CY7C1565KV18

72-Mbit QDR® II+ SRAM Four-Word Burst Architecture (2.5 Cycle Read Latency)

Features

- Separate independent read and write data ports
  - Supports concurrent transactions
- 550-MHz clock for high bandwidth
- Four-word burst for reducing address bus frequency
- Double data rate (DDR) Interfaces on both read and write ports (data transferred at 1100 MHz) at 550 MHz
- Available in 2.5-clock cycle latency
- Two input clocks (K and $\overline{K}$) for precise DDR timing
  - SRAM uses rising edges only
- Echo clocks (CQ and $\overline{CQ}$) simplify data capture in high speed systems
- Data valid pin (QVLD) to indicate valid data on the output
- Single multiplexed address input bus latches address inputs for read and write ports
- Separate port selects for depth expansion
- Synchronous internally self-timed writes
- Quad data rate (QDR®) II+ operates with 2.5-cycle read latency when $\overline{DOFF}$ is asserted HIGH
- Operates similar to QDR I device with one cycle read latency when $\overline{DOFF}$ is asserted LOW
- Available in × 36 configurations
- Full data coherency, providing most current data
- Core V_{DD} = 1.8 V± 0.1 V; I/O V_{DDQ} = 1.4 V to V_{DD} [1]
  - Supports both 1.5 V and 1.8 V I/O supply
- High-speed transceiver logic (HSTL) inputs and variable drive HSTL output buffers
- Available in 165-ball fine pitch ball grid array (FBGA) package (13 × 15 × 1.4 mm)
- Offered in both Pb-free and non Pb-free packages
- JTAG 1149.1 compatible test access port
- Phase-locked loop (PLL) for accurate data placement

Configurations

With Read Cycle Latency of 2.5 cycles

CY7C1565KV18: 2M × 36

Functional Description

The CY7C1565KV18 is1.8-V synchronous pipelined SRAM, equipped with QDR II+ architecture. Similar to QDR II architecture, QDR II+ architecture consists of two separate ports: the read port and the write port to access the memory array. The read port has dedicated data outputs to support read operations and the write port has dedicated data inputs to support write operations. QDR II+ architecture has separate data inputs and data outputs to completely eliminate the need to “turnaround” the data bus that exists with common I/O devices. Each port is accessed through a common address bus. Addresses for read and write addresses are latched on alternate rising edges of the input (K) clock. Accesses to the QDR II+ read and write ports are completely independent of one another. To maximize data throughput, both read and write ports are equipped with DDR interfaces. Each address location is associated with four 36-bit words (CY7C1565KV18) that burst sequentially into or out of the device. Because data is transferred into and out of the device on every rising edge of both input clocks (K and $\overline{K}$), memory bandwidth is maximized while simplifying system design by eliminating bus “turnarounds”.

Depth expansion is accomplished with port selects, which enables each port to operate independently.

All synchronous inputs pass through input registers controlled by the K or $\overline{K}$ input clocks. All data outputs pass through output registers controlled by the K or $\overline{K}$ input clocks. Writes are conducted with on-chip synchronous self-timed write circuitry.

For a complete list of related documentation, click here.

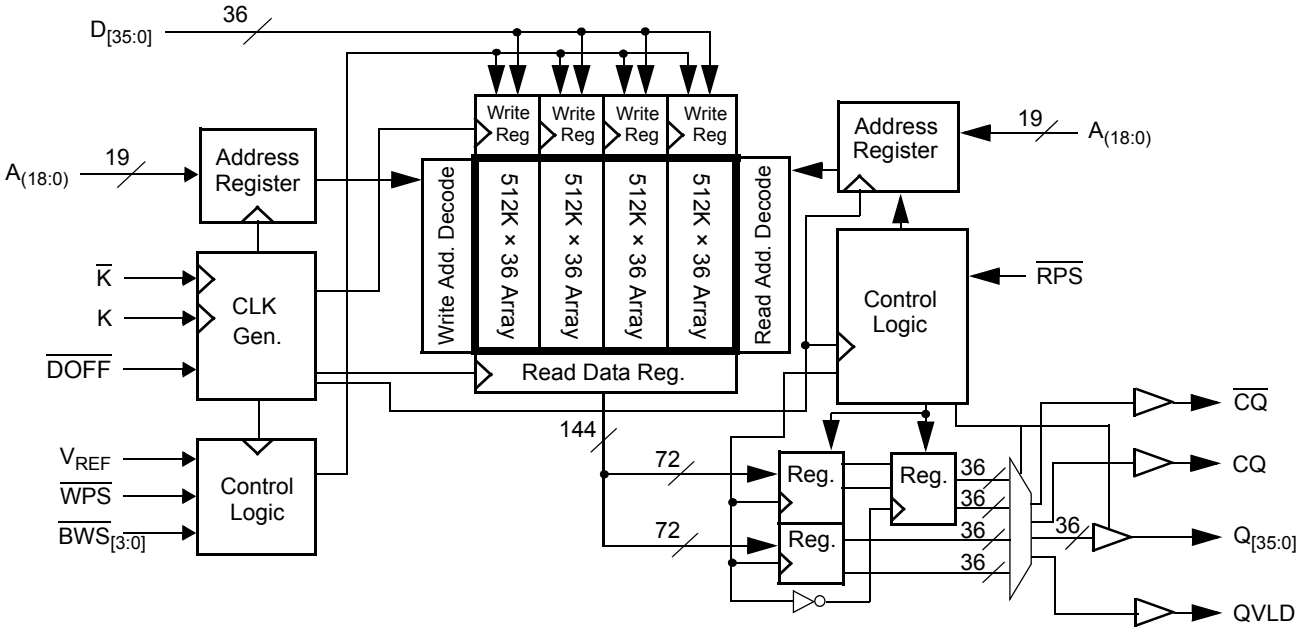
Selection Guide

Description		550 MHz	500 MHz	450 MHz	400 MHz	Unit
Maximum operating frequency		550	500	450	400	MHz
Maximum operating current	× 36	1310	1210	1100	1000	mA

Note
1. The Cypress QDR II+ devices surpass the QDR consortium specification and can support V_{DDQ} = 1.4 V to V_{DD}.

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Logic Block Diagram – CY7C1565KV18



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Pin Configuration

The pin configurations for CY7C1565KV18 follows. [2]

Figure 1. 165-ball FBGA (13 × 15 × 1.4 mm) pinout

CY7C1565KV18 (2M × 36)

	1	2	3	4	5	6	7	8	9	10	11
A	$\overline{CQ}$	NC/288M	A	$\overline{WPS}$	$\overline{BWS_2}$	$\overline{K}$	$\overline{BWS_1}$	$\overline{RPS}$	A	NC/144M	CQ
B	Q27	Q18	D18	A	$\overline{BWS_3}$	K	$\overline{BWS_0}$	A	D17	Q17	Q8
C	D27	Q28	D19	V_{SS}	A	NC	A	V_{SS}	D16	Q7	D8
D	D28	D20	Q19	V_{SS}	V_{SS}	V_{SS}	V_{SS}	V_{SS}	Q16	D15	D7
E	Q29	D29	Q20	V_{DDQ}	V_{SS}	V_{SS}	V_{SS}	V_{DDQ}	Q15	D6	Q6
F	Q30	Q21	D21	V_{DDQ}	V_{DD}	V_{SS}	V_{DD}	V_{DDQ}	D14	Q14	Q5
G	D30	D22	Q22	V_{DDQ}	V_{DD}	V_{SS}	V_{DD}	V_{DDQ}	Q13	D13	D5
H	$\overline{DOFF}$	V_{REF}	V_{DDQ}	V_{DDQ}	V_{DD}	V_{SS}	V_{DD}	V_{DDQ}	V_{DDQ}	V_{REF}	ZQ
J	D31	Q31	D23	V_{DDQ}	V_{DD}	V_{SS}	V_{DD}	V_{DDQ}	D12	Q4	D4
K	Q32	D32	Q23	V_{DDQ}	V_{DD}	V_{SS}	V_{DD}	V_{DDQ}	Q12	D3	Q3
L	Q33	Q24	D24	V_{DDQ}	V_{SS}	V_{SS}	V_{SS}	V_{DDQ}	D11	Q11	Q2
M	D33	Q34	D25	V_{SS}	V_{SS}	V_{SS}	V_{SS}	V_{SS}	D10	Q1	D2
N	D34	D26	Q25	V_{SS}	A	A	A	V_{SS}	Q10	D9	D1
P	Q35	D35	Q26	A	A	QVLD	A	A	Q9	D0	Q0
R	TDO	TCK	A	A	A	NC	A	A	A	TMS	TDI

Note
2. NC/144M and NC/288M are not connected to the die and can be tied to any voltage level.

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Pin Definitions

Pin Name	I/O	Pin Description
$D_{[x:0]}$	Input-Synchronous	**Data input signals**. Sampled on the rising edge of K and $\overline{K}$ clocks when valid write operations are active. CY7C1565KV18 – $D_{[35:0]}$
$\overline{WPS}$	Input-Synchronous	**Write port select – Active LOW**. Sampled on the rising edge of the K clock. When asserted active, a write operation is initiated. Deasserting deselects the write port. Deselecting the write port ignores $D_{[x:0]}$.
$\overline{BWS_0}$, $\overline{BWS_1}$, $\overline{BWS_2}$, $\overline{BWS_3}$	Input-Synchronous	**Byte write select (BWS) 0, 1, 2, and 3 – Active LOW**. Sampled on the rising edge of the K and $\overline{K}$ clocks when write operations are active. Used to select which byte is written into the device during the current portion of the write operations. Bytes not written remain unaltered. CY7C1565KV18 – $\overline{BWS_0}$ controls $D_{[8:0]}$, $\overline{BWS_1}$ controls $D_{[17:9]}$, $\overline{BWS_2}$ controls $D_{[26:18]}$ and $\overline{BWS_3}$ controls $D_{[35:27]}$. All the byte write selects are sampled on the same edge as the data. Deselecting a BWS ignores the corresponding byte of data and it is not written into the device.
A	Input-Synchronous	**Address inputs**. Sampled on the rising edge of the K clock during active read and write operations. These address inputs are multiplexed for both read and write operations. Internally, the device is organized as 2M × 36 (4 arrays each of 512K × 36) for CY7C1565KV18. Therefore, 19 address inputs for CY7C1565KV18. These inputs are ignored when the appropriate port is deselected.
$Q_{[x:0]}$	Outputs-Synchronous	**Data output signals**. These pins drive out the requested data when the read operation is active. Valid data is driven out on the rising edge of the K and $\overline{K}$ clocks during read operations. On deselecting the read port, $Q_{[x:0]}$ are automatically tri-stated. CY7C1565KV18 – $Q_{[35:0]}$
$\overline{RPS}$	Input-Synchronous	**Read port select – Active LOW**. Sampled on the rising edge of positive input clock (K). When active, a read operation is initiated. Deasserting deselects the read port. When deselected, the pending access is allowed to complete and the output drivers are automatically tri-stated following the next rising edge of the $\overline{K}$ clock. Each read access consists of a burst of four sequential transfers.
QVLD	Valid output indicator	**Valid output indicator**. The Q Valid indicates valid output data. QVLD is edge aligned with CQ and $\overline{CQ}$.
K	Input Clock	**Positive input clock input**. The rising edge of K is used to capture synchronous inputs to the device and to drive out data through $Q_{[x:0]}$. All accesses are initiated on the rising edge of K.
$\overline{K}$	Input Clock	**Negative input clock input**. $\overline{K}$ is used to capture synchronous inputs being presented to the device and to drive out data through $Q_{[x:0]}$.
CQ	Echo Clock	**Synchronous echo clock outputs**. This is a free running clock and is synchronized to the input clock (K) of the QDR II+. The timings for the echo clocks are shown in the Switching Characteristics on page 23.
$\overline{CQ}$	Echo Clock	**Synchronous echo clock outputs**. This is a free running clock and is synchronized to the input clock (K) of the QDR II+.The timings for the echo clocks are shown in the Switching Characteristics on page 23.
ZQ	Input	**Output impedance matching input**. This input is used to tune the device outputs to the system data bus impedance. CQ, $\overline{CQ}$, and $Q_{[x:0]}$ output impedance are set to 0.2 × RQ, where RQ is a resistor connected between ZQ and ground. Alternatively, this pin can be connected directly to V_{DDQ}, which enables the minimum impedance mode. This pin cannot be connected directly to GND or left unconnected.
$\overline{DOFF}$	Input	**PLL turn-off – Active LOW**. Connecting this pin to ground turns off the PLL inside the device. The timings in the PLL turned off operation differs from those listed in this datasheet. For normal operation, this pin can be connected to a pull-up through a 10 KΩ or less pull-up resistor. The device behaves in QDR I mode when the PLL is turned off. In this mode, the device can be operated at a frequency of up to 167 MHz with QDR I timing.
TDO	Output	**Test data out (TDO) pin for JTAG**
TCK	Input	**Test clock (TCK) pin for JTAG**
TDI	Input	**Test data in (TDI) pin for JTAG**
TMS	Input	**Test mode select (TMS) pin for JTAG**
NC	N/A	**Not Connected to the die**. Can be tied to any voltage level.

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Pin Definitions (continued)

Pin Name	I/O	Pin Description
NC/144M	N/A	**Not Connected to the die**. Can be tied to any voltage level.
NC/288M	N/A	**Not Connected to the die**. Can be tied to any voltage level.
V_{REF}	Input-Reference	**Reference voltage input**. Static input used to set the reference level for HSTL inputs, outputs, and AC measurement points.
V_{DD}	Power Supply	**Power supply inputs to the core of the device**
V_{SS}	Ground	**Ground for the device**
V_{DDQ}	Power Supply	**Power supply inputs for the outputs of the device**

Functional Overview

The CY7C1565KV18 is synchronous pipelined Burst SRAM equipped with a read port and a write port. The read port is dedicated to read operations and the write port is dedicated to write operations. Data flows into the SRAM through the write port and flows out through the read port. These devices multiplex the address inputs to minimize the number of address pins required. By having separate read and write ports, the QDR II+ completely eliminates the need to “turnaround” the data bus and avoids any possible data contention, thereby simplifying system design. Each access consists of four 36-bit data transfers in the case of CY7C1565KV18, in two clock cycles.

These devices operate with a read latency of two and half cycles when DOFF pin is tied HIGH. When DOFF pin is set LOW or connected to V_{SS} then device behaves in QDR I mode with a read latency of one clock cycle.

Accesses for both ports are initiated on the positive input clock (K). All synchronous input and output timing are referenced from the rising edge of the input clocks (K and K).

All synchronous data inputs ($D_{[x:0]}$) pass through input registers controlled by the input clocks (K and K). All synchronous data outputs ($Q_{[x:0]}$) outputs pass through output registers controlled by the rising edge of the input clocks (K and K) as well.

All synchronous control ($\overline{RPS}$, $\overline{WPS}$, $\overline{BWS}_{[x:0]}$) inputs pass through input registers controlled by the rising edge of the input clocks (K and K).

CY7C1565KV18 is described in the following sections.

Read Operations

The CY7C1565KV18 is organized internally as four arrays of 512K × 36. Accesses are completed in a burst of four sequential 36-bit data words. Read operations are initiated by asserting $\overline{RPS}$ active at the rising edge of the positive input clock (K). The address presented to the address inputs is stored in the read address register. Following the next two K clock rise, the corresponding lowest order 36-bit word of data is driven onto the $Q_{[35:0]}$ using K as the output timing reference. On the subsequent rising edge of K, the next 36-bit data word is driven onto the $Q_{[35:0]}$. This process continues until all four 36-bit data words have been driven out onto $Q_{[35:0]}$. The requested data is valid 0.45 ns from the rising edge of the input clock (K or K). To maintain the internal logic, each read access must be allowed to complete. Each read access consists of four 36-bit data words and takes two clock cycles to complete. Therefore, read accesses to the device can not be initiated on two consecutive K clock rises. The internal logic of the device ignores the second read request. Read accesses can be initiated on every other K clock rise. Doing so pipelines the data flow such that data is transferred out of the device on every rising edge of the input clocks (K and K).

When the read port is deselected, the CY7C1565KV18 first completes the pending read transactions. Synchronous internal circuitry automatically tri-states the outputs following the next rising edge of the negative input clock (K). This enables for a seamless transition between devices without the insertion of wait states in a depth expanded memory.

Write Operations

Write operations are initiated by asserting $\overline{WPS}$ active at the rising edge of the positive input clock (K). On the following K clock rise the data presented to $D_{[35:0]}$ is latched and stored into the lower 36-bit write data register, provided $\overline{BWS}_{[3:0]}$ are both asserted active. On the subsequent rising edge of the negative input clock (K) the information presented to $D_{[35:0]}$ is also stored into the write data register, provided $\overline{BWS}_{[3:0]}$ are both asserted active. This process continues for one more cycle until four 36-bit words (a total of 144 bits) of data are stored in the SRAM. The 144 bits of data are then written into the memory array at the specified location. Therefore, write accesses to the device can not be initiated on two consecutive K clock rises. The internal logic of the device ignores the second write request. Write accesses can be initiated on every other rising edge of the positive input clock (K). Doing so pipelines the data flow such that 36 bits of data can be transferred into the device on every rising edge of the input clocks (K and K).

When deselected, the write port ignores all inputs after the pending write operations have been completed.

Byte Write Operations

Byte write operations are supported by the CY7C1565KV18. A write operation is initiated as described in the Write Operations section. The bytes that are written are determined by $\overline{BWS}_0$, $\overline{BWS}_1$, $\overline{BWS}_2$, and $\overline{BWS}_3$ which are sampled with each set of 36-bit data words. Asserting the appropriate Byte Write Select input during the data portion of a write latches the data being presented and writes it into the device. Deasserting the Byte Write Select input during the data portion of a write enables the data stored in the device for that byte to remain unaltered. This feature can be used to simplify read, modify, or write operations to a byte write operation.

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Concurrent Transactions

The read and write ports on the CY7C1565KV18 operates completely independently of one another. As each port latches the address inputs on different clock edges, the user can read or write to any location, regardless of the transaction on the other port. If the ports access the same location when a read follows a write in successive clock cycles, the SRAM delivers the most recent information associated with the specified address location. This includes forwarding data from a write cycle that was initiated on the previous K clock rise.

Read access and write access must be scheduled such that one transaction is initiated on any clock cycle. If both ports are selected on the same K clock rise, the arbitration depends on the previous state of the SRAM. If both ports are deselected, the read port takes priority. If a read was initiated on the previous cycle, the write port takes priority (as read operations cannot be initiated on consecutive cycles). If a write was initiated on the previous cycle, the read port takes priority (as write operations can not be initiated on consecutive cycles). Therefore, asserting both port selects active from a deselected state results in alternating read or write operations being initiated, with the first access being a read.

Depth Expansion

The CY7C1565KV18 has a port select input for each port. This enables for easy depth expansion. Both port selects are sampled on the rising edge of the positive input clock only (K). Each port select input can deselect the specified port. Deselecting a port does not affect the other port. All pending transactions (read and write) are completed before the device is deselected.

Programmable Impedance

An external resistor, RQ, must be connected between the ZQ pin on the SRAM and V_{SS} to allow the SRAM to adjust its output driver impedance. The value of RQ must be 5X the value of the intended line impedance driven by the SRAM, the allowable range of RQ to guarantee impedance matching with a tolerance of ±15% is between 175 Ω and 350 Ω, with V_{DDQ} = 1.5 V. The output impedance is adjusted every 1024 cycles upon power-up to account for drifts in supply voltage and temperature.

Echo Clocks

Echo clocks are provided on the QDR II+ to simplify data capture on high-speed systems. Two echo clocks are generated by the QDR II+. CQ is referenced with respect to K and $\overline{CQ}$ is referenced with respect to $\overline{K}$. These are free running clocks and are synchronized to the input clock of the QDR II+. The timing for the echo clocks is shown in Switching Characteristics on page 23.

Valid Data Indicator (QVLD)

QVLD is provided on the QDR II+ to simplify data capture on high speed systems. The QVLD is generated by the QDR II+ device along with data output. This signal is also edge-aligned with the echo clock and follows the timing of any data pin. This signal is asserted half a cycle before valid data arrives.

PLL

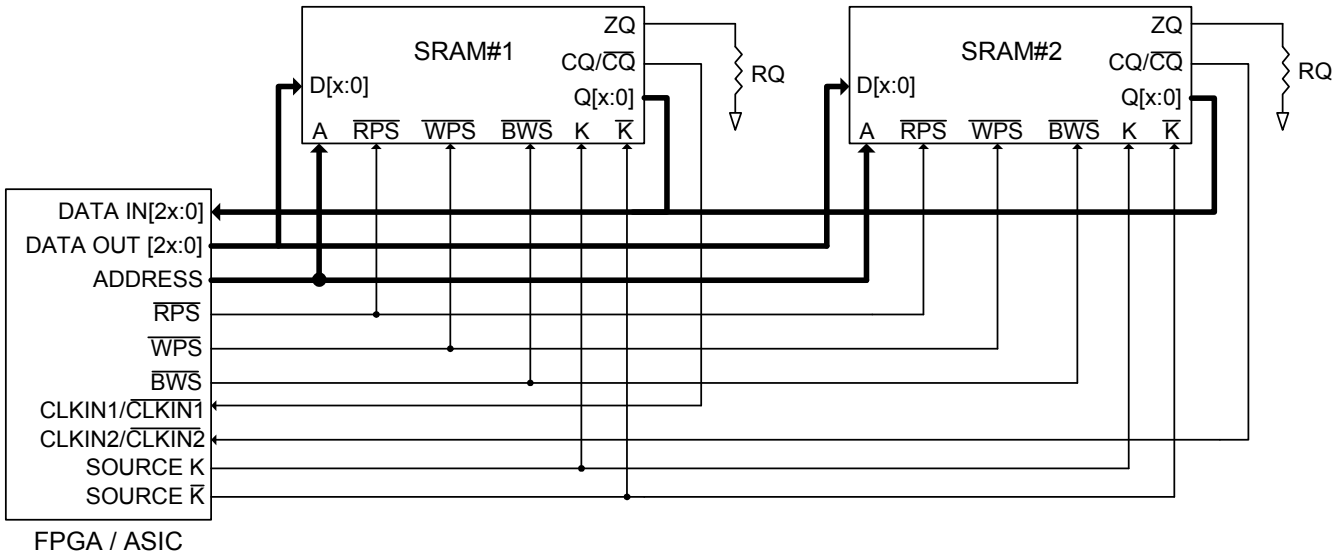
These chips use a PLL that is designed to function between 120 MHz and the specified maximum clock frequency. During power-up, when the $\overline{DOFF}$ is tied HIGH, the PLL is locked after 20 µs of stable clock. The PLL can also be reset by slowing or stopping the input clocks K and $\overline{K}$ for a minimum of 30 ns. However, it is not necessary to reset the PLL to lock to the desired frequency. The PLL automatically locks 20 µs after a stable clock is presented. The PLL may be disabled by applying ground to the $\overline{DOFF}$ pin. When the PLL is turned off, the device behaves in QDR I mode (with one cycle latency and a longer access time). For information, refer to the application note, *PLL Considerations in QDRII/DDRII/QDRII+/DDRII+*.

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Application Example

Figure 2 shows two QDR II+ used in an application.

Figure 2. Application Example (Width Expansion)



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Truth Table

The truth table for CY7C1565KV18 follows. [3, 4, 5, 6, 7, 8]

Operation	K	$\overline{RPS}$	$\overline{WPS}$	DQ	DQ	DQ	DQ
Write cycle: Load address on the rising edge of K; input write data on two consecutive K and $\overline{K}$ rising edges.	L–H	H [9]	L [10]	D(A) at K(t + 1)↑	D(A + 1) at $\overline{K}$(t + 1)↑	D(A + 2) at K(t + 2)↑	D(A + 3) at $\overline{K}$(t + 2)↑
Read cycle: (2.5 cycle Latency) Load address on the rising edge of K; wait two and half cycles; read data on two consecutive $\overline{K}$ and K rising edges.	L–H	L [10]	×	Q(A) at $\overline{K}$(t + 2)↑	Q(A + 1) at K(t + 3)↑	Q(A + 2) at $\overline{K}$(t + 3)↑	Q(A + 3) at K(t + 4)↑
NOP: No operation	L–H	H	H	D = X Q = High Z	D = X Q = High Z	D = X Q = High Z	D = X Q = High Z
Standby: Clock stopped	Stopped	X	X	Previous state	Previous state	Previous state	Previous State

Notes

3. X = "Don't Care," H = Logic HIGH, L = Logic LOW, ↑represents rising edge.
4. Device powers up deselected with the outputs in a tri-state condition.
5. "A" represents address location latched by the devices when transaction was initiated. A + 1, A + 2, and A + 3 represents the address sequence in the burst.
6. "t" represents the cycle at which a read/write operation is started. t + 1, t + 2, and t + 3 are the first, second and third clock cycles respectively succeeding the "t" clock cycle.
7. Data inputs are registered at K and $\overline{K}$ rising edges. Data outputs are delivered on K and $\overline{K}$ rising edges as well.
8. It is recommended that K = $\overline{K}$ = HIGH when clock is stopped. This is not essential, but permits most rapid restart by overcoming transmission line charging symmetrically.
9. If this signal was LOW to initiate the previous cycle, this signal becomes a "Don't Care" for this operation.
10. This signal was HIGH on previous K clock rise. Initiating consecutive read or write operations on consecutive K clock rises is not permitted. The device ignores the second read or write request.

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Write Cycle Descriptions

The write cycle description table for CY7C1565KV18 follows. [11, 12]

$\overline{BWS_0}$	$\overline{BWS_1}$	$\overline{BWS_2}$	$\overline{BWS_3}$	K	$\overline{K}$	Comments
L	L	L	L	L–H	–	During the data portion of a write sequence, all four bytes ($D_{[35:0]}$) are written into the device.
L	L	L	L	–	L–H	During the data portion of a write sequence, all four bytes ($D_{[35:0]}$) are written into the device.
L	H	H	H	L–H	–	During the data portion of a write sequence, only the lower byte ($D_{[8:0]}$) is written into the device. $D_{[35:9]}$ remains unaltered.
L	H	H	H	–	L–H	During the data portion of a write sequence, only the lower byte ($D_{[8:0]}$) is written into the device. $D_{[35:9]}$ remains unaltered.
H	L	H	H	L–H	–	During the data portion of a write sequence, only the byte ($D_{[17:9]}$) is written into the device. $D_{[8:0]}$ and $D_{[35:18]}$ remains unaltered.
H	L	H	H	–	L–H	During the data portion of a write sequence, only the byte ($D_{[17:9]}$) is written into the device. $D_{[8:0]}$ and $D_{[35:18]}$ remains unaltered.
H	H	L	H	L–H	–	During the data portion of a write sequence, only the byte ($D_{[26:18]}$) is written into the device. $D_{[17:0]}$ and $D_{[35:27]}$ remains unaltered.
H	H	L	H	–	L–H	During the data portion of a write sequence, only the byte ($D_{[26:18]}$) is written into the device. $D_{[17:0]}$ and $D_{[35:27]}$ remains unaltered.
H	H	H	L	L–H	–	During the data portion of a write sequence, only the byte ($D_{[35:27]}$) is written into the device. $D_{[26:0]}$ remains unaltered.
H	H	H	L	–	L–H	During the data portion of a write sequence, only the byte ($D_{[35:27]}$) is written into the device. $D_{[26:0]}$ remains unaltered.
H	H	H	H	L–H	–	No data is written into the device during this portion of a write operation.
H	H	H	H	–	L–H	No data is written into the device during this portion of a write operation.

Notes

11. X = “Don't Care,” H = Logic HIGH, L = Logic LOW, ↑represents rising edge.
12. Is based on a write cycle that was initiated in accordance with Write Cycle Descriptions. $\overline{BWS_0}$, $\overline{BWS_1}$, $\overline{BWS_2}$, and $\overline{BWS_3}$ can be altered on different portions of a write cycle, as long as the setup and hold requirements are achieved.

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IEEE 1149.1 Serial Boundary Scan (JTAG)

These SRAMs incorporate a serial boundary scan Test Access Port (TAP) in the FBGA package. This part is fully compliant with IEEE Standard #1149.1-2001. The TAP operates using JEDEC standard 1.8 V I/O logic levels.

Disabling the JTAG Feature

It is possible to operate the SRAM without using the JTAG feature. To disable the TAP controller, TCK must be tied LOW (V_{SS}) to prevent clocking of the device. TDI and TMS are internally pulled up and may be unconnected. They may alternatively be connected to V_{DD} through a pull-up resistor. TDO must be left unconnected. Upon power-up, the device comes up in a reset state, which does not interfere with the operation of the device.

Test Access Port

Test Clock

The test clock is used only with the TAP controller. All inputs are captured on the rising edge of TCK. All outputs are driven from the falling edge of TCK.

Test Mode Select (TMS)

The TMS input is used to give commands to the TAP controller and is sampled on the rising edge of TCK. This pin may be left unconnected if the TAP is not used. The pin is pulled up internally, resulting in a logic HIGH level.

Test Data-In (TDI)

The TDI pin is used to serially input information into the registers and can be connected to the input of any of the registers. The register between TDI and TDO is chosen by the instruction that is loaded into the TAP instruction register. For information on loading the instruction register, see TAP Controller State Diagram on page 13. TDI is internally pulled up and can be unconnected if the TAP is unused in an application. TDI is connected to the most significant bit (MSB) on any register.

Test Data-Out (TDO)

The TDO output pin is used to serially clock data out from the registers. The output is active, depending upon the current state of the TAP state machine (see Instruction Codes on page 17). The output changes on the falling edge of TCK. TDO is connected to the least significant bit (LSB) of any register.

Performing a TAP Reset

A Reset is performed by forcing TMS HIGH (V_{DD}) for five rising edges of TCK. This Reset does not affect the operation of the SRAM and can be performed while the SRAM is operating. At power-up, the TAP is reset internally to ensure that TDO comes up in a High Z state.

TAP Registers

Registers are connected between the TDI and TDO pins to scan the data in and out of the SRAM test circuitry. Only one register can be selected at a time through the instruction registers. Data is serially loaded into the TDI pin on the rising edge of TCK. Data is output on the TDO pin on the falling edge of TCK.

Instruction Register

Three-bit instructions can be serially loaded into the instruction register. This register is loaded when it is placed between the TDI and TDO pins, as shown in TAP Controller Block Diagram on page 14. Upon power-up, the instruction register is loaded with the IDCODE instruction. It is also loaded with the IDCODE instruction if the controller is placed in a reset state, as described in the previous section.

When the TAP controller is in the Capture-IR state, the two least significant bits are loaded with a binary “01” pattern to allow for fault isolation of the board level serial test path.

Bypass Register

To save time when serially shifting data through registers, it is sometimes advantageous to skip certain chips. The bypass register is a single-bit register that can be placed between TDI and TDO pins. This enables shifting of data through the SRAM with minimal delay. The bypass register is set LOW (V_{SS}) when the BYPASS instruction is executed.

Boundary Scan Register

The boundary scan register is connected to all of the input and output pins on the SRAM. Several No Connect (NC) pins are also included in the scan register to reserve pins for higher density devices.

The boundary scan register is loaded with the contents of the RAM input and output ring when the TAP controller is in the Capture-DR state and is then placed between the TDI and TDO pins when the controller is moved to the Shift-DR state. The EXTEST, SAMPLE/PRELOAD, and SAMPLE Z instructions can be used to capture the contents of the input and output ring.

The section Boundary Scan Order on page 18 shows the order in which the bits are connected. Each bit corresponds to one of the bumps on the SRAM package. The MSB of the register is connected to TDI, and the LSB is connected to TDO.

Identification (ID) Register

The ID register is loaded with a vendor-specific, 32-bit code during the Capture-DR state when the IDCODE command is loaded in the instruction register. The IDCODE is hardwired into the SRAM and can be shifted out when the TAP controller is in the Shift-DR state. The ID register has a vendor code and other information described in Identification Register Definitions on page 17.

TAP Instruction Set

Eight different instructions are possible with the three-bit instruction register. All combinations are listed in Instruction Codes on page 17. Three of these instructions are listed as RESERVED and must not be used. The other five instructions are described in this section in detail.

Instructions are loaded into the TAP controller during the Shift-IR state when the instruction register is placed between TDI and TDO. During this state, instructions are shifted through the instruction register through the TDI and TDO pins. To execute the instruction after it is shifted in, the TAP controller must be moved into the Update-IR state.

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IDCODE

The IDCODE instruction loads a vendor-specific, 32-bit code into the instruction register. It also places the instruction register between the TDI and TDO pins and shifts the IDCODE out of the device when the TAP controller enters the Shift-DR state. The IDCODE instruction is loaded into the instruction register at power-up or whenever the TAP controller is supplied a Test-Logic-Reset state.

SAMPLE Z

The SAMPLE Z instruction connects the boundary scan register between the TDI and TDO pins when the TAP controller is in a Shift-DR state. The SAMPLE Z command puts the output bus into a High Z state until the next command is supplied during the Update IR state.

SAMPLE/PRELOAD

SAMPLE/PRELOAD is a 1149.1 mandatory instruction. When the SAMPLE/PRELOAD instructions are loaded into the instruction register and the TAP controller is in the Capture-DR state, a snapshot of data on the input and output pins is captured in the boundary scan register.

The TAP controller clock can only operate at a frequency up to 20 MHz, while the SRAM clock operates more than an order of magnitude faster. Because there is a large difference in the clock frequencies, it is possible that during the Capture-DR state, an input or output undergoes a transition. The TAP may then try to capture a signal while in transition (metastable state). This does not harm the device, but there is no guarantee as to the value that is captured. Repeatable results may not be possible.

To guarantee that the boundary scan register captures the correct value of a signal, the SRAM signal must be stabilized long enough to meet the TAP controller's capture setup plus hold times (t_{CS} and t_{CH}). The SRAM clock input might not be captured correctly if there is no way in a design to stop (or slow) the clock during a SAMPLE/PRELOAD instruction. If this is an issue, it is still possible to capture all other signals and simply ignore the value of the CK and $\overline{CK}$ captured in the boundary scan register.

After the data is captured, it is possible to shift out the data by putting the TAP into the Shift-DR state. This places the boundary scan register between the TDI and TDO pins.

PRELOAD places an initial data pattern at the latched parallel outputs of the boundary scan register cells before the selection of another boundary scan test operation.

The shifting of data for the SAMPLE and PRELOAD phases can occur concurrently when required, that is, while the data captured is shifted out, the preloaded data can be shifted in.

BYPASS

When the BYPASS instruction is loaded in the instruction register and the TAP is placed in a Shift-DR state, the bypass register is placed between the TDI and TDO pins. The advantage of the BYPASS instruction is that it shortens the boundary scan path when multiple devices are connected together on a board.

EXTEST

The EXTEST instruction drives the preloaded data out through the system output pins. This instruction also connects the boundary scan register for serial access between the TDI and TDO in the Shift-DR controller state.

EXTEST OUTPUT BUS TRI-STATE

IEEE Standard 1149.1 mandates that the TAP controller be able to put the output bus into a tri-state mode.

The boundary scan register has a special bit located at bit #108. When this scan cell, called the “extest output bus tri-state,” is latched into the preload register during the Update-DR state in the TAP controller, it directly controls the state of the output (Q-bus) pins, when the EXTEST is entered as the current instruction. When HIGH, it enables the output buffers to drive the output bus. When LOW, this bit places the output bus into a High Z condition.

This bit can be set by entering the SAMPLE/PRELOAD or EXTEST command, and then shifting the desired bit into that cell, during the Shift-DR state. During Update-DR, the value loaded into that shift-register cell latches into the preload register. When the EXTEST instruction is entered, this bit directly controls the output Q-bus pins. Note that this bit is preset HIGH to enable the output when the device is powered up, and also when the TAP controller is in the Test-Logic-Reset state.

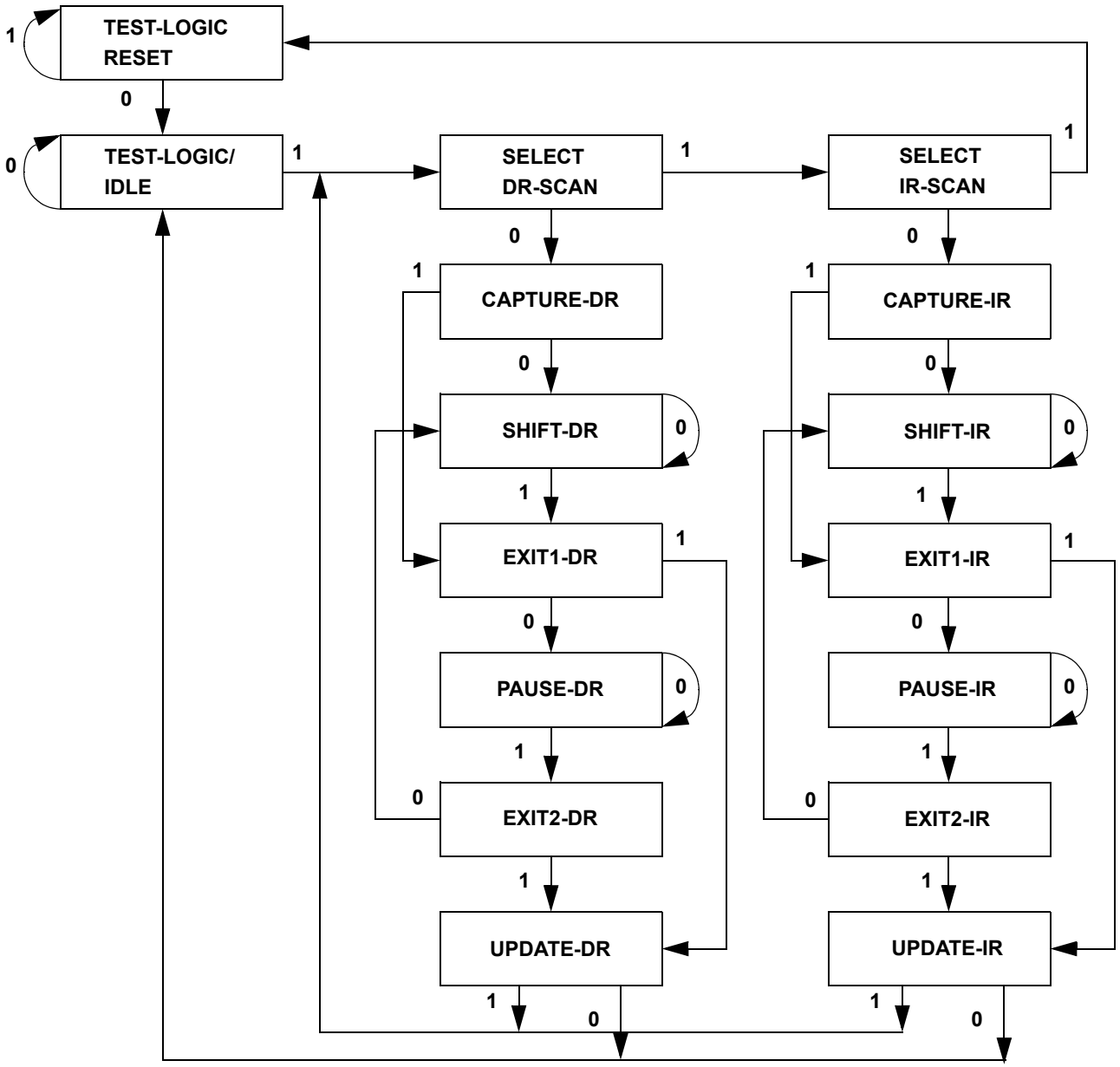
Reserved

These instructions are not implemented but are reserved for future use. Do not use these instructions.

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TAP Controller State Diagram

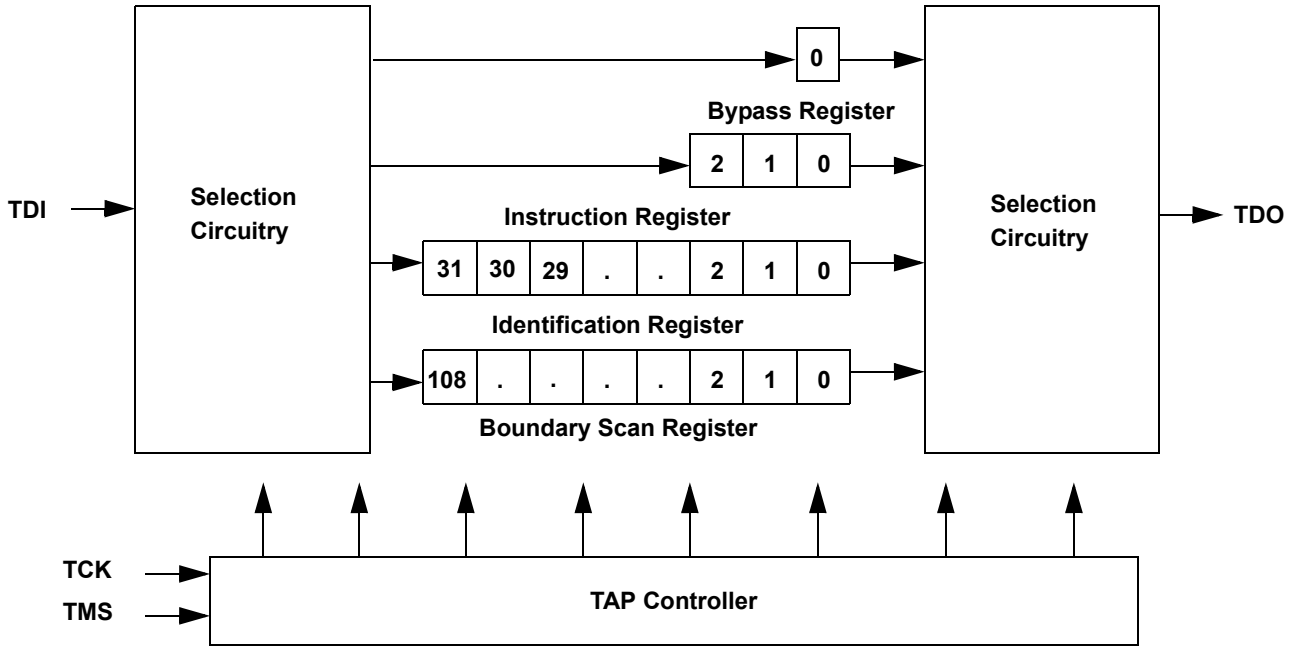
The state diagram for the TAP controller follows. [13]



Note
13. The 0/1 next to each state represents the value at TMS at the rising edge of TCK.

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TAP Controller Block Diagram



TAP Electrical Characteristics

Over the Operating Range

Parameter [14, 15, 16]	Description	Test Conditions	Min	Max	Unit
V_{OH1}	Output HIGH voltage	I_{OH} = –2.0 mA	1.4	–	V
V_{OH2}	Output HIGH voltage	I_{OH} = –100 µA	1.6	–	V
V_{OL1}	Output LOW voltage	I_{OL} = 2.0 mA	–	0.4	V
V_{OL2}	Output LOW voltage	I_{OL} = 100 µA	–	0.2	V
V_{IH}	Input HIGH voltage		0.65 × V_{DD}	V_{DD} + 0.3	V
V_{IL}	Input LOW voltage		–0.3	0.35 × V_{DD}	V
I_X	Input and output load current	GND ≤ V_I ≤ V_{DD}	–5	5	µA

Notes
14. These characteristics pertain to the TAP inputs (TMS, TCK, TDI and TDO). Parallel load levels are specified in the Electrical Characteristics on page 20.
15. Overshoot: $V_{IH}(AC) \leq V_{DDQ} + 0.35V$ (Pulse width less than $t_{CYC}/2$), Undershoot: $V_{IL}(AC) \geq -0.3$ V (Pulse width less than $t_{CYC}/2$).
16. All Voltage referenced to Ground.

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TAP AC Switching Characteristics

Over the Operating Range

Parameter [17, 18]	Description	Min	Max	Unit
t_{TCYC}	TCK clock cycle time	50	–	ns
t_{TF}	TCK clock frequency	–	20	MHz
t_{TH}	TCK clock HIGH	20	–	ns
t_{TL}	TCK clock LOW	20	–	ns
Setup Times				
t_{TMSS}	TMS setup to TCK clock rise	5	–	ns
t_{TDIS}	TDI setup to TCK clock rise	5	–	ns
t_{CS}	Capture setup to TCK rise	5	–	ns
Hold Times				
t_{TMSH}	TMS hold after TCK clock rise	5	–	ns
t_{TDIH}	TDI hold after clock rise	5	–	ns
t_{CH}	Capture hold after clock rise	5	–	ns
Output Times				
t_{TDOV}	TCK clock LOW to TDO valid	–	10	ns
t_{TDOX}	TCK clock LOW to TDO invalid	0	–	ns

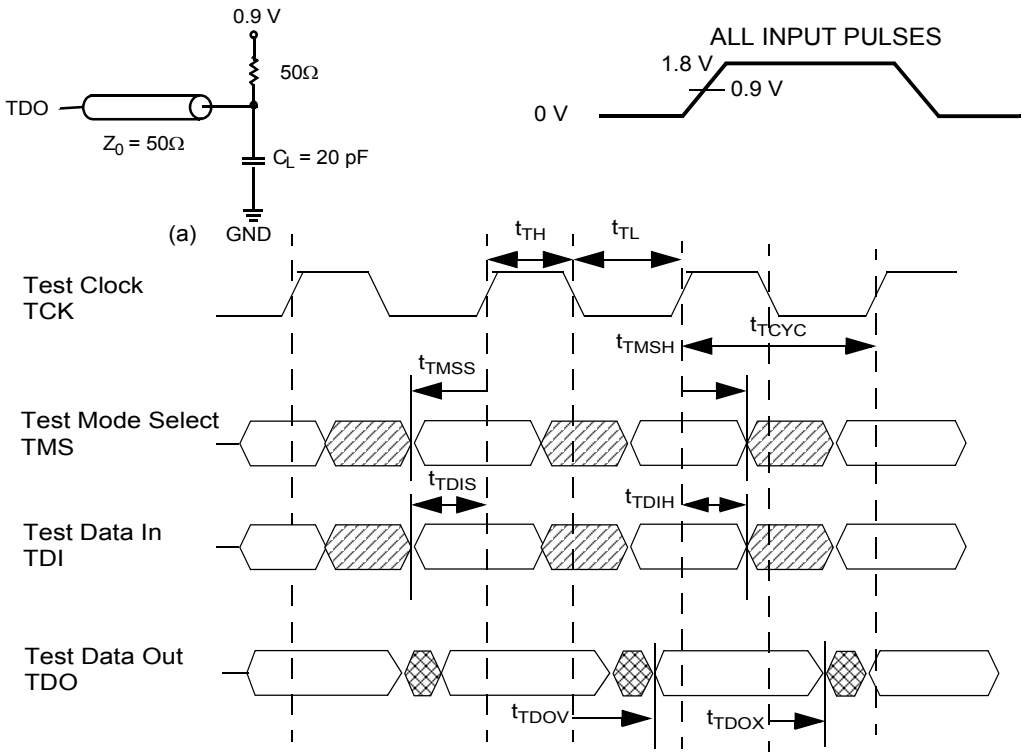
Notes
17. t_{CS} and t_{CH} refer to the setup and hold time requirements of latching data from the boundary scan register.
18. Test conditions are specified using the load in TAP AC Test Conditions. t_R/t_F = 1 ns.

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TAP Timing and Test Conditions

Figure 3 shows the TAP timing and test conditions. [19]

Figure 3. TAP Timing and Test Conditions



Note
19. Test conditions are specified using the load in TAP AC Test Conditions. t_R/t_F = 1 ns.

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Identification Register Definitions

Instruction Field	Value	Description
	CY7C1565KV18	
Revision number (31:29)	000	Version number.
Cypress device ID (28:12)	11010010001100100	Defines the type of SRAM.
Cypress JEDEC ID (11:1)	00000110100	Allows unique identification of SRAM vendor.
ID register presence (0)	1	Indicates the presence of an ID register.

Scan Register Sizes

Register Name	Bit Size
Instruction	3
Bypass	1
ID	32
Boundary scan	109

Instruction Codes

Instruction	Code	Description
EXTEST	000	Captures the input and output ring contents.
IDCODE	001	Loads the ID register with the vendor ID code and places the register between TDI and TDO. This operation does not affect SRAM operation.
SAMPLE Z	010	Captures the input and output contents. Places the boundary scan register between TDI and TDO. Forces all SRAM output drivers to a High Z state.
RESERVED	011	Do Not Use: This instruction is reserved for future use.
SAMPLE/PRELOAD	100	Captures the input and output ring contents. Places the boundary scan register between TDI and TDO. Does not affect the SRAM operation.
RESERVED	101	Do Not Use: This instruction is reserved for future use.
RESERVED	110	Do Not Use: This instruction is reserved for future use.
BYPASS	111	Places the bypass register between TDI and TDO. This operation does not affect SRAM operation.

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Boundary Scan Order

Bit #	Bump ID	Bit #	Bump ID	Bit #	Bump ID	Bit #	Bump ID
0	6R	28	10G	56	6A	84	1J
1	6P	29	9G	57	5B	85	2J
2	6N	30	11F	58	5A	86	3K
3	7P	31	11G	59	4A	87	3J
4	7N	32	9F	60	5C	88	2K
5	7R	33	10F	61	4B	89	1K
6	8R	34	11E	62	3A	90	2L
7	8P	35	10E	63	2A	91	3L
8	9R	36	10D	64	1A	92	1M
9	11P	37	9E	65	2B	93	1L
10	10P	38	10C	66	3B	94	3N
11	10N	39	11D	67	1C	95	3M
12	9P	40	9C	68	1B	96	1N
13	10M	41	9D	69	3D	97	2M
14	11N	42	11B	70	3C	98	3P
15	9M	43	11C	71	1D	99	2N
16	9N	44	9B	72	2C	100	2P
17	11L	45	10B	73	3E	101	1P
18	11M	46	11A	74	2D	102	3R
19	9L	47	10A	75	2E	103	4R
20	10L	48	9A	76	1E	104	4P
21	11K	49	8B	77	2F	105	5P
22	10K	50	7C	78	3F	106	5N
23	9J	51	6C	79	1G	107	5R
24	9K	52	8A	80	1F	108	Internal
25	10J	53	7A	81	3G		
26	11J	54	7B	82	2G		
27	11H	55	6B	83	1H		

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Power-Up Sequence in QDR II+ SRAM

QDR II+ SRAMs must be powered up and initialized in a predefined manner to prevent undefined operations.

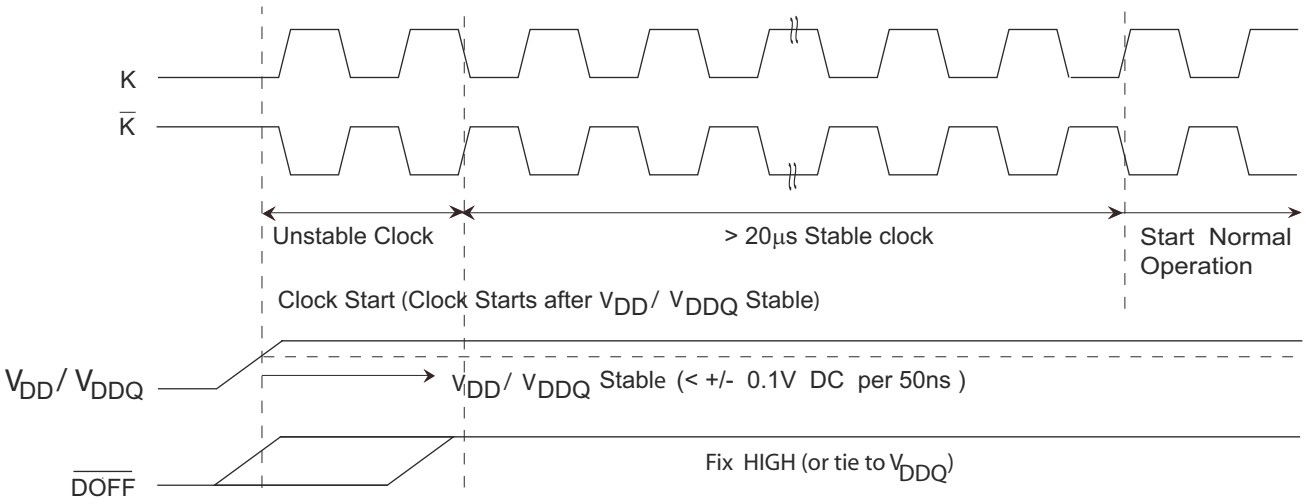
Power-Up Sequence

- Apply power and drive $\overline{DOFF}$ either HIGH or LOW (All other inputs can be HIGH or LOW).
  - Apply V_{DD} before V_{DDQ}.
  - Apply V_{DDQ} before V_{REF} or at the same time as V_{REF}.
  - Drive $\overline{DOFF}$ HIGH.
- Provide stable $\overline{DOFF}$ (HIGH), power and clock (K, $\overline{K}$) for 20 µs to lock the PLL.

PLL Constraints

- PLL uses K clock as its synchronizing input. The input must have low phase jitter, which is specified as $t_{KC\ Var}$.
- The PLL functions at frequencies down to 120 MHz.
- If the input clock is unstable and the PLL is enabled, then the PLL may lock onto an incorrect frequency, causing unstable SRAM behavior. To avoid this, provide 20 µs of stable clock to relock to the desired clock frequency.

Figure 4. Power-Up Waveforms



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Maximum Ratings

Exceeding maximum ratings may shorten the useful life of the device. User guidelines are not tested.

Storage temperature ..... –65 °C to +150 °C

Ambient temperature with power applied ..... –55 °C to +125 °C

Supply voltage on V_{DD} relative to GND .....–0.5 V to +2.9 V

Supply voltage on V_{DDQ} relative to GND ..... –0.5 V to +V_{DD}

DC applied to outputs in High Z .....–0.5 V to V_{DDQ} + 0.3 V

DC input voltage [20] .....–0.5 V to V_{DD} + 0.3 V

Current into outputs (LOW) ..... 20 mA

Static discharge voltage (MIL-STD-883, M. 3015) ..... > 2,001 V

Latch up current ..... > 200 mA

Operating Range

Range	Ambient Temperature (T_A)	V_{DD} [21]	V_{DDQ} [21]
Commercial	0 °C to +70 °C	1.8 ± 0.1 V	1.4 V to V_{DD}
Industrial	–40 °C to +85 °C		

Neutron Soft Error Immunity

Parameter	Description	Test Conditions	Typ	Max*	Unit
LSBU	Logical single-bit upsets	25 °C	197	216	FIT/Mb
LMBU	Logical multi-bit upsets	25 °C	0	0.01	FIT/Mb
SEL	Single-event latch up	85 °C	0	0.1	FIT/Dev

* *No LMBU or SEL events occurred during testing*; this column represents a statistical χ^2, 95% confidence limit calculation. For more details refer to Application Note AN54908 "Accelerated Neutron SER Testing and Calculation of Terrestrial Failure Rates"

Electrical Characteristics

Over the Operating Range

DC Electrical Characteristics

Over the Operating Range

Parameter [22]	Description	Test Conditions			Min	Typ	Max	Unit
V_{DD}	Power supply voltage				1.7	1.8	1.9	V
V_{DDQ}	I/O supply voltage				1.4	1.5	V_{DD}	V
V_{OH}	Output HIGH voltage	Note 23			$V_{DDQ}/2 – 0.12$	–	$V_{DDQ}/2 + 0.12$	V
V_{OL}	Output LOW voltage	Note 24			$V_{DDQ}/2 – 0.12$	–	$V_{DDQ}/2 + 0.12$	V
$V_{OH(LOW)}$	Output HIGH voltage	I_{OH} = –0.1 mA, nominal impedance			$V_{DDQ} – 0.2$	–	V_{DDQ}	V
$V_{OL(LOW)}$	Output LOW voltage	I_{OL} = 0.1 mA, nominal impedance			V_{SS}	–	0.2	V
V_{IH}	Input HIGH voltage				$V_{REF} + 0.1$	–	$V_{DDQ} + 0.15$	V
V_{IL}	Input LOW voltage				–0.15	–	$V_{REF} – 0.1$	V
I_X	Input leakage current	$GND \le V_I \le V_{DDQ}$			–2	–	2	µA
I_{OZ}	Output leakage current	$GND \le V_I \le V_{DDQ}$, output disabled			–2	–	2	µA
V_{REF}	Input reference voltage [25]	Typical Value = 0.75V			0.68	0.75	0.95	V
I_{DD} [26]	V_{DD} operating supply	V_{DD} = Max, I_{OUT} = 0 mA, $f = f_{MAX} = 1/t_{CYC}$	550 MHz	(× 36)	–	–	1310	mA
			500 MHz	(× 36)	–	–	1210	mA
			450 MHz	(× 36)	–	–	1100	mA
			400 MHz	(× 36)	–	–	1000	mA

Notes

20. Overshoot: $V_{IH}(AC) \le V_{DDQ} + 0.35$ V (Pulse width less than $t_{CYC}/2$), Undershoot: $V_{IL}(AC) \ge -0.3$ V (Pulse width less than $t_{CYC}/2$).
21. Power-up: Assumes a linear ramp from 0 V to V_{DD}(min) within 200 ms. During this time $V_{IH} < V_{DD}$ and $V_{DDQ} \le V_{DD}$.
22. All Voltage referenced to Ground.
23. Output are impedance controlled. $I_{OH} = -(V_{DDQ}/2)/(RQ/5)$ for values of 175 ohms <= RQ <= 350 ohms.
24. Output are impedance controlled. $I_{OL} = (V_{DDQ}/2)/(RQ/5)$ for values of 175 ohms <= RQ <= 350 ohms.
25. V_{REF} (min) = 0.68V or 0.46V_{DDQ}, whichever is larger, V_{REF} (max) = 0.95V or 0.54V_{DDQ}, whichever is smaller.
26. The operation current is calculated with 50% read cycle and 50% write cycle.

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Electrical Characteristics (continued)

Over the Operating Range

DC Electrical Characteristics (continued)

Over the Operating Range

Parameter [22]	Description	Test Conditions			Min	Typ	Max	Unit
I_{SB1}	Automatic power-down current	Max V_{DD}, Both Ports Deselected, $V_{IN} \geq V_{IH}$ or $V_{IN} \leq V_{IL}$ f = f_{MAX} = $1/t_{CYC}$, Inputs Static	550 MHz	(× 36)	–	–	380	mA
			500 MHz	(× 36)	–	–	360	mA
			450 MHz	(× 36)	–	–	340	mA
			400 MHz	(× 36)	–	–	320	mA

AC Electrical Characteristics

Over the Operating Range

Parameter [27]	Description	Test Conditions	Min	Typ	Max	Unit
V_{IH}	Input HIGH voltage		V_{REF} + 0.2	–	V_{DDQ} + 0.24	V
V_{IL}	Input LOW voltage		–0.24	–	V_{REF} – 0.2	V

Capacitance

Parameter [28]	Description	Test Conditions	Max	Unit
C_{IN}	Input capacitance	T_A = 25 °C, f = 1 MHz, V_{DD} = 1.8 V, V_{DDQ} = 1.5 V	4	pF
C_O	Output capacitance		4	pF

Thermal Resistance

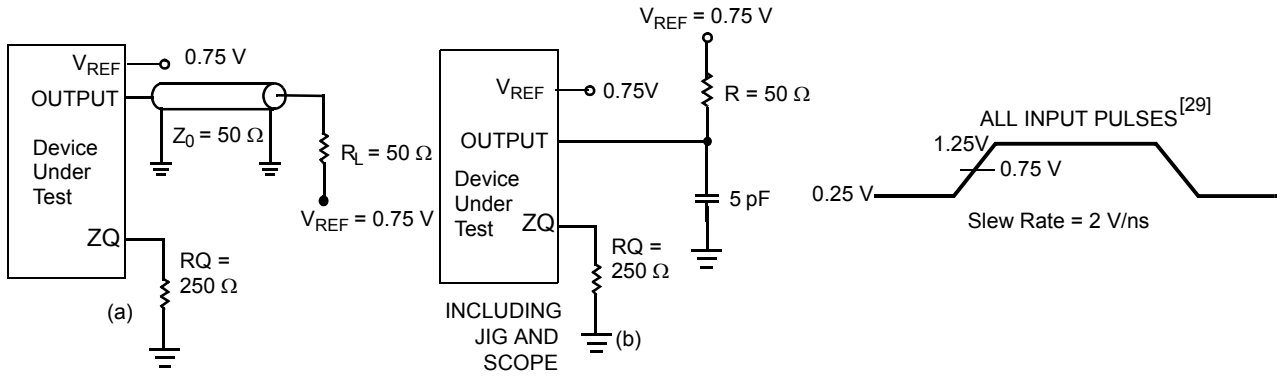
Parameter [28]	Description	Test Conditions	165-ball FBGA Package	Unit
Θ_{JA} (0 m/s)	Thermal resistance (junction to ambient)	Socketed on a 170 × 220 × 2.35 mm, eight-layer printed circuit board	14.43	°C/W
Θ_{JA} (1 m/s)			13.40	°C/W
Θ_{JA} (3 m/s)			12.66	°C/W
Θ_{JB}	Thermal resistance (junction to board)		11.38	°C/W
Θ_{JC}	Thermal resistance (junction to case)		3.30	°C/W

Notes
27. Overshoot: $V_{IH}(AC) \leq V_{DDQ}$ + 0.35 V (Pulse width less than $t_{CYC}/2$), Undershoot: $V_{IL}(AC) \geq -0.3$ V (Pulse width less than $t_{CYC}/2$).
28. Tested initially and after any design or process change that may affect these parameters.

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AC Test Loads and Waveforms

Figure 5. AC Test Loads and Waveforms



Note

29. Unless otherwise noted, test conditions are based on signal transition time of 2 V/ns, timing reference levels of 0.75 V, Vref = 0.75V, RQ = 250 Ω, V_{DDQ} = 1.5 V, input pulse levels of 0.25 V to 1.25 V, and output loading of the specified I_{OL}/I_{OH} and load capacitance shown in (a) of Figure 5.

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Switching Characteristics

Over the Operating Range [29, 30]

Cypress Parameter	Consortium Parameter	Description	550 MHz		500 MHz		450 MHz		400 MHz		Unit
			Min	Max	Min	Max	Min	Max	Min	Max	
t_{POWER}		V_{DD}(typical) to the first access [31]	1	–	1	–	1	–	1	–	ms
t_{CYC}	t_{KHKH}	K clock cycle time	1.81	8.4	2.0	8.4	2.2	8.4	2.5	8.4	ns
t_{KH}	t_{KHKL}	Input clock (K/$\overline{K}$) HIGH	0.4	–	0.4	–	0.4	–	0.4	–	ns
t_{KL}	t_{KLKH}	Input clock (K/$\overline{K}$) LOW	0.4	–	0.4	–	0.4	–	0.4	–	ns
$t_{KH\overline{K}H}$	$t_{KH\overline{K}H}$	K clock rise to $\overline{K}$ clock rise (rising edge to rising edge)	0.77	–	0.85	–	0.94	–	1.06	–	ns
Setup Times											
t_{SA}	t_{AVKH}	Address setup to K clock rise	0.23	–	0.25	–	0.275	–	0.4	–	ns
t_{SC}	t_{IVKH}	Control setup to K clock rise ($\overline{RPS}$, $\overline{WPS}$)	0.23	–	0.25	–	0.275	–	0.4	–	ns
t_{SCDDR}	t_{IVKH}	Double data rate control setup to clock (K/$\overline{K}$) rise ($\overline{BWS_0}$, $\overline{BWS_1}$, $\overline{BWS_2}$, $\overline{BWS_3}$)	0.18	–	0.20	–	0.22	–	0.28	–	ns
t_{SD}	t_{DVKH}	$D_{[X:0]}$ setup to clock (K/$\overline{K}$) rise	0.18	–	0.20	–	0.22	–	0.28	–	ns
Hold Times											
t_{HA}	t_{KHAX}	Address hold after K clock rise	0.23	–	0.25	–	0.275	–	0.4	–	ns
t_{HC}	t_{KHIX}	Control hold after K clock rise ($\overline{RPS}$, $\overline{WPS}$)	0.23	–	0.25	–	0.275	–	0.4	–	ns
t_{HCDDR}	t_{KHIX}	Double data rate control hold after clock (K/$\overline{K}$) rise ($\overline{BWS_0}$, $\overline{BWS_1}$, $\overline{BWS_2}$, $\overline{BWS_3}$)	0.18	–	0.20	–	0.28	–	0.28	–	ns
t_{HD}	t_{KHDX}	$D_{[X:0]}$ hold after clock (K/$\overline{K}$) rise	0.18	–	0.20	–	0.28	–	0.28	–	ns
Output Times											
t_{CO}	t_{CHQV}	K/$\overline{K}$ clock rise to data valid	–	0.45	–	0.45	–	0.45	–	0.45	ns
t_{DOH}	t_{CHQX}	Data output hold after output K/$\overline{K}$ clock rise (active to active)	–0.45	–	–0.45	–	–0.45	–	–0.45	–	ns
t_{CCQO}	t_{CHCQV}	K/$\overline{K}$ clock rise to echo clock valid	–	0.45	–	0.45	–	0.45	–	0.45	ns
t_{CQOH}	t_{CHCQX}	Echo clock hold after K/$\overline{K}$ clock rise	–0.45	–	–0.45	–	–0.45	–	–0.45	–	ns
t_{CQD}	t_{CQHQV}	Echo clock high to data valid		0.15		0.15		0.15		0.20	ns
t_{CQDOH}	t_{CQHQX}	Echo clock high to data invalid	–0.15	–	–0.15	–	–0.15	–	–0.20	–	ns
t_{CQH}	t_{CQHCQL}	Output clock (CQ/$\overline{CQ}$) HIGH [32]	0.655	–	0.75	–	0.85	–	1.0	–	ns
$t_{CQH\overline{CQ}H}$	$t_{CQH\overline{CQ}H}$	CQ clock rise to $\overline{CQ}$ clock rise (rising edge to rising edge) [32]	0.655	–	0.75	–	0.85	–	1.0	–	ns
t_{CHZ}	t_{CHQZ}	Clock (K/$\overline{K}$) rise to high Z (active to High Z) [33, 34]	–	0.45	–	0.45	–	0.45	–	0.45	ns
t_{CLZ}	t_{CHQX1}	Clock (K/$\overline{K}$) rise to Low Z [33, 34]	–0.45	–	–0.45	–	–0.45	–	–0.45	–	ns
t_{QVLD}	$t_{CQHQVLD}$	Echo clock high to QVLD Valid [35]	–0.15	0.15	–0.15	0.15	–0.15	0.15	–0.20	0.20	ns
PLL Timing											
$t_{KC\ Var}$	$t_{KC\ Var}$	Clock phase jitter	–	0.15	–	0.15	–	0.15	–	0.20	ns
$t_{KC\ lock}$	$t_{KC\ lock}$	PLL lock time (K)	20	–	20	–	20	–	20	–	µs
$t_{KC\ Reset}$	$t_{KC\ Reset}$	K static to PLL reset [36]	30		30		30		30		ns

Notes

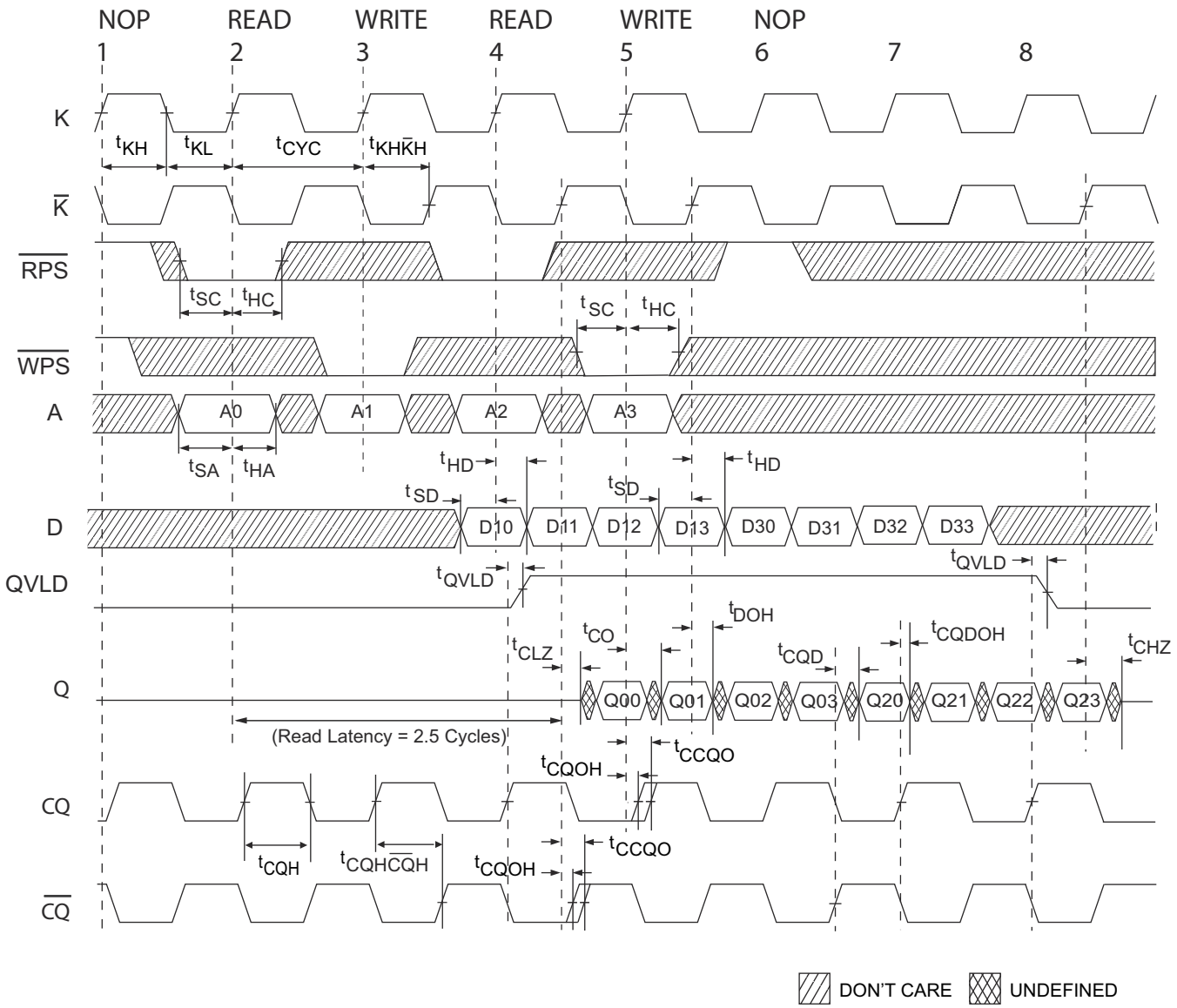
30. When a part with a maximum frequency above 400 MHz is operating at a lower clock frequency, it requires the input timings of the frequency range in which it is being operated and outputs data with the output timings of that frequency range.
31. This part has a voltage regulator internally; t_{POWER} is the time that the power must be supplied above V_{DD} minimum initially before a read or write operation can be initiated.
32. These parameters are extrapolated from the input timing parameters (t_{CYC}/2 - 250 ps, where 250 ps is the internal jitter). These parameters are only guaranteed by design and are not tested in production.
33. t_{CHZ}, t_{CLZ}, are specified with a load capacitance of 5 pF as in (b) of Figure 5 on page 22. Transition is measured ± 100 mV from steady-state voltage.
34. At any voltage and temperature t_{CHZ} is less than t_{CLZ} and t_{CHZ} less than t_{CO}.
35. t_{QVLD} spec is applicable for both rising and falling edges of QVLD signal.
36. Hold to >V_{IH} or <V_{IL}.

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Switching Waveforms

Read/Write/Deselect Sequence

Figure 6. Waveform for 2.5 Cycle Read Latency [37, 38, 39]



Notes

37. Q00 refers to output from address A0. Q01 refers to output from the next internal burst address following A0, that is, A0+1.
38. Outputs are disabled (High Z) one clock cycle after a NOP.
39. In this example, if address A2 = A1, then data Q20 = D10, Q21 = D11, Q22 = D12, and Q23 = D13. Write data is forwarded immediately as read results. This note applies to the whole diagram.

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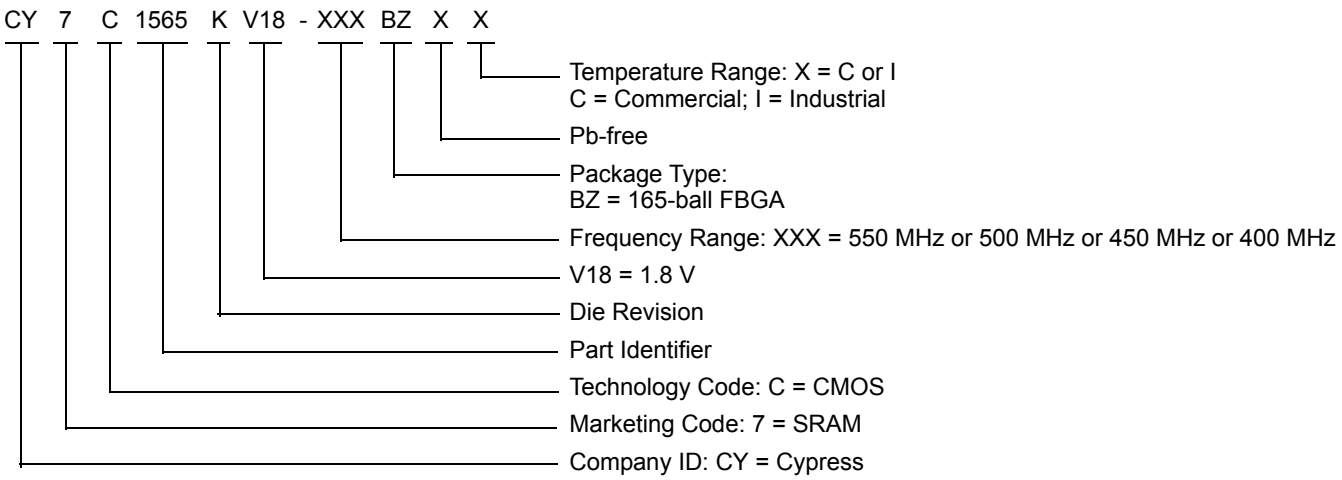
Ordering Information

The following table contains only the parts that are currently available. If you do not see what you are looking for, contact your local sales representative. For more information, visit the Cypress website at www.cypress.com and refer to the product summary page at http://www.cypress.com/products

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Speed (MHz)	Ordering Code	Package Diagram	Part and Package Type	Operating Range
550	CY7C1565KV18-550BZXC	51-85180	165-ball FBGA (13 × 15 × 1.4 mm) Pb-free	Commercial
500	CY7C1565KV18-500BZXC	51-85180	165-ball FBGA (13 × 15 × 1.4 mm) Pb-free	Commercial
	CY7C1565KV18-500BZXI	51-85180	165-ball FBGA (13 × 15 × 1.4 mm) Pb-free	Industrial
450	CY7C1565KV18-450BZXC	51-85180	165-ball FBGA (13 × 15 × 1.4 mm) Pb-free	Commercial
	CY7C1565KV18-450BZI	51-85180	165-ball FBGA (13 × 15 × 1.4 mm)	Industrial
400	CY7C1565KV18-400BZXC	51-85180	165-ball FBGA (13 × 15 × 1.4 mm) Pb-free	Commercial
	CY7C1565KV18-400BZI	51-85180	165-ball FBGA (13 × 15 × 1.4 mm)	Industrial
	CY7C1565KV18-400BZXI		165-ball FBGA (13 × 15 × 1.4 mm) Pb-free	

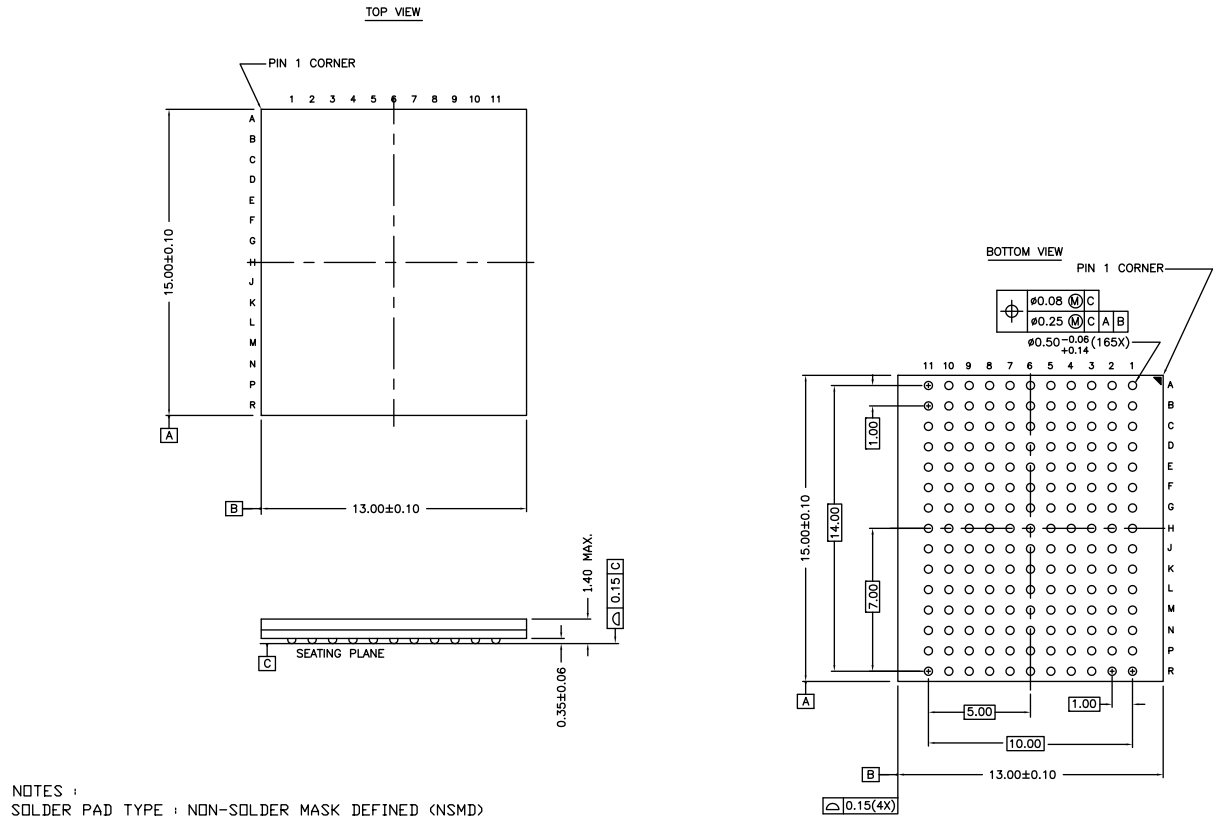
Ordering Code Definitions



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Package Diagram

Figure 7. 165-ball FBGA (13 × 15 × 1.4 mm) BB165D/BW165D (0.5 Ball Diameter) Package Outline, 51-85180



NOTES :
SOLDER PAD TYPE : NON-SOLDER MASK DEFINED (NSMD)
JEDEC REFERENCE : MO-216 / ISSUE E
PACKAGE CODE : BB0AC/BW0AC
PACKAGE WEIGHT : SEE CYPRESS PACKAGE MATERIAL DECLARATION DATASHEET (PMDD) POSTED ON THE CYPRESS WEB.

51-85180 *G

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Acronyms

Acronym	Description
DDR	Double Data Rate
FBGA	Fine-Pitch Ball Grid Array
HSTL	High-Speed Transceiver Logic
JEDEC	Joint Electron Device Engineering Council
JTAG	Joint Test Action Group
ODT	On-Die Termination
PLL	Phase-Locked Loop
QDR	Quad Data Rate
TAP	Test Access Port
TCK	Test Clock
TDI	Test Data In
TDO	Test Data Out
TMS	Test Mode Select

Document Conventions

Units of Measure

Symbol	Unit of Measure
°C	degree Celsius
MHz	megahertz
µA	microampere
µs	microsecond
mA	milliampere
mm	millimeter
ms	millisecond
ns	nanosecond
Ω	ohm
%	percent
pF	picofarad
V	volt
W	watt

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Document History Page

Document Title: CY7C1565KV18, 72-Mbit QDR® II+ SRAM Four-Word Burst Architecture (2.5 Cycle Read Latency) Document Number: 001-15878				
Revision	**ECN**	**Orig. of Change**	**Submission Date**	**Description of Change**
**	1120252	VKN	06/01/07	New data sheet.
*A	1246904	VKN / AESA	See ECN	Added 550 MHz and 500 MHz speed bins related information in all instances across the document. Removed 375 MHz, 333 MHz, and 300 MHz speed bins related information in all instances across the document. Updated Pin Configuration: Added Note 2 and referred the same note in description below heading.
*B	1739283	VKN / AESA	See ECN	Changed status from Advance Information to Preliminary.
*C	2065806	VKN / AESA	See ECN	Updated Electrical Characteristics: Added Note 26 and referred the same note in I_{DD} parameter. Updated Switching Characteristics: Updated Note 32 (Corrected typo). Changed minimum value of PLL Lock Time (K) from 2048 cycles to 20 µs.
*D	2612383	VKN / AESA	11/25/08	Updated Identification Register Definitions: Changed Revision Number (31:29) from 001 to 000. Updated Power-Up Sequence in QDR II+ SRAM: Updated description. Updated Figure 4. Updated Thermal Resistance: Included values. Updated Package Diagram: Removed spec 51-85195 *A. Added spec 51-85180 *A. Updated to new template.
*E	2683451	VKN / PYRS	04/03/09	Updated Ordering Information: No change in part numbers. Added disclaimer on top of the Ordering Information table. Post to external web.
*F	2746858	VKN	07/31/09	Changed status from Preliminary to Final. Added Neutron Soft Error Immunity. Updated Switching Characteristics: Updated maximum value of t_{CO}, t_{CCQO}, t_{CHZ} parameters to 0.45 ns for 550 MHz, 500 MHz and 450 MHz speed bins. Updated minimum value of t_{DOH}, t_{CQOH}, t_{CLZ} parameters to –0.45 ns for 550 MHz, 500 MHz and 450 MHz bins Updated Ordering Information: Updated part numbers. Modified the disclaimer on top of the Ordering Information table.
*G	2870201	NJY	02/01/10	Added Contents. Updated to new template.
*H	2894069	NJY	03/17/10	Updated Ordering Information: Updated part numbers.
*I	2931775	VKN	05/13/10	Updated Document Title. Removed CY7C1563KV18 parts related information in all instances across the document. Updated Capacitance: Changed Input Capacitance (C_{IN}) from 2 pF to 4 pF, and Output Capacitance (C_O) from 3 pF to 4 pF, Updated Ordering Information: Updated part numbers.

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Document History Page (continued)

Document Title: CY7C1565KV18, 72-Mbit QDR® II+ SRAM Four-Word Burst Architecture (2.5 Cycle Read Latency) Document Number: 001-15878				
Revision	**ECN**	**Orig. of Change**	**Submission Date**	**Description of Change**
*J	2981269	NJY	07/15/10	Added Pin Configuration and Pin Definitions sections that were missing in the previous revision.
*K	3055013	NJY	10/12/10	Updated Ordering Information: Updated part numbers.
*L	3216622	NJY	04/05/2011	Updated to new template.
*M	3506265	PRIT	02/14/2012	Removed CY7C1561KV18 and CY7C1576KV18 parts related information in all instances across the document. Updated Ordering Information. Updated Package Diagram.
*N	3862706	PRIT	01/09/2013	Updated Package Diagram: spec 51-85180 – Changed revision from *E to *F.
*O	4371936	PRIT	05/06/2014	Updated Application Example: Updated Figure 2. Updated Thermal Resistance: Updated values of Θ_{JA} parameter. Included Θ_{JB} parameter and its details. Updated to new template.
*P	4567085	PRIT	11/11/2014	Updated Functional Description: Added “For a complete list of related documentation, click here.” at the end. Updated Ordering Information: Updated part numbers.
*Q	4675860	PRIT	03/04/2015	Updated Ordering Information: Updated part numbers. Completing Sunset Review.
*R	5097330	PRIT	01/21/2016	Updated Ordering Information: Updated part numbers. Updated Package Diagram: spec 51-85180 – Changed revision from *F to *G. Updated to new template. Completing Sunset Review.
*S	5980585	AESATMP9	11/30/2017	Updated logo and copyright.

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