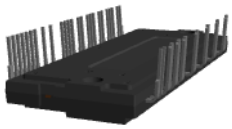
ASPM34 Series Automotive 3-Phase 1200 V 50 A IGBT Intelligent Power Module

NFVA25012NP2T



3D Package Drawing
(Click to Activate 3D Content)

DIP34 80x33, AUTOMOTIVE MODULE CASE MODGL

General Description

NFVA25012NP2T is an advanced Auto IPM module providing a fully−featured, high−performance inverter output stage for hybrid and electric vehicles. These modules integrate optimized gate drive of the built−in IGBTs to minimize EMI and losses, while also providing multiple on−module protection features including under−voltage lockouts, over−current shutdown, thermal monitoring of drive IC, and fault reporting. The built−in, high−speed HVIC requires only a single supply voltage and translates the incoming logic−level gate inputs to the high−voltage, high−current drive signals required to properly drive the module's internal IGBTs. Separate negative IGBT terminals are available for each phase to support the widest variety of control algorithms.

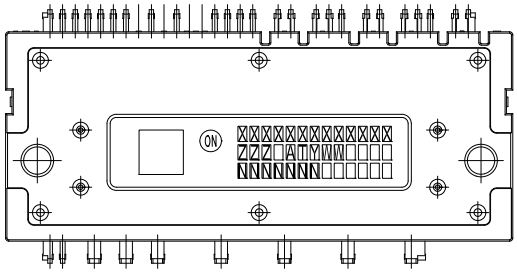
Features

- Automotive SPM® in 34 Pin DIP Package
- AEC & AQG324 Qualified and PPAP Capable
- 1200 V − 50 A 3−Phase IGBT Inverter with Integral Gate Drivers and Protection
- Low−Loss, Short−Circuit Rated IGBTs
- Very Low Thermal Resistance Using AlN DBC Substrate
- Built−In Bootstrap Diodes and Dedicated Vs Pins Simplify PCB Layout
- Separate Open−Emitter Pins from Low−Side IGBTs for Three−Phase Current Sensing
- Single−Grounded Power Supply Supported
- Built−In NTC Thermistor for Temperature Monitoring and Management
- Adjustable Over−Current Protection via Integrated Sense−IGBTs
- Isolation Rating of 2500 Vrms / 1 min
- This is a Pb−Free Device

Applications

- Automotive High Voltage Auxiliary Motors
  - Climate E−Compressors
  - Oil / Water Pumps
  - Super / Turbo Chargers
  - Variety Fans
- Motion Control
  - Industrial Motor

MARKING DIAGRAM



XXXXXXXXXXXX	= Specific Device Code
ZZZ	= Lot ID
AT	= Assembly & Test Location
Y	= Year
W	= Work Week
NNN	= Serial Number

ORDERING INFORMATION

See detailed ordering and shipping information on page 14 of this data sheet.

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Related Resources

- AN−9075 − Users Guide for 1200V SPM® 2 Series
- AN−9076 − Mounting Guide for New SPM® 2 Package
- AN−9079 − Thermal Performance of 1200 V Motion SPM® 2 Series by Mounting Torque
- Integrated Power Functions
- Integrated Drive, Protection, and System Control Functions

Integrated Power Functions

- 1200 V - 50 A IGBT inverter for three−phase DC / AC power conversion (Please refer to Figure 1)

Integrated Drive, Protection and System Control Functions

- For inverter high−side IGBTs: gate drive circuit, high−voltage isolated high−speed level shifting control circuit Under−Voltage Lock−Out Protection (UVLO)
- For inverter low−side IGBTs: gate drive circuit, Short−Circuit Protection (SCP) control supply circuit Under−Voltage Lock−Out Protection (UVLO)
- Fault signaling: corresponding to UVLO (low−side supply) and SC faults
- Input interface: active−HIGH interface, works with 3.3 / 5 V logic, Schmitt−trigger input

PIN CONFIGURATION

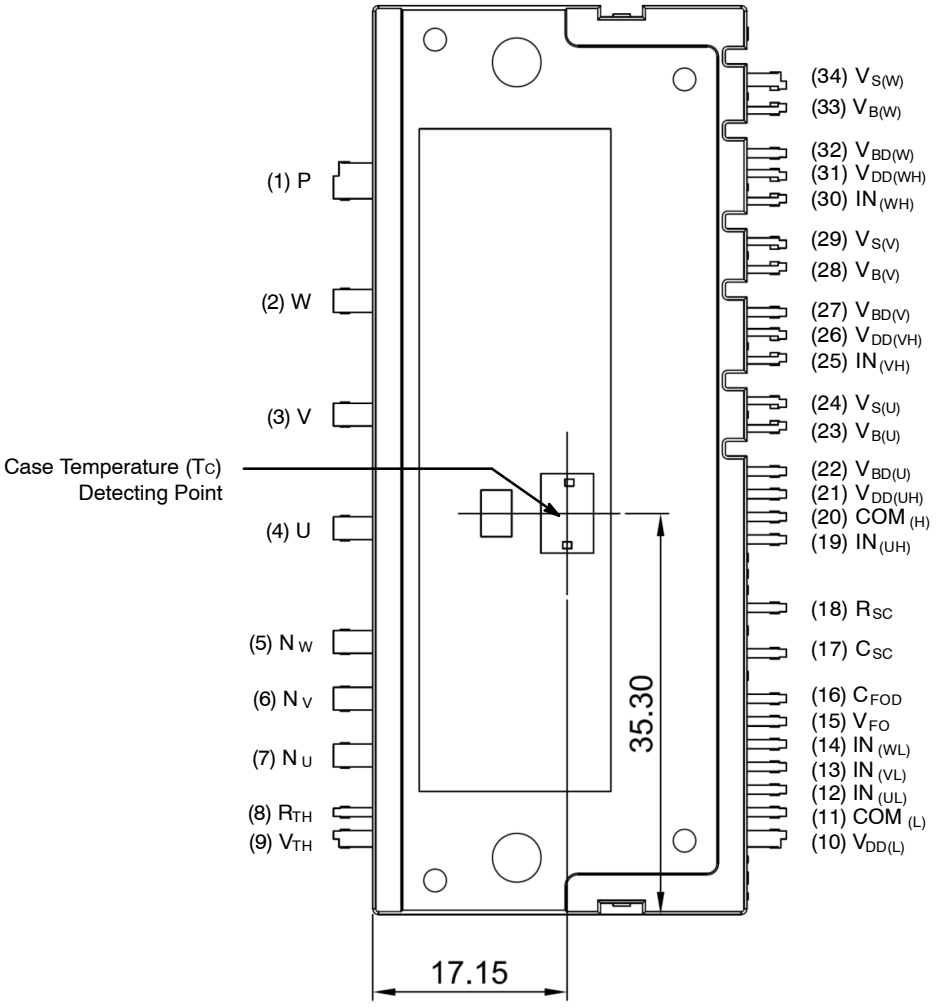


Figure 1. Pin Configuration − Top View

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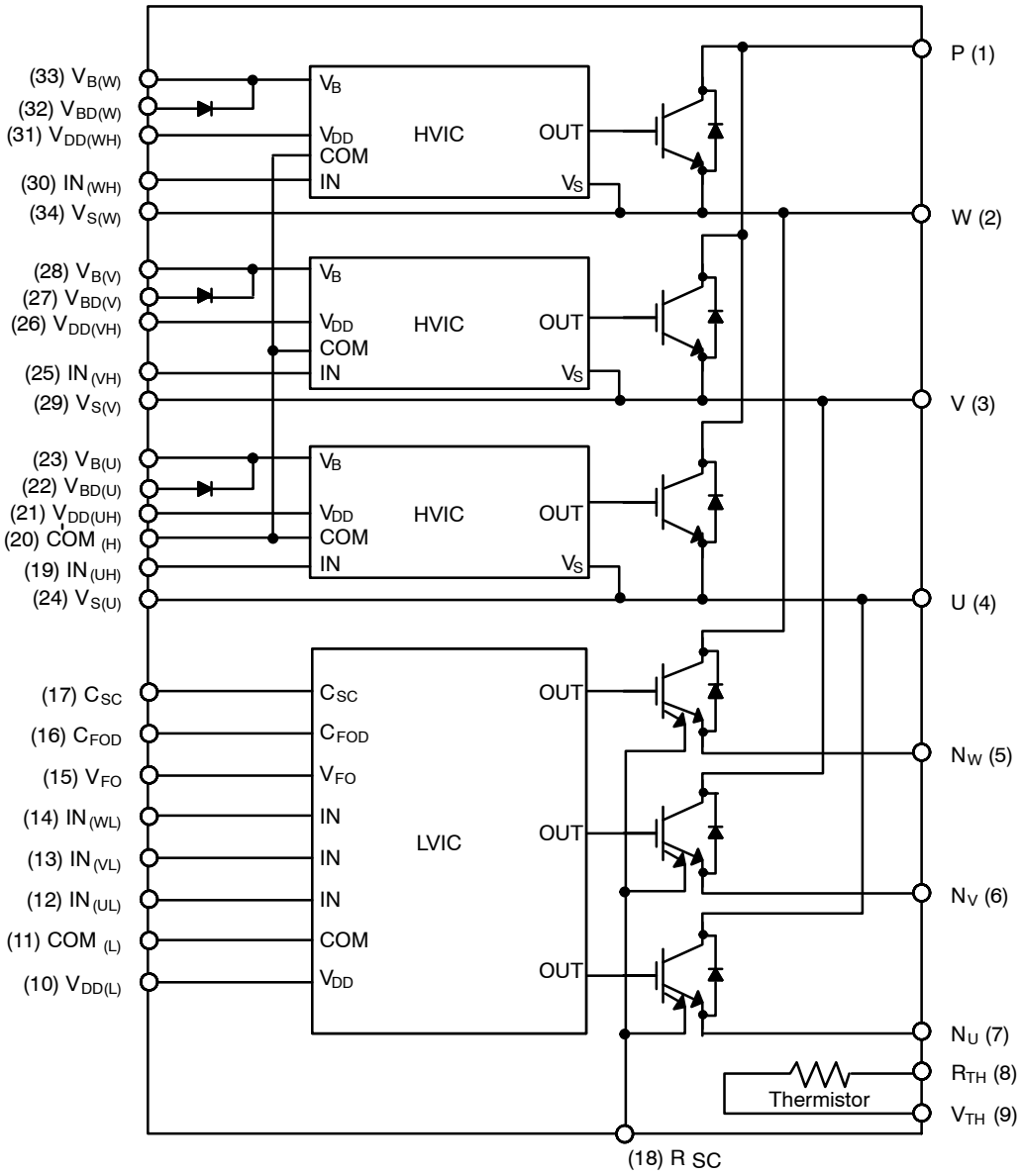
PIN DESCRIPTION

Pin Number	Pin Name	Pin Description
1	P	Positive DC−Link Input
2	W	Output for W Phase
3	V	Output for V Phase
4	U	Output for U Phase
5	N_W	Negative DC−Link Input for W Phase
6	N_V	Negative DC−Link Input for V Phase
7	N_U	Negative DC−Link Input for U Phase
8	R_{TH}	Series Resistor for Thermistor (Temperature Detection)
9	V_{TH}	Thermistor Bias Voltage
10	$V_{DD(L)}$	Low−Side Bias Voltage for IC and IGBTs Driving
11	$COM_{(L)}$	Low−Side Common Supply Ground
12	$IN_{(UL)}$	Signal Input for Low−Side U Phase
13	$IN_{(VL)}$	Signal Input for Low−Side V Phase
14	$IN_{(WL)}$	Signal Input for Low−Side W Phase
15	V_{FO}	Fault Output
16	C_{FOD}	Capacitor for Fault Output Duration Selection
17	C_{SC}	Shut Down Input for Short−Circuit Current Detection Input
18	R_{SC}	Resistor for Short−Circuit Current Detection
19	$IN_{(UH)}$	Signal Input for High−Side U Phase
20	$COM_{(H)}$	High−Side Common Supply Ground
21	$V_{DD(UH)}$	High−Side Bias Voltage for U Phase IC
22	$V_{BD(U)}$	Anode of Bootstrap Diode for U Phase High−Side Bootstrap Circuit
23	$V_{B(U)}$	High−Side Bias Voltage for U Phase IGBT Driving
24	$V_{S(U)}$	High−Side Bias Voltage Ground for U Phase IGBT Driving
25	$IN_{(VH)}$	Signal Input for High−Side V Phase
26	$V_{DD(VH)}$	High−Side Bias Voltage for V Phase IC
27	$V_{BD(V)}$	Anode of Bootstrap Diode for V Phase High−Side Bootstrap Circuit
28	$V_{B(V)}$	High−Side Bias Voltage for V Phase IGBT Driving
29	$V_{S(V)}$	High−Side Bias Voltage Ground for V Phase IGBT Driving
30	$IN_{(WH)}$	Signal Input for High−Side W Phase
31	$V_{DD(WH)}$	High−Side Bias Voltage for W Phase IC
32	$V_{BD(W)}$	Anode of Bootstrap Diode for W Phase High−Side Bootstrap Circuit
33	$V_{B(W)}$	High−Side Bias Voltage for W Phase IGBT Driving
34	$V_{S(W)}$	High−Side Bias Voltage Ground for W Phase IGBT Driving

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INTERNAL EQUIVALENT CIRCUIT AND INPUT/OUTPUT PINS



NOTES:

1. nverter low−side is composed of three IGBTs, freewheeling diodes for each IGBT, and one control IC. It has gate drive and protection functions.
2. nverter power side is composed of four inverter DC−link input terminals and three inverter output terminals.
3. Inverter high−side is composed of three IGBTs, freewheeling diodes, and three drive ICs for each IGBT.

Figure 2. Internal Block Diagram

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ABSOLUTE MAXIMUM RATINGS (Tj = 25°C unless otherwise noted)

Symbol	Rating	Conditions	Rating	Unit
INVERTER PART				
V_{PN}	Supply Voltage	Applied between P – N_U, N_V, N_W	900	V
$V_{PN(Surge)}$	Supply Voltage (Surge)	Applied between P – N_U, N_V, N_W	1000	V
V_{CES}	Collector – Emitter Voltage		1200	V
$\pm I_C$	Each IGBT Collector Current	T_C = 100°C, $T_J \le$ 150°C, $V_{DD} \ge$ 15 V	50	A
$\pm I_{CP}$	Each IGBT Collector Current (Peak)	T_C = 25°C, $T_J \le$ 150°C, Under 1 ms Pulse Width	75	A
P_C	Collector Dissipation	T_C = 25°C per One Chip	347	W
T_J	Operating Junction Temperature	V_{CES} = 960 V	–40~150	°C
		V_{CES} = 1200 V	–40~125	°C
CONTROL PART				
V_{DD}	Control Supply Voltage	Applied between $V_{DD(H)}$, $V_{DD(L)}$ – COM	20	V
V_{BS}	High–Side Control Bias Voltage	Applied between $V_{B(U)} - V_{S(U)}$, $V_{B(V)} - V_{S(V)}$, $V_{B(W)} - V_{S(W)}$	20	V
V_{IN}	Input Signal Voltage	Applied between $IN_{(UH)}$, $IN_{(VH)}$, $IN_{(WH)}$, $IN_{(UL)}$, $IN_{(VL)}$, $IN_{(WL)}$ – COM	–0.3~V_{DD} + 0.3	V
V_{FO}	Fault Output Supply Voltage	Applied between V_{FO} – COM	–0.3~V_{DD} + 0.3	V
I_{FO}	Fault Output Current	Sink Current at V_{FO} pin	2	mA
V_{SC}	Current Sensing Input Voltage	Applied between C_{SC} – COM	–0.3~V_{DD} + 0.3	V
BOOSTSTRAP DIODE PART				
V_{RRM}	Maximum Repetitive Reverse Voltage		1200	V
I_F	Forward Current	T_C = 25°C, $T_J \le$ 150°C	1.0	A
I_{FP}	Forward Current (Peak)	T_C = 25°C, $T_J \le$ 150°C, Under 1 ms Pulse Width	2.0	A
T_J	Operating Junction Temperature		–40~150	°C
TOTAL SYSTEM				
t_{SC}	Short Circuit Withstand Time	$V_{DD} = V_{BS} \le$ 16.5 V, $V_{PN} \le$ 800 V, T_J = 150°C Non–repetitive	3	μs
T_{STG}	Storage Temperature		–40~150	°C
V_{ISO}	Isolation Voltage	60 Hz, Sinusoidal, AC 1 minute, Connection Pins to Heat Sink Plate	2500	V_{rms}

Stresses exceeding those listed in the Maximum Ratings table may damage the device. If any of these limits are exceeded, device functionality should not be assumed, damage may occur and reliability may be affected.

THERMAL RESISTANCE

Symbol	Parameter	Conditions	Min	Typ	Max	Unit
$R_{th(j-c)Q}$	Junction to Case Thermal Resistance (Note 4)	Inverter IGBT part (per 1 / 6 module)	–	–	0.36	°C/W
$R_{th(j-c)F}$		Inverter FWD part (per 1 / 6 module)	–	–	0.66	°C/W
L_σ	Package Stray Inductance	P to N_U, N_V, N_W (Note 5)	–	32	–	nH

4. For the measurement point of case temperature (T_C), please refer to Figure 1. DBC discoloration and Picker Circle Printing allowed, please refer to application note AN–9190 (Impact of DBC Oxidation on SPM® Module Performance).
5. Stray inductance per phase measured per IEC 60747–15.

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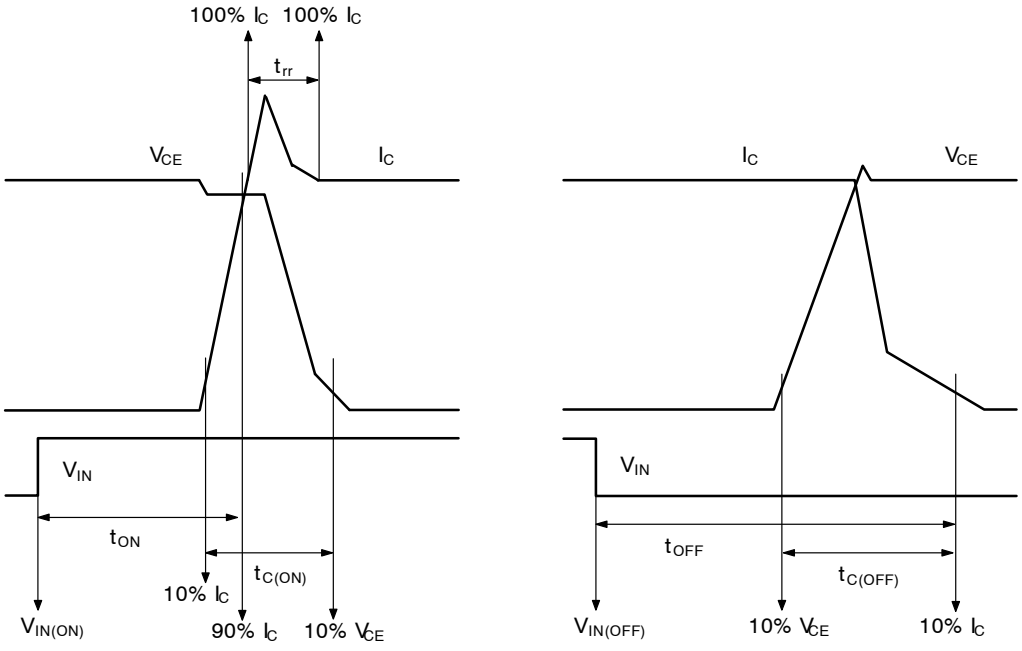
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ELECTRICAL CHARACTERISTICS

Symbol		Parameter	Conditions	Min	Typ	Max	Unit
INVERTER PART (Tj as specified)							
$V_{CE(SAT)}$		Collector –Emitter Saturation Voltage	V_{DD} = V_{BS} = 15 V, V_{IN} = 5 V, I_C = 50 A, T_J = 25°C	–	2.20	2.80	V
			V_{DD} = V_{BS} = 15 V, V_{IN} = 5 V, I_C = 50 A, T_J = 150°C	–	2.75	3.25	V
V_F		FWDi Forward Voltage	V_{IN} = 0 V, I_F = 50 A, T_J = 25°C	–	2.40	3.00	V
			V_{IN} = 0 V, I_F = 50 A, T_J = 150°C	–	2.25	2.85	V
HS	t_{ON}	High Side Switching Times	V_{PN} = 600 V, V_{DD} = 15 V, I_C = 50 A, T_J = 25°C V_{IN} = 0 V ↔ 5 V, Inductive Load See Figure 4 (Note 6)	0.90	1.40	2.00	μs
	$t_{C(ON)}$			–	0.50	0.95	μs
	t_{OFF}			–	1.10	1.70	μs
	$t_{C(OFF)}$			–	0.15	0.55	μs
	t_{rr}			–	0.20	–	μs
LS	t_{ON}	Low Side Switching Times	V_{PN} = 600 V, V_{DD} = 15 V, I_C = 50 A, T_J = 25°C V_{IN} = 0 V ↔ 5 V, Inductive Load See Figure 4 (Note 6)	0.50	1.00	1.60	μs
	$t_{C(ON)}$			–	0.50	0.95	μs
	t_{OFF}			–	1.10	1.70	μs
	$t_{C(OFF)}$			–	0.15	0.55	μs
	t_{rr}			–	0.25	–	μs
I_{CES}		Collector – Emitter Leakage Current	Tj = 25°C, V_{CE} = V_{CES}	–	–	3	mA

Product parametric performance is indicated in the Electrical Characteristics for the listed test conditions, unless otherwise noted. Product performance may not be indicated by the Electrical Characteristics if operated under different conditions.

6. t_{ON} and t_{OFF} include the propagation delay time of the internal drive IC. $t_{C(ON)}$ and $t_{C(OFF)}$ are the switching time of IGBT itself under the given gate driving condition internally. For the detailed information, *please see Figure 3*.



(a) turn-on **(b) turn-off**

Figure 3. Switching Time Definition

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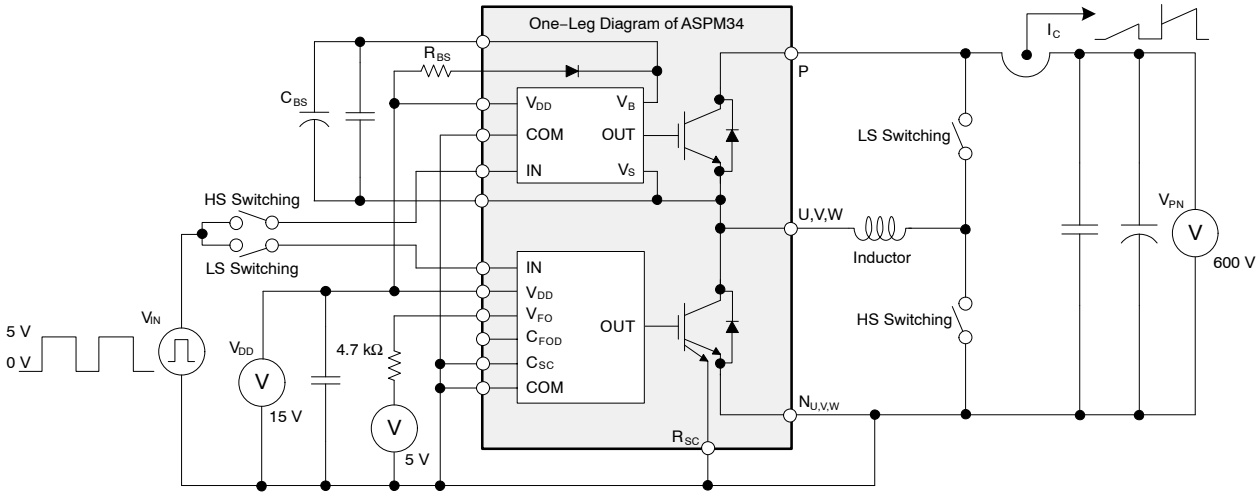


Figure 4. Example Circuit of Switching Test

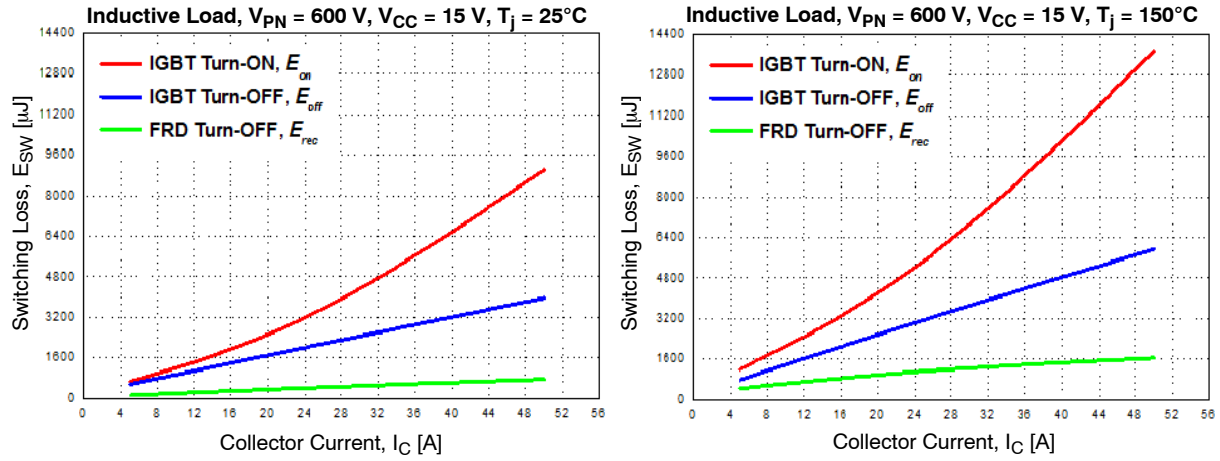


Figure 5. Switching Loss Characteristics

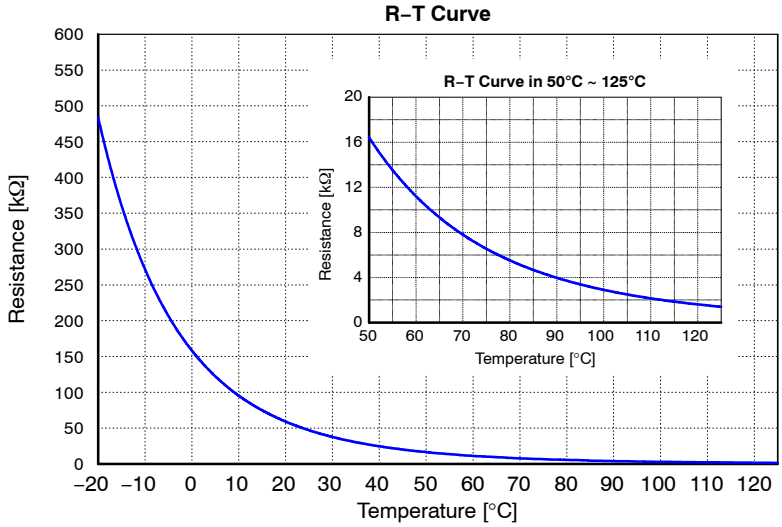


Figure 6. R−T Curve of Built−in Thermistor

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ELECTRICAL CHARACTERISTICS

Symbol	Parameter	Conditions		Min	Typ	Max	Unit
BOOTSTRAP DIODE PART (Tj as specified)							
V_F	Forward Voltage	I_F = 1.0 A, T_J = 25°C		–	2.2	–	V
t_{rr}	Reverse−Recovery Time	I_F = 1.0 A, dI_F / dt = 50 A/ms, T_J = 25°C		–	80	–	ns
CONTROL PART (T_j = 25°C unless otherwise noted)							
I_{QDDH}	Quiescent V_{DD} Supply Current	$V_{DD(UH,VH,WH)}$ = 15 V, $IN_{(UH,VH,WH)}$ = 0 V	$V_{DD(UH)} - COM_{(H)}$, $V_{DD(VH)} - COM_{(H)}$, $V_{DD(WH)} - COM_{(H)}$	–	–	0.15	mA
I_{QDDL}		$V_{DD(L)}$ = 15 V, $IN_{(UL,VL,WL)}$ = 0 V	$V_{DD(L)} - COM_{(L)}$	–	–	4.80	mA
I_{PDDH}	Operating V_{DD} Supply Current	$V_{DD(UH,VH,WH)}$ = 15 V, f_{PWM} = 20 kHz, Duty = 50%, Applied to one PWM Signal Input for High−Side	$V_{DD(UH)} - COM_{(H)}$, $V_{DD(VH)} - COM_{(H)}$, $V_{DD(WH)} - COM_{(H)}$	–	–	0.30	mA
I_{PDDL}		$V_{DD(L)}$ = 15V, f_{PWM} = 20 kHz, Duty = 50%, Applied to one PWM Signal Input for Low−Side	$V_{DD(L)} - COM_{(L)}$	–	–	15.5	mA
I_{QBS}	Quiescent V_{BS} Supply Current	V_{BS} = 15 V, $IN_{(UH,VH,WH)}$ = 0 V	$V_{B(U)} - V_{S(U)}$, $V_{B(V)} - V_{S(V)}$, $V_{B(W)} - V_{S(W)}$	–	–	0.30	mA
I_{PBS}	Operating V_{BS} Supply Current	V_{DD} = V_{BS} = 15 V, f_{PWM} = 20 kHz, Duty = 50%, Applied to one PWM Signal Input for High−Side	$V_{B(U)} - V_{S(U)}$, $V_{B(V)} - V_{S(V)}$, $V_{B(W)} - V_{S(W)}$	–	–	12.0	mA
V_{FOH}	Fault Output Voltage	V_{DD} = 15 V, V_{SC} = 0 V, V_{FO} Circuit: 4.7 kΩ to 5 V Pull−up		4.5	–	–	V
V_{FOL}		V_{DD} = 15 V, V_{SC} = 1 V, V_{FO} Circuit: 4.7 kΩ to 5 V Pull−up		–	–	0.5	V
I_{SEN}	Sensing Current of Each Sense IGBT	V_{DD} = 15 V, V_{IN} = 5 V, R_{SC} = 0 Ω, No Connection of Shunt Resistor at $N_{U,V,W}$ Terminal	I_C = 50 A	–	43	–	mA
$V_{SC(ref)}$	Short Circuit Trip Level	V_{DD} = 15 V (Note 7)	$C_{SC} - COM_{(L)}$	0.43	0.50	0.57	V
I_{SC}	Short Circuit Current Level for Trip	R_{SC} = 13 Ω (±1%), No Connection of Shunt Resistor at $N_{U,V,W}$ Terminal (Note 7)		–	75	–	A
UV_{DDD}	Supply Circuit Under−Voltage Protection	Detection Level		10.3	–	12.8	V
UV_{DDR}		Reset Level		10.8	–	13.3	V
UV_{BSD}		Detection Level		9.5	–	12.0	V
UV_{BSR}		Reset Level		10.0	–	12.5	V
t_{FOD}	Fault−Out Pulse Width	C_{FOD} = Open	(Note 8)	50	–	–	μs
		C_{FOD} = 2.2 nF		1.7	–	–	ms
$V_{IN(ON)}$	ON Threshold Voltage	Applied between $IN_{(UH,VH,WH)} - COM_{(H)}$, $IN_{(UL,VL,WL)} - COM_{(L)}$		–	–	2.6	V
$V_{IN(OFF)}$	OFF Threshold Voltage			0.8	–	–	V
R_{TH}	Resistance of Thermistor	at T_{TH} = 25°C	*See Figure 6* (Note 9)	–	47	–	kΩ
		at T_{TH} = 100°C		–	2.9	–	kΩ

Product parametric performance is indicated in the Electrical Characteristics for the listed test conditions, unless otherwise noted. Product performance may not be indicated by the Electrical Characteristics if operated under different conditions.

7. Short−circuit current protection functions only at the low−sides because the sense current is divided from main current at low−side IGBTs. Inserting the shunt resistor for monitoring the phase current at N_U, N_V, N_W terminal, the trip level of the short−circuit current is changed.
8. The fault−out pulse width t_{FOD} depends on the capacitance value of C_{FOD} according to the following approximate equation: t_{FOD} = 0.8 x 10^6 x C_{FOD} [s].
9. T_{TH} is the temperature of thermistor itself. To know case temperature (T_C), conduct experiments considering the application.

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RECOMMENDED OPERATING RANGES

Symbol	Parameter	Conditions	Min	Typ	Max	Unit
V_{PN}	Supply Voltage	Applied between P – N_U, N_V, N_W	300	600	800	V
V_{DD}	Control Supply Voltage	Applied between $V_{DD(UH,VH,WH)}$ – $COM_{(H)}$, $V_{DD(L)}$ – $COM_{(L)}$	14.0	15.0	16.5	V
V_{BS}	High−Side Bias Voltage	Applied between $V_{B(U)}$ – $V_{S(U)}$, $V_{B(V)}$ – $V_{S(V)}$, $V_{B(W)}$ – $V_{S(W)}$	13.0	15.0	18.5	V
dV_{DD} / dt, dV_{BS} / dt	Control Supply Variation		−1	–	1	V/μs
t_{dead}	Blanking Time for Preventing Arm – Short	For Each Input Signal	2.0	–	–	μs
f_{PWM}	PWM Input Signal	−40°C ≤ T_C ≤ 125°C, −40°C ≤ T_J ≤ 150°C	–	–	20	kHz
V_{SEN}	Voltage for Current Sensing	Applied between N_U, N_V, N_W – $COM_{(H, L)}$ (Including Surge Voltage)	−5	–	5	V
$PW_{IN(ON)}$	Minimum Input Pulse Width	V_{DD} = V_{BS} = 15 V, I_C ≤ 75 A, Wiring Inductance between $N_{U,V,W}$ and DC Link N < 10 nH (Note 10)	2.5	–	–	μs
$PW_{IN(OFF)}$			2.5	–	–	
T_J	Junction Temperature		−40	–	150	°C

Functional operation above the stresses listed in the Recommended Operating Ranges is not implied. Extended exposure to stresses beyond the Recommended Operating Ranges limits may affect device reliability.
10. This product might not make output response if input pulse width is less than the recommended value.

MECHANICAL CHARACTERISTICS AND RATINGS

Parameter	Conditions		Min	Typ	Max	Unit
Device Flatness	*See Figure 7*		0	–	+200	μm
Mounting Torque	Mounting Screw: M4 *See Figure 8*	Recommended 1.0 N · m	0.9	1.0	1.5	N · m
		Recommended 10.1 kg · cm	9.1	10.1	15.1	kg · cm
Terminal Pulling Strength	Load 19.6 N		10	–	–	s
Terminal Bending Strength	Load 9.8 N, 90 degrees Bend		2	–	–	times
Weight			–	50	–	g

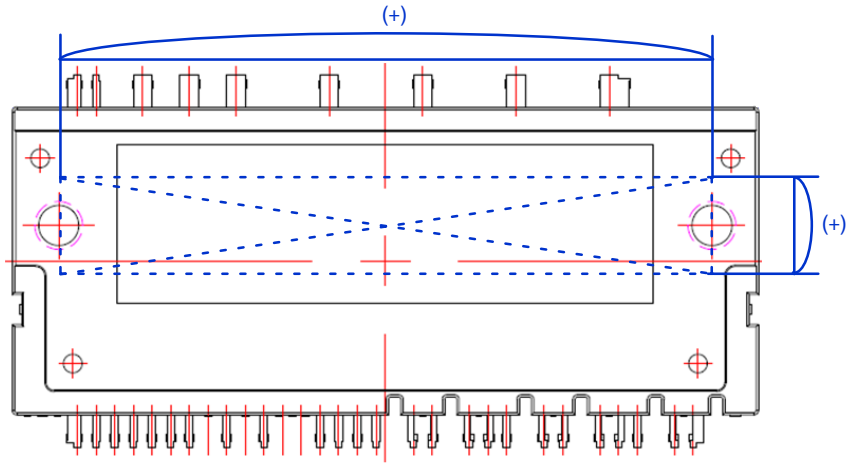
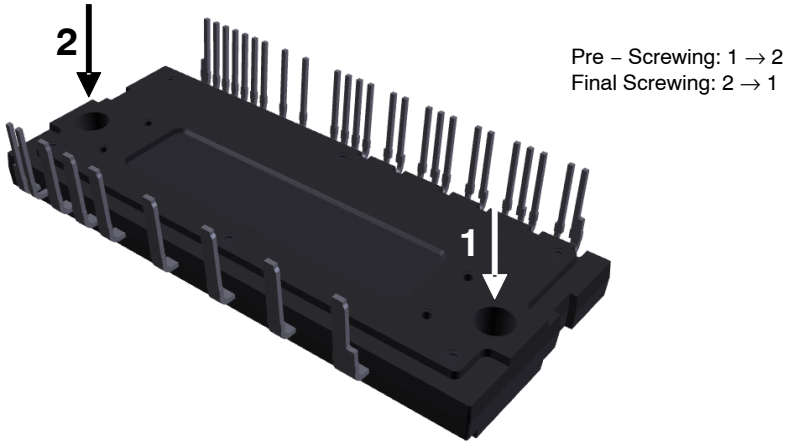


Figure 7. Flatness Measurement Position

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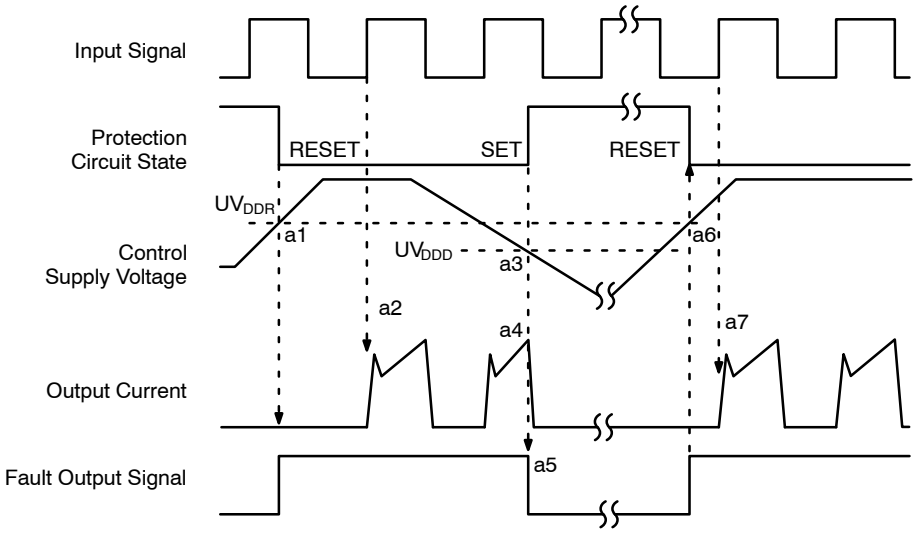
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NOTES:
11. Do not over torque when mounting screws. Too much mounting torque may cause DBC cracks, as well as bolts and Al heat−sink destruction.
12. Avoid one−sided tightening stress. Figure 8 shows the recommended torque order for the mounting screws. Uneven mounting can cause the DBC substrate of package to be damaged. The pre−screwing torque is set to 20~30% of maximum torque rating.

Figure 8. Mounting Screws Torque Order

TIME CHARTS OF SPMS PROTECTIVE FUNCTION

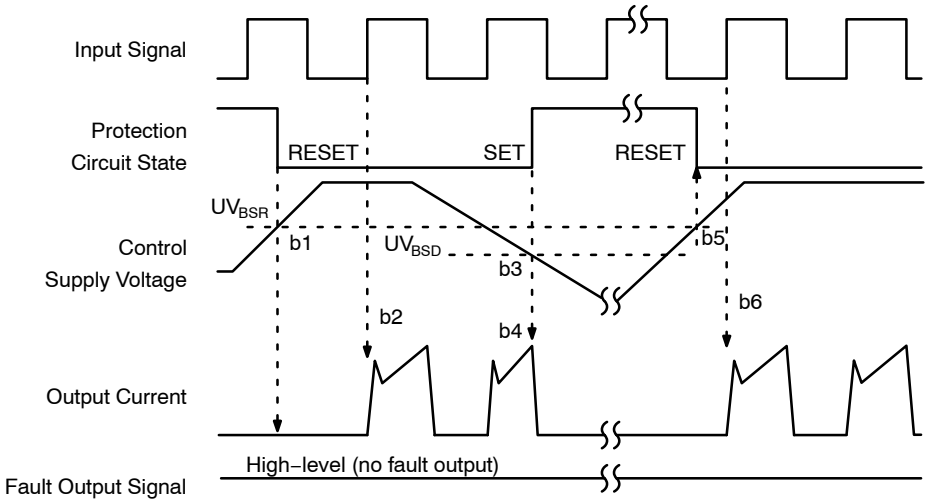


a1: Control supply voltage rises: after the voltage rises UV_{DDR}, the circuits start to operate when the next input is applied.
a2: Normal operation: IGBT ON and carrying current.
a3: Under-voltage detection (UV_{DDD}).
a4: IGBT OFF in spite of control input condition.
a5: Fault output operation starts with a fixed pulse width according to the condition of the external capacitor C_{FOD}.
a6: Under-voltage reset (UVDDR).
a7: Normal operation: IGBT ON and carrying current by triggering next signal from LOW to HIGH.

Figure 9. Under-Voltage Protection (Low-Side)

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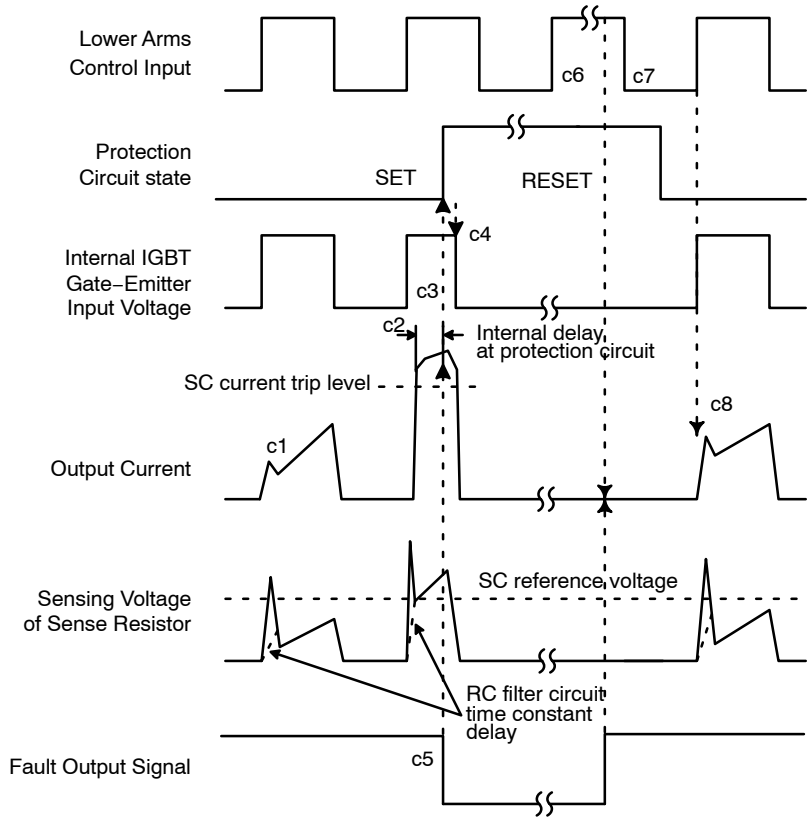


b1: Control supply voltage rises: after the voltage reaches UV_{BSR}, the circuits start to operate when the next input is applied.
b2: Normal operation: IGBT ON and carrying current.
b3: Under-voltage detection (UV_{BSD}).
b4: IGBT OFF in spite of control input condition, but there is no fault output signal.
b5: Under-voltage reset (UV_{BSR}).
b6: Normal operation: IGBT ON and carrying current by triggering next signal from LOW to HIGH.

Figure 10. Under-Voltage Protection (High-Side)

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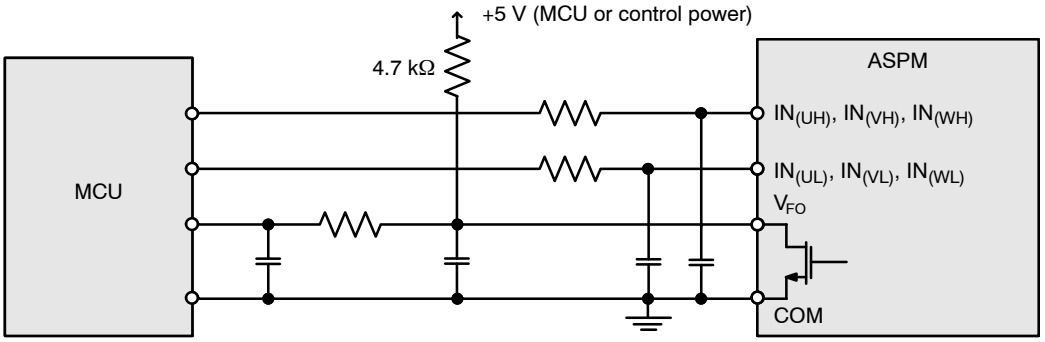
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(With the external sense resistance and RC filter connection)
c1: Normal operation: IGBT ON and carrying current.
c2: Short-circuit current detection (SC trigger).
c3: All low-side IGBTs gate are hard interrupted.
c4: All low-side IGBTs turn OFF.
c5: Fault output operation starts with a fixed pulse width according to the condition of the external capacitor C_{FOD}.
c6: Input HIGH: IGBT ON state, but during the active period of fault output, the IGBT doesn't turn ON.
c7: Fault output operation finishes, but IGBT doesn't turn on until triggering the next signal from LOW to HIGH.
c8: Normal operation: IGBT ON and carrying current.

Figure 11. Short-Circuit Current Protection (Low-Side Operation Only)

INPUT/OUTPUT INTERFACE CIRCUIT

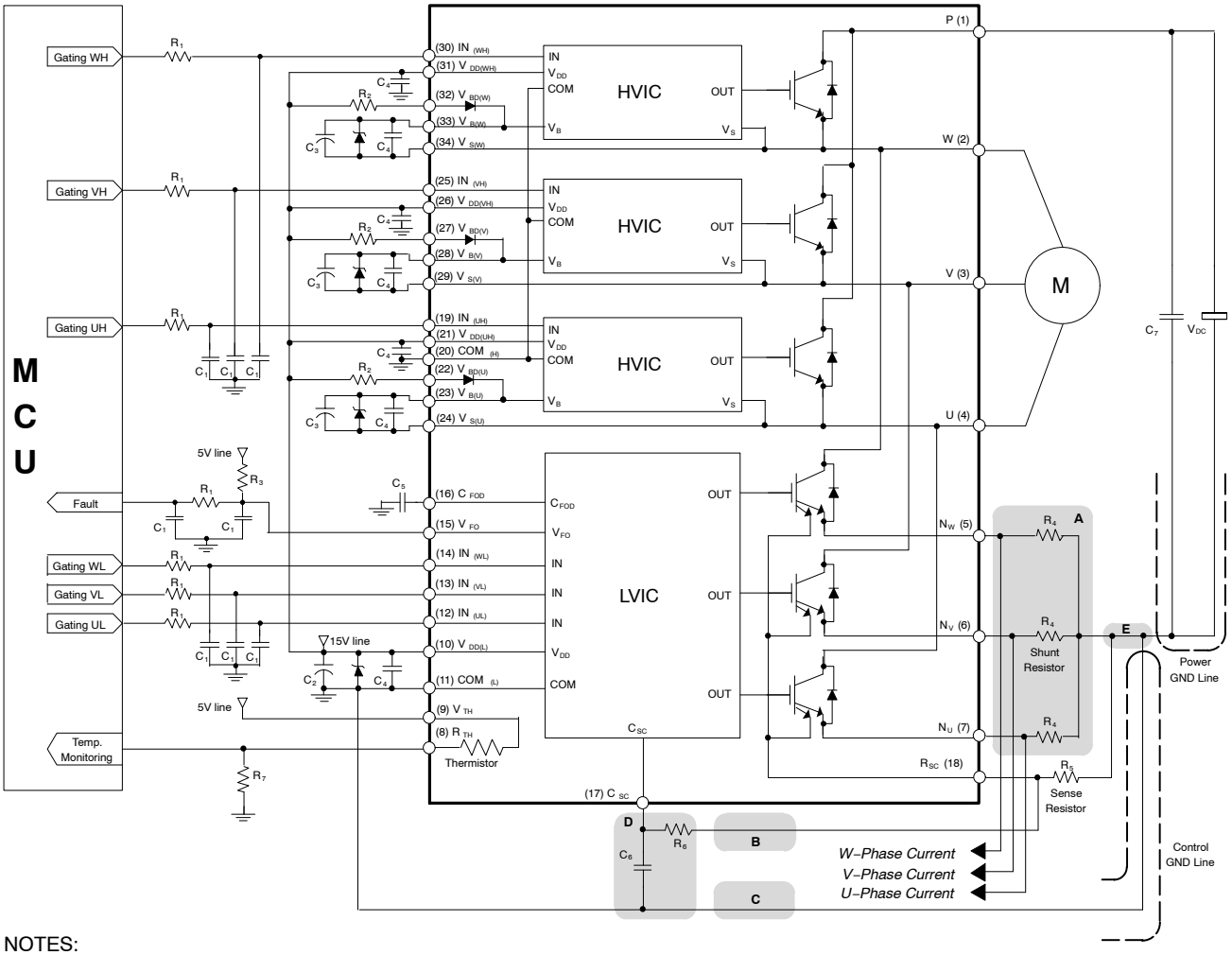


NOTE:
13. RC coupling at each input might change depending on the PWM control scheme used in the application and the wiring impedance of the application's printed circuit board. The input signal section of the Motion SPM 2 product integrates 5 kΩ (typ.) pull−down resistor. Therefore, when using an external filtering resistor, please pay attention to the signal voltage drop at input terminal.

Figure 12. Recommended MCU I/O Interface Circuit

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NOTES:

14. To avoid malfunction, the wiring of each input should be as short as possible (less than 2 – 3 cm).
15. V_{FO} output is an open−drain type. This signal line should be pulled up to the positive side of the MCU or control power supply with a resistor that makes I_{FO} up to 2 mA. *Please refer to Figure 13*.
16. Fault out pulse width can be adjust by capacitor C_5 connected to the C_{FOD} terminal.
17. Input signal is active−HIGH type. There is a 5 kΩ resistor inside the IC to pull−down each input signal line to GND. RC coupling circuits should be adopted for the prevention of input signal oscillation. R_1C_1 time constant should be selected in the range 50~ 50 ns (recommended R_1 = 100 Ω, C_1 = 1 nF).
18. Each wiring pattern inductance of point A should be minimized (recommend less than 10 nH). Use the shunt resistor R_4 of surface mounted (SMD) type to reduce wiring inductance. To prevent malfunction, wiring of point E should be connected to the terminal of the shunt resistor R_4 as close as possible.
19. To insert the shunt resistor to measure each phase current at N_U, N_V, N_W terminal, it makes to change the trip level I_{SC} about the short−circuit current.
20. To prevent errors of the protection function, the wiring of points B, C, and D should be as short as possible. The wiring of B between C_{SC} filter and R_{SC} terminal should be divided at the point that is close to the terminal of sense resistor R_5.
21. For stable protection function, use the sense resistor R_5 with resistance variation within 1% and low inductance value.
22. In the short−circuit protection circuit, select the R_6C_6 time constant in the range 1.0~1.5 μs. R_6 should be selected with a minimum of 10 times larger resistance than sense resistor R_5. Do enough evaluaiton on the real system because short−circuit protection time may vary wiring pattern layout and value of the R_6C_6 time constant.
23. Each capacitor should be mounted as close to the pins of the ASPM34 product as possible.
24. To prevent surge destruction, the wiring between the smoothing capacitor C_7 and the P & GND pins should be as short as possible. The use of a high−frequency non−inductive capacitor of around 0.1~0.22 μF between the P & GND pins is recommended.
25. Relays are used in most systems of electrical equipments in industrial application. In these cases, there should be sufficient distance between the MCU and the relays.
26. The Zener diode or transient voltage suppressor should be adapted for the protection of ICs from the surge destruction between each pair of control supply terminals (recommended Zener diode is 22 V / 1 W, which has the lower Zener impedance characteristic than about 15 Ω).
27. C_2 of around seven times larger than bootstrap capacitor C_3 is recommended.
28. Please choose the electrolytic capacitor with good temperature characteristic in C_3. Choose 0.1~0.2 μF R−category ceramic capacitors with good temperature and frequency characteristics in C_4.

Figure 13. Typical Application Circuit

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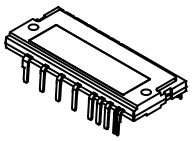
PACKAGE MARKING AND ORDERING INFORMATION

Device	Device Marking	Package	Shipping
NFVA25012NP2T	NFVA25012NP2T	ASPM34−CAA (Pb−Free)	6 Units/Tube

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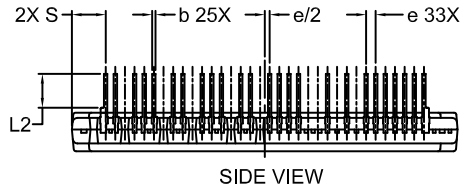
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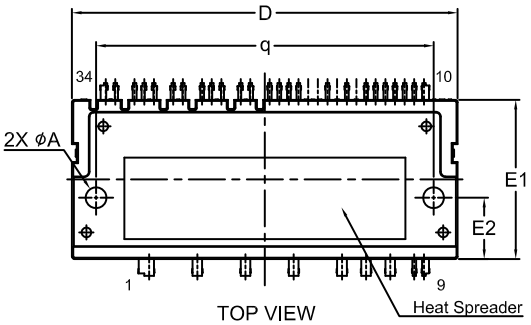


DIP34 80x33, AUTOMOTIVE MODULE
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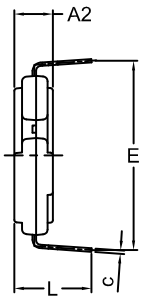
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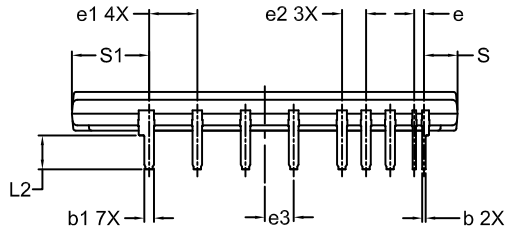
SIDE VIEW



TOP VIEW



END VIEW



SIDE VIEW

NOTES:

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2. CONTROLLING DIMENSION: MILLIMETERS
3. DIMENSIONS ARE EXCLUSIVE OF BURRS, MOLD FLASH AND TIE BAR EXTRUSIONS.

	MILLIMETERS		
DIM	MIN.	NOM.	MAX.
A2	7.90	8.00	8.10
b	0.60	0.70	0.80
b1	1.90	2.00	2.10
c	0.65	0.70	0.80
D	79.70	80.00	80.30
E	38.76	39.26	39.76
E1	32.70	33.00	33.30
E2	12.50	12.70	12.90
e	1.70	2.00	2.30
e1	9.70	10.00	10.30
e2	4.70	5.00	5.30
e3	5.70	6.00	6.30
L	15.50	16.00	16.50
L2	6.70	7.00	7.30
q	69.75	70.00	70.25
S	7.00 REF		
S1	16.00 REF		
ϕA	4.10	4.30	4.50

GENERIC MARKING DIAGRAM*



XXXX = Specific Device Code
ZZZ = Lot ID
AT = Assembly & Test Location
Y = Year
W = Work Week
NNN = Serial Number

*This information is generic. Please refer to device data sheet for actual part marking. Pb−Free indicator, “G” or microdot “▪”, may or may not be present. Some products may not follow the Generic Marking.

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