

4A Rated μ POL™ DC-DC Converter with Integrated Inductor, Telemetry and Digital Power System Management

Features

- μ POL™ package with output inductor included
- Small size: 3.3mm x 3.3mm x 1.35mm
- Continuous 4A load capability
- Plug and play: no external compensation required
- Programmable operation using the I²C serial bus
- Voltage, current and temperature telemetry
- Operating temperature from -40°C to +125°C
- Preset output voltage: 2.5V or 3.3V
- Wide input voltage range: 4.2V–16V
- Enable input with precise threshold
- Open-drain power-good indicator
- Built-in protection features
- Servo loop for high precision Vout regulation
- Offers 8 i2c addresses
- Lead-free and halogen-free

Applications

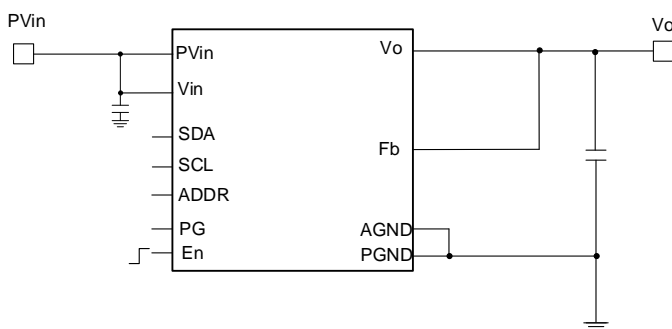
- Telecom and networking applications
- Data Center Applications
- 5G, AI Applications
- Industrial applications
- Storage applications
- Distributed point-of-load power architectures

Description

The FS1604 is an easy-to-use, fully integrated and highly efficient micro-point-of-load (μ POL™) voltage regulator. The on-chip pulse-width modulation (PWM) controller and integrated MOSFETs, plus incorporated inductor and capacitors, result in an extremely compact and accurate regulator. The low-profile package is suitable for automated assembly using standard surface-mount equipment.

Developed by a cross-functional engineering team, the design exemplifies best practice and uses class-leading technologies. From early in the integrated circuit design phase, designers worked with application and packaging engineers to select compatible technologies and implement them in ways that reduce compromise. The ability to program aspects of the FS1604's operation using the Inter-Integrated Circuit (I²C) protocol is unique in this class of product. Developing and optimizing all these elements together has yielded the smallest, most efficient and fully featured 4A μ POL™ currently available.

The built-in protection features include pre-biased start-up, soft-start protection, over-voltage protection, thermally compensated over-current protection with hiccup mode, thermal shut-down with auto-recovery.



Pin configuration

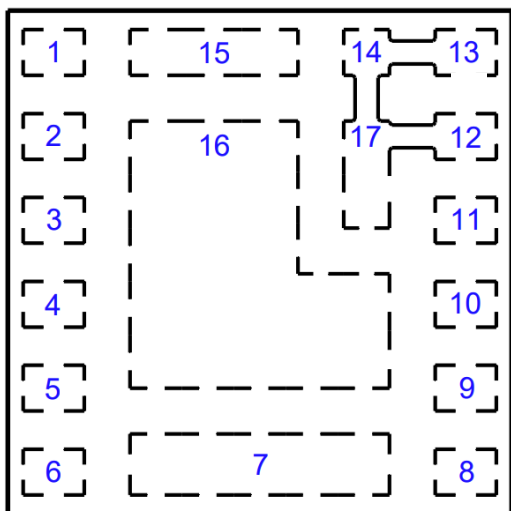


Figure 1 Pin layout (top view)

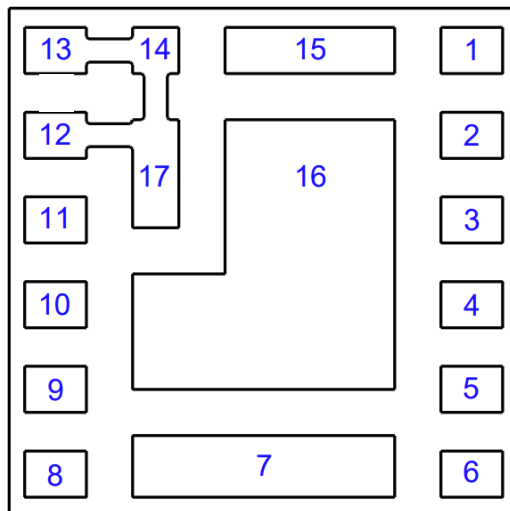


Figure 2 Pin layout (bottom view)

Pin functions

Pin Number	Name	Description
1	SDA	I²C Data Serial Input/Output line. Pull up to bus voltage with a 4.99k Ω resistor. If unused tie to AGnd.
2	PG	Power Good status. Open drain of an internal MOSFET. Pull up to V _{CC} – pin 10 or an external bias voltage – with a 49.9k Ω resistor.
3	En	Enable. Switches the FS1604 on and off. Can be used with two external resistors to set an external UVLO (Figure 6).
4	SCL	I²C Clock line. Pull up to bus voltage with a 4.99k Ω resistor. If unused, tie to AGnd.
5	Fb	V_{OUT} sense pin. Connect directly to V _{OUT} on the application board
6	ADDR	Address. Connect to AGnd through a resistor to program FS1604 address (pages 16 and 18).
7	V _{OUT}	Regulator output voltage. Place output capacitors between this pin and PGnd (pin 8).
8, 16	PGnd	Power ground. Serves as a separate ground for the MOSFETs. Connect to the power ground plane in the application.
9	AGnd	Signal ground. Serves as the ground for the internal reference and control circuitry.
10	V _{CC}	Supply voltage. This is the output of the internal LDO regulator; it may also be used to apply an external V _{CC} voltage. If used in this way, it is recommended to tie Vin and V _{CC} .
11	V _{IN}	Input voltage. Input for the internal LDO regulator. For single supply applications, this should be tied to PV _{IN} using a 2.7 Ω resistor. If a separate V _{CC} is used it is recommended to tie this to V _{CC} .
12,13,14,17	PV _{IN}	Power input voltage. Input for the MOSFETs.
15	V _{SW}	Test point for internal V_{SW}. Connect to an isolated pad on the PCB.

Block diagram

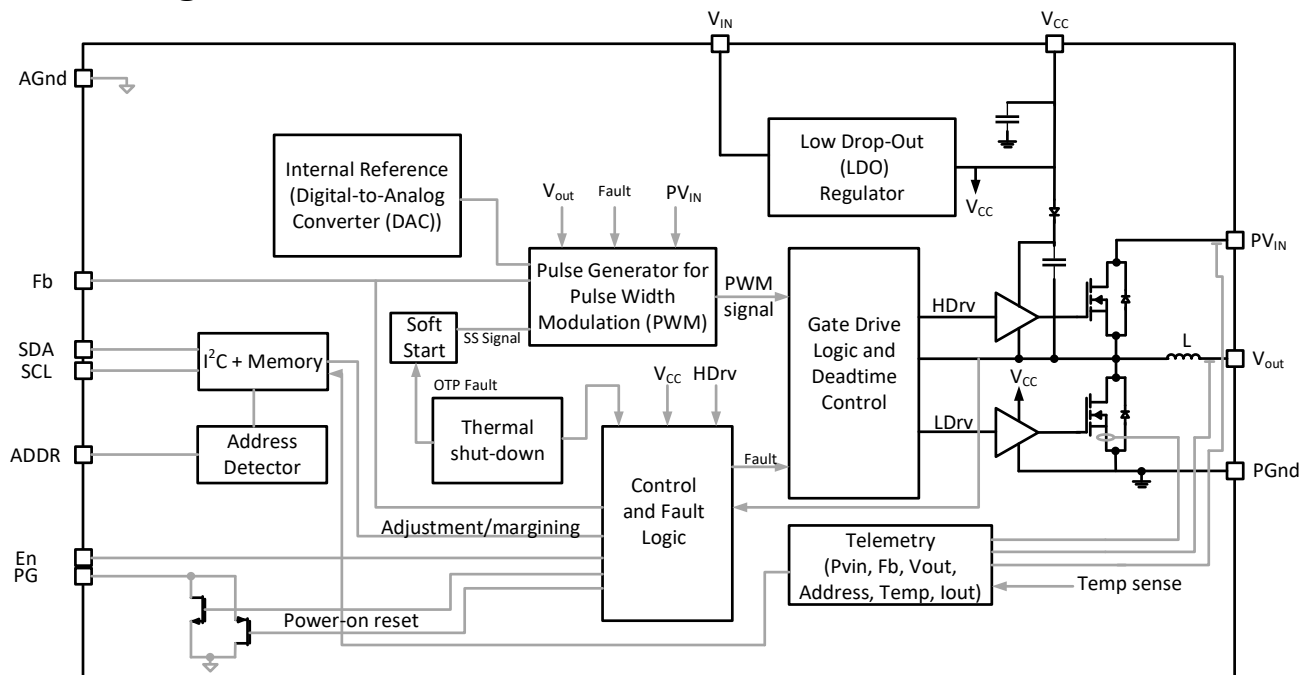


Figure 3 FS1604 μ POL™

Typical applications

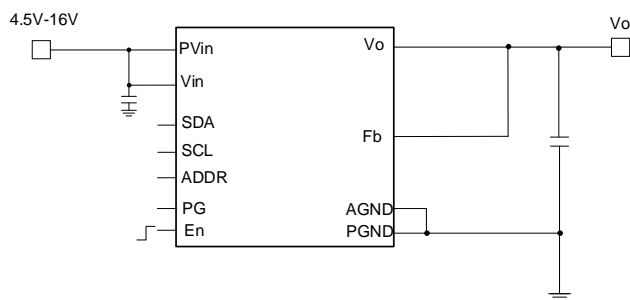


Figure 4 Single supply applications circuit

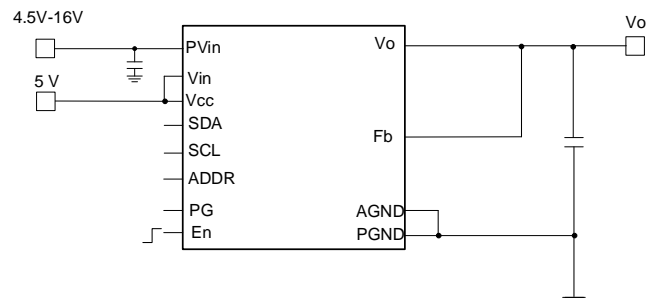


Figure 5 Dual supply applications circuit

Absolute maximum ratings

Warning: Stresses beyond those shown may cause permanent damage to the FS1604.

Note: Functional operation of the FS1604 is not implied under these or any other conditions beyond those stated in the FS1604 specification.

Reference	Range
PV _{IN} , V _{IN} , En to PGnd	-0.3V to 18V (Note 1, page 11)
V _{CC} to PGnd	-0.3V to 6V (Note 2, page 11)
FB to AGnd	-0.3V to V _{CC} (Note 2, page 11)
PG to AGnd	-0.3V to V _{CC} (Note 2, page 11)
PGnd to AGnd	-0.3V to +0.3V
ESD Classification	2kV (HBM JESD22-A114)
Moisture Sensitivity Level	MSL 3 (JEDEC J-STD-020D)

Thermal Information	Range
Junction-to-Ambient Thermal Resistance Θ_{JA}	22.6°C/W
Junction to PCB Thermal Resistance Θ_{J-c} (bottom)	2.36°C/W
Storage Temperature Range	-55°C to 150°C
Junction Temperature Range	-40°C to 150°C
Note: Θ_{JA} : FS1604 evaluation board and JEDEC specifications JESD 51-2A Θ_{J-c} (bottom) : JEDEC specification JESD 51-8	

Order information

Package details

The FS1604 uses a μ POL™ 3.3 mm x 3.3 mm package delivered in tape-and-reel format (Figure 20).

For more information on the tape-and-reel specification, go to:

<https://product.tdk.com/en/products/power/switching-power/micro-pol/designtool.html>

Standard part numbers

Output voltages of 2.5V and 3.3V are available.

Part Number	VOUT (V)	Quantity per Reel	Package Description	Package Designator
FS1604-2500-AL	2.5	3000	17L Open Top LGA SiP	A01
FS1604-3300-AL	3.3	3000	17L Open Top LGA SiP	A01

*Compliant with EU Directives REACH and RoHS. RoHS is defined as semiconductor products that are compliant with the current EU RoHS requirements for all 10 RoHS substances.

	FS1604	
	PRODUCT	MARKING CODE
	FS1604-2500	M
	FS1604-3300	N

Recommended operating conditions

2.5V output voltage

Definition	Symbol	Min	Max	Units
Input Voltage Range with External V_{CC} (Note 3, Note 5)	PV_{IN}	4.2	16	V
Input Voltage Range with Internal LDO (Note 4, Note 5)	PV_{IN}, V_{IN}	4.5	16	
Supply Voltage Range (Note 2)	V_{CC}	4.5	5.5	
Continuous Output Current Range	I_O	0	4	A
Operating Junction Temperature	T_J	-40	125	°C

3.3V output voltage

Definition	Symbol	Min	Max	Units
Input Voltage Range with External V_{CC} (Note 3, Note 5)	PV_{IN}	4.75	16	V
Input Voltage Range with Internal LDO (Note 4, Note 5)	PV_{IN}, V_{IN}	4.75	16	
Supply Voltage Range (Note 2)	V_{CC}	4.5	5.5	
Continuous Output Current Range	I_O	0	4	A
Operating Junction Temperature	T_J	-40	125	°C

Electrical characteristics

ELECTRICAL CHARACTERISTICS						
Unless otherwise stated, these specifications apply over: 4.5V < PV _{IN} = V _{IN} < 16V, 0°C < T _J < 125°C						
Typical values are specified at T _A = 25°C						
Parameter	Symbol	Conditions	Min	Typ	Max	Unit
Supply Current						
V _{IN} Supply Current (Standby)	I _{IN} (STANDBY)	Enable low		2.1		mA
V _{IN} Supply Current (Static)	I _{IN} (STATIC)	No switching, En = 2V		3.6		
V _{IN} Supply Current (Dynamic)	I _{IN} (DYN)	En high, V _{IN} = PV _{IN} =12V, V _{OUT} = 3.3V, F _{SW} =1.5MHz		15	18.6	
Soft-Start						
Soft-Start Rate	SS _{RATE} (default)	(Note 7)		1		V/ms
Output Voltage						
Output Voltage Range	V _{OUT} (default)			3.3		V
	V (resolution)			10		mV
Accuracy		T _J = 25°C, PV _{IN} = 12V, V _{OUT} = 3.3V (Note 6)		±0.5		%
		25°C < T _J < 125°C, PV _{IN} = 12V, V _{OUT} =2.5V, 3.3V (Note 6)	-1		1	
On-Time Timer Control						
On Time	T _{ON}	PV _{IN} = 12V, V _{OUT} = 3.3V, F _{SW} =1.5MHz	170	190	210	ns
Minimum On-Time	T _{ON(MIN)}	(Note 7)		30		
Minimum Off-Time	T _{OFF(MIN)}			120	150	
Internal Low Drop-Out (LDO) Regulator						
LDO Regulator Output Voltage	V _{CC}	5.5V < V _{IN} = 16V, 0 – 20mA	4.9	5.2	5.5	V
		4.5V <= V _{IN} < 5.5V, 0 – 20mA	4.3			
Thermal Shut-Down						
Thermal Shut-Down	TSD (default)	(Note 7)		145		°C
Hysteresis				25		
Under-Voltage Lock-Out						
V _{CC} Start Threshold	V _{CC_UVLO} (START)	V _{CC} Rising Trip Level	3.95	4.15	4.35	V
V _{CC} Stop Threshold	V _{CC_UVLO} (STOP)	V _{CC} Falling Trip Level	3.6	3.8	3.95	
Enable Threshold	En(HIGH)	Ramping Up	1.1	1.2	1.3	
	En(LOW)	Ramping Down	0.9	1	1.06	
Input Impedance	R _{EN}		500	1000	1500	kΩ
Current Limit						
Current Limit Threshold	I _{OC} (default)	T _J = 25°C	5.4	6	6.6	A
Hiccup Blanking Time	T _{BLK} (HICCUP)			20		ms
Over-Voltage Protection						
Output Over-Voltage Protection Threshold	V _{OVP} (default)	OVP Detect (Note 7)	115	120	125	Fb%
Output Over-voltage Protection Delay	T _{OVPDEL}			5		μs

ELECTRICAL CHARACTERISTICS						
Unless otherwise stated, these specifications apply over: $4.5V < PV_{IN} = V_{IN} < 16V$, $0^{\circ}C < T_J < 125^{\circ}C$						
Typical values are specified at $T_A = 25^{\circ}C$						
Parameter	Symbol	Conditions	Min	Typ	Max	Unit
Power Good (PG)						
Power Good Upper Threshold	$V_{PG(UPPER)}$ (default)	V_{OUT} Rising to 0.6V	85	90	95	Fb%
Power Good Hysteresis	$V_{PG(LOWER)}$	V_{OUT} Falling from 0.6V		5		
Power Good Sink Current	I_{PG}	$PG = 0.5V$, $En = 2V$		9		mA
Telemetry						
Input voltage reporting accuracy	$PV_{IN_report_pc}$	$PV_{IN}=12V$, $T_J = 25^{\circ}C$	-0.8		0.8	%
		$5V < PV_{IN} < 16V$, $-40^{\circ}C < T_J < 125^{\circ}C$	-1.5		1.5	%
Output voltage reporting accuracy	$V_{OUT_report_pc}$	$V_{OUT} = V_{FB} = 2.5V$, $-40^{\circ}C < T_J < 125^{\circ}C$	-1.5		1.5	%
		$V_{OUT} = V_{FB} = 3.3V$, $-40^{\circ}C < T_J < 125^{\circ}C$	-2		2	%
Output current reporting accuracy	$I_{OUT_report_acc}$	$PV_{IN}=12V$, $T_J = 25^{\circ}C$, $V_{OUT} = 2.5V$, $3.3V$, $I_{OUT}=0A$	0		500	mA
		$PV_{IN}=12V$, $-40^{\circ}C < T_J < 125^{\circ}C$, $V_{OUT} = 2.5V$, $3.3V$, $I_{OUT}=0A$ (Note 6)	0		600	mA
Temperature reporting accuracy	T_report_acc	$-40^{\circ}C < T_J < 125^{\circ}C$ (Note 7)	-10		10	$^{\circ}C$

ELECTRICAL CHARACTERISTICS							
Unless otherwise stated, these specifications apply over: 4.5V < PV _{IN} = V _{IN} < 16V, 0°C < T < 125°C							
Typical values are specified at T _A = 25°C							
Parameter	Symbol	Conditions	Fast-mode		Fast-mode Plus		Unit
I ² C parameters		(Note 7 for all parameters)	Min	Max	Min	Max	
I ² C bus voltage	V _{BUS}		1.8	5.5	1.8	5.5	V
LOW-level input voltage	V _{IL}		−0.5	0.3V _{BUS}	−0.5	0.3V _{BUS}	
HIGH-level input voltage	V _{IH}		0.7V _{BUS}		0.7V _{BUS}		
Hysteresis	V _{HYS}		0.05V _{BUS}		0.05V _{BUS}		
LOW-level output voltage 1	V _{OL1}	(open-drain or open-collector) at 3mA sink current; V _{DD} > 2 V,	0	0.4	0	0.4	
LOW-level output voltage 2	V _{OL2}	(open-drain or open-collector) at 2mA sink current; V _{DD} ≤ 2 V,	0	0.2V _{BUS}	0	0.2V _{BUS}	
LOW-level output current	I _{OL}	V _{OL} = 0.4 V,	3	−	3	−	mA
		V _{OL} = 0.6 V	6	−	6	−	
Output fall time	T _{OF}	From V _{IHmin} to V _{ILmax}	20 × (V _{BUS} /5.5 V)	250	20 × (V _{BUS} /5.5 V)	125	ns
Pulse width of spikes that must be suppressed by the input filter	T _{SP}		0	50	0	50	
Input current each I/O pin	I _I		−10	10	−10	10	μA

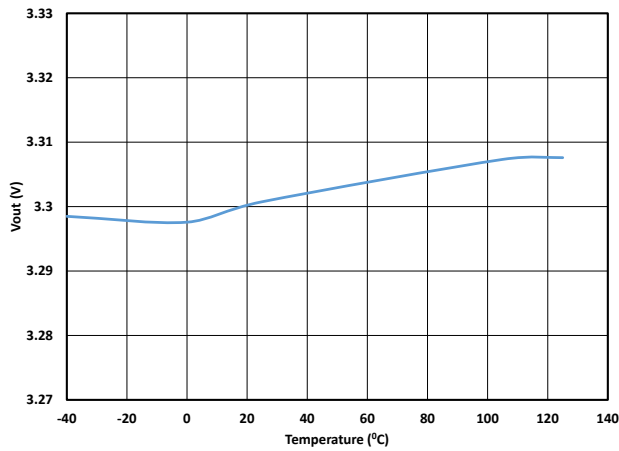
ELECTRICAL CHARACTERISTICS							
Unless otherwise stated, these specifications apply over: $4.5V < PV_{IN} = V_{IN} < 16V$, $0^{\circ}C < T < 125^{\circ}C$							
Typical values are specified at $T_A = 25^{\circ}C$							
Parameter	Symbol	Conditions	Fast-mode		Fast-mode Plus		Unit
I ² C parameters		(Note 7 for all parameters)	Min	Max	Min	Max	
Capacitance for each I/O pin	C_I		-	10	-	10	pF
SCL clock frequency	F_{SCL}		0	400	0	1000	kHz
Hold time (repeated) START condition	$T_{HD;STA}$	After this time, the first clock pulse is generated	0.6	-	0.26	-	μs
LOW period of the SCL clock	T_{LOW}		1.3	-	0.5	-	
HIGH period of the SCL clock	T_{HIGH}		0.6	-	0.26	-	
Set-up time for a repeated START condition	$T_{SU;STA}$		0.6	-	0.26	-	
Data hold time	$T_{HD;DAT}$	I ² C-bus devices	0	-	0	-	
Data set-up time	$T_{SU;DAT}$		100	-	50	-	ns
Rise time of SDA and SCL signals	T_R		20	300	-	120	
Fall time of SDA and SCL signals	T_F		$20 \times (V_{DD}/5.5 V)$	300	$20 \times (V_{DD}/5.5 V)$	120	
Set-up time for STOP condition	$T_{SU;STO}$		0.6	-	0.26	-	μs
Bus free time between a STOP and START condition	T_{BUF}		1.3	-	0.5	-	
Capacitive load for each bus line	C_B		-	400	-	550	pF
Data valid time	$T_{VD;DAT}$		-	0.9	-	0.45	μs
Data valid acknowledge time	$T_{VD;ACK}$		-	0.9	-	0.45	
Noise margin at the LOW level	V_{NL}	For each connected device, including hysteresis	$0.1V_{DD}$	-	$0.1V_{DD}$	-	V
Noise margin at the HIGH level	V_{NH}		$0.2V_{DD}$	-	$0.2V_{DD}$	-	
SDA timeout	T_{TO}		200		200		μs

Notes

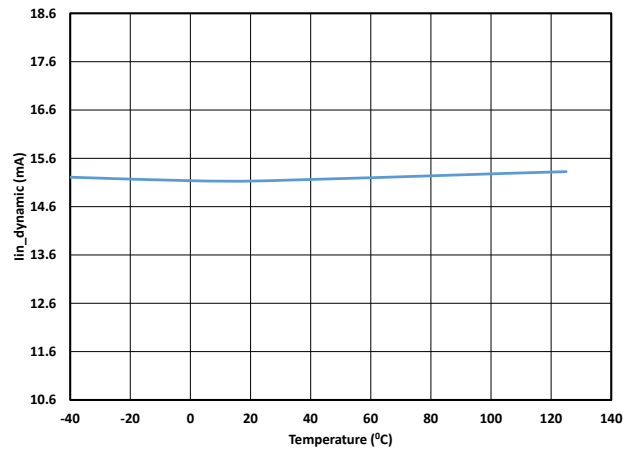
- 1 PGnd pin and AGnd pin are connected together
- 2 Must not exceed 6V
- 3 V_{IN} is connected to V_{CC} to bypass the internal Low Drop-Out (LDO) regulator
- 4 V_{IN} is connected to PV_{IN} (for single-rail applications with $PV_{IN}=V_{IN}=4.5V-16V$)
- 5 Maximum switch node voltage should not exceed 22V
- 6 Hot and cold temperature performance is assured by correlation using statistical quality control, but not tested in production; performance at 25°C is tested and guaranteed in production environment
- 7 Guaranteed by design but not tested in production

Temperature characteristics

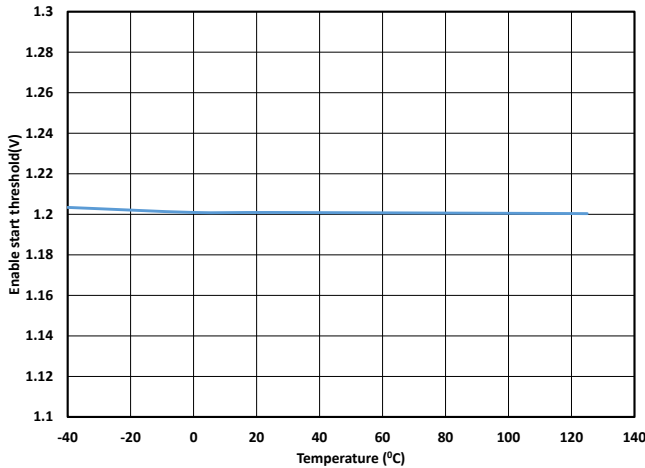
Output Voltage



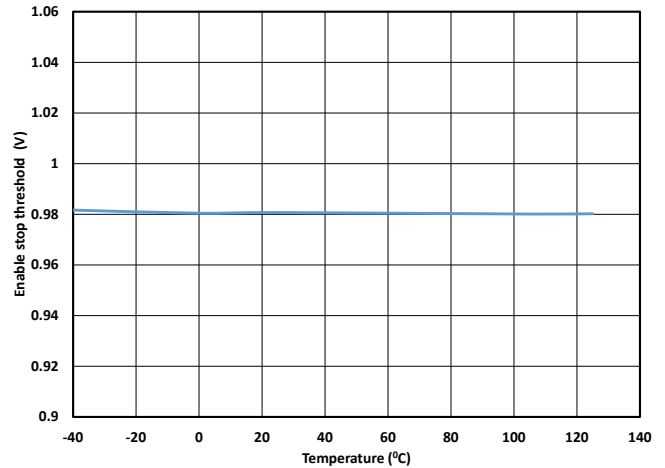
V_{IN} Supply Current (Dynamic)



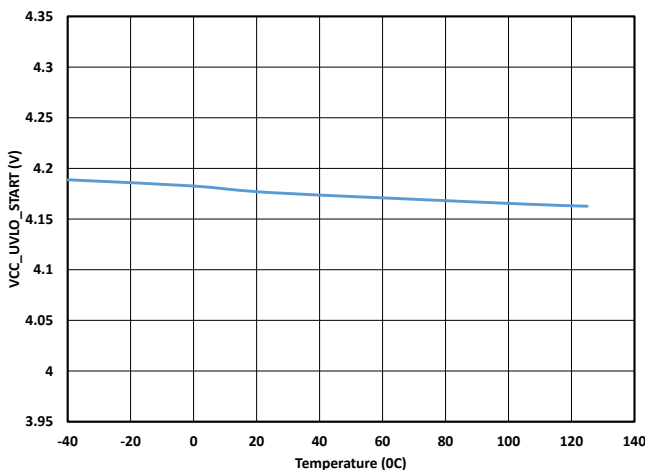
Enable Start Threshold



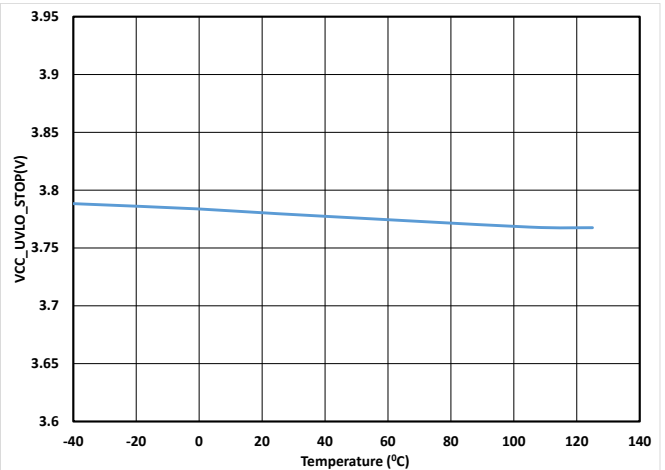
Enable Stop Threshold



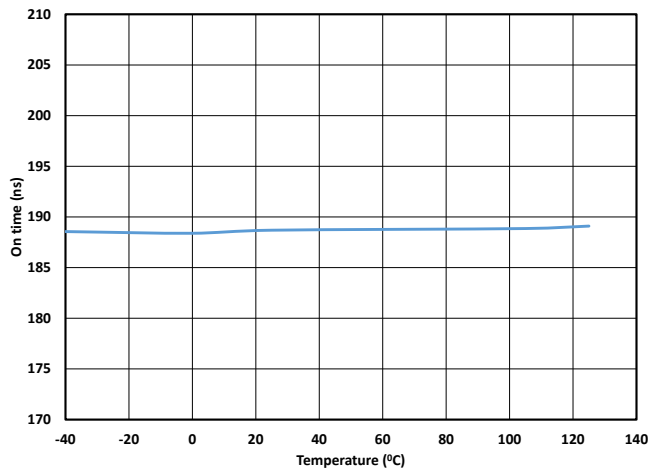
V_{CC} Start Threshold



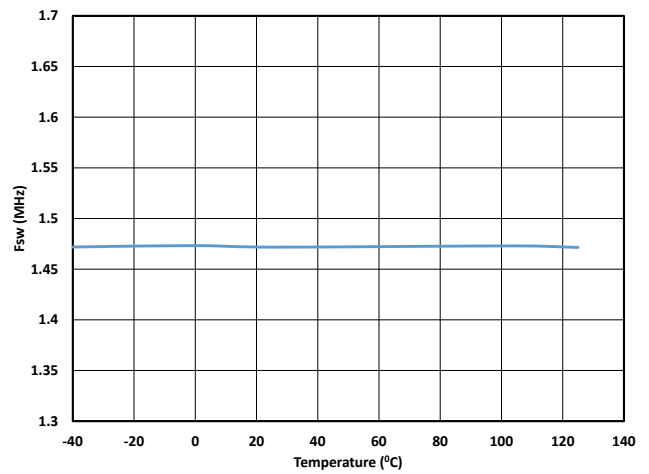
V_{CC} Stop Threshold



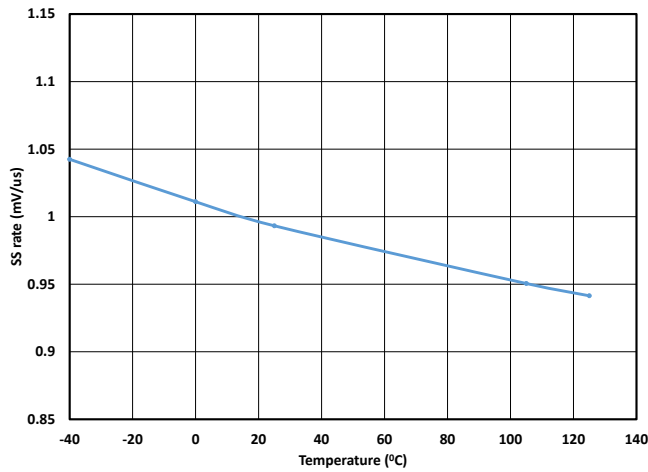
On Time



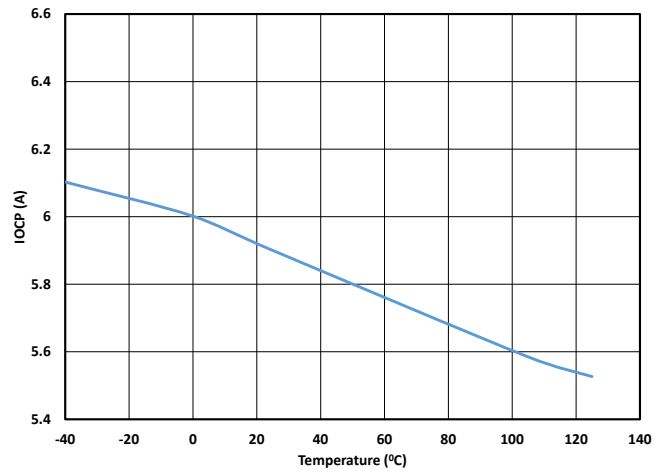
Switching Frequency



Soft-Start Rate



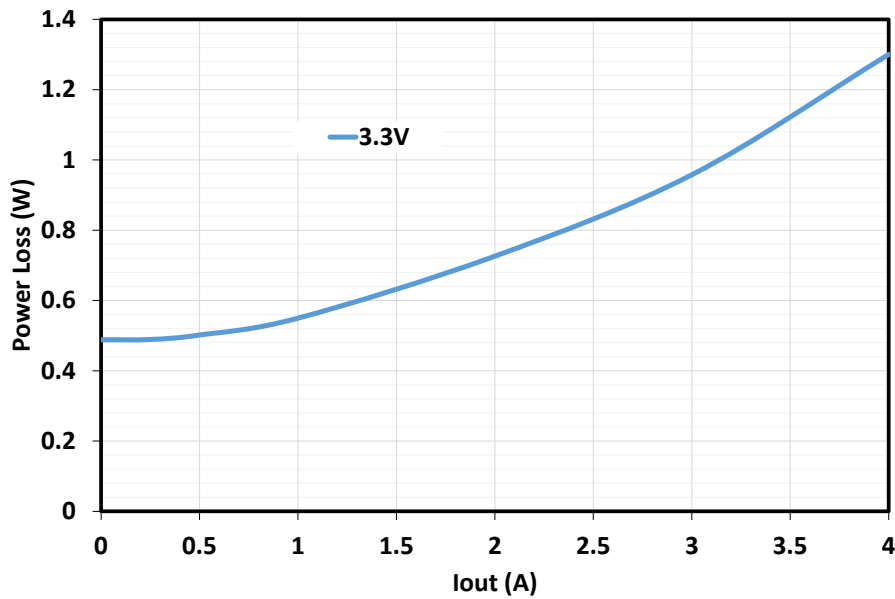
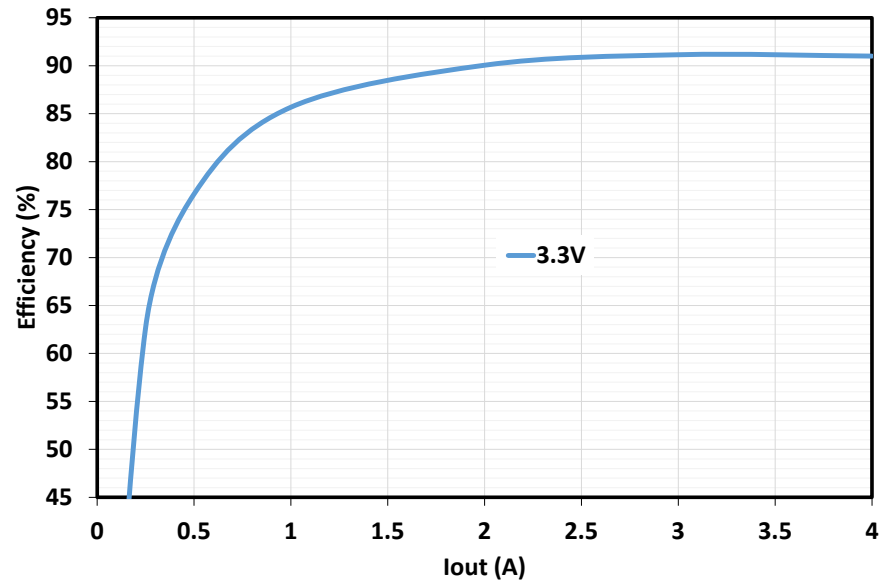
Current Limit Threshold



Efficiency characteristics

Typical efficiency and power loss at $P_{VIN} = 12V$

$P_{VIN} = 12V$, Internal LDO used, $I_O = 0A-4A$, room temperature, no air flow, all losses included



Applications information

Overview

The FS1604 is an easy-to-use, fully integrated, and highly efficient DC/DC regulator. Aspects of its operation, including output voltage and system optimization parameters, can be programmed using the I²C protocol. It uses a proprietary modulator to deliver fast transient responses. The modulator has internal stability compensation so that it can be used in a wide range of applications, with various types of output capacitors, without loop stability issues. An added servo loop ensures precise output voltage regulation.

Bias voltage

The FS1604 has an integrated Low Drop-Out (LDO) regulator, providing the DC bias voltage for the internal circuitry. The typical LDO regulator output voltage is 5.2V. For internally biased single-rail operation, the V_{IN} pin should be connected to the PV_{IN} pin (Figure 6). If an external bias voltage is used, the V_{IN} pin should be connected to the V_{CC} pin to bypass the internal LDO regulator (Figure 7).

The supply voltage (internal or external) rises with V_{IN} and does not need to be enabled using the En pin. Consequently, I²C communication can begin as soon as:

- V_{CC_UVLO} start threshold is exceeded
- Memory contents are loaded
- Initialization is complete
- Address offset is read

The I²C bus may be pulled up either to V_{CC} or to a system I²C bus voltage. The FS1604 offers two ranges for the I²C bus voltage, defined by the user register bit **Bus_voltage_sel**.

Register	Bits	Name/Description
0x1A	[1]	Bus_voltage_sel 0:1.8–2.5V, 1: 3.3–5V

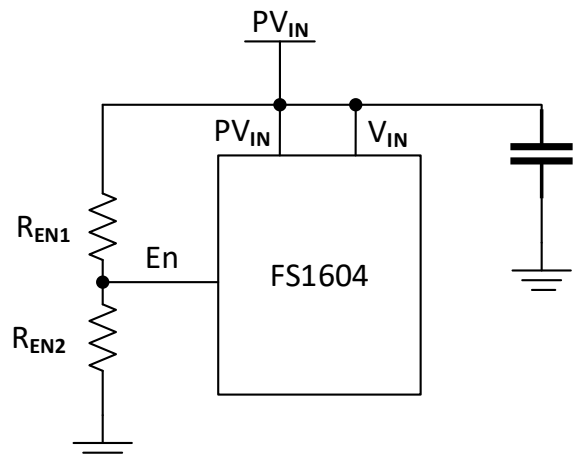


Figure 6 Single supply configuration: internal LDO regulator, adjustable PV_{IN_UVLO}

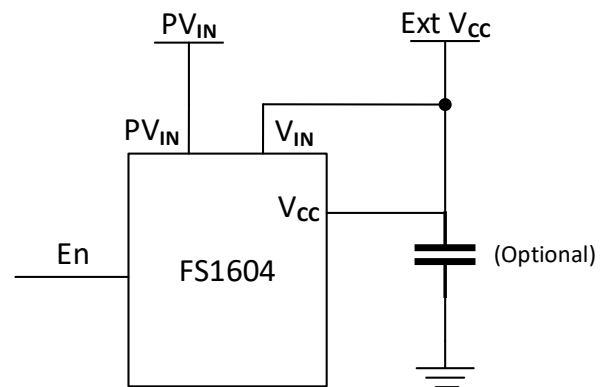


Figure 7 Using an external bias voltage

I²C base address and offsets

The FS1604 has a user register called **Base_address[7:0]** stored in memory that sets its base I²C address. The default base address is 0x08. An offset of 0-7 is then defined by connecting the ADDR pin to the AGnd pin either directly or through a resistor (1% or better). An address detector reads the resistance of the connection at startup and uses it to set the offset, which is added to the base I²C address to set the address at which the I²C master device will communicate with the FS1604.

To select offsets of 0 to 7, connect the pins as follows:

- **0** – 0 Ω (short ADDR to AGnd)
- **+1** – 1.13k Ω
- **+2** – 1.87k Ω
- **+3** – 2.61k Ω
- **+4** – 3.4k Ω
- **+5** – 4.12k Ω
- **+6** – 4.87k Ω
- **+7** – 5.62k Ω

Soft-start and target output voltage

The FS1604 has an internal digital soft-start circuit to control output voltage rise-time and limit current surge at start-up. When V_{CC} exceeds its start threshold ($V_{CC_UVLO(START)}$), the FS1604 exits reset mode; this initiates loading of the contents of the non-volatile memory into the working registers and calculates the address offset as described above.

Once initialization is complete and the Enable (En) pin has been asserted (Figure 8), the internal reference soft-starts to the target output voltage at the rate defined by the user register bit **SS_rate**.

Register	Bits	Name/Description
0x14	[3]	SS_rate 0:1 mV/ μ s (default), 1: 2 mV/ μ s

During initial start-up, the FS1604 operates with a minimum width high-drive (HDrv) pulses until the

output voltage increases (see Switching frequency and minimum values for on-time, off-time on page 17). On-time is increased until V_{OUT} reaches the target value defined by the user register bit **Vout_high_byte** and user register **Vout_low_byte[7:0]**.

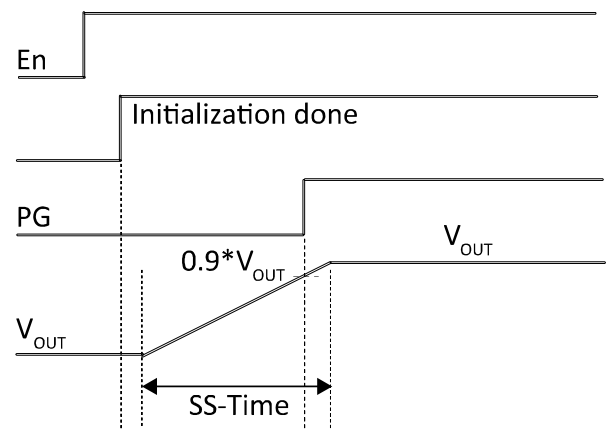


Figure 8 Theoretical operational waveforms during soft-start

Register	Bits	Name/Description
0x12	[0]	Vout_high_byte
0x13	[7:0]	Vout_low_byte

V_{OUT} is set in 10mV increments. Use the following equation to calculate the V_{OUT} code to enter into **Vout_high_byte** and **Vout_low_byte[7:0]**:

$$Vout_{code} = \frac{Vout_{target} - \frac{0.4 \times resolution}{0.005}}{resolution}$$

All voltages and resolutions are in Volts.

For example:

To set V_{OUT} = 3.3V (resolution of 10mV):

$$Vout_{code} = \frac{3.3 - \frac{0.4 \times 0.01}{0.005}}{0.01} = 250$$

250 is 0FA in hexadecimal, therefore:

Set **Vout_high_byte** to 0

Set **Vout_low_byte** to FA or (11111010)_b

Over-current protection (OCP) and over-voltage protection (OVP) is enabled during soft-start to protect the FS1604 from short circuits and excess voltages respectively.

For maximum system accuracy, the recommended way to set the output voltage is by programming the user registers with the appropriate code. For optimum performance when using this approach, the change in output voltage should not exceed $\pm 20\%$ of the pre-set default output voltage.

Pre-biased start-up

The FS1604 can start up into a pre-charged output smoothly, without causing oscillations and disturbances of the output voltage. When it starts up in this way, the Control and Synchronous MOSFETs are forced off until the internal Soft-Start (SS) signal exceeds the sensed output voltage at the FB pin. Only then is the first gate signal of the Control MOSFET generated, followed by complementary turn on of the Synchronous MOSFET. The Power Good (PG) function is not active until this point.

Shut-down mechanisms

The FS1604 has two shut-down mechanisms:

- *Hard shut-down or decay according to load*
Initiated by de-asserting the En pin. Both drivers switch off and the digital-to-analog converter (DAC) and soft-start are pulled down instantaneously.
- *Soft-Stop or controlled ramp down*
Initiated by setting user register bit **SoftStopEnable** to 1 **and** user register bit **SoftDisable** to 1. The SS signal falls to 0 at the same rate as it rises during start-up; the drivers are disabled only when it reaches 0. The output voltage then follows the SS signal down to 0.

The **SoftDisable** bit must not be toggled while the part is enabled and switching. Instead, for

applications requiring soft-stop, this bit must be set to 1 and, with the En pin asserted, the **SoftStopEnable** bit must be toggled to soft-start or soft-stop the device.

By default, both the **SoftDisable** bit and the **SoftStopEnable** bit are 0, which means that soft-stop operation is disabled by default.

Register	Bits	Name/Description
0x14	[2]	SoftStopEnable
0x1C	[3]	SoftDisable

Switching frequency and minimum values for on-time, off-time and PV_{IN}

The switching frequency of the FS1604 depends on the output voltage. For an output voltage of 2.5V, the switching frequency is nominally 1.3MHz; for an output voltage of 3.3V, the switching frequency is nominally 1.5MHz. These are set at the factory.

As a result, system designers need not concern themselves with selecting the switching frequency and have one fewer design task to manage.

When input voltage is high relative to target output voltage, the Control MOSFET is switched on for shorter periods. The shortest period for which it can reliably be switched on is defined by minimum on-time ($T_{ON(MIN)}$), nominally 30ns. During start-up, when the output voltage is very small, the FS1604 operates with minimum on-time.

When input voltage is low relative to target output voltage, the Control MOSFET is switched on for longer periods. The shortest period for which it can be switched off is defined by minimum off-time ($T_{OFF(MIN)}$). The Synchronous MOSFET stays on during this period and its current is detected for over-current protection. This dictates the minimum input voltage that can still allow the device to regulate its output at the target voltage.

The minimum input voltage required to support these output voltages over the entire load range is 4.2V for 2.5V output and 4.75V for 3.3V output. However, as these values are affected by both efficiency and dynamic load requirements, system designers should validate them in their own applications.

If designers still would like to tune the switching frequency for further optimization, the ADDR pin offers a way to lower the switching frequency of the device from 1.3 MHz to 1 MHz for FS1604-2500 and from 1.5 MHz to 1.2 MHz for FS1604-3300. While the address resistor selection on Page 16 corresponds to default switching frequency setting, the table below indicates the acceptable values of the ADDR pin resistor to change the switching frequency lower. For instance, an address resistor value of 1.13 k Ω sets the address offset to +1 and the default switching frequency, while an address resistor value of 7.15 k Ω sets the same address offset (+1), but sets the switching frequency to 1 MHz for FS1604-2500 and to 1.2 MHz for FS1604-3300

* If $T_j < 20^\circ\text{C}$, use 10.9 k Ω \pm 0.5% resistor

Resistor	Addr Offset	Switching Frequency Set for FS1604-2500	Switching Frequency Set for FS1604-3300
6.34k Ω	0	1 MHz	1.2 MHz
7.15k Ω	+1		
7.87k Ω	+2		
8.66k Ω	+3		
9.42k Ω	+4		
10.2k Ω	+5		
11k Ω *	+6		
12.1k Ω	+7		

Enable (En) pin

The Enable (En) pin has several functions:

- It is used to switch the FS1604 on and off. It has a precise threshold, which is internally monitored by the UVLO circuit. If it is left floating, an internal 1M Ω resistor pulls it down to prevent the FS1604 being switched on unintentionally.
- It can be used to implement a precise input voltage UVLO. The input of the En pin is derived from the PV_{IN} voltage by a set of resistive dividers, R_{EN1} and R_{EN2} (Figure 6). Users can program the UVLO threshold voltage by selecting different ratios. This is a useful feature that stops the FS1604 regulating when PV_{IN} is lower than the desired voltage.
- It can be used to monitor other rails for a specific power sequencing scheme (Figure 9).

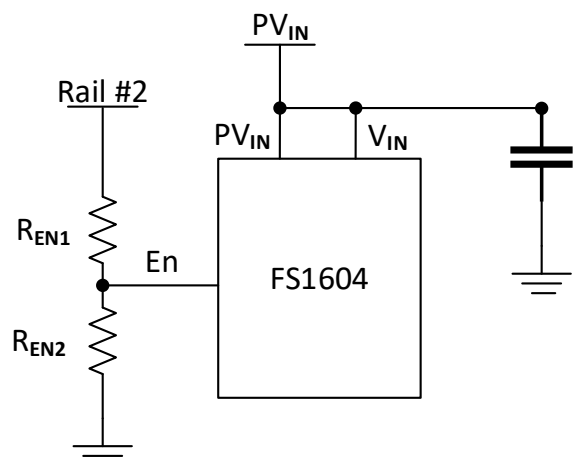


Figure 9 En pin used to monitor other rails for sequencing purposes

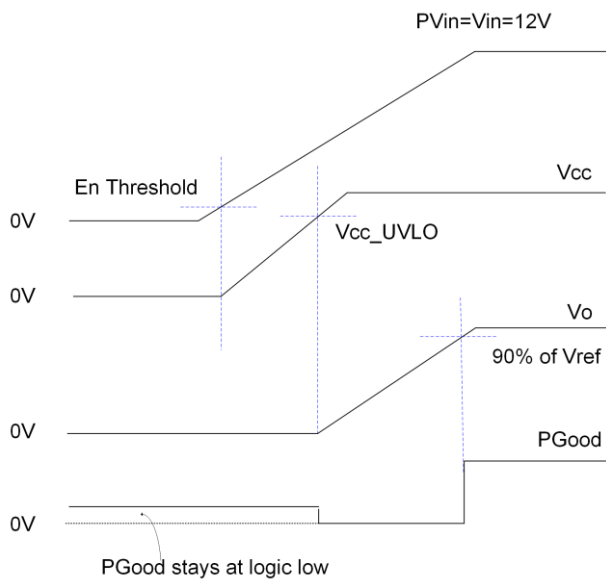


Figure 10 Start-up: PV_{IN} , V_{IN} tied together, En connected to resistor divider from PV_{IN} , PG pin pulled up to an external supply

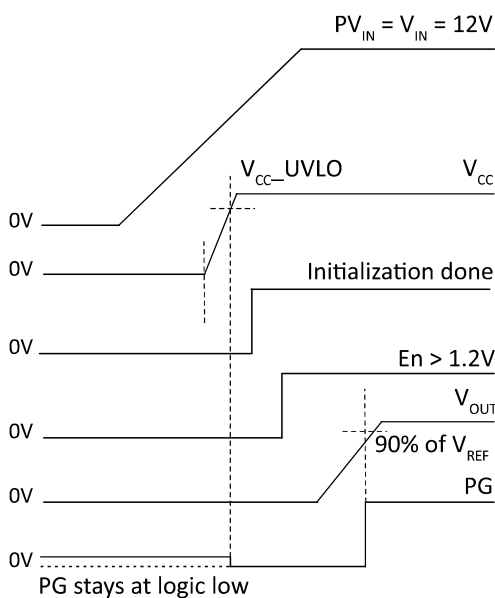


Figure 11 Start-up: En pin asserted after PV_{IN} and V_{IN} , PG pin pulled up to an external supply

For V_{OUT} to start up as defined by the soft-start rate requires correct sequencing:

- PV_{IN} must start up before V_{CC} and/or Enable.
- PV_{IN} must ramp down only after V_{CC} has ramped down below its UVLO threshold and/or Enable has been de-asserted.

Over-current protection (OCP)

Over-current protection (OCP) is provided by sensing the current through the $R_{DS(on)}$ of the Synchronous MOSFET. When this current exceeds the OCP threshold, a fault condition is generated. This method provides several benefits:

- Provides accurate overcurrent protection without reducing converter efficiency (the current sensing is lossless)
- Reduces cost by eliminating a current-sense resistor
- Reduces any layout-related noise issues.

The OCP threshold is set to 6A. The threshold is internally compensated so that it remains almost constant at different ambient temperatures.

When the current exceeds the OCP threshold, the PG and SS signals are pulled low. The Synchronous MOSFET remains on until the current falls to 0, then the FS1604 enters hiccup mode (Figure 12). Both the Control MOSFET and the Synchronous MOSFET remain off for the hiccup-blanking time. After this time, the FS1604 tries to restart. If an over-current fault is still detected, the preceding actions are repeated. The FS1604 remains in hiccup mode until the over-current fault is remedied.

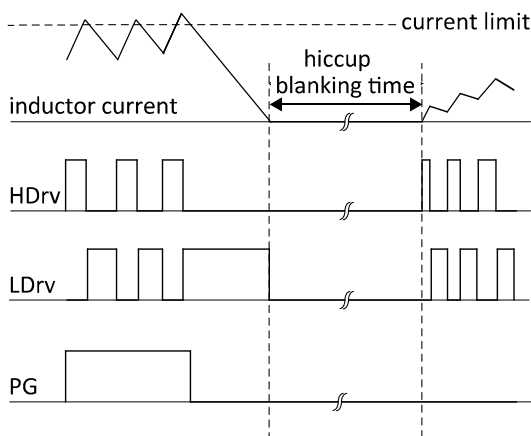


Figure 12 Illustration of OCP in hiccup mode

Over-voltage protection (OVP)

Over-voltage protection (OVP) is provided by sensing the voltage at the FB pin. When FB exceeds the output OVP threshold for longer than the output OVP delay (typically 5 μ s), a fault condition is generated.

The OVP threshold is defined by the user register bits **OV_Threshold**.

Register	Bits	Name/Description
0x17	[1:0]	OV_Threshold 0:105% of V_{OUT} 1:110% of V_{OUT} 2:115% of V_{OUT} 3:120% of V_{OUT} (default)

The Control MOSFET is switched off immediately and the PG pin is pulled low. The Synchronous MOSFET is switched on to discharge the output capacitor.

The Control MOSFET remains latched off until reset by cycling either VCC or En. The voltage at the FB pin falling below the output OVP threshold (with 5% hysteresis) does not switch on the Control MOSFET but it does switch off the Synchronous MOSFET to prevent build-up of negative current.

Figure 13 shows a timing diagram for over-voltage protection.

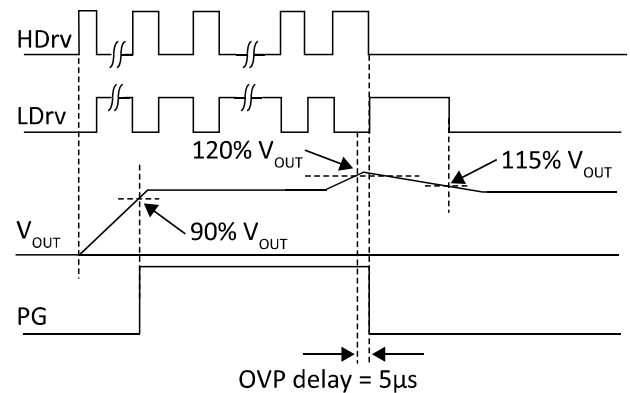


Figure 13 Illustration of latched OVP

Over-temperature protection (OTP)

Temperature sensing is provided inside the FS1604. The OTP threshold is defined by the user register bits **OT_Threshold**.

Register	Bits	Name/Description
0x19	[1:0]	OT_Threshold 0:75°C 1: 85°C 2: 125°C 3: 145°C (default)

When the threshold is exceeded, thermal shut-down switches off both MOSFETs and resets the internal soft-start, but the internal LDO regulator is still in operation.

Automatic restart is initiated when the sensed temperature drops within the operating range. There is a 20°C hysteresis in the OTP threshold.

Telemetry (ADC)

FS1604 has telemetry through I²C. Parameters P_{Vin}, V_{OUT}, F_b, I_{OUT}, and Temperature can be read through the telemetry.

Telemetry	Resolution	Min	Max
V_{in} (V)	1/16	0	15.9375
V_{out} (V)	0.02	2.5	3.3
I_{out} (A)	0.03125	0	4
Temperature ($^{\circ}$ C)	1	-40	145

Servo loop and precision output voltage

FS1604 has an internal servo loop to minimize V_{out} error at steady state. Load and line regulation of +/- 1% is achieved.

Power Good (PG)

Power Good (PG) behavior is defined by the user register bits **PGControl** and **PG_Threshold**.

Register	Bits	Name/Description
0x18	[1:0]	PG_Threshold 0: 80% of V_{out} 1: 85% of V_{out} 2: 90% of V_{out} (default) 3: 95% of V_{out}
0x14	[0]	PG_Control 1: Threshold based (default) 0: DAC based

PG_Threshold bit

The user register bit **PG_Threshold** defines the PG threshold as a percentage of V_{out} . Hysteresis of 5% is applied to this, giving a lower threshold.

When FB rises above the upper threshold, the PG signal is pulled high. When FB drops below the lower threshold, the PG signal is pulled low.

PGControl bit set to 1 (default)

Figure 14 shows PG behavior in this situation.

The behavior is the same at start-up and during normal operation. The PG signal is asserted when:

- E_n and V_{CC} are both above their thresholds
- No fault has occurred (including over-current, over-voltage and over-temperature)
- V_{out} is within the target range (determined by continuously monitoring whether FB is above the PG threshold)

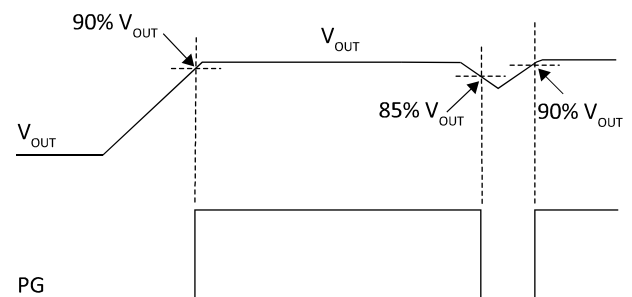


Figure 14 PG signal when PGControl bit=1

PGControl bit set to 0

Figure 15 shows PG behavior in this situation.

In normal operation, the PG signal behaves in the same way as when the **PGControl** bit is 1.

At start-up, however, the PG signal is asserted after soft-start is within 2% of target output voltage, not when FB exceeds the upper PG threshold.

For pre-biased start-up, the PG signal is not active until the first gate signal of the Control MOSFET is generated.

FS1604 also integrates an additional PMOS in parallel to the NMOS internally connected to the PG pin (Figure 3). This PMOS allows the PG signal to stay at logic low, even if V_{CC} is low and the PG pin is pulled up to an external voltage not V_{CC} (Figure 10 and Figure 11).

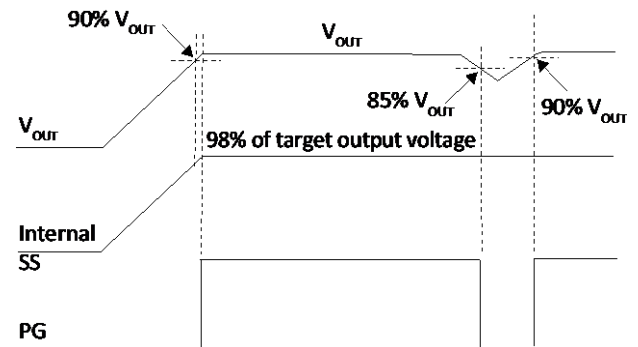


Figure 15 PG signal when PGControl bit=0

Design example

Let us now consider a simple design example, using the FS1604 for the following design parameters:

- $PV_{IN} = V_{IN} = 12V$
- $V_{OUT} = 3.3V$
- $F_{SW} = 1.5MHz$
- $C_{OUT} = 2 \times 22\mu F$
- $C_{IN} = 2 \times 22\mu F$
- Ripple Voltage = $\pm 1\% \times V_{OUT}$
- $\Delta V_{OUT(MAX)} = \pm 3\% \times V_{OUT}$
(for 50% load transient)

Input capacitor

The input capacitor selected for this design must:

- Handle the peak and root mean square (RMS) input currents required by the FS1604
- Have low equivalent series resistance and inductance (ESR and ESL) to reduce input voltage ripple

MLCCs (multi-layer ceramic capacitors) are ideal. Typically, in 0805 case size, they can handle 2A RMS current with less than 5°C temperature rise.

For a buck converter operating at duty cycle D and output current I_O , the RMS value of the input current is:

$$I_{RMS} = I_O \sqrt{D(1-D)}$$

In this application, $I_O = 4A$ and $D = \frac{V_{OUT}}{PV_{IN}} = 0.275$

Therefore, $I_{RMS} = 1.78A$ and we can select two 22 μF 16V ceramic capacitors for the input capacitors (C2012X5R1C226K125AC from TDK).

If the FS1604 is not located close to the 12V power supply, a bulk capacitor (68–330 μF) may be used in addition to the ceramic capacitors.

For V_{IN} , which is the input to the LDO, it is recommended to use a 1 μF capacitor very close to the pin. The V_{IN} pin should be connected to PV_{IN} through a 2.7 Ω resistor. Together, the 2.7 Ω resistor and 1 μF capacitor filter noise on PV_{IN} .

Output voltage and output capacitor

The FS1604 is supplied pre-programmed and factory-trimmed in a closed loop to the target voltage specified for the part number. As a result, no external resistor divider is required, and resistor tolerances are eliminated from the error budget.

The design requires minimal output capacitance to meet the target output voltage ripple and target maximum output voltage deviation under load transient conditions.

For the FS1604, the minimum number of output capacitors required to achieve target peak-to-peak V_{OUT} ripple is:

$$N_{MIN} = 4.68 \times \frac{\frac{(1-D)}{8CF_{SW}} + ESR(1-D) + \frac{ESL \times F_{SW} \times (1-D)^2}{D}}{\Delta V_{OUTripple(p-p)}}$$

where:

- N_{MIN} = minimum number of output capacitors
- D = duty cycle
- C = equivalent capacitance of each output capacitor
- F_{SW} = switching frequency
- ESR = equivalent series resistance of each output capacitor
- ESL = equivalent series inductance of each output capacitor
- $\Delta V_{OUTripple(p-p)}$
= target peak-to-peak V_{OUT} ripple

This design uses C2012X5R0J226K125AB from TDK; this is a 22 μF MLCC, 0805 case size, rated at 6.3V. At 3.3V, accounting for DC bias and AC ripple derating, it has an equivalent capacitance of 7 μF (C). Equivalent series resistance is 3m Ω (ESR) and equivalent series inductance is 0.44nH (ESL).

Putting these parameters into the equation gives:

$$N_{MIN} = 0.85$$

To meet the maximum voltage deviation $\Delta V_{O_{max}}$ under a ΔI_o load transient, the minimum required number of output capacitors is:

$$\frac{235 \times 10^{-9} \times \Delta I_o^2}{\Delta V_{O_{max}} \times V_{OUT} \times C}$$

where:

- ΔI_o = load step
- $\Delta V_{O_{max}}$ = target maximum voltage deviation
- V_{OUT} = output voltage
- C = equivalent capacitance of each output capacitor

Again, using $C = 7\mu\text{F}$, it can be seen that the minimum number of output capacitors required is 0.41.

In our design intended for space-constrained applications, therefore, we use two C2012X5R0J226K125AB capacitors.

It should be noted here that the calculation for the minimum number of output capacitors under a load transient makes some assumptions:

- a) No ESR or ESL
- b) Converter can saturate its duty cycle instantly

- c) No latency
- d) Step load (infinite slew rate)

Assumptions (a), (b) and (c) are liberal, whereas (d) is conservative. Therefore, in a real application, additional capacitance may be required to meet transient requirements and should be carefully considered by the system designer.

It should be noted that even in the absence of a target V_{OUT} ripple or target maximum voltage deviation under load transient, at least one $22\mu\text{F}$ capacitor is still required in order to ensure stable operation without excessive jitter.

Up to six $22\mu\text{F}$ capacitors may be used in the design. If more capacitance is required, it is recommended to use a capacitor with relatively high ESR ($>3\text{m}\Omega$) such as POSCAP or specialty polymer capacitors.

V_{CC} capacitor selection

FS1604 uses an on-package V_{CC} capacitor to ensure effective high-frequency bypassing. However, especially for applications that use an external V_{CC} supply, it is recommended that system designers place a $2.2\mu\text{F}/0603/\text{X7R}/10\text{V}$ capacitor on the application board as close as possible to the V_{CC} pin.

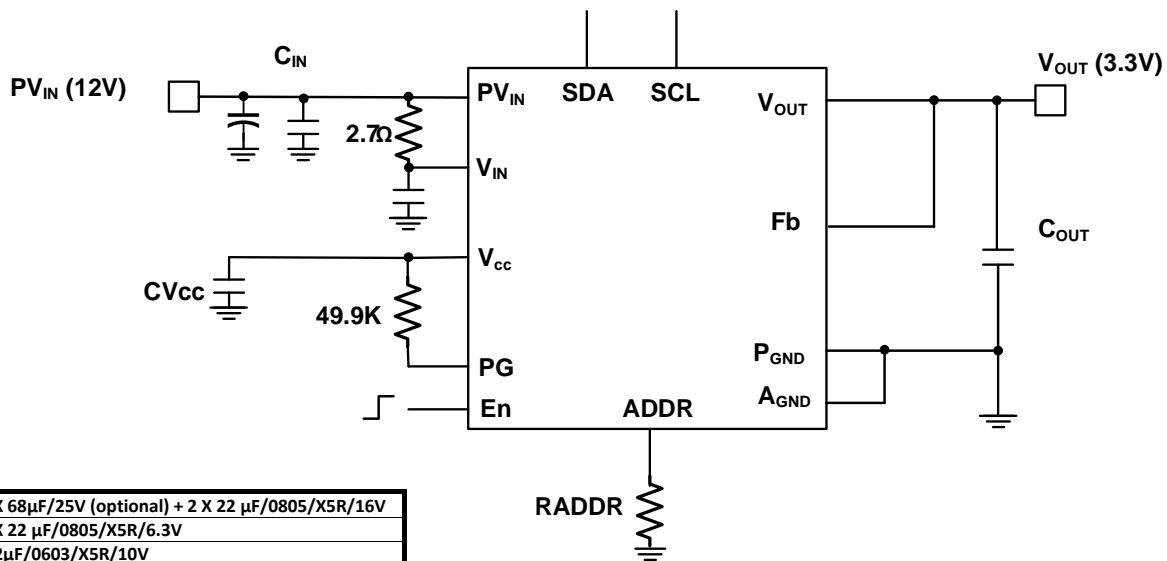


Figure 16 Application circuit for a single supply,
 $PV_{IN}=12V$, $V_{OUT}=3.3V$, 4A

Layout recommendations

FS1604 is a highly integrated device with very few external components, which simplifies PCB layout. However, to achieve the best performance, these general PCB design guidelines should be followed:

- Bypass capacitors, including input/output capacitors and the V_{CC} bypass capacitor (if used), should be placed as close as possible to the FS1604 pins.
- Output voltage should be sensed with a separated trace directly from the output capacitor.
- Analog ground and power ground are connected through a single-point connection.
- To aid thermal dissipation, the PGnd pad should be connected to the power ground plane using vias. Copper-filled vias are preferred but plated-through-hole vias are acceptable, provided that they are not filled with resin or covered with solder mask.
- Adequate numbers of vias should be used to make connections between layers, especially for the power traces.
- To minimize power losses and thermal dissipation, wide copper polygons should be used for input and output power connections.
- SCL and SDA traces must be at least 10mil wide, with 20–30mil spacing between them.

Thermal considerations

The FS1604 has been thermally tested and modelled in accordance with JEDEC specifications JESD 51-2A and JESD 51-8. It has been tested using a 4-layer application PCB, with thermal vias under the device to assist cooling (for details of the PCB, refer to the application notes).

The FS1604 has two significant sources of heat:

- The power MOSFET section of the IC
- The inductor

The IC is well coupled to the PCB, which provides its primary cooling path. Although the inductor is also connected to the PCB, its primary cooling path is through convection. The cooling process for both heat sources is ultimately through convection. The PCB can be seen as a heat-spreader or, to some degree, a heat-sink.

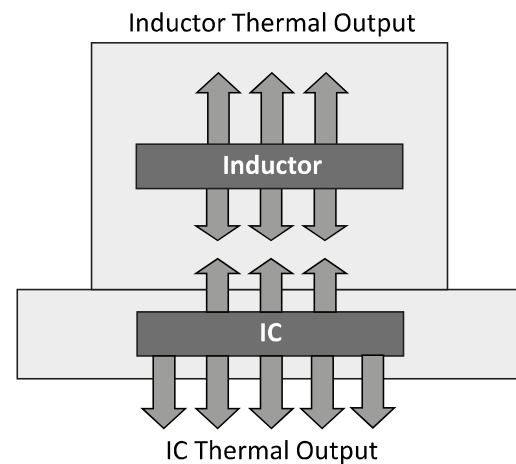


Figure 17 Heat sources in the FS1604

Figure 18 shows the thermal resistances in the FS1604, where:

- Θ_{JA} is the measure of natural convection from the assembled test sample within a confined enclosure of approximately 30x30x30cm. The air is passive within this environment and the only air movement is due to convection from the device on test.
- $\Theta_{JCbottom}$ is the heat flow from the IC to the bottom of the package, to which it is well coupled. The testing method adopts the method outlined in JESD 51-8, where the test PCB is clamped between cold plates at defined distances from the device.
- Θ_{JCtop} is theoretically the heat flow from the IC to the top of the package. This is not representative for the FS1604 for two reasons: firstly, it is not the primary conduction path of the IC and, more importantly, the inductor is positioned directly over the IC. As the inductor is a heat source, generating a similar amount

of heat to the IC, a meaningful value for junction-to-case (top) cannot be derived.

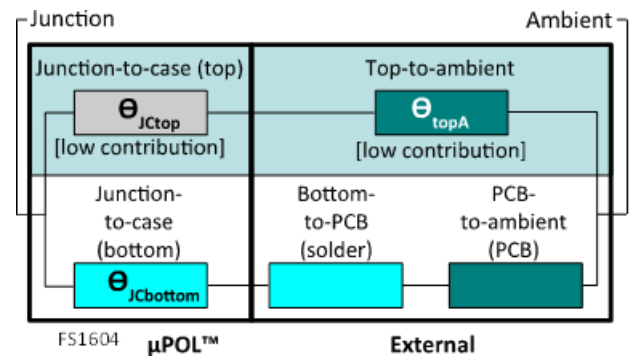


Figure 18 Thermal resistances of the FS1604

The values of the thermal resistances are:

- $\Theta_{JA} = 22.6^{\circ}\text{C/W}$
- $\Theta_{JCbottom} = 2.36^{\circ}\text{C/W}$

Although these values indicate how the FS1604 compares with similar point-of-load products tested using the same conditions and specifications, they cannot be used to predict overall thermal performance. For accurate modeling of the μ POL™'s interaction with its environment, computational fluid dynamics (CFD) simulation software is needed to calculate combined routes of conduction and convection simultaneously.

Note: In all tests, airflow has been considered as passive or static; applications using forced air may achieve a greater cooling effect.

I²C protocol

S = Start bit
P = Stop bit
A = Ack
N = Nack

W = Write bit ('0')
R = Read ('1')
Sr = Repeated start

White bits = Issued by master
Grey bits = Sent by slave (FS160x)

Write transaction

1 7 1 1 8 1 8 1 1

S	Slave Address	W	A	Register Address	A	Data Byte	A	P
---	---------------	---	---	------------------	---	-----------	---	---

Read transaction

1 7 1 1 8 1 1 7 1 1 8 1 1

S	Slave Address	W	A	Register Address	A	Sr	Slave Address	R	A	Data Byte	N	P
---	---------------	---	---	------------------	---	----	---------------	---	---	-----------	---	---

Package description

The FS1604 is designed for use with standard surface-mount technology (SMT) population techniques. It has a positive (raised) footprint, with the pads being higher than the surrounding substrate.

As a result of these properties, the FS1604 works extremely well in lead-free environments. The

surface wets easily and the positive footprint accommodates processing variations.

Note: Refer to the Design Guidelines for more information about TDK's μ POL™ package series P11F1, including important guidance on checking the compatibility of manufacturing processes such as cleanable flux systems.

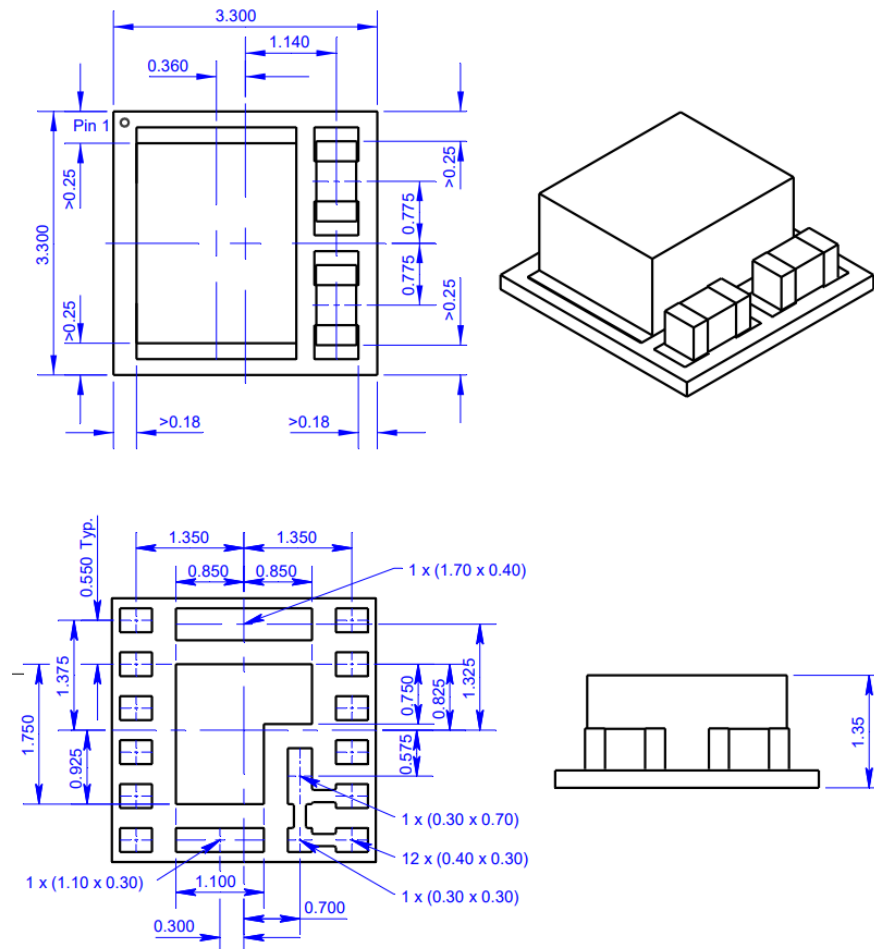
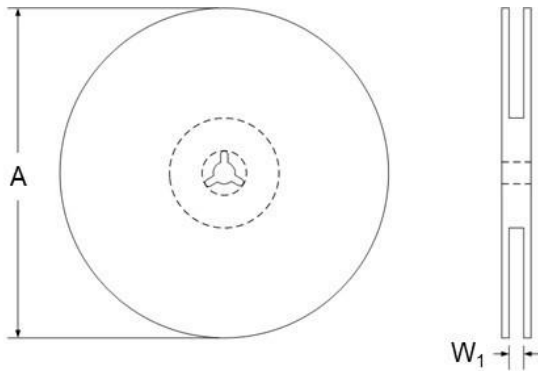


Figure 19 Dimensioned drawings

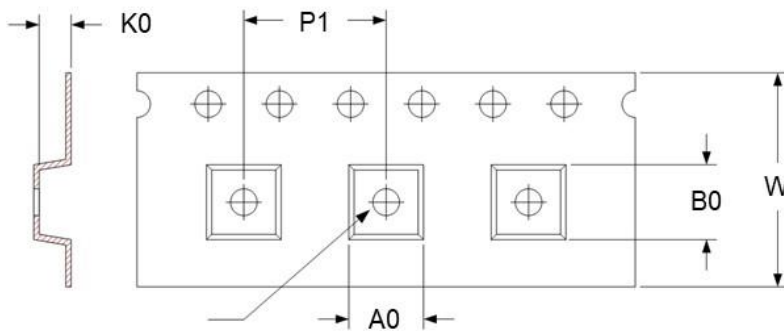
Tape and Reel Information

Reel Dimensions



Reel Diameter (mm)	Reel Width W1 (mm)
330	12.8

Tape Dimensions



P1	8.00
W	12.00
A0	3.60
B0	3.60
K0	1.80

Pin 1 Orientation in Tape

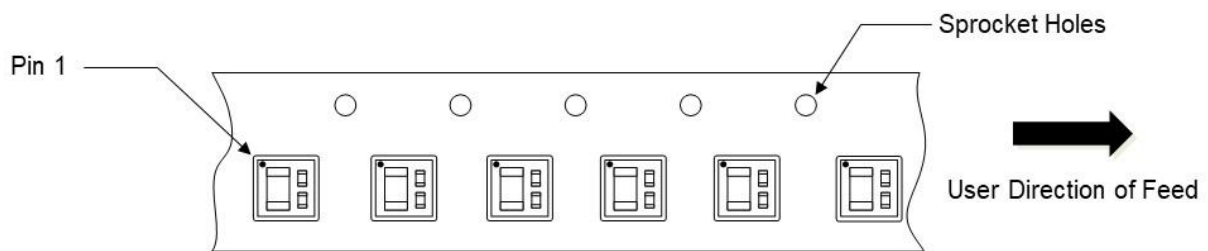


Figure 20 Tape and reel pack

REMINDERS FOR USING THESE PRODUCTS

Before using these products, be sure to request the delivery specifications.

SAFETY REMINDERS

Please pay sufficient attention to the warnings for safe designing when using these products.

REMINDER

The products listed on this specification sheet are intended for use in general electric equipment (AV equipment, telecommunication equipment, home appliances, amusement equipment, computer equipment, personal equipment, office equipment, measurement equipment, industrial robots) under a normal condition and use condition.

The products are not designed or warranted to meet the requirements of the applications listed below, whose performance and/or quality require a more stringent level of safety or reliability, or whose failure, malfunction or trouble could cause serious damage to sociality, person or property. Please understand that we are not responsible for any damage or liability caused by use of the products in any of the applications below or for any other use exceeding the range or conditions set forth in this specification sheet.

1. Aerospace/Aviation equipment
2. Transportation equipment (cars, electric trains, ships, etc.)
3. Medical equipment
4. Power-generation control equipment
5. Atomic energy related equipment
6. Seabed equipment
7. Transportation control equipment
8. Public Information-processing equipment
9. Military equipment
10. Electric heating apparatus, burning equipment
11. Disaster prevention/crime prevention equipment
12. Safety equipment
13. Other applications that are not considered general-purpose applications

When using this product in general-purpose application, you are kindly requested to take into consideration securing protection circuit/ equipment or providing backup circuits, etc., to ensure higher safety.

This product is subject to a license from Power One, Inc. related to digital power technology patents owned by Power One, Inc. Power One, Inc. technology is protected by patents including:

AU 3287379M 3287437AA 3290643AA 3291357AA

CN 10371856C 10452610C 10458656C 10459360C 10465848C 1069332A 11124619A 11346682A 1685299A 1685459A 1685582A 1685583A 1698023A 1802619A

EP 1561156A1 1561268A2 1576710A1 1576711A1 1604254A4 1604264A4 1714369A2 1745536A4 1769382A4 1899789A2 1984801A2

US 20040246754 2004090219A1 2004093533A1 2004123164A1 2004123167A1 2004178780A1 2004179382A1 20050200344 20050223252 2005209373A1 20060061214 2006015619A1 20060174145 20070226526 20070234095 20070240000 20080052551 20080072080 20080186006 6741099 6788036 6936999 6949916 7000125 7049798 7069021 7080265 7249267 7266709 7315156 7372682 7373527 7394445 7456617 7459892 7493504 7526660

WO 04044718A1 04045042A3 04045042C1 04062061A1 04062062A1 04070780A3 04084390A3 04084391A3 05079227A3 05081771A3 06019569A3 2007001584A3 2007094935A3