

# 20-Output DB2000QL Buffer with Ultra-Low Additive Jitter

#### **Features**

- Fully Compliant with Intel DB2000QL Specification
- 20 Low-Power Push-Pull HCSL PCIe Outputs
- Supports Latest PCIe Gen 5.0 and Gen 6.0 Spec
- Ultra-Low Additive Jitter: 14 fs Typical in DB2000QL Band
- Supports Clock Frequencies from 0 MHz to 250 MHz
- · Supports 3.3V Power Supplies
- Embedded Low Drop Out (LDO) Voltage Regulator Provides Superior Power Supply Noise Rejection
- · Maximum Output to Output Skew of 50 ps
- · SMBus Interface
- · Side-Band Interface (SBI)
- · Eight OE Pins
- Embedded Series Termination Resistors for  $85\Omega$  Differential Transmission Line
- · Transparent for Spread Spectrum Clock

# **Applications**

- PCI Express generation 1/2/3/4/5/6 clock distribution
- Intel QPI
- Servers
- · Storage and Data Centers
- · Switches and Routers

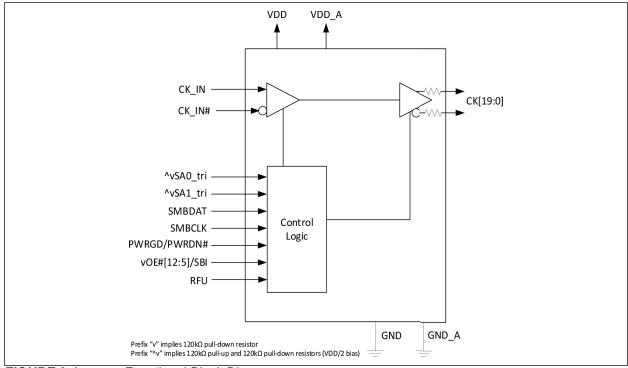


FIGURE 0-1: Functional Block Diagram

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# 1.0 PIN DIAGRAM

The device is packaged in a 6 mm × 6 mm 80-lead WGQFN.

	1	2	3	4	5	6	7	8	9	10	11	12	_
Α	CK17	CK16#	СК16	CK15#	CK15	CK14#	CK14	CK13#	CK13	CK12#	CK12	CK11#	А
В	CK17#	VDD	NC	^vSA0_tri	NC	VDD	NC	^vSA1_tri	NC	vOE12#	VDD	CK11	В
С	CK18	NC								_	vOE11#	CK10#	С
D	CK18#	NC									NC	CK10	D
E	CK19	vSBEN									vOE10# vSHFT_LD#	vOE9#	E
F	CK19#	NC			DB 2000QL 80-pin GQFN						NC	СК9#	F
G	CK_IN	NC				top	x 6mm view is GND				NC	СК9	G
Н	CK_IN#	VDD_A									vOE8#	CK8#	Н
J	СКО	NC									NC	СК8	J
K	СКО#	NC									vOE7#	CK7#	К
L	CK1	VDD	NC	SMBDAT	SMBCLK	NC	NC	vOE5# vDATA	NC	vOE6# vCLK	VDD	СК7	L
М	CK1#	CK2	CK2#	СКЗ	СК3#	vPWRGN/ PWRDWN#	CK4	CK4#	CK5	CK5#	CK6	CK6#	М
	1	2	3	4	5	6	7	8	9	10	11	12	J

FIGURE 1-1: Pin Diagram.

# 2.0 PIN DESCRIPTIONS

The I/O column in Table 2-1 uses the following symbols:

- I = Input
- $I_{PU}$  = Input with 120 k $\Omega$  internal pull-up resistor
- $I_{PD}$  = Input with 120 k $\Omega$  internal pull-down resistor
- O = Output
- I/O = Input/Output Drain pin
- NC = No connect pin
- P = Power supply pin
- I<sub>TRI</sub> = Tri-level input pin biased to VDD/2 by internal 120  $k\Omega$  pull-up and 120  $k\Omega$  pull-down resistors

# TABLE 2-1: PIN DESCRIPTIONS

Pin Number	Pin Name	Input/Output	Description
Input Reference			<u> </u>
G1	CK_IN		Input Differential or Single Ended Reference.
H1	CK_IN#	- I	Input frequency range 0 Hz to 250 MHz.
<b>Output Clocks</b>			
J1	CK0		
K1	CK0#		
L1	CK1		
M1	CK1#		
M2	CK2		
M3	CK2#		
M4	CK3		
M5	CK3#		
M7	CK4		
M8	CK4#		
M9	CK5		
M10	CK5#		
M11	CK6		
M12	CK6#		
L12	CK7	0	Ultra-Low Additive Jitter Differential Outputs 0 to 19. Output frequency range 0 MHz to 250 MHz
K12	CK7#		Output frequency range of while to 200 while
J12	CK8		
H12	CK8#		
G12	CK9		
F12	CK9#		
D12	CK10		
C12	CK10#		
B12	CK11		
A12	CK11#		
A11	CK12		
AS10	CK12#		
A9	CK13		
A8	CK13#		
A7	CK14		

TABLE 2-1: PIN DESCRIPTIONS (CONTINUED)

Pin Number	Pin Name	Input/Output	Description
A6	CK14#		-
A5	CK15		
A4	CK15#		
A3	CK16		
A2	CK16#		
A1	CK17	О	Ultra-Low Additive Jitter Differential Outputs 0 to 19
B1	CK17#		Output frequency range 0 MHz to 250 MHz
C1	CK18		
D1	CK18#		
E1	CK19		
F1	CK19#		
Hardware Contr	ol		1
K11	vOE7#		Output English Lagis lavel on those pine anables/disables
H11	vOE8#		Output Enable. Logic level on these pins enables/disables corresponding output:
E12	vOE9#	I <sub>PD</sub>	OE_n# = CKn/n#
C11	vOE11#		0 = Active
B10	vOE12#		1 = Low/Low both pulled low by $42.5Ω$
L8	vOE5#/vDATA	I <sub>PD</sub>	Output Enable/Data Input for Side Band Interface. When Side-Band interface is disabled (pin SBEN pulled low) this pin is Output Enable and the description is the same as for pin K11 above. When Side-Band interface is enabled (pin SBEN pulled high) this pin is Data Input for Side-Band interface.
L10	vOE6#/vCLK	I <sub>PD</sub>	Output Enable/Clock Input for Side Band Interface. When Side-Band Interface is disabled (pin SBEN pulled low) this pin is Output Enable and the description is the same as for pin K11 above. When Side-Band interface is enabled (pin SBEN pulled high) this pin is Clock Input for Side-Band interface.
E11	vOE10#/ vSHFT_LD#	I <sub>PD</sub>	Output Enable/Shift Load for Side Band Interface. When Side-Band interface is disabled (pin SBEN pulled low) this pin is Output Enable and the description is the same as for pin K11 above. When Side-Band interface is enabled (pin SBEN pulled high) this pin is Shift Load Input for Side-Band interface. A falling edge on this pin transfers the sideband shift register content to the output register.
E2	vSBEN	I <sub>PD</sub>	Side-Band Interface Enable. When this pin is low, the Side-Band interface is disabled and OE pins and OE Register bits (via SMBus) can be used to enable/disable outputs. When this pin is high, the Side-Band interface can be used to enable/disable outputs, and OE pins and OE Register bits (via SMBus) are disabled.
M6	PWRGD/ PWRDN#	I	Power Up/Power Down.
SMBus Control			
L5	SMBCLK	I	SMBus Client Clock Input.
L4	SMBDAT	I/O	Input/Open Drain SMBus Data.

TABLE 2-1: PIN DESCRIPTIONS (CONTINUED)

Pin Number		ABLE 2-1: PIN DESCRIPTIONS (CONTINUED)							
No Connect Pins	Pin Number	Pin Name	Input/Output	Description					
B8	B4	^vSA0_tri		Tri Level Address Selection Inputs					
B2	B8	^vSA1_tri	ITRI	Til Level Address Selection inputs.					
B6	Power and Grou	nd							
B111	B2								
L2	B6								
L11	B11	VDD	Р	Positive Supply Voltage. Connect to 3.3V supply.					
H2	L2								
No Connect Pins	L11								
No Connect Pins	H2	VDD_A	Р						
B3	E-Pad	GND	Р	Ground. Connect to ground.					
B5	No Connect Pins	5							
B7	В3								
B9	B5								
C2	B7								
D2 D11 F2 F11 N/C F11 N/C F2 F11 R12 G2 G11 J2 J11 K2 L3 L6 L7 L9 Pin Number Pin Name Input/Output Description  Input Reference G1 CK_IN H1 CK_IN# CK_IN# H1 CK_IN# CK_IN* H1 CK_IN# CK_IN* L1 CK_IN* L1 CK_IN* L1 CK_IN* L1 CK_IN* L1 CK_IN* CK_IN* L1 CK_IN* C	B9								
D11	C2								
F2 F11 N/C No Connect. These pins are not connected to the die. Leave them open. One of these pins might be used for future modifications of DB2000QL spec. The current DB2000QL v1.0 standard calls for Reserved for Future Upgrades (RFU) pin but does not assign it to any pin number. F12 N/C F13 N/C F14 N/C F15 No Connect. These pins are not connected to the die. Leave them open. One of these pins might be used for future modifications of DB2000QL spec. The current DB2000QL v1.0 standard calls for Reserved for Future Upgrades (RFU) pin but does not assign it to any pin number.  F15 F16 F17 F17 F18	D2								
F2	D11			No Connect Those nine are not connected to the die Leave					
F11 G2 G11 G11 J2 J11 K2 L3 L6 L7 L9 Pin Number Pin Name Input/Output Description  F11 CK_IN# H1 CK_IN#  CK_IN#  CK_IN#  J1 CKO K11 CK0# K11 CK1# M1 CK1# M1 CK1# M1 CK1# M2 CK2 M3 CK2# M4 CK3  One of these pins might be used for future modifications of DB2000QL spec. The current DB2000QL v1.0 standard calls for Reserved for Future Upgrades (RFU) pin but does not assign it to any pin number.  Description  Description  Input Differential or Single Ended Reference Input frequency range 0 Hz to 250 MHz  Ultra-Low Additive Jitter Differential Outputs 0 to 19 Output frequency range 0 MHz to 250 MHz  Ultra-Low Additive Jitter Differential Outputs 0 to 19 Output frequency range 0 MHz to 250 MHz	F2								
B2000QL spec. The current DB200QL V1.0 standard calls for Reserved for Future Upgrades (RFU) pin but does not assign it to any pin number.    Comparison	F11	N/C		One of these pins might be used for future modifications of					
Sin   J2	G2	IN/C							
11	G11								
K2	J2		assign to any pinnamber.	assign to any pin number.					
L3	J11								
L6	K2								
L7	L3								
L9	L6								
Pin Number         Pin Name         Input/Output         Description           Input Reference         61         CK_IN         Input Differential or Single Ended Reference Input frequency range 0 Hz to 250 MHz           H1         CK_IN#         Input Differential or Single Ended Reference Input frequency range 0 Hz to 250 MHz           Output Clocks         J1         CK0           K1         CK0#         CK0#           L1         CK1         CK1           M1         CK1#         Output frequency range 0 MHz to 250 MHz           M3         CK2#           M4         CK3	L7								
Input Reference	L9								
G1	Pin Number	Pin Name	Input/Output	Description					
H1	Input Reference	•	•						
H1	G1	CK_IN		Input Differential or Single Ended Reference					
J1         CK0           K1         CK0#           L1         CK1           M1         CK1#           M2         CK2           M3         CK2#           M4         CK3    Ultra-Low Additive Jitter Differential Outputs 0 to 19 Output frequency range 0 MHz to 250 MHz	H1		7 '						
K1         CK0#           L1         CK1           M1         CK1#           M2         CK2           M3         CK2#           M4         CK3    Ultra-Low Additive Jitter Differential Outputs 0 to 19 Output frequency range 0 MHz to 250 MHz	<b>Output Clocks</b>	•	•						
L1 CK1 M1 CK1# O Ultra-Low Additive Jitter Differential Outputs 0 to 19 Output frequency range 0 MHz to 250 MHz M3 CK2# M4 CK3	J1	CK0							
M1 CK1# M2 CK2 M3 CK2# M4 CK3  Ultra-Low Additive Jitter Differential Outputs 0 to 19 Output frequency range 0 MHz to 250 MHz	K1	CK0#							
M2 CK2 M3 CK2# M4 CK3  Output frequency range 0 MHz to 250 MHz	L1	CK1							
M2 CK2 M3 CK2# M4 CK3  Output frequency range 0 MHz to 250 MHz	M1	CK1#	1	Ultra-Low Additive Jitter Differential Outputs 0 to 19					
M4 CK3	M2	CK2	] 0						
	M3	CK2#	1						
M5 CK3#	M4	СКЗ	1						
	M5	CK3#							

# **ZL40294**

NOTES:

### 3.0 FUNCTIONAL DESCRIPTION

The ZL40294 is an ultra-low additive jitter, low power 1 to 20 fanout buffer which is fully compliant with Intel DB2000QL Standard.

The device operates from 3.3V+/-5% supply as per Intel spec. Its operation is guaranteed over the industrial temperature range -40°C to +85°C.

#### 3.1 Clock Inputs

The following blocks diagram shows how to terminate different signals fed to the inputs of ZL40294 device.

The device input can be fed with transmission lines of any impedance. Examples below show only  $50\Omega$  single ended,  $85\Omega$  differential and  $100\Omega$  differential which are the most common ones in practice. Figure 3-1 and Figure 3-2 show how to terminate the input when driven from a push-pull and traditional HCSL drivers respectively.

Figure 3-3 shows how to terminate a single ended output such as LVCMOS. This example assumes  $50\Omega$  transmission line which is the most common for single ended CMOS signaling. Ideally, resistors R1 and R2 should be  $100\Omega$  each and  $R_O + R_S$  should be  $50\Omega$  so that the transmission line is terminated at both ends with characteristic impedance.

If the driving strength of the output driver is not sufficient to drive low impedance, the value of series resistor RS should be increased. This will reduce the voltage swing at the input but this should be fine as long as the input voltage swing requirement is not violated (Figure 3-3). The source resistors of  $R_S = 270\Omega$  could be used for standard LVCMOS driver. This will provide 516 mV of voltage swing for 3.3V LVCMOS driver with load current of (3.3V/2) \*(1/(270 $\Omega$  + 50 $\Omega$ )) = 5.16 mA.

For optimum performance both differential input pins (\_p and \_n) need to be DC biased to the same voltage. Hence, the ratio R1/R2 should be equal to the ratio R3/R4.

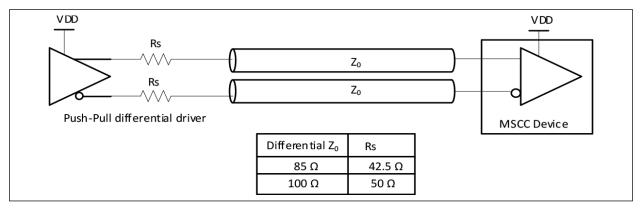


FIGURE 3-1: INPUT DRIVEN BY A PUSH-PULL DIFFERENTIAL OUTPUT.

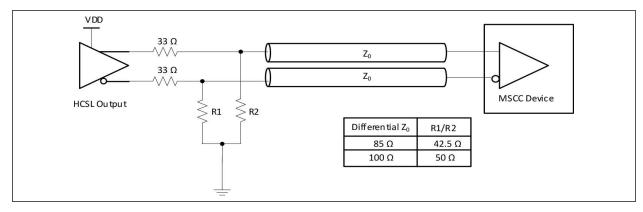


FIGURE 3-2: INPUT DRIVEN BY AN HCSL OUTPUT.

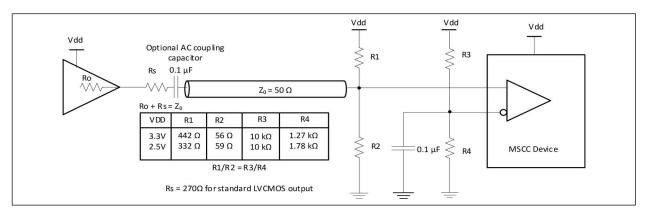


FIGURE 3-3: INPUT DRIVEN BY A SINGLE ENDED OUTPUT.

### 3.2 Clock Outputs

Differential outputs have embedded termination resistors as shown in Figure 3-4. This provides significant saving relative to traditional current based HCSL outputs which require four resistors per differential pair (80 resistors for 20 outputs).

Embedded termination resistors in ZL40294 are matched for  $85\Omega$  differential transmission line.

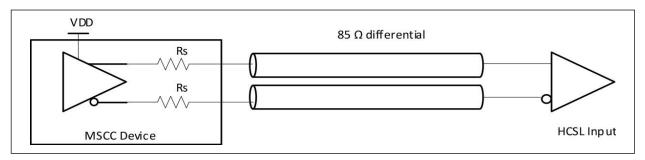


FIGURE 3-4: TERMINATING DIFFERENTIAL OUTPUTS.

# 3.3 Termination of unused outputs

Unused outputs should be left unconnected.

# 3.4 Power Supply Filtering

Each power pin (VDDA and VDD) should be decoupled with 0.1  $\mu$ F capacitor with minimum equivalent series resistance (ESR) and minimum series inductance (ESL). For example, 0402 X5R Ceramic Capacitors with 6.3V minimum rating could be used. These capacitors should be placed as close as possible to the power pins. To reduce the power noise from adjacent digital components on the board each power supply could be further insulated with low DC resistance ferrite bead with two capacitors. The ferrite bead will also insulate adjacent component from the noise generated from the device. Figure 3-5 shows recommended decoupling.

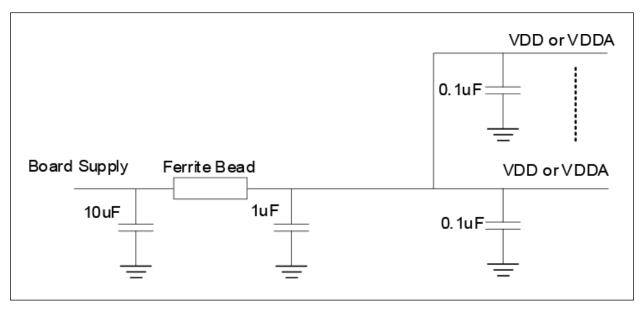


FIGURE 3-5: POWER SUPPLY FILTERING.

### 3.5 OE# and Output Enables (Control Register)

Each output can be individually enabled or disabled by SMBus control register bits or via OE# pin. The OE# pins are asynchronous asserted-low signals. The Output Enable bits in the SMBus registers are active high and are set to enable by default.

OE# pins are mapped to CK[12:5] outputs.

Note that the logic level for assertion or de-assertion is different in software than it is on hardware. This follows hardware default nomenclature for communication channels (e.g., output is enabled if OE# pin is pulled low) and still maintains software programming logic (e.g., output is enabled if OE register is true).

Refer to Table 3-1 for the truth table for enabling and disabling outputs via hardware and software. Note that both the control register bit must be a '1' AND the OE# pin must be a '0' for the output to be active.

Inpu	ıts	OI	E# Hardwai	re Pins and Control Reg	ister Bits
PWRGD/ PWRDN#	CK_IN/ CK_IN#	SMBUS Enable Bit	OE# Pin	CK/CK# [12:5]	CK/CK# [4:0] and [19:13]
0	Х	Х	Х	0	0
		0	Х	0	0
1	Running	1	0	Running	Running

0

**TABLE 3-1: OE FUNCTIONALITY** 

#### 3.6 OE# Assertion (Transition from '1' to '0')

All differential outputs that were disabled are to resume normal operation in a glitch free manner. The latency from the assertion to active outputs is 0 - 10 CK clock periods.

# 3.7 OE# De-Assertion (Transition from '0' to '1')

The impact of de-asserting OE# is each corresponding output will transition from normal operation to disabled in a glitch free manner. A minimum of four valid clocks will be provided after the de-assertion of OE#. The maximum latency from the de-assertion to disabled outputs is 10 CK clock periods.

Running

#### 3.8 PWRGD / PWRDN#

PWRGD is asserted high and de-asserted low. De-assertion of PWRGD (pulling the signal low) is equivalent to indicating a powerdown condition. PWRGD (assertion) is used by the ZL40294 to sample initial configurations such as SA selections.

After PWRGD has been asserted high for the first time, the pin becomes a PWRDN# (Power Down) pin which is used to disable (drive low/low) all clocks cleanly and instruct the device to invoke power savings mode. PWRDN# is a completely asynchronous active low input. When entering power savings mode, PWRDN# should be asserted low prior to shutting off the input clock or power to ensure all clocks shut down in a glitch free manner. When PWRDN# is deasserted high, all clocks will start and stop without any abnormal behavior and will meet all AC and DC parameters.

The assertion and de-assertion of PWRDN# is asynchronous.

Disabling of the CK\_IN input clock prior to assertion of PWRDN# is an undefined mode and not recommended. Operation in this mode may result in glitches.

TABLE 3-2: PWRGD/PWRDN# FUNCTIONALITY

PWRGD/PWRDN#	СК	CK#
0	Low	Low
1	1 Normal	

#### 3.9 PWRDN# Assertion

When PWRDN# is sampled low by two consecutive rising edges of CK#, all differential outputs will be disabled on the next CK# high to low transition.

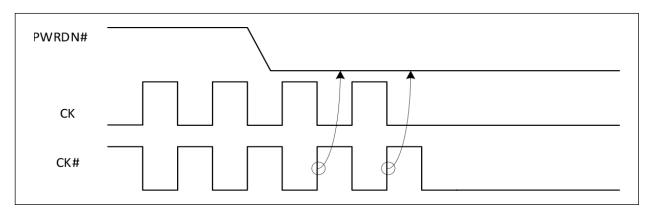


FIGURE 3-6: PWRDN# ASSERTION

#### 3.10 PWRGD Assertion

PWRGD to the clock buffer should not be asserted before  $V_{DD}$  reaches  $V_{DD(MIN)}$ . Prior to  $V_{DD(MIN)}$  it is recommended to hold PWRGD low (less than 0.5 V).

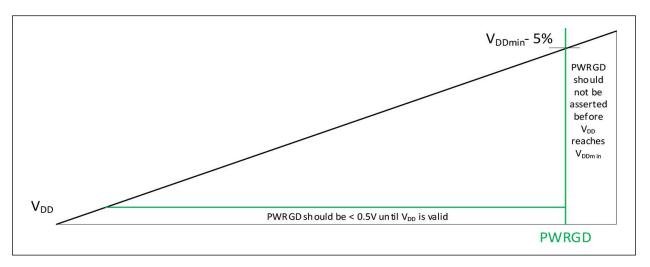


FIGURE 3-7: PWRGD AND VDD RELATIONSHIP DIAGRAM.

The power-up latency  $T_{STABLE}$  is to be less than 1.8 ms. This is the time from the valid CK\_IN input clocks and the assertion of the PWRGD signal to the time that stable clocks are output from the buffer chip. All differential outputs stopped in a disabled condition resulting from power down must be driven high in less than 300  $\mu$ s of PWRGD assertion to a voltage greater than 200 mV.

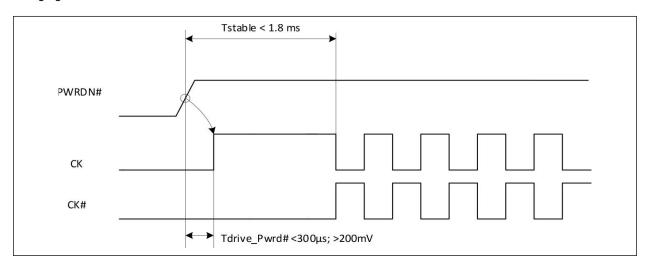


FIGURE 3-8: PWRGD ASSERTION.

#### 3.11 Programming via SMBus

The address selection is done via SA\_0 and SA\_1 tri-level hardware pins, which select the appropriate address for the device

The two tri-level input pins that can configure the ZL40294 to nine different addresses (refer to Table 5-2 for VIL\_Tri, VIM Tri, VIH Tri signal level).

TABLE 3-3: SMBUS ADDRESS TABLE

SA_1	SA_0	SMBus Address
L	L	D8
L	M	DA
L	Н	DE
M	L	C2
M	M	C4
M	Н	C6
Н	L	CA
Н	M	CC
Н	Н	CE

# 3.12 SMBus Byte Read/Write

Reading or writing a register in a SMBus client device in byte mode always involves specifying the register number.

**Read.** The standard byte read is as shown in Figure 11. It is an extension of the byte write. The write start condition is repeated then the client device starts sending data and the host acknowledges it until the last byte is sent. The host terminates the transfer with a NAK then a stop condition. For byte operation, the 2\*7th bit of the command byte must be set. For block operations, the 2\*7th bit must be reset. If the bit is not set, the next byte must be the byte transfer count.

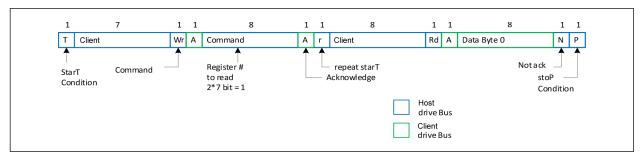


FIGURE 3-9: SMBUS BYTE READ.

**Write.** Figure 12 illustrates a simple typical byte write. For byte operation the 2\*7th bit of the command byte must be set. For block operations, the 2\*7th bit must be reset. If the bit is not set the next byte must be the byte transfer count. The count can be between 1 and 32. It cannot be zero or exceed 32.

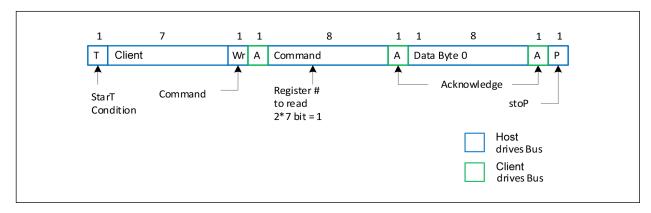


FIGURE 3-10: SMBUS BYTE WRITE.

#### 3.13 SMBus Block Read/Write

**Read.** After the client address is sent with the r/w condition bit set, the command byte is sent with the MSB = 0. The client Ack's the register index in the command byte. The host sends a repeat start function. After the client Ack's this the client sends the number of bytes it wants to transfer (>0 and <33). The host Ack's each byte except the last and sends a stop function.

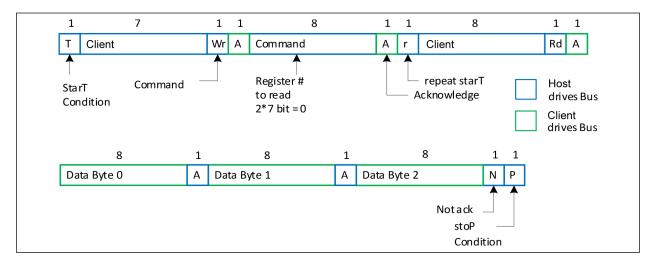


FIGURE 3-11: SMBUS BLOCK READ.

Write. After the client address is sent with the r/w condition bit not set, the command byte is sent with the MSB = 0. The lower seven bits indicate what register to start the transfer at. If the command byte is 00h, the client device will be compatible with existing block mode client devices. The next byte of a write must be the count of bytes that the host will transfer to the client device. The byte count must be greater than zero and less than 33. Following this byte are the data bytes to be transferred to the client device. The client device always acknowledges each byte received. The transfer is terminated after the client sends the Ack and the host sends a stop function.

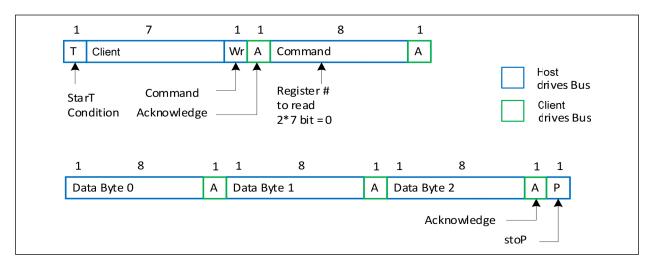


FIGURE 3-12: SMBUS BLOCK WRITE.

#### 3.14 Side-Band Interface

Besides OE pins and SMBUS interface the device outputs can be enabled/disabled via a simple 3-wire serial interface referred to as the Side-Band Interface (SBI).

This interface consists of the vDATA, vCLK and vSHFT\_LD# pins. When the vSHFT\_LD# pin is high, the rising edge of vCLK can shift vDATA into the shift register. After shifting data, the falling edge of vSHFT\_LD# clocks the shift register contents to the Output register.

Both the SBI and the traditional interface feed common output enable/disable synchronization logic ensuring glitch free enable and disable outputs, regardless of the method used.

Both interfaces are not active at the same time, and the SBEN pin selects which interface is active. Tying the SBEN high enables the SBI. Tying the SBEN pin low enables the traditional OE# pin/SMBus output-enable interface.

When the SBI is enabled, OE[7:9, 11,12]# are disabled and vDATA, vCLK and vSHFT\_LD# are enabled on vOE5#, vOE6# and vOE10# respectively. Additionally, SMBus registers for masking off the disable function of the shift register (0 value of a bit) become active.

When set to 1, the mask register forces its respective output to 'enabled.' This prevents accidentally disabling critical outputs when using the SBI.

A SMBus read-back bit in Byte 4 indicates which output-enable control interface is enabled.

When the SBI is enabled and power has been applied, the SBI is active, even if the PWRGD/PWRDN# pin indicates the part is in powerdown. This allows loading the shift register and transferring the contents to the output register before the assertion of PWRGD.

Note that the mask registers are part of the normal SMBus interface and cannot be accessed when the PWRGD/PWRDN# is low. Figure 3-13 provides a functional description of the SBI.

The SBI and the traditional SMBus output-enable registers both default to the 'output enabled' state at power-up. The mask registers default to zero at power-up, allowing the shift register bits to disable their respective output.

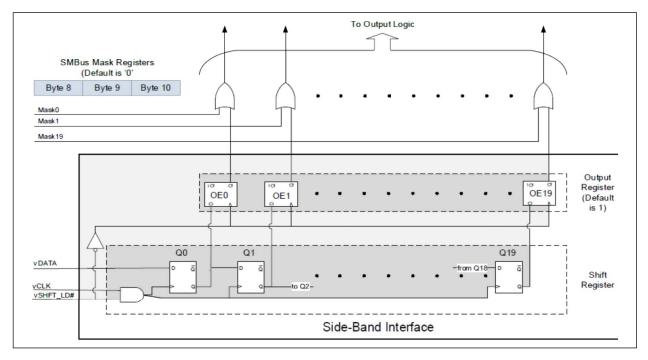


FIGURE 3-13: SIDE-BAND INTERFACE CONTROL LOGIC (FUNCTIONAL DESCRIPTION).

Figure 3-14 shows the basic timing of the side-band interface. The vSHFT\_LD# pin goes high to enable the CLK input. Next, the rising edge of CLK clocks enable vDATA into the shift register. After the 20th clock cycle for output 19, it stops the clock low and drive the vSHFT\_LD# pin low. The falling edge of vSHFT\_LD# clocks the shift register contents to the output register, enabling or disabling the outputs. It always shifts 20 bits of data into the shift register to control the outputs.

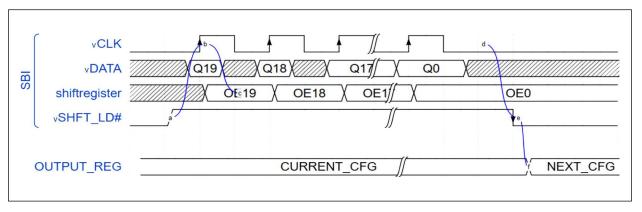


FIGURE 3-14: SIDE-BAND INTERFACE FUNCTIONAL TIMING.

The SBI supports clock rates up to 10 MHz. Multiple devices may share vCLK and vDATA pins. Dedicating a vSH-FT\_LD# pin to each device allows its use as a chip-select pin. When the vSHFT\_LD# pin is low, the ZL40292 device ignores any activity on the vCLK and vDATA pins.

# 3.15 Typical Jitter Performance

Following figure shows a typical phase noise for 100 MHz clock frequency. The light blue curve is the phase noise of the input clock fed to the device and the dark blue is the phase noise of the output clock. The plot shows phase noise floor of -163 dBc/Hz and an absolute jitter in 12 kHz to 20 MHz of 76 fs. The additive jitter is calculated as:

#### **EQUATION 3-1:**

$$J_{ADD} = \sqrt{(J^2_{OUT} - J^2_{IN})}$$

Where  $J_{OUT}$  and  $J_{IN}$  are respectively output and input jitter in the band of interest.

The additive jitter for 12 kHz to 20 MHz band is 58 fs. The additive jitter in the bandpass filter specified by DB2000QL is 14 fs.

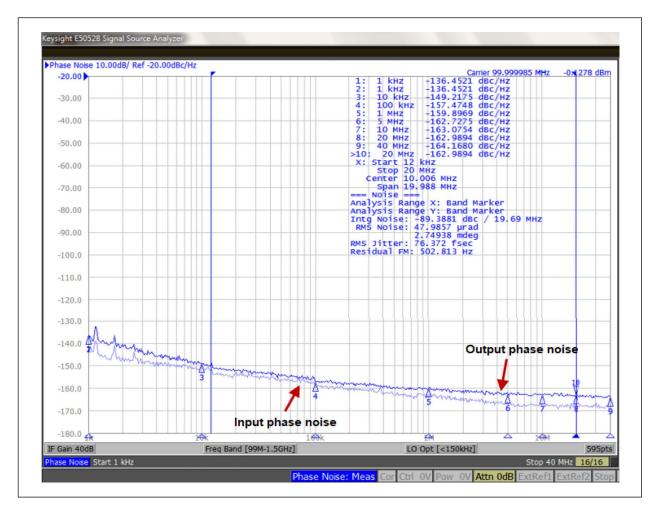


FIGURE 3-15: ZL40294 TYPICAL PHASE NOISE.

# 4.0 REGISTER MAPS

TABLE 4-1: BYTE 0: OUTPUT ENABLE

Bit	Description	If Bit = 0	If Bit = 1	Туре	Default	Output(s) Affected
0	Reserved	_	_	_	0	_
1	Reserved	_	_	_	0	_
2	Reserved	_	_	_	0	_
3	Output Enable CK_16	Low	Enabled	RW	1	CK[16]
4	Output Enable CK_17	Low	Enabled	RW	1	CK[17]
5	Output Enable CK_18	Low	Enabled	RW	1	CK[18]
6	Output Enable CK_19	Low	Enabled	RW	1	CK[19]
7	Reserved	_	_	_	0	_

TABLE 4-2: BYTE 1: OUTPUT ENABLE CONTROL REGISTER

Bit	Description	If Bit = 0	If Bit = 1	Туре	Default	Output(s) Affected
0	Output Enable CK_0	Low	Enabled	RW	1	CK[0]
1	Output Enable CK_1	Low	Enabled	RW	1	CK[1]
2	Output Enable CK_2	Low	Enabled	RW	1	CK[2]
3	Output Enable CK_3	Low	Enabled	RW	1	CK[3]
4	Output Enable CK_4	Low	Enabled	RW	1	CK[4]
5	Output Enable CK_5	Low	Enabled	RW	1	CK[5]
6	Output Enable CK_6	Low	Enabled	RW	1	CK[6]
7	Output Enable CK_7	Low	Enabled	RW	1	CK[7]

TABLE 4-3: BYTE 2: OUTPUT ENABLE CONTROL REGISTER

Bit	Description	If Bit = 0	If Bit = 1	Туре	Default	Output(s) Affected
0	Output Enable CK_8	Low	Enabled	RW	1	CK[8]
1	Output Enable CK_9	Low	Enabled	RW	1	CK[9]
2	Output Enable CK_10	Low	Enabled	RW	1	CK[10]
3	Output Enable CK_11	Low	Enabled	RW	1	CK[11]
4	Output Enable CK_12	Low	Enabled	RW	1	CK[12]
5	Output Enable CK_13	Low	Enabled	RW	1	CK[13]
6	Output Enable CK_14	Low	Enabled	RW	1	CK[14]
7	Output Enable CK_15	Low	Enabled	RW	1	CK[15]

TABLE 4-4: BYTE 3: OE# PIN REALTIME READBACK CONTROL REGISTER

Bit	Description	If Bit = 0	If Bit = 1	Type	Default	Output(s) Affected
0	Realtime Readback of OE_5#	OE_5# Low	OE_5# High	R	Realtime	CK[5]
1	Realtime Readback of OE_6#	OE_6# Low	OE_6# High	R	Realtime	CK[6]
2	Realtime Readback of OE_7#	OE_7# Low	OE_7# High	R	Realtime	CK[7]
3	Realtime Readback of OE_8#	OE_8# Low	OE_8# High	R	Realtime	CK[8]
4	Realtime Readback of OE_9#	OE_9# Low	OE_9# High	R	Realtime	CK[9]
5	Realtime Readback of OE_10#	OE_10# Low	OE_10# High	R	Realtime	CK[10]
6	Realtime Readback of OE_11#	OE_11# Low	OE_11# High	R	Realtime	CK[11]
7	Realtime Readback of OE_12#	OE_12# Low	OE_12# High	R	Realtime	CK[12]

# TABLE 4-5: BYTE 4: RESERVED CONTROL REGISTER

Bit	Description	If Bit = 0	If Bit = 1	Туре	Default	Output(s) Affected
0	RB_SBEN	Pin Low	Pin High	Input	Input Real time	
1	Reserved	_	_	_	_	_
2	Reserved	_	_	_	_	_
3	Reserved	_	_	_		_
4	Reserved	_	_	_	_	_
5	Reserved	_	_	_	_	_
6	Reserved	_	_	_		_
7	Reserved	_	_	_	_	_

TABLE 4-6: BYTE 6: DEVICE ID CONTROL REGISTER

Bit	Description	If Bit = 0	If Bit = 1	Type	Default	Output(s) Affected
0	Device ID 0	_	_	R	0	_
1	Device ID 1	_	_	R	1	_
2	Device ID 2	_	_	R	1	_
3	Device ID 3	_	_	R	1	_
4	Device ID 4	_	_	R	1	_
5	Device ID 5	_	_	R	0	_
6	Device ID 6	_	_	R	1	_
7	Device ID 7 (MSB)	_	_	R	0	_

TABLE 4-7: BYTE 7: BYTE COUNT REGISTER

Bit	Description	If Bit = 0	If Bit = 1	Type	Default	Output(s) Affected
0	BC0 - Writing to this register configures how many bytes will be read back	_	_	RW	0	_
1	BC1 - Writing to this register configures how many bytes will be read back	_	_	RW	0	_
2	BC2 - Writing to this register configures how many bytes will be read back	_	_	RW	0	_
3	BC2 - Writing to this register configures how many bytes will be read back	_	_	RW	1	_
4	BC3 - Writing to this register configures how many bytes will be read back	_	_	RW	0	_
5	BC4 - Writing to this register configures how many bytes will be read back	_	_	RW	0	_
6	Reserved			RW	0	_
7	Reserved	_	_	RW	0	_

TABLE 4-8: BYTE 8: SBI MASK (FUNCTIONAL ONLY WHEN SBEN = 1)

Bit	Description	If Bit = 0	If Bit = 1	Туре	Default	Output(s) Affected
0	Mask0 - Masks off Side Band Disable		Forces output to always be enabled regardless of Side Band shift register	RW	0	CK[0]
1	Mask1 - Masks off Side Band Disable			RW	0	CK[1]
2	Mask2 - Masks off Side Band Disable			RW	0	CK[2]
3	Mask3 - Masks off Side Band Disable	Side Band shift register may		RW	0	CK[3]
4	Mask4 - Masks off Side Band Disable	disable the output		RW	0	CK[4]
5	Mask5 - Masks off Side Band Disable		value	RW	0	CK[5]
6	Mask6 - Masks off Side Band Disable			RW	0	CK[6]
7	Mask7 - Masks off Side Band Disable			RW	0	CK[7]

# TABLE 4-9: BYTE 9: SBI MASK (FUNCTIONAL ONLY WHEN SBEN = 1)

Bit	Description	If Bit = 0	If Bit = 1	Туре	Default	Output(s) Affected
0	Mask8 - Masks off Side Band Disable			RW	0	CK[8]
1	Mask9 - Masks off Side Band Disable			RW	0	CK[9]
2	Mask10 - Masks off Side Band Disable		Forces output to always be enabled regardless of Side Band shift register value	RW	0	CK[10]
3	Mask11 - Masks off Side Band Disable	Side Band shift register may		RW	0	CK[11]
4	Mask12 - Masks off Side Band Disable	disable the output		RW	0	CK[12]
5	Mask13 - Masks off Side Band Disable			RW	0	CK[13]
6	Mask14 - Masks off Side Band Disable			RW	0	CK[14]
7	Mask15 - Masks off Side Band Disable			RW	0	CK[15]

TABLE 4-10: BYTE 10: SBI MASK (FUNCTIONAL ONLY WHEN SBEN = 1)

Bit	Description	If Bit = 0	If Bit = 1	Туре	Default	Output(s) Affected
0	Mask16 - Masks off Side Band Disable			RW	0	CK[16]
1	Mask17 - Masks off Side Band Disable		Forces output to always be enabled regardless of Side Band	RW	0	CK[17]
2	Mask18 - Masks off Side Band Disable	Side Band shift register may		RW	0	CK[18]
3	Mask19 - Masks off Side Band Disable	disable the output		RW	0	CK[19]
4	Reserved		shift register value	RW	0	_
5	Reserved			RW	0	_
6	Reserved			RW	0	_
7	Reserved			RW	0	_

# **ZL40294**

NOTES:

# 5.0 AC AND DC ELECTRICAL CHARACTERISTICS

# 5.1 Absolute Maximum Ratings

TABLE 5-1: ABSOLUTE MAXIMUM RATINGS\*

Bit	Parameter	Symbol	Min.	Max.	Units	Notes
1	3.3V Core Supply Voltage	$V_{DD\_A}$	_	4.6	V	Note 3
2	3.3V I/O Supply Voltage	$V_{DD}$	_	4.6	V	Note 3
5	3.3V Input High Voltage	V <sub>IH</sub>	_	4.6	V	Note 1, Note 3
_	3.3V Input Low Voltage	V <sub>IL</sub>	-0.5	_	V	Note 3
_	Storage Temperature	T <sub>S</sub>	-65	150	°C	Note 3
6	Input ESD Protection	V <sub>DD-IN</sub>	_	_	V	Note 2

<sup>\*</sup> Exceeding these values may cause permanent damage.

Note 1: Maximum VIH is not to exceed maximum VDD.

2: Human body model.

3: Consult manufacturer regarding extended operation in excess of normal DC operating parameters.

# 5.2 DC Electrical Specification

TABLE 5-2: DC OPERATING CHARACTERISTICS\*

Bit	Parameter	Symbol	Min.	Тур.	Max.	Units	Notes
1	3.3V Core Supply Voltage	V <sub>DD_A</sub>	3.135	3.3	3.465	V	_
2	3.3V I/O Supply Voltage	$V_{DD}$	3.135	3.3	3.465	V	_
3	3.3V Input High Voltage	V <sub>IH</sub>	2.0	_	VDD+0.3	V	_
4	3.3V Input Low Voltage	V <sub>IL</sub>	VSS-0.3	_	0.8	V	_
5	Input Leakage Curent	I <sub>IL</sub>	<b>-</b> 5	_	+5	μA	_
6	Input Low Voltage, 3-Level CMOS Input	V <sub>IL3</sub>	VSS-0.3	_	0.9	V	Note 1
7	Input Midrange Voltage, 3-Level CMOS Input	V <sub>IM3</sub>	1.3	_	1.8	V	Note 1
8	Input High Voltage, 3-Level CMOS Input	V <sub>IH3</sub>	2.4	_	VDD	V	Note 1
9	Input Capacitance	C <sub>IN</sub>	_	_	4.5	pF	Note 2
10	Output Capacitance	C <sub>OUT</sub>	_	_	4.5	pF	Note 2
11	Ambient Temperature	T <sub>A</sub>	-40	_	85	°C	_

<sup>\*</sup> Voltages are with respect to ground (GND) unless otherwise stated.

2: For parasitic simulation, use IBIS model.

<sup>\*</sup> Functional operation under these conditions is not implied.

<sup>\*</sup> Voltages are with respect to ground (GND) unless otherwise stated.

Note 1: Spec assumption based on Intel DB1900Z specification revision 1.5 Table 3-2. VIL3,min changed from 0 to VSS-0.3 to match DB2000QL VIL,min spec.

TABLE 5-3: DIFFERENTIAL DC OUTPUT CHARACTERISTICS\*

Bit	Parameter	Symbol	Min.	Тур.	Max.	Units	Notes
1	Maximum Voltage (Overshoot)	V <sub>OVS</sub>	_	_	V <sub>HIGH</sub> + 75	mV	Note 1
2	Maximum Voltage (Undershoot)	V <sub>UDS</sub>	_	_	V <sub>LOW</sub> – 75	mV	Note 1
3	Voltage High	V <sub>HIGH</sub>	225	_	270	mV	Note 1
4	Voltage Low	$V_{LOW}$	10	_	150	mV	Note 1
5	Absolute Crossing Point Voltages	V <sub>CROSS_ABSOLUTE</sub>	130	_	200	mV	Note 1
6	Relative Crossing Point Voltages	V <sub>CROSS_RELATIVE</sub>	_	_	35	mV	Note 1
7	Output Buffer Differential Impedance	Z <sub>DIFF</sub>	85 – 5%	_	85+5%	Ω	Note 2
8	Output Buffer Differential Impedance	Z <sub>DIFF_CROSSING</sub>	85 – 20%	_	85+20%	Ω	Note 3

<sup>\*</sup> Voltages are with respect to ground (GND) unless otherwise stated.

Note 1: Measured into DC test load, see Figure 5-1.

2: Measured at VOL / VOH.

3: Measured during a transition.

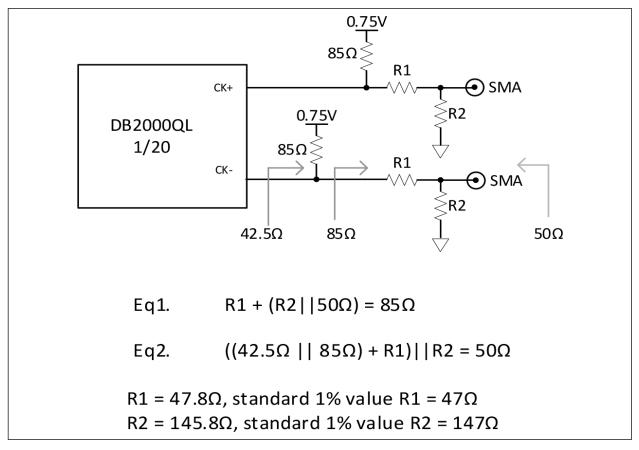


FIGURE 5-1: DC TEST LOAD (AS PER DB2000QL).

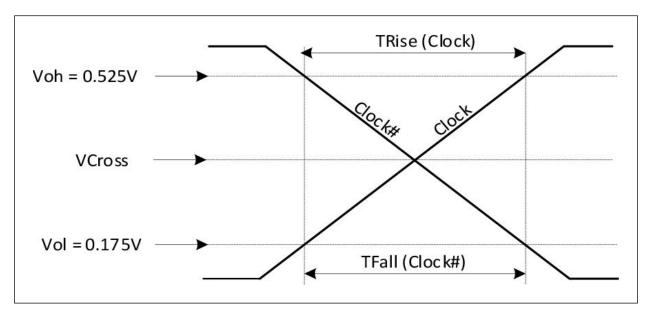


FIGURE 5-2: SINGLE-ENDED MEASUREMENT POINTS FOR  $T_{RISE}$  AND  $T_{FALL}$ .

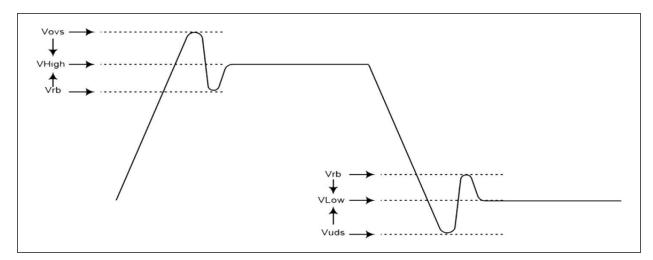


FIGURE 5-3: SINGLE-ENDED MEASUREMENT POINTS FOR V<sub>OVS</sub>, V<sub>UDS</sub>, V<sub>RB</sub>.

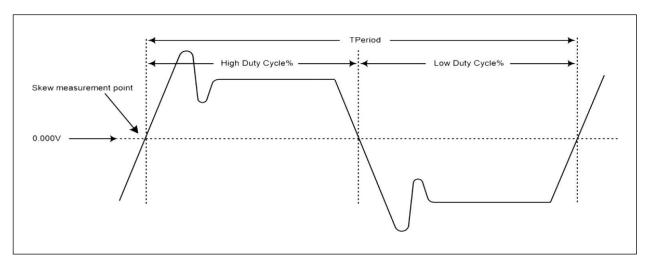


FIGURE 5-4: DIFFERENTIAL (CK, CK#) MEASUREMENT POINTS.

# 5.3 AC Electrical Specification

TABLE 5-4: POWER NOISE TOLERANCE\*

	VDD Electrical Noise Range	Symbol	Min.	Тур.	Max.	Units	Notes
1	f <sub>NOISE</sub> = 12 kHz to 20 MHz	N <sub>VDD_MID</sub>	100	_	_	mV,p-p	Note 1, Note 2
2	f <sub>NOISE</sub> > 20 MHz	N <sub>VDD_HIGH</sub>	50	_	_	mV,p-p	Note 1, Note 2
3	f <sub>NOISE</sub> = 12 kHZ to 20 MHz	N <sub>VDD_A_MID</sub>	40	_	_	mV,p-p	Note 1, Note 2
4	f <sub>NOISE</sub> > 20 MHz	N <sub>VDD A HIGH</sub>	20	_	_	mV,p-p	Note 1, Note 2

<sup>\*</sup> The device meets all specification in the presence of noise specified in this table.

Note 1: Jitter and electrical characteristics are met with specified AC noise present on any of the power pins.

2: Over the specified frequency range, a single sinusoid tone should be assumed swept as the worst case.

TABLE 5-5: SKEW AND JITTER

	Parameter	Symbol	Min.	Тур.	Max.	Units	Notes
1	Input-to-Output Delay	I/O <sub>DELAY</sub>	1.1	1.3	1.5	ns	Note 1, Note 3
2	Input-to-Output Delay Variation as Temperature Changes from –40°C to +85°C	I/O <sub>DELAY</sub> Δ(T)	0.13	0.14	0.15	ns	Note 1, Note 3
3	Output-to-Output Skew	O/O <sub>DELAY</sub>	_	_	50	ps	Note 1, Note 2
4	RMS Additive Jitter as per DB2000QL Spec	AJ <sub>RMS</sub>	_	14	17	fs RMS	Note 1, Note 2, Note 4
5	Peak-to-Peak Additive Jitter	p-pAJ <sub>RMS</sub>	_	_	1	ps	Note 1, Note 2
6	Additive Jitter as per PCIe 1.0 (1.5 MHz to 22 MHz)	T <sub>jPCle_1.0</sub>	_	0.73	0.90	ps pp	Note 1, Note 2
7	Additive Jitter as per PCIe 2.0 high band (1.5 MHz to 50 MHz)	T <sub>jPCle_2.0_high</sub>	_	82	100	fs RMS	Note 1, Note 2

Note 1: Measured into AC test load as per Figure 5-5.

- 2: Measured from differential crossing point to differential crossing point.
- 3: Input-to-output specs refer to the timing between an input edge and the specific output edge created by it.
- **4:** Integrated after the measurement filter. See Intel DB2000QL specification Jitter Measurement section for the measurement filter details.

TABLE 5-5: SKEW AND JITTER (CONTINUED)

	Parameter	Symbol	Min.	Тур.	Max.	Units	Notes
8	Additive Jitter as per PCIe 2.0 low band (10 kHz to 1.5 MHz)	T <sub>jPCle_2.0_low</sub>	_	20	30	fs RMS	Note 1, Note 2
9	Additive Jitter as per PCIe 2.0 low band (5 MHz to 16 MHz)	T <sub>jPCle_2.0_mid</sub>	_	64	79	fs RMS	Note 1, Note 2
10	Additive Jitter as per PCIe 3.0 (PLL_BW = 2 MHz to 5 MHz, CDR = 10 MHz)	T <sub>jPCle_3.0</sub>	_	20	25	fs RMS	Note 1, Note 2
11	Additive Jitter as per PCIe 4.0 (PLL_BW = 2 MHz to 5 MHz, CDR = 10 MHz)	T <sub>jPCle_4.0</sub>	_	20	25	fs RMS	Note 1, Note 2
12	Additive Jitter as per PCle 5.0 (PLL_BW = 0.5 MHz to 1.8 MHz, CDR for 32 GT/s CC)	T <sub>jPCle_5.0</sub>	_	8	10	fs RMS	Note 1, Note 2
13	Additive Jitter as per PCle 6.0 (PLL_BW = 0.5 MHz to 1.0 MHz, CDR for 64 GT/s CC)	T <sub>jPCle_6.0</sub>	_	5	6.5	fs RMS	Note 1, Note 2
14	Additive Jitter as per Intel QPI 9.6 Gbps	T <sub>jQPI</sub>	_	29	36	fs RMS	Note 1, Note 2
15	Additive RMS Jitter in 1 MHz to	т	_	55	67	fs RMS	Note 1, Note 2 (100 MHz clock)
15	20 MHz band	T <sub>j_1M_20M</sub>	_	42	51	fs RMS	Note 1, Note 2 (133 MHz clock)
16	Additive RMS Jitter in 12 kHz	т	_	58	71	fs RMS	Note 1, Note 2 (100 MHz clock)
10	to 20 MHz band	T <sub>j_12k_20M</sub>	_	44	67	fs RMS	Note 1, Note 2 (133 MHz clock)
17	Noise Floor	N	_	-163	-162	dBc/Hz	Note 1, Note 2 (100 MHz clock)
17	NOISE FIOOI	N <sub>F</sub>	_	-163	-162	dBc/Hz	Note 1, Note 2 (133 MHz clock)

- **Note 1:** Measured into AC test load as per Figure 5-5.
  - 2: Measured from differential crossing point to differential crossing point.
  - 3: Input-to-output specs refer to the timing between an input edge and the specific output edge created by it.
  - **4:** Integrated after the measurement filter. See Intel DB2000QL specification Jitter Measurement section for the measurement filter details.

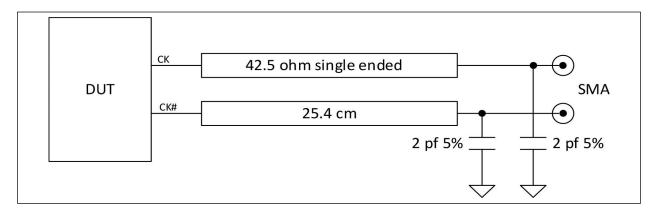


FIGURE 5-5: AC TEST LOAD AS PER DB2000QL SPECIFICATIONS.

TABLE 5-6: DIFFERENTIAL OUTPUT CLOCK AC CHARACTERISTICS

	Parameter	Symbol	Min.	Тур.	Max.	Units	Notes
1	Clock Stabilization Time from PWRGD	T <sub>STAB</sub>	_	_	1.8	ms	Note 1
2	Edge Rate at V <sub>CROSS</sub>	EDGE_RATE	2	_	20	V/ns	Note 1
3	Slew Rate at V <sub>CROSS</sub>	RISE_MATCHING/ FALL_MATCHING	_	_	20%	V	Note 1

Note 1: Measured into Figure 5-5 AC test load.

#### TABLE 5-7: DIFFERENTIAL INPUT CLOCK AC CHARACTERISTICS

	Parameter	Symbol	Min.	Тур.	Max.	Units
1	Edge Rate	Input_Slew_Rate	0.7	_	_	V/ns
2	Total Variation of V <sub>CROSS</sub> over All Edges	Total_Δ_Vcross	_	_	140	mV
3	Input Voltage	Input_Voltage	200	_	_	mv diff

#### TABLE 5-8: CURRENT CONSUMPTION

	Parameter	Parameter Condition	Symbol	Min.	Тур.	Max.	Units	Notes
1		fIN = 100 MHz, All CKx/ CKx# outputs enabled	- I <sub>DDPG</sub>	_	205	240		Note 1, Note 2
2	Active Mode Supply Current	fIN = 100 MHz, All CKx/ CKx# outputs disabled		_	48	55	- mA	Note 1, Note 3
3		fIN = 133 MHz, All CKx/ CKx# outputs enabled		_	257	308		Note 1, Note 2
4		fIN = 133 MHz, All CKx/ CKx# outputs disabled		_	49	56		Note 1, Note 3
5	Power Down Mode fIN = 100 MHz			_	21	25	mA	Note 1, Note 4
6	Supply Current	fIN = 133 MHz	IDDPD	_	22	26		Note 1, Note 4

**Note 1:** VDD = 3.3V + 5%.

- 2: Device operating in active mode (Pin PWRGD/PWRDN# = 1) with all 20 CKx/CKx# outputs enabled (all OE\_xN pin = 0, all OCR1, OCR2, OCR3 register OEx bits = 1).
- **3:** Device operating in active mode (Pin PWRGD/PWRDN# = 1) with all 20 CKx/CKx# outputs disabled (all OCR1, OCR2, OCR3 register OEx bits = 0).
- 4: Device operating in low power mode (Pin PWRGD/PWRDN# = 0).

#### TABLE 5-9: SMBUS ELECTRICAL CHARACTERISTICS

	Parameter	Symbol	Min.	Тур.	Max.	Units	Notes
1	Nominal Bus Voltage	V <sub>DD(SMB)</sub>	2.7	_	5.5	V	1
2	Input Low Voltage	V <sub>IL</sub>	_	_	0.8	V	_
3	Input High Voltage	V <sub>IH</sub>	2.1	_	V <sub>DD(SMB)</sub>	V	_
4	Output Low Voltage	V <sub>OL</sub>	_	_	0.4	V	At I <sub>PULLUP(MAX)</sub>

**Note 1:** 3V to 5V ±10%.

- 2: Rise and fall time is defined as follows: TR =  $(V_{IL(MAX)} 0.15)$  to  $(V_{IH(MIN)} + 0.15)$ , TF =  $(V_{IH(MIN)} + 0.15)$  to  $(V_{IL(MAX)} 0.15)$ .
- 3: Devices must provide a means to reject noise spikes of a duration up to the maximum specified value.

TABLE 5-9: SMBUS ELECTRICAL CHARACTERISTICS (CONTINUED)

	Parameter	Symbol	Min.	Тур.	Max.	Units	Notes
5	Input Leakage Current	I <sub>LEAK</sub>	_	_	±10	μA	_
6	Current Sinking at V <sub>OL(MAX)</sub>	I <sub>PULLUP</sub>	4	_	_	mA	_
7	Pin Capacitive Load	C <sub>I</sub>	_	_	10	pF	
8	Signal Noise Immunity from 10 MHz to 100 MHz	$V_{NOISE}$	300		_	mV <sub>p-p</sub>	
9	Noise Spike Suppression Time	T <sub>SPIKE</sub>	0	_	50	ns	3
10	SMBus Operating Frequency	F <sub>SMB</sub>	10	_	400	kHz	
11	Bus Free Time Between Stop and Start Condition	T <sub>BUF</sub>	4.7	_	_	μs	1
12	Hold Time After (Repeated) Start Condition	T <sub>HD-STA</sub>	4.0		_	μs	After this period, the first clock is generated.
13	Repeated Start Condition Setup Time	T <sub>SU-STA</sub>	4.7	_	_	μs	_
14	Stop Condition Setup Time	T <sub>SU-STO</sub>	4.0	_	_	μs	_
15	Data Hold Time	$T_{HD\text{-}DAT}$	300	_		ns	
16	Data Setup Time	T <sub>SU-DAT</sub>	250	_	_	ns	
17	Clock Low Period	$T_{LOW}$	4.7	_	_	μs	
18	Clock High Period	T <sub>HIGH</sub>	4.0	_	50	μs	_
19	Clock/Data Fall Time	$T_F$	_	_	300	ns	2
20	Clock/Data Rise Time	$T_{R}$	_	_	1000	ns	2

**Note 1:** 3V to 5V ±10%.

- 2: Rise and fall time is defined as follows: TR =  $(V_{IL(MAX)} 0.15)$  to  $(V_{IH(MIN)} + 0.15)$ , TF =  $(V_{IH(MIN)} + 0.15)$  to  $(V_{IL(MAX)} 0.15)$ .
- 3: Devices must provide a means to reject noise spikes of a duration up to the maximum specified value.

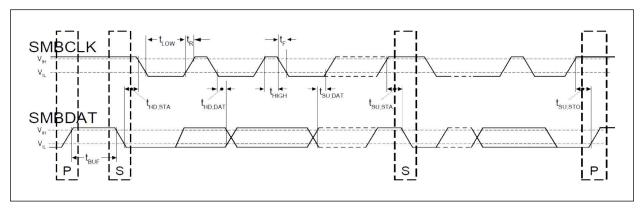


FIGURE 5-6: SMBUS TIMING.

# 5.4 Side-Band Interface Characteristics

TABLE 5-10: SIDE-BAND INTERFACE

	Parameter	Symbol	Min.	Тур.	Max.	Units	Notes	
1	Clock Period	t <sub>PERIOD</sub>	40	_	_	ns	_	
2	vSHFT_LD# setup to vCLK ris- ing edge	t <sub>SETUP</sub>	15	_	_	ns	_	
3	vDATA setup to vCLK rising edge	t <sub>DSU</sub>	10	_	_	ns	_	
4	vDATA hold after vCLK rising edge	t <sub>DSHOLD</sub>	5	_	_	ns	_	
5	Delay from vCLK rising edge to vSHFT_LD# falling edge	t <sub>DELAY</sub>	15	_	_	ns	_	
6	Delay from vSHFT_LD# falling edge to next output configura- tion taking effect	t <sub>PD</sub>	4	_	10	clocks	Note 1	
7	Slew Rate	t <sub>SLEW</sub>	0.5	_	_	V/ns	_	
Note	lote 1: Refers to device differential input clock							

Note 1: Refers to device differential input clock.

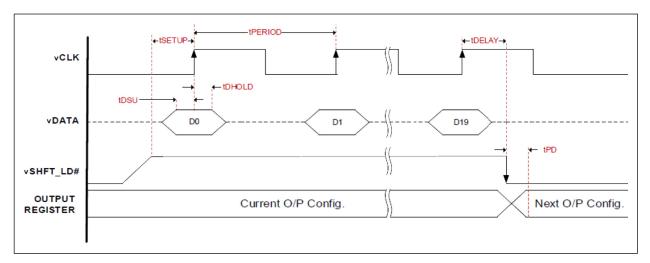


FIGURE 5-7: SIDE-BAND INTERFACE TIMING.

TABLE 5-11: 6 MM × 6 MM GQFN PACKAGE THERMAL PROPERTIES

Parameter	Symbol	Conditions	Value	Units
Maximum Ambient Temperature	T <sub>A</sub>	_	85	°C
Maximum Junction Temperature	T <sub>J(MAX)</sub>	_	125	°C
Junction to Ambient Thermal Resis- tance (Note 1)		Still air	32.42	
	$\theta_{JA}$	1 m/s airflow	27.67	°C/W
tance (Note 1)		2.5 m/s airflow	26.33	
Junction to Board Thermal Resistance	$\theta_{JB}$	_	12.48	°C/W
Junction to Case Thermal Resistance	$\theta_{JC}$	_	21.33	°C/W
Junction to Pad Thermal Resistance (Note 2)	$\theta_{JT}$	Still air	1.83	°C/W
Junction to Top-Center Thermal Characterization Parameter	ΨЈТ	Still air	0.25	°C/W

**Note 1:** Theta-JA (θ<sub>JA</sub>) is the thermal resistance from junction to ambient when the package is mounted on a 4-layer JEDEC standard test board and dissipating maximum power

<sup>2:</sup> Theta-JP  $(\theta_{JP})$  is the thermal resistance from junction to the center exposed pad on the bottom of the package)

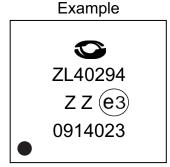
# **ZL40294**

NOTES:

### 6.0 PACKAGING INFORMATION

# 6.1 Package Marking Information





Legend: XX...X Product code customer-specific information Year code (last digit of calendar year) YY Year 2 year) code (last digits of calendar ww January 1 Week '01') code (week of is week NNN Alphanumeric traceability code  $\mathsf{JEDEC}^{\mathbb{R}}$ Pb-free designator for Matte Tin (Sn) (e3) This package is Pb-free. The Pb-free JEDEC designat(e3) can be found on the outer packaging for this package.

•, ▲, ▼ Pin one index is identified by a dot, delta up, or delta down (triangle mark).

**Note**: In the event the full Microchip part number cannot be marked on one line, it will be carried over to the next line, thus limiting the number of available characters for customer-specific information. Package may or may not include the corporate logo.

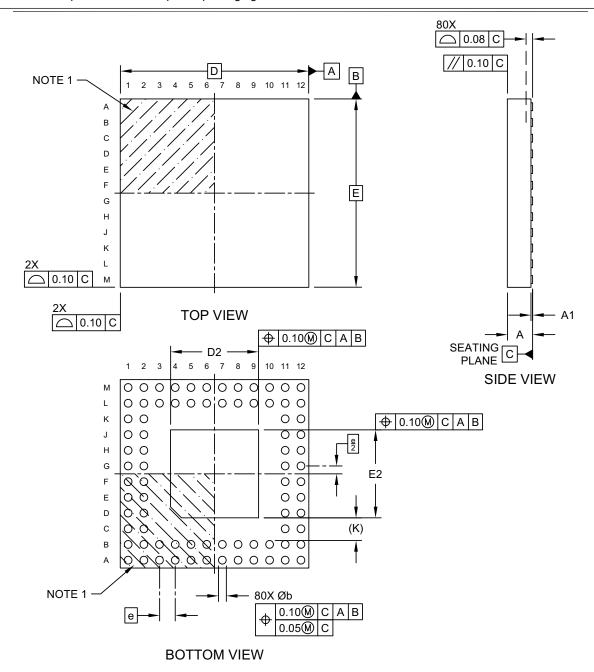
Underbar (\_) and/or Overbar (\_) symbol may not be to scale.

Note: If the full seven-character YYWWNNN code cannot fit on the package, the following truncated codes are used based on the available marking space: 6 Characters = YWWNNN; 5 Characters = WWNNN; 4 Characters = WNNN; 3 Characters = NNN; 2 Characters = NN; 1 Character = N

### 7.0 PACKAGE OUTLINE

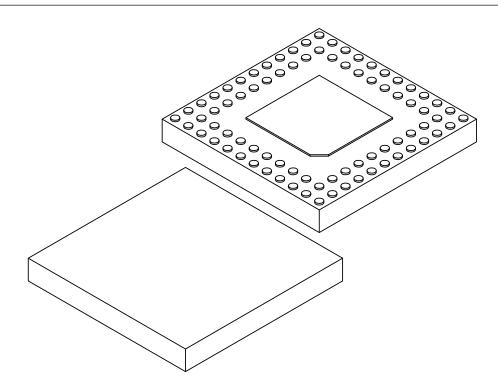
# 80-Lead 6 mm x 6 mm WGQFN Package Outline and Recommended Land Pattern

**Note:** For the most current package drawings, please see the Microchip Packaging Specification located at https://www.microchip.com/packaging/



Microchip Technology Drawing C04-25442 Rev A Sheet 1 of 2

**Note:** For the most current package drawings, please see the Microchip Packaging Specification located at https://www.microchip.com/packaging/



	Units			MILLIMETERS			
Dimension	Dimension Limits			MAX			
Number of Terminals	N		80				
Pitch	е		0.50 BSC				
Overall Height	Α	0.70 0.75 0.80					
Standoff	A1	0.00	0.02	0.05			
Overall Length	D		6.00 BSC				
Exposed Pad Length	D2	2.70	2.80	2.90			
Overall Width	Е		6.00 BSC				
Exposed Pad Width	E2	2.70 2.80 2.90					
Terminal Width	b	b 0.20 0.25 0.30					
Terminal-to-Exposed-Pad	K	0.725 REF					

#### Notes:

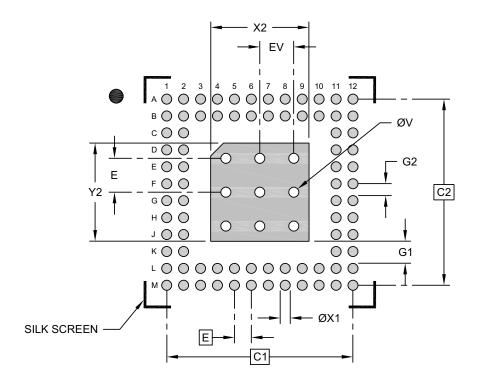
- 1. Terminal A1 visual index feature may vary, but must be located within the hatched area.
- 2. Package is saw singulated
- Package Type WGQFN is not registered with JEDEC as of April, 2019 Dimensioning and tolerancing per ASME Y14.5M

BSC: Basic Dimension. Theoretically exact value shown without tolerances.

REF: Reference Dimension, usually without tolerance, for information purposes only.

Microchip Technology Drawing C04-25442 Rev A Sheet 2 of 2

**Note:** For the most current package drawings, please see the Microchip Packaging Specification located at https://www.microchip.com/packaging/



#### RECOMMENDED LAND PATTERN

	MILLIMETERS			
Dimensior	Dimension Limits			MAX
Contact Pitch	E		0.50 BSC	
Optional Center Pad Width	X2	2.90		
Optional Center Pad Length	Y2	2.90		
Contact Pad Spacing	C1	5.50 BSC		
Contact Pad Spacing	C2		5.50 BSC	
Contact Pad Diameter (X80)	X1			0.30
Contact Pad to Center Pad (X32)	G1	0.65		
Contact Pad to Contact Pad	G2	0.20		
Thermal Via Diameter	V	0.30		
Thermal Via Pitch	EV		1.00	

#### Notes:

- 1. Dimensioning and tolerancing per ASME Y14.5M
  - BSC: Basic Dimension. Theoretically exact value shown without tolerances.
- 2. For best soldering results, thermal vias, if used, should be filled or tented to avoid solder loss during reflow process

Microchip Technology Drawing C04-27442 Rev A

# 8.0 ACRONYMS AND ABBREVIATIONS

APLL analog phase locked loop

CML current mode logic
GbE gigabit Ethernet

HCSL high-speed current steering logic
HSTL high-speed transceiver logic

I/O input/output LOS loss of signal

LVDS low-voltage differential signal

LVPECL low-voltage positive emitter-coupled logic

PFD phase/frequency detector

PLL phase locked loop
ppb parts per billion
ppm parts per million
pk-pk peak-to-peak
RMS root-mean-square

RO read-only R/W read/write

SS or SSM spread spectrum modulation

TCXO temperature-compensated crystal oscillator

UI unit interval

UI<sub>PP</sub> or UI<sub>P-P</sub> unit interval, peak to peak

XO crystal oscillator

# **ZL40294**

# APPENDIX A: DATA SHEET REVISION HISTORY

# **TABLE A-1: REVISION HISTORY**

Revision	Section/Figure/Entry	Correction
DS20006768A (02-09-23)	_	Converted Microsemi data sheet ZL40294 to Microchip DS20006768A. Minor text changes throughout.
DS20006768B (05-12-23)	_	Added PCIe 6.0 compliance to data sheet. Corrected detail in Table 5-10.

# APPENDIX B: PRODUCT IDENTIFICATION SYSTEM

To order or obtain information, e.g., on pricing or delivery, refer to the factory or the listed sales office.

<b>Device</b> Part C  Number	<b>X</b> Chip Carrier Type	<u>X</u> Package	<b>X</b> Media Type	<u>X</u> Finish	a) ZL402	
Device:	ZL40294: 20-Output DB2000QL Buffer with Ultra-Low Additive Jitter				b) ZL402	294LI
Chip Carrier Type:	L = Leadles	s Chip Carrier				
Package:	D = 80-Lead 6 mm × 6 mm WGQFN				Note 1:	Tap
Media Type:	G = 490/Tray F = 4000/Reel					cat ide is r wit
Finish:	6 = Pb-Free	9				pa

DG6:

ZL40294, 20-Output DB2000QL Buffer with Ultra-Low Additive Jitter, Leadless Chip Carrier, 80-Lead WGQFN, 490/Tray, Pb-Free

DF6:

ZL40294, 20-Output DB2000QL Buffer with Ultra-Low Additive Jitter, Leadless Chip Carrier, 80-Lead WGQFN, 4000/Reel, Pb-Free

pe and Reel identifier only appears in the abe and Reen the finite only appears in the attalog part number description. This dentifier is used for ordering purposes and sonot printed on the device package. Check with your Microchip Sales Office for ackage availability with the Tape and Reel paties. otion.

# **ZL40294**

NOTES:

#### Note the following details of the code protection feature on Microchip products:

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