



SGM8602

2.2mA, 12MHz, Low Noise, Rail-to-Rail I/O Tiny Package, CMOS Operational Amplifier

GENERAL DESCRIPTION

The SGM8602 is a dual, low voltage, low noise and low power operational amplifier, which can operate from 2.1V to 5.5V single supply, while consuming only 2.2mA quiescent current at 5V.

The minimum input common mode voltage is within 0.1V below the negative rail, and the output swing is rail-to-rail with heavy loads. The SGM8602 exhibits a high gain-bandwidth product of 12MHz and a slew rate of 9V/ μ s. These specifications make the operational amplifier appropriate for various applications.

The SGM8602 is available in Green SOT-23-8 and TDFN-2 $\times$ 3-8L packages. It is specified over the extended -40°C to +125°C industrial temperature range.

FEATURES

- **Input Offset Voltage: 5.1mV (MAX)**
- **High Gain-Bandwidth Product: 12MHz**
- **High Slew Rate: 9V/ μ s**
- **Settling Time to 0.1% with 2V Step: 0.2 μ s**
- **Overload Recovery Time: 0.4 μ s**
- **Low Noise: 9nV/ $\sqrt{\text{Hz}}$ at 10kHz**
- **Rail-to-Rail Input and Output**
- **Supply Voltage Range: 2.1V to 5.5V**
- **Input Voltage Range: -0.1V to +5.6V with $V_S = 5.5V$**
- **Low Power: 2.2mA (TYP) Supply Current**
- **-40°C to +125°C Operating Temperature Range**
- **Available in Green SOT-23-8 and TDFN-2 $\times$ 3-8L Packages**

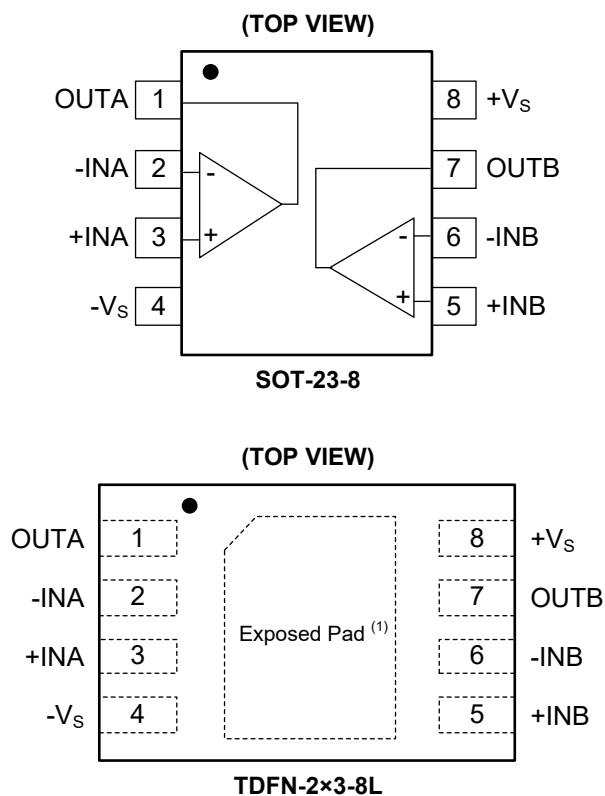
APPLICATIONS

Sensors
Audio
Active Filters
A/D Converters
Communications
Test Equipment
Cellular and Cordless Phones
Laptops and PDAs
Photodiode Amplification
Battery-Powered Instrumentation

2.2mA, 12MHz, Low Noise, Rail-to-Rail I/O Tiny Package, CMOS Operational Amplifier

MODEL	PACKAGE DESCRIPTION	SPECIFIED TEMPERATURE RANGE	ORDERING NUMBER	PACKAGE MARKING	PACKING OPTION
SGM8602	SOT-23-8	-40°C to +125°C	SGM8602XN8G/TR	SUDXX	Tape and Reel, 3000
	TDFN-2×3-8L	-40°C to +125°C	SGM8602XTDC8G/TR	8602 XXXX	Tape and Reel, 3000

PIN CONFIGURATIONS



NOTE: 1. Exposed pad can be connected to -Vs or left floating.

ELECTRICAL CHARACTERISTICS

(At $T_A = +25^\circ\text{C}$, $V_S = 5\text{V}$, $V_{CM} = V_S/2$ and $R_L = 600\Omega$, unless otherwise noted.)

PARAMETER	CONDITIONS	MIN	TYP	MAX	UNITS
Input Characteristics					
Input Offset Voltage (V _{OS})			1.2	5.1	mV
	T _A = -40°C to +125°C			5.5	
Input Bias Current (I _B)			1		pA
Input Offset Current (I _{OS})			1		pA
Input Common Mode Voltage Range (V _{CM})	V _S = 5.5V	-0.1		5.6	V
Common Mode Rejection Ratio (CMRR)	V _S = 5.5V, V _{CM} = -0.1V to 4V	67	84		dB
	T _A = -40°C to +125°C	66			
	V _S = 5.5V, V _{CM} = -0.1V to 5.6V	60	75		dB
	T _A = -40°C to +125°C	59			
Open-Loop Voltage Gain (A _{OL})	R _L = 10kΩ, V _{OUT} = 0.05V to 4.95V	97	104		dB
	T _A = -40°C to +125°C	72			
	R _L = 600Ω, V _{OUT} = 0.15V to 4.85V	84	92		dB
	T _A = -40°C to +125°C	64			
Input Offset Voltage Drift (ΔV _{OS} /ΔT)			4.7		μV/°C
Output Characteristics					
Output Voltage Swing from Rail (V _{OL})	R _L = 10kΩ		6	12	mV
	T _A = -40°C to +125°C			17	
	R _L = 600Ω		75	100	mV
	T _A = -40°C to +125°C			144	
Output Current (I _{OUT})		52	65		mA
	T _A = -40°C to +125°C	36			
Power Supply					
Operating Voltage Range		2.1		5.5	V
Power Supply Rejection Ratio (PSRR)	V _S = +2.1V to +5.5V, V _{CM} = (-V _S) + 0.5V	68	82		dB
	T _A = -40°C to +125°C	63			
Quiescent Current (I _Q)	I _{OUT} = 0		2.2	2.8	mA
	T _A = -40°C to +125°C			3.6	
Dynamic Performance					
Gain-Bandwidth Product (GBP)	R _L = 600Ω		12		MHz
Slew Rate (SR)	G = 1, 2V output step		9.0		V/μs
Settling Time to 0.1% (t _s)	G = 1, 2V output step		0.2		μs
Overload Recovery Time	V _{IN} × Gain = V _S		0.4		μs
Phase Margin (φ _O)	R _L = 600Ω		65		°
Noise Performance					
Input Voltage Noise Density (e _n)	f = 1kHz		13		nV/√Hz
	f = 10kHz		9		

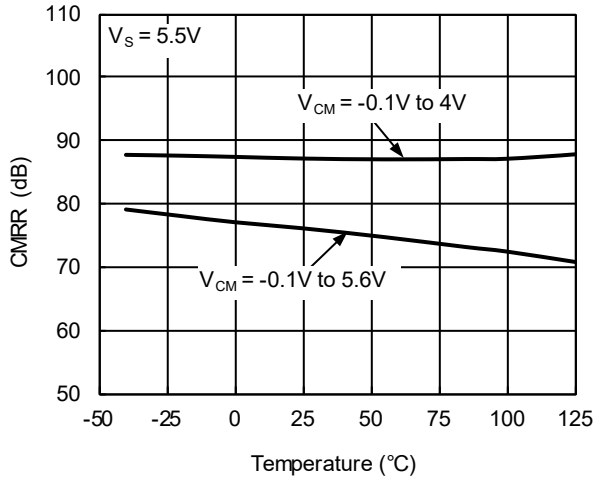
ELECTRICAL CHARACTERISTICS (continued)(At $T_A = +25^\circ\text{C}$, $V_S = 2.1\text{V}$, $V_{CM} = V_S/2$ and $R_L = 600\Omega$, unless otherwise noted.)

PARAMETER	CONDITIONS	MIN	TYP	MAX	UNITS
Input Characteristics					
Input Offset Voltage (V _{OS})			1.2	5.5	mV
	T _A = -40°C to +125°C			5.9	
Input Bias Current (I _B)			1		pA
Input Offset Current (I _{OS})			1		pA
Input Common Mode Voltage Range (V _{CM})	V _S = 2.1V	-0.1		2.2	V
Common Mode Rejection Ratio (CMRR)	V _S = 2.1V, V _{CM} = - 0.1V to 0.6V	60	77		dB
	T _A = -40°C to +125°C	51			
	V _S = 2.1V, V _{CM} = - 0.1V to 2.2V	53	68		dB
	T _A = -40°C to +125°C	46			
Open-Loop Voltage Gain (A _{OL})	R _L = 10kΩ , V _{OUT} = 0.05V to 2.05V	90	100		dB
	T _A = -40°C to +125°C	68			
	R _L = 600Ω , V _{OUT} = 0.15V to 1.95V	75	88		dB
	T _A = -40°C to +125°C	63			
Input Offset Voltage Drift (ΔV _{OS} /ΔT)			4.5		μV/°C
Output Characteristics					
Output Voltage Swing from Rail (V _{OL})	R _L = 10kΩ		4	10	mV
	T _A = -40°C to +125°C			12	
	R _L = 600Ω		36	51	mV
	T _A = -40°C to +125°C			67	
Output Current (I _{OUT})		15	30		mA
	T _A = -40°C to +125°C	7			
Power Supply					
Quiescent Current (I _Q)	I _{OUT} = 0		2.2	2.8	mA
	T _A = -40°C to +125°C			3.6	
Dynamic Performance					
Gain-Bandwidth Product (GBP)	R _L = 600Ω		11.5		MHz
Slew Rate (SR)	G = 1, 2V output step		8.6		V/μs
Settling Time to 0.1% (t _S)	G = 1, 2V output step		0.2		μs
Overload Recovery Time	V _{IN} × Gain = V _S		0.7		μs
Phase Margin (φ _O)	R _L = 600Ω		65		°
Noise Performance					
Input Voltage Noise Density (e _n)	f = 1kHz		15		nV/√Hz
	f = 10kHz		9		

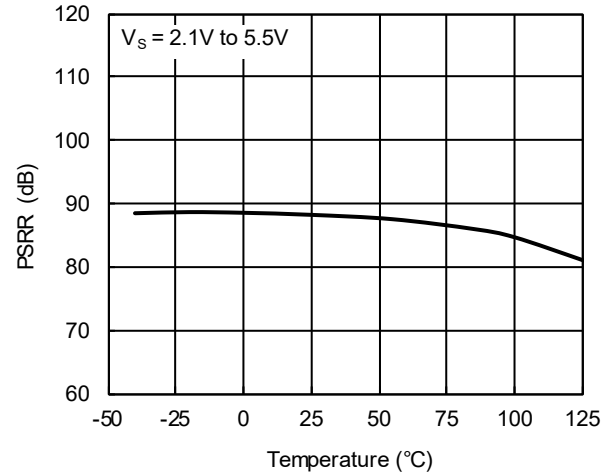
TYPICAL PERFORMANCE CHARACTERISTICS

At $T_A = +25^\circ\text{C}$, $V_{CM} = V_S/2$ and $R_L = 600\Omega$, unless otherwise noted.

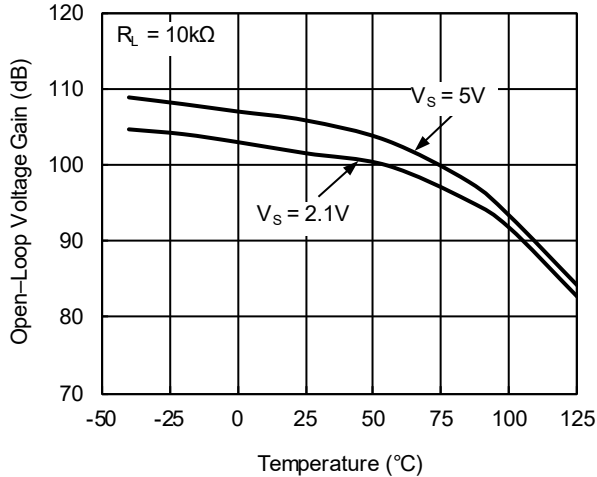
CMRR vs. Temperature



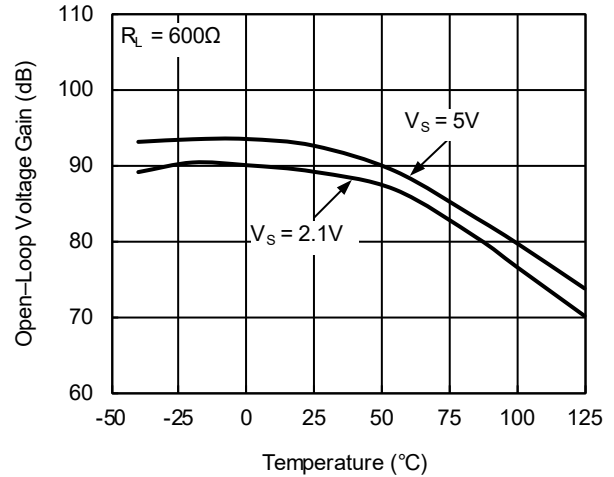
PSRR vs. Temperature



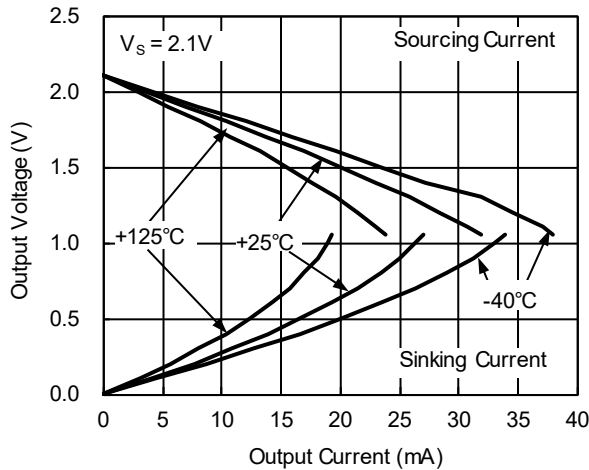
Open-Loop Voltage Gain vs. Temperature



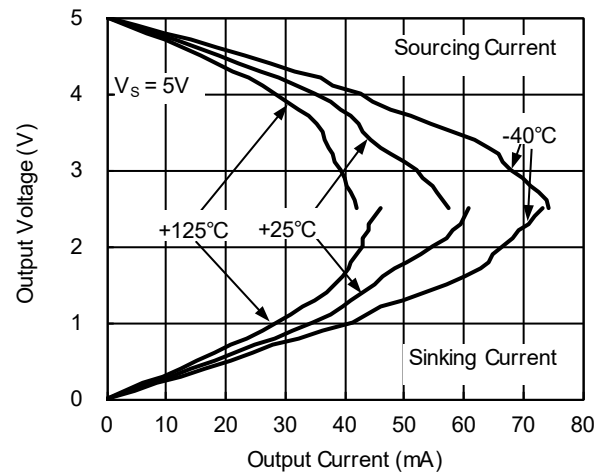
Open-Loop Voltage Gain vs. Temperature



Output Voltage Swing vs. Output Current



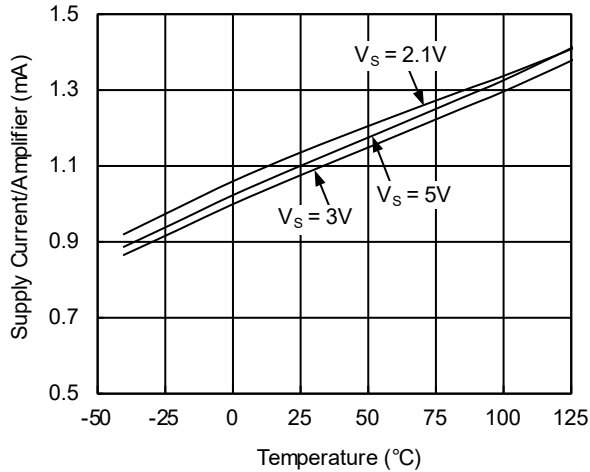
Output Voltage Swing vs. Output Current



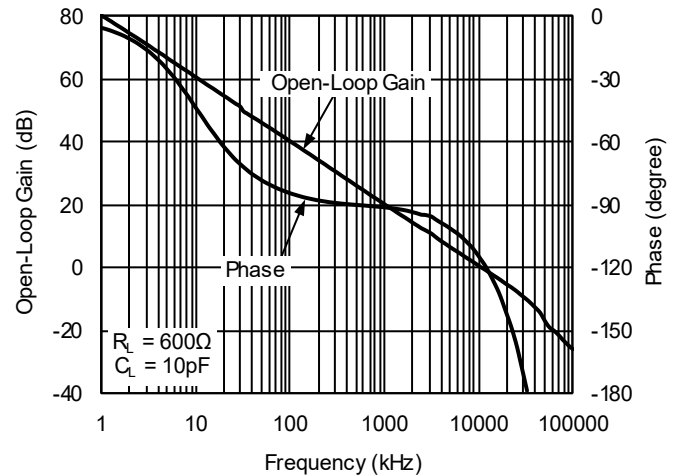
TYPICAL PERFORMANCE CHARACTERISTICS (continued)

At $T_A = +25^\circ\text{C}$, $V_{CM} = V_S/2$ and $R_L = 600\Omega$, unless otherwise noted.

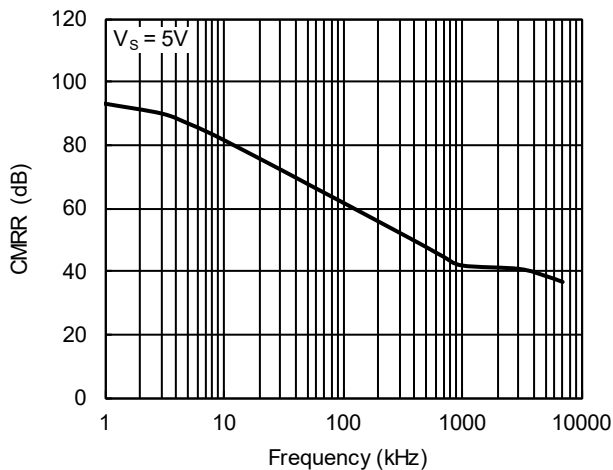
Supply Current vs. Temperature



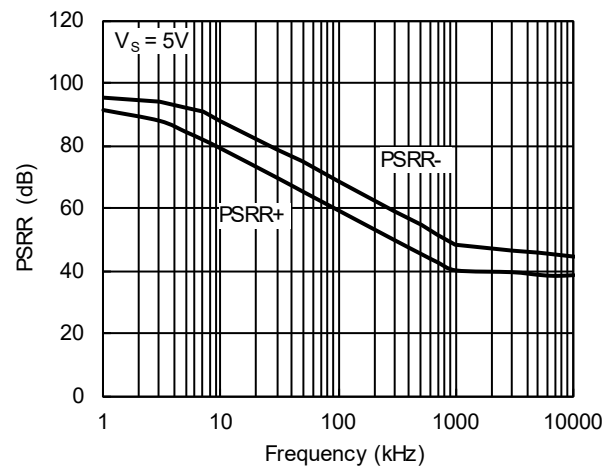
Open-Loop Gain and Phase vs. Frequency



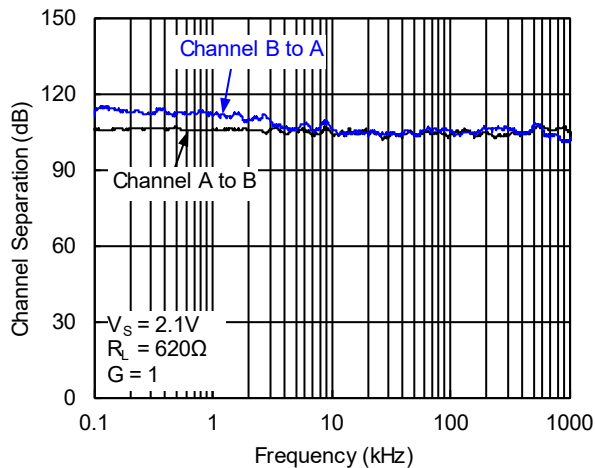
CMRR vs. Frequency



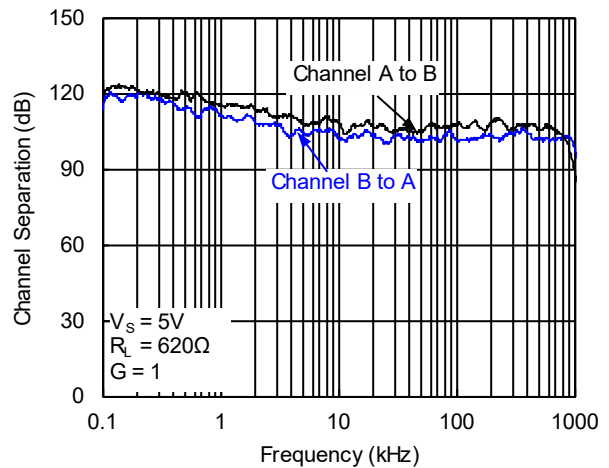
PSRR vs. Frequency



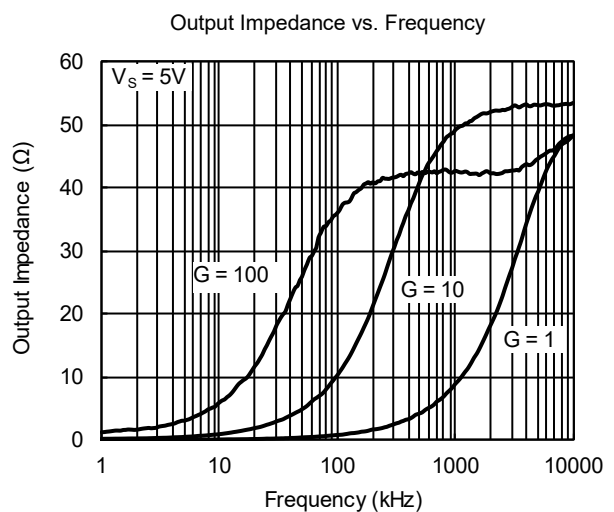
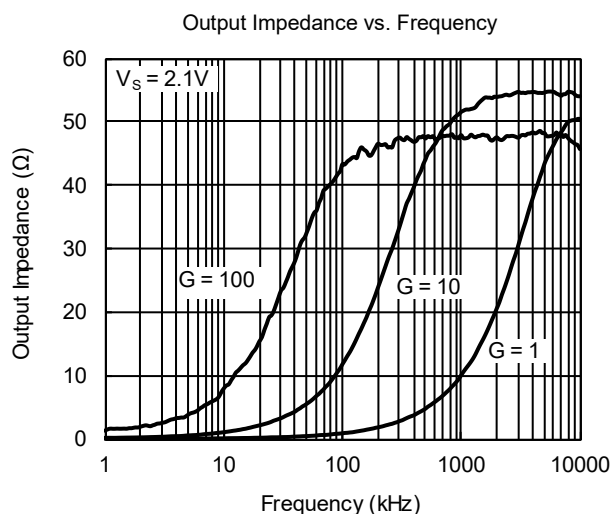
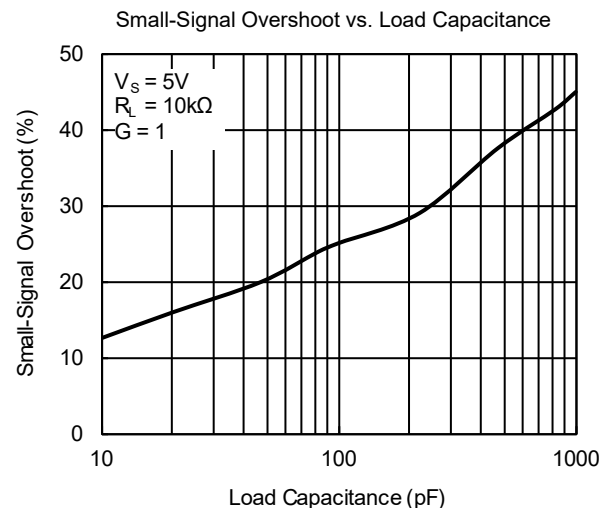
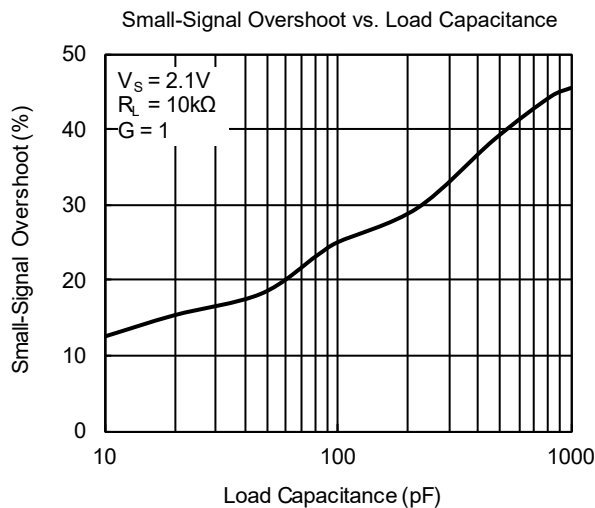
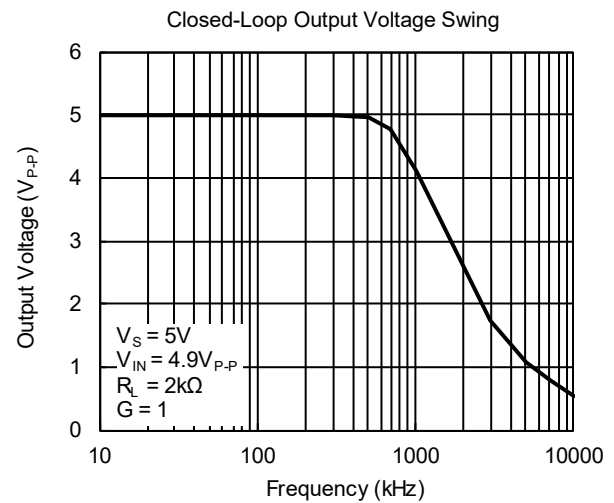
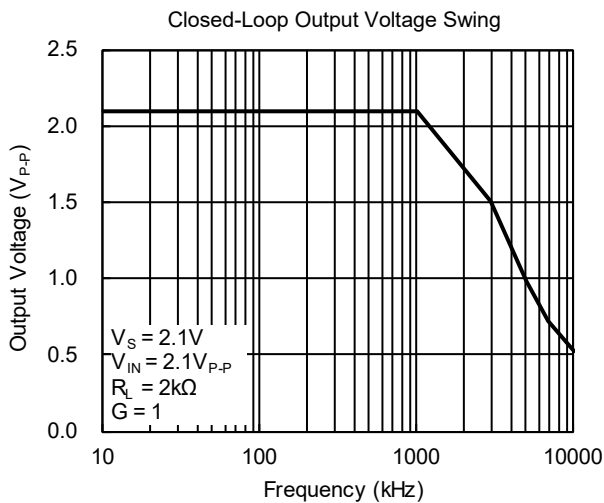
Channel Separation vs. Frequency



Channel Separation vs. Frequency



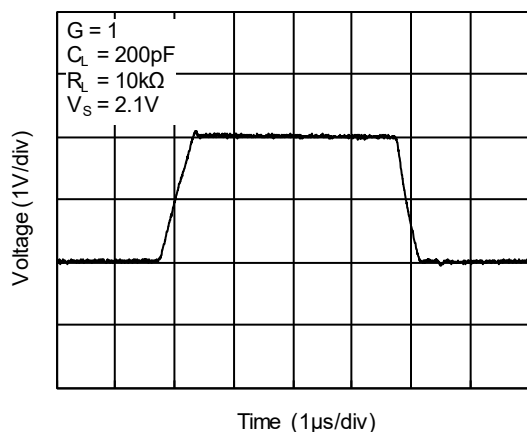
TYPICAL PERFORMANCE CHARACTERISTICS (continued)

At $T_A = +25^\circ\text{C}$, $V_{CM} = V_S/2$ and $R_L = 600\Omega$, unless otherwise noted.

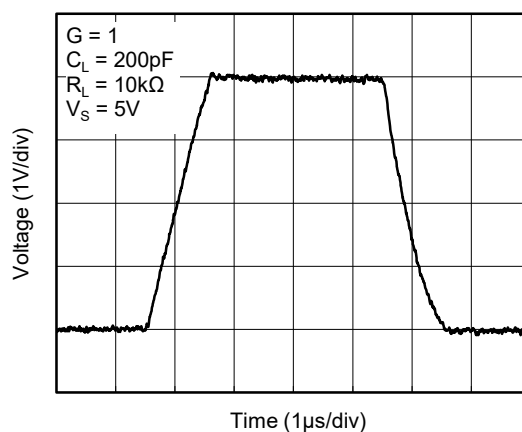
TYPICAL PERFORMANCE CHARACTERISTICS (continued)

At $T_A = +25^\circ\text{C}$, $V_{CM} = V_S/2$ and $R_L = 600\Omega$, unless otherwise noted.

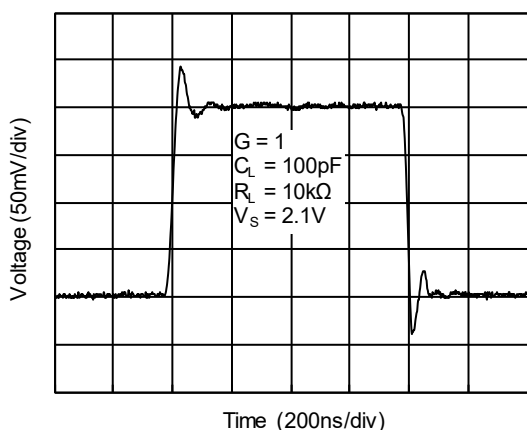
Large-Signal Step Response



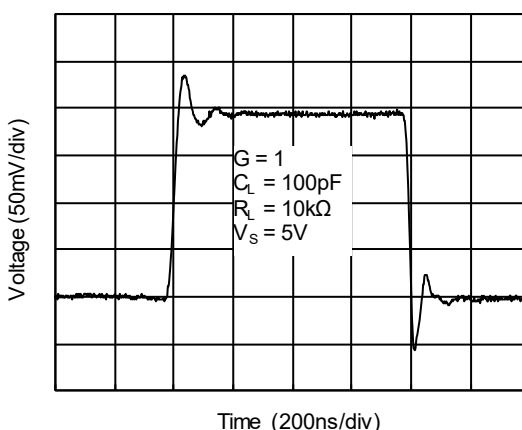
Large-Signal Step Response



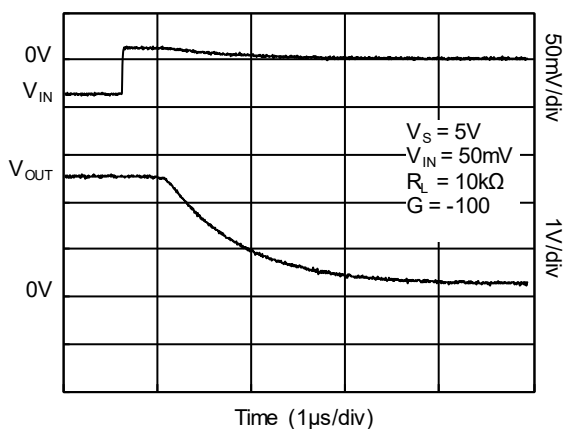
Small-Signal Step Response



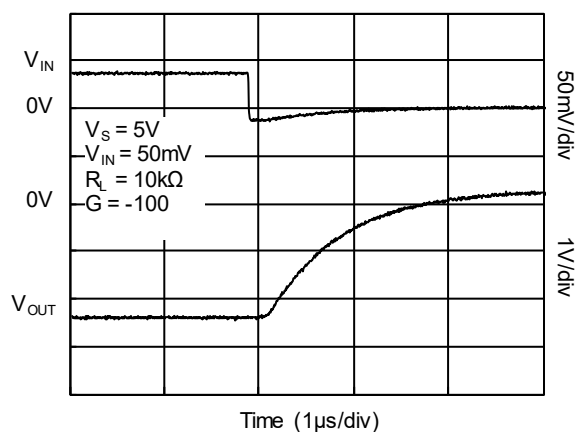
Small-Signal Step Response



Positive Overload Recovery

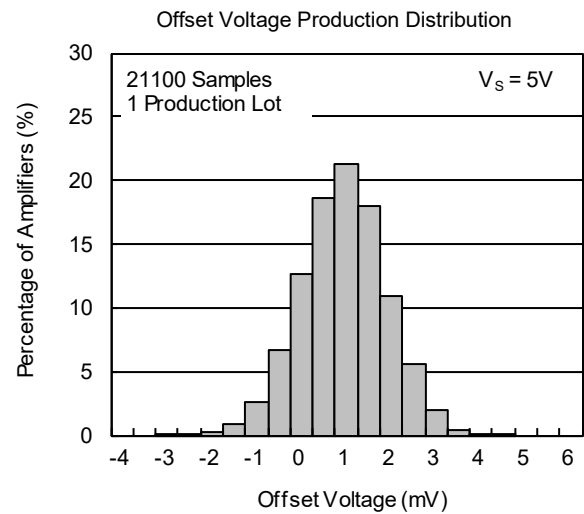
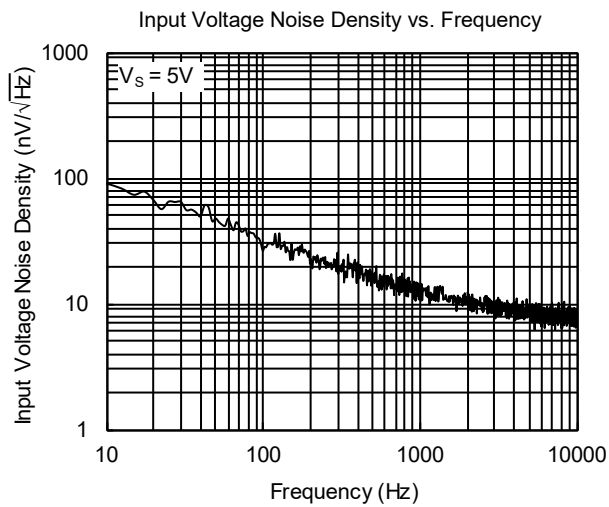


Negative Overload Recovery



TYPICAL PERFORMANCE CHARACTERISTICS (continued)

At $T_A = +25^\circ\text{C}$, $V_{CM} = V_S/2$ and $R_L = 600\Omega$, unless otherwise noted.



APPLICATION INFORMATION

Rail-to-Rail Input

When SGM8602 works at the power supply between 2.1V and 5.5V, the input common mode voltage range is from $(-V_S) - 0.1V$ to $(+V_S) + 0.1V$. In Figure 1, the ESD diodes between the inputs and the power supply rails will clamp the input voltage not to exceed the rails.

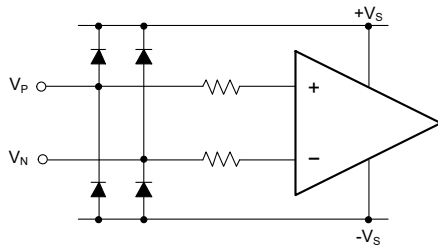


Figure 1. Input Equivalent Circuit

Rail-to-Rail Output

The SGM8602 supports rail-to-rail output operation. In single power supply application, for example, when $+V_S = 5V$, $-V_S = GND$, 10k Ω load resistor is tied from OUT pin to ground, the typical output swing range is from 0.006V to 4.994V.

Driving Capacitive Loads

The SGM8602 is designed for unity-gain stable for capacitive load up to 4700pF. If greater capacitive load must be driven in application, the circuit in Figure 2 can be used. In this circuit, the IR drop voltage generated by R_{ISO} is compensated by feedback loop.

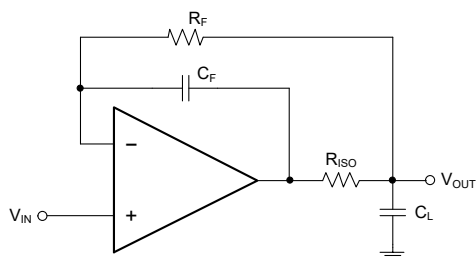


Figure 2. Circuit to Drive Heavy Capacitive Load

Power Supply Decoupling and Layout

A clean and low noise power supply is very important in amplifier circuit design, besides of input signal noise, the power supply is one of important source of noise to the amplifier through $+V_S$ and $-V_S$ pins. Power supply bypassing is an effective method to clear up the noise at power supply, and the low impedance path to ground of decoupling capacitor will bypass the noise to GND. In application, 10 μF ceramic capacitor paralleled with 0.1 μF or 0.01 μF ceramic capacitor is used in Figure 3. The ceramic capacitors should be placed as close as possible to $+V_S$ and $-V_S$ power supply pins.

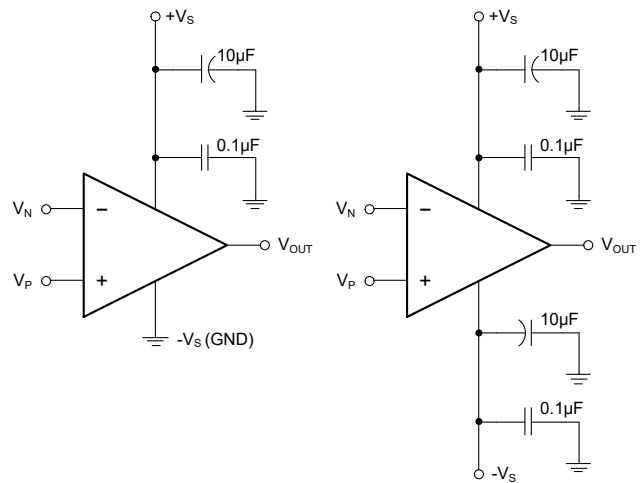


Figure 3. Amplifier Power Supply Bypassing

Grounding

In low speed application, one node grounding technique is the simplest and most effective method to eliminate the noise generated by grounding. In high speed application, the general method to eliminate noise is to use a complete ground plane technique, and the whole ground plane will help distribute heat and reduce EMI noise pickup.

Reduce Input-to-Output Coupling

To reduce the input-to-output coupling, the input traces must be placed as far away from the power supply or output traces as possible. The sensitive trace must not be placed in parallel with the noisy trace in same layer. They must be placed perpendicularly in different layers to reduce the crosstalk. These PCB layout techniques will help to reduce unwanted positive feedback and noise.

APPLICATION INFORMATION (continued)

Typical Application Circuits

Difference Amplifier

The circuit in Figure 4 is a design example of classical difference amplifier. If $R_4/R_3 = R_2/R_1$, then $V_{OUT} = (V_P - V_N) \times R_2/R_1 + V_{REF}$.

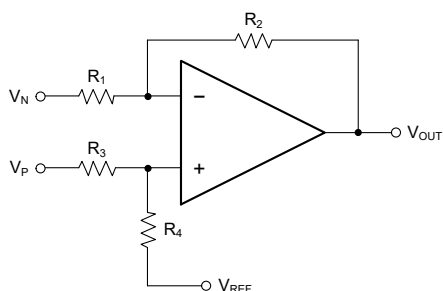


Figure 4. Difference Amplifier

High Input Impedance Difference Amplifier

The circuit in Figure 5 is a design example of high input impedance difference amplifier, the added amplifiers at the input are used to increase the input impedance and eliminate drawback of low input impedance in Figure 4.

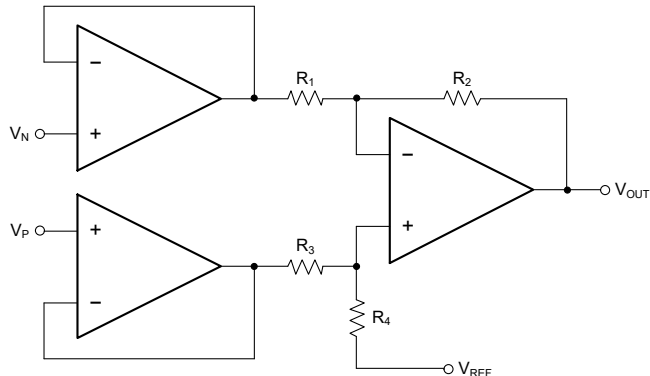


Figure 5. High Input Impedance Difference Amplifier

Active Low-Pass Filter

The circuit in Figure 6 is a design example of active low-pass filter, the DC gain is equal to $-R_2/R_1$ and the -3dB corner frequency is equal to $1/2\pi R_2 C$. In this design, the filter bandwidth must be less than the bandwidth of the amplifier, the resistor values must be selected as low as possible to reduce ringing or oscillation generated by the parasitic parameters in PCB layout.

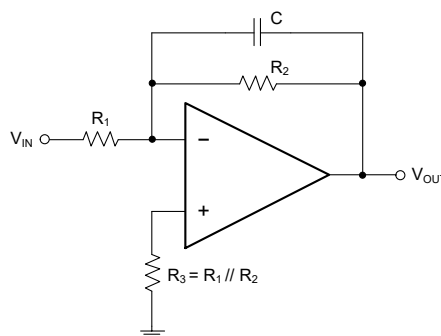


Figure 6. Active Low-Pass Filter

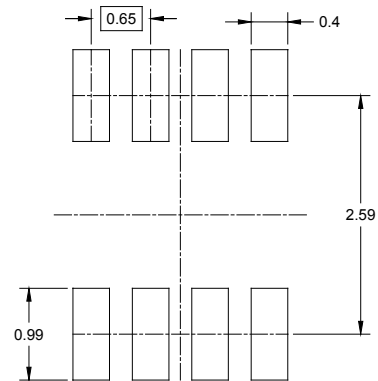
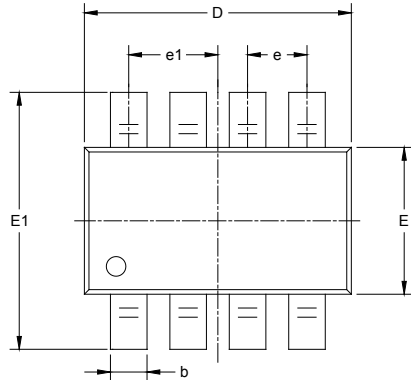
REVISION HISTORY

NOTE: Page numbers for previous revisions may differ from page numbers in the current version.

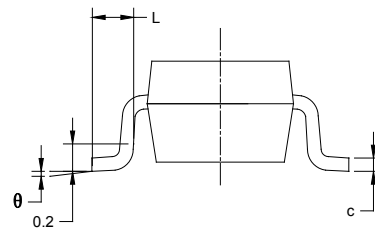
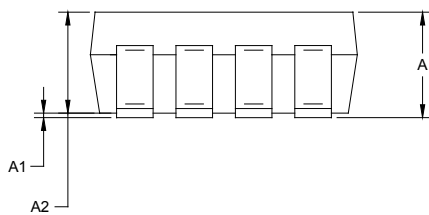
JANUARY 2018 – REV.A to REV.A.1		Page
Added Open-Loop Gain and Phase vs. Frequency		7
Changes from Original (AUGUST 2015) to REV.A		Page
Changed from product preview to production data.....		All

PACKAGE OUTLINE DIMENSIONS

SOT-23-8



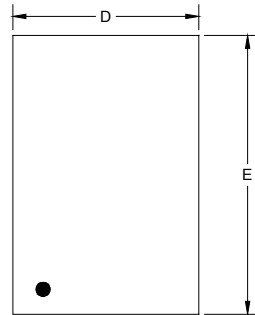
RECOMMENDED LAND PATTERN (Unit: mm)



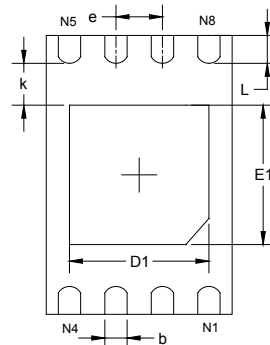
Symbol	Dimensions In Millimeters		Dimensions In Inches	
	MIN	MAX	MIN	MAX
A	1.050	1.250	0.041	0.049
A1	0.000	0.100	0.000	0.004
A2	1.050	1.150	0.041	0.045
b	0.300	0.500	0.012	0.020
c	0.100	0.200	0.004	0.008
D	2.820	3.020	0.111	0.119
E	1.500	1.700	0.059	0.067
E1	2.650	2.950	0.104	0.116
e	0.650 BSC		0.026 BSC	
e1	0.975 BSC		0.038 BSC	
L	0.300	0.600	0.012	0.024
θ	0°	8°	0°	8°

PACKAGE OUTLINE DIMENSIONS

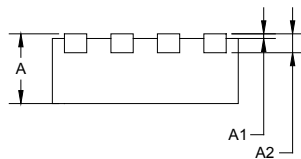
TDFN-2×3-8L



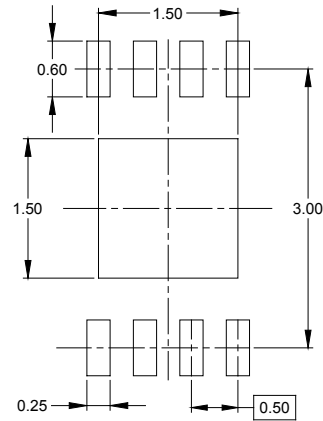
TOP VIEW



BOTTOM VIEW



SIDE VIEW



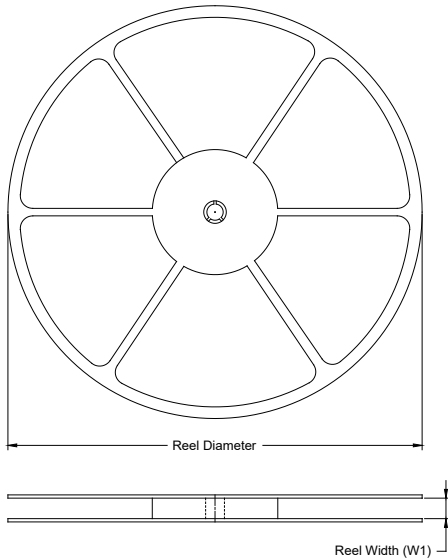
RECOMMENDED LAND PATTERN (Unit: mm)

Symbol	Dimensions In Millimeters		Dimensions In Inches	
	MIN	MAX	MIN	MAX
A	0.700	0.800	0.028	0.031
A1	0.000	0.050	0.000	0.002
A2	0.203 REF		0.008 REF	
D	1.924	2.076	0.076	0.082
D1	1.400	1.600	0.055	0.063
E	2.924	3.076	0.115	0.121
E1	1.400	1.600	0.055	0.063
k	0.200 MIN		0.008 MIN	
b	0.200	0.300	0.008	0.012
e	0.500 TYP		0.020 TYP	
L	0.224	0.376	0.009	0.015

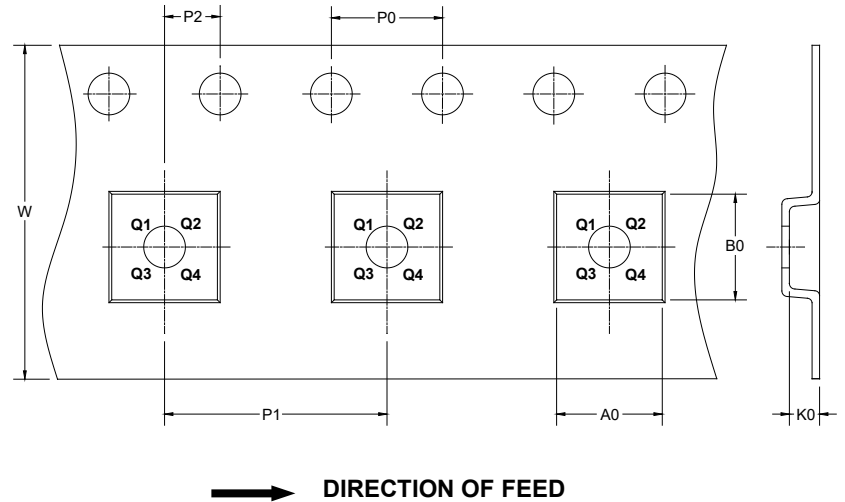
PACKAGE INFORMATION

TAPE AND REEL INFORMATION

REEL DIMENSIONS



TAPE DIMENSIONS



NOTE: The picture is only for reference. Please make the object as the standard.

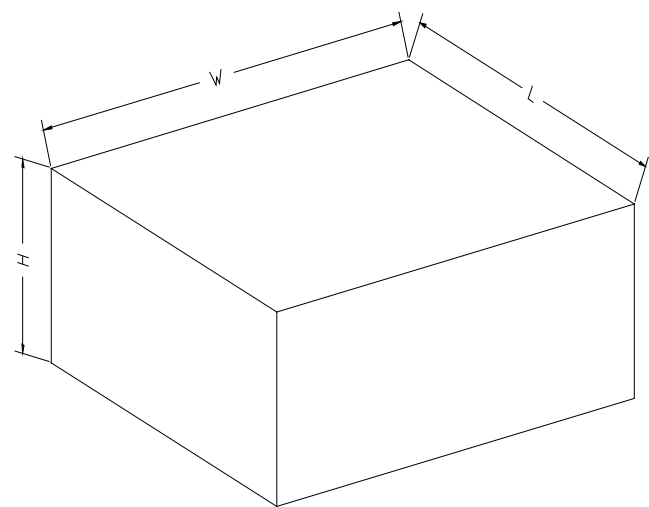
KEY PARAMETER LIST OF TAPE AND REEL

Package Type	Reel Diameter	Reel Width W1 (mm)	A0 (mm)	B0 (mm)	K0 (mm)	P0 (mm)	P1 (mm)	P2 (mm)	W (mm)	Pin1 Quadrant
SOT-23-8	7"	9.5	3.17	3.23	1.37	4.0	4.0	2.0	8.0	Q3
TDFN-2×3-8L	7"	9.5	2.30	3.30	1.10	4.0	4.0	2.0	8.0	Q2

DD0001

PACKAGE INFORMATION

CARTON BOX DIMENSIONS



NOTE: The picture is only for reference. Please make the object as the standard.

KEY PARAMETER LIST OF CARTON BOX

Reel Type	Length (mm)	Width (mm)	Height (mm)	Pizza/Carton
7" (Option)	368	227	224	8
7"	442	410	224	18

DD0002