{\text{CC}}$  falls below the reset threshold, thus functioning as a low-line output.

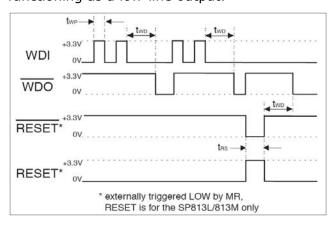


Fig. 18: Watchdog Timing Waveforms



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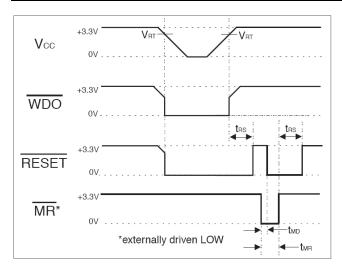


Fig. 19: Timing Diagrams with WDI tri-stated.

#### **POWER-FAIL COMPARATOR**

The power-fail comparator can be used for various purposes because its output and non inverting input are not internally connected. The inverting input is internally connected to a 1.25V reference.

To build an early-warning circuit for power failure, connect the PFI pin to a voltage divider as shown in Figure 20. Choose the voltage divider ratio so that the voltage at PFI falls below 1.25V just before the +5V regulator drops out. Use PFO to interrupt the  $\mu P$  so it can prepare for an orderly power-down.

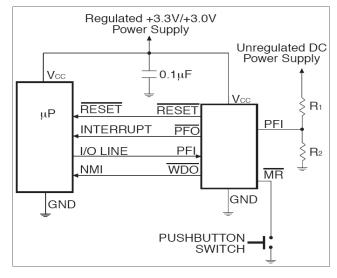


Fig. 20: Typical Operating Circuit

#### MANUAL RESET

The manual-reset input (MR) allows RESET to be triggered by a pushbutton switch. The switch is effectively debounced by the 140ms minimum RESET pulse width. MR is TTL/CMOS logic compatible, so it can be driven by an external logic line. MR can be used to force a watchdog timeout to generate a RESET pulse in the SP706R/S/T-SP708R/S/T. Simply connect WDO to MR.

## Ensuring a Valid Reset Output Down to Vcc=0V

When  $V_{CC}$  falls below 1V, the RESET output no longer sinks current, it becomes an open circuit. High-impedance CMOS logic inputs can drift to undetermined voltages if left undriven. If a pull-down resistor is added to the RESET pin, any stray charge or leakage currents will be shunted to ground, holding RESET LOW. The resistor value is not critical. It should be about  $100 \mathrm{K}\Omega$ , large enough not to load RESET and small enough to pull RESET to ground.

## MONITORING VOLTAGES OTHER THAN THE UNREGULATED DC INPUT

Monitor voltages other than the unregulated DC by connecting a voltage divider to PFI and adjusting the ratio appropriately. If required, add hysteresis by connecting a resistor (with a value approximately 10 times the sum of the two resistors in the potential divider network) between PFI and PFO. A capacitor between PFI and GND will reduce the power-fail circuit's sensitivity to high-frequency noise on the line being monitored. RESET can be used to monitor voltages other than the +3.3V/+3.0V Vcc line. Connect PFO to MR to initiate a RESET pulse when PFI drops below 1.25V. Figure 21 shows the SP706R/S/T-SP708R/S/T series configured to assert RESET when the +3.3V/3.0V supply falls below the RESET threshold, or when the +12V supply falls below approximately 11V.

#### MONITORING A NEGATIVE VOLTAGE SUPPLY

The power-fail comparator can also monitor a negative supply rail, shown in Figure 22. When the negative rail is good (a negative voltage of large magnitude), PFO is LOW. By adding the resistors and transistor as shown, a HIGH PFO

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triggers RESET. As long as PFO remains HIGH, the SP706R/S/T-SP708R/S/T series will keep RESET asserted (where RESET = LOW and RESET = HIGH). Note that this circuit's accuracy depends on the PFI threshold tolerance, the  $V_{CC}$  line, and the resistors.

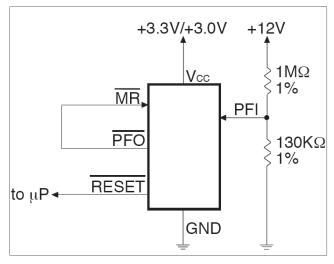


Fig. 21: Monitoring +3.3V/+3.0V and +12V Power Supplies

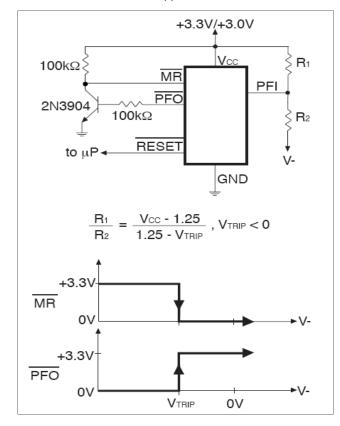


Fig. 22: Monitoring a Negative Voltage Supply

## INTERFACING TO µPS WITH BIDIRECTIONAL RESET PINS

μPs with bidirectional RESET pins, such as the Motorola 68HC11 series, can contend with the SP706/708 RESET output. If, for example, the RESET output is driven HIGH and the μP wants to pull it LOW, indeterminate logic levels may result. To correct this, connect a 4.7K $\Omega$  resistor between the RESET output and the μP reset I/O, as shown if Figure 23. Buffer the RESET output to other system components.

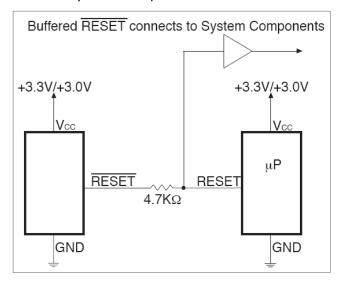


Fig. 23: Interfacing to Microprocessors with Bidirectional RESET I/O (SP706)

#### **NEGATIVE-GOING VCC TRANSIENT**

While issuing resets to the  $\mu P$  during power-up, power-down, and brownout conditions, these supervisors are relatively immune to short duration negative-going VCC transients (glitches). It is usually undesirable to reset the  $\mu P$  when  $V_{CC}$  experiences only small glitches.

Figure 24 shows maximum transient duration vs. reset-comparator overdrive, for which reset pulses are not generated. The data was generated using negative-going  $V_{\text{CC}}$  pulses, starting at 3.3V and ending below the reset threshold by the magnitude indicated (reset comparator overdrive). The graph shows the maximum pulse width a negative-going  $V_{\text{CC}}$  transient may typically have without causing a reset pulse to be issued. As the amplitude of the transient increases (i.e. goes farther below the reset threshold), the maximum allowable



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pulse width decreases. Typically, a  $V_{\text{CC}}$  transient that goes 100mV below the reset threshold and lasts for 40µs or less will not cause a reset pulse to be issued. A 100nF bypass capacitor mounted close to the VCC pin provides additional transient immunity.

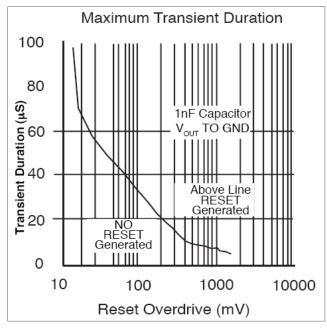


Fig. 24: Maximum Transient Duration without Causing a Reset Pulse vs. Reset Comparator Overdrive

#### **APPLICATIONS**

The SP706P/R/S/T-SP708R/S/T series offers unmatched performance and the lowest power consumption for these industry standard devices. Refer to Figures 25 and 26 for supply current performance characteristics rated against temperature and supply voltages.

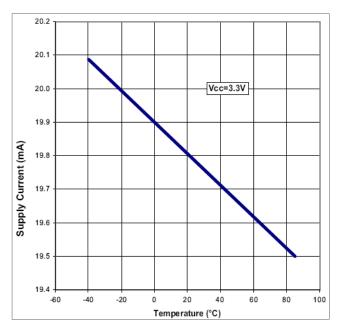


Fig. 25: Supply Current vs. Temperature

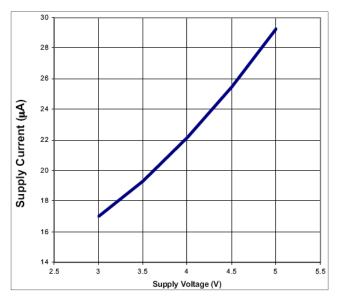


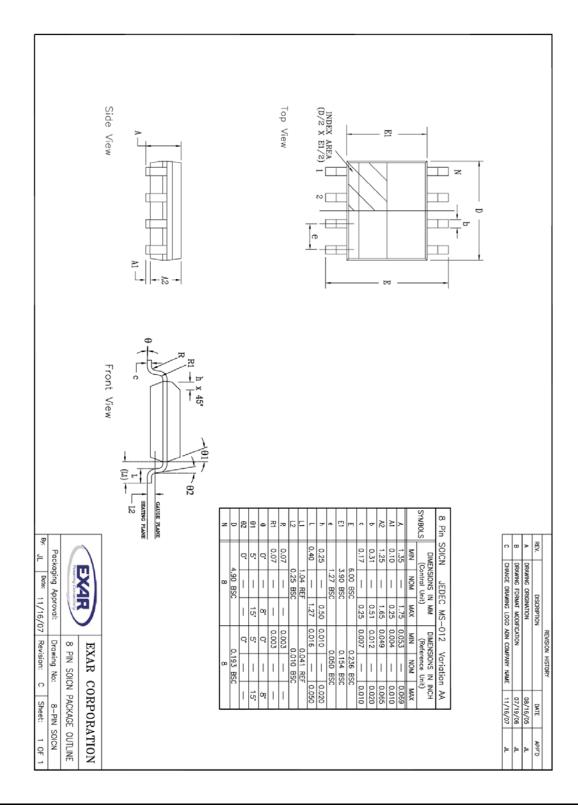
Fig. 26: Supply Current vs. Supply Voltage

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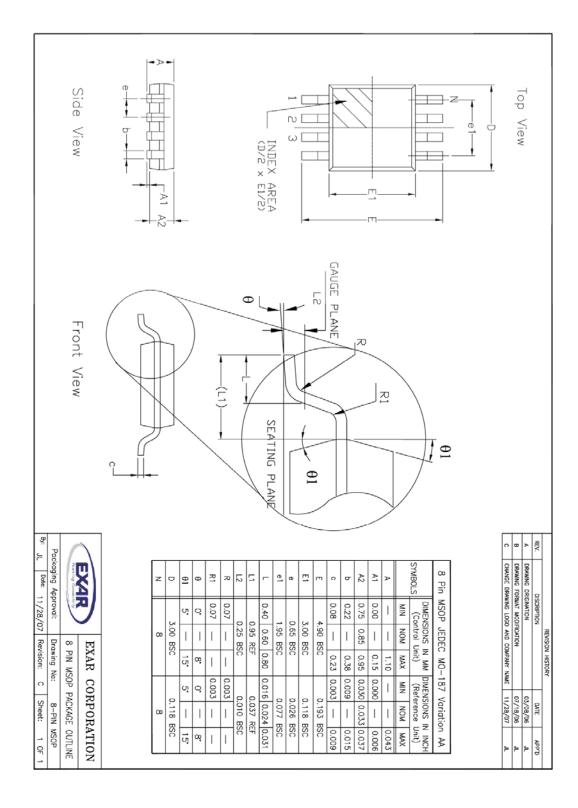
#### PACKAGE SPECIFICATION

#### 8-PIN NSOIC



# 3.0V/3.3V Low Power Microprocessor Supervisory Circuits

#### 8-PIN MSOP





## 3.0V/3.3V Low Power Microprocessor Supervisory Circuits

#### REVISION HISTORY

Revision	Date	Description					
2.0.0	06/04/2010	Reformat of datasheet					
3.0.0	04/14/2015	Change of specs to match industry standards [ECN 1517-08]					

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