4.2V to 18V Input 1ch Boost Controller

BD9306AFVM

General Description

BD9306AFVM is a 1-channel DC/DC converter controller. A Step-up DC/DC converter can be configured. In addition, it has a built-in master-slave function which improves synchronization.

Key Specifications

- Power Supply Voltage Range: 4.2V to 18V
- Error Amplifier Feed Back Voltage: 1.25±1.6%
- Oscillating Frequency Range: 100kHz to 800kHz
- Standby Current: 0μA(Typ)
- Operating Temperature Range: -40°C to +85°C

Features

- 1ch PWM Controlled DC/DC Converter Controller
- Built-in Soft Start Function
- Built-in Master / Slave Function
- Protection Circuits:
  - Under Voltage Lockout Protection Circuit
  - Thermal Shutdown Circuit
  - Short Protection Circuit of Timer Latch type

Package

W(Typ) x D(Typ) x H(Max)



MSOP8
2.90mm x 4.00mm x 0.90mm

Applications

- TV, Power Supply for the TFT-LCD Panels used for LCD TVs, Back Lights
- DSC, DVC, Printer, DVD ,DVD Recorder, Generally Consumer Equipments etc.

Typical Application Circuit

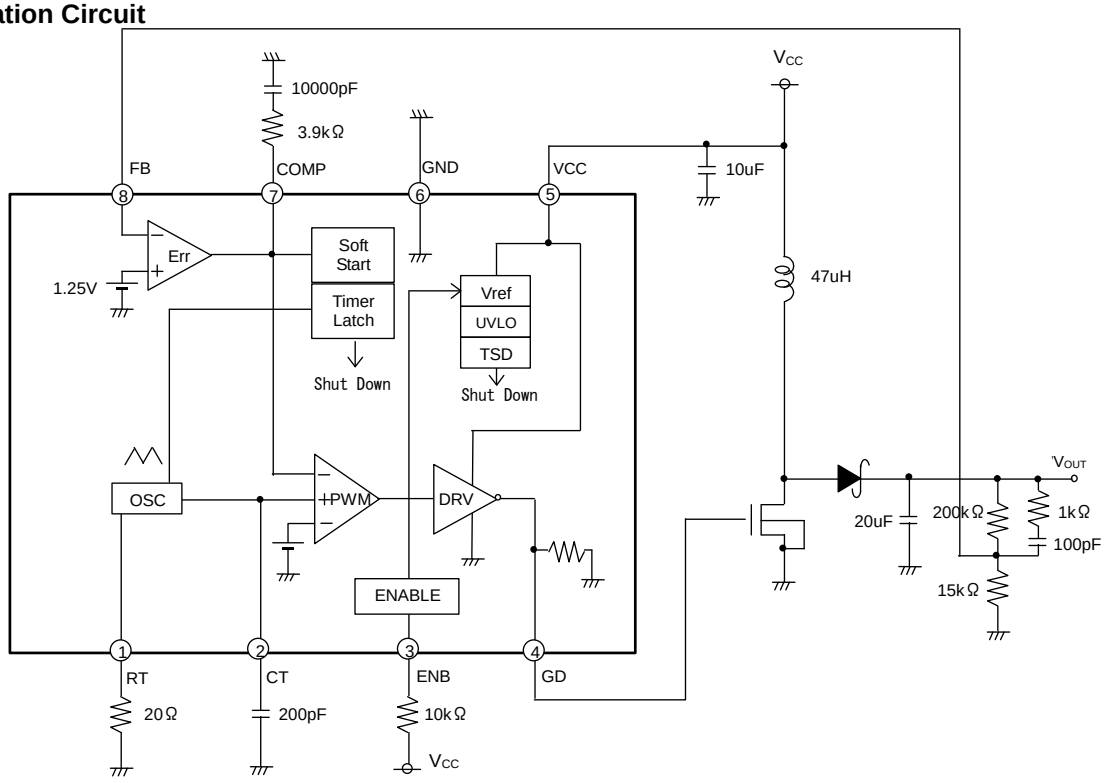
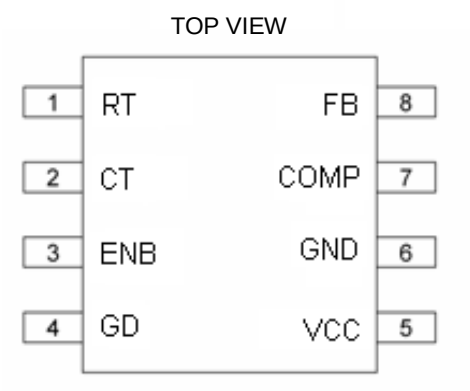


Figure 1. Typical Application Circuit

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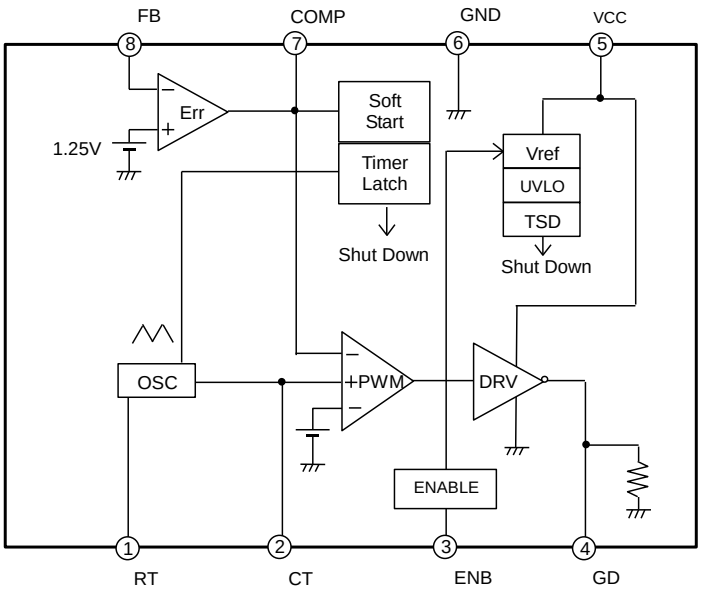
Pin Configuration



Pin Descriptions

Pin No	Pin Name	Function
1	RT	External timing resistor pin
2	CT	External timing capacitor pin
3	ENB	Control pin
4	GD	Gate drive output pin
5	VCC	Power supply pin
6	GND	Ground pin
7	COMP	Error amp output pin
8	FB	Error amp inversion input pin

Block Diagram



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Block Operation

1. Error Amplifier (Err)
   It compares a reference voltage of 1.25V (TYP) and the output feedback voltage.
   This block produces the COMP terminal voltage that determines the duty cycle.

2. Oscillator (OSC)
   This block determines the switching frequency by RT and CT values. The triangular wave is determined by RT and CT.

3. PWM
   The duty cycle is determined by comparing the output of Error amplifier and the angular wave of Oscillator.
   The switching Duty of BD9306AFVM is limited by the maximum duty ratio that is determined by the internal part, and will not be up to 100%.

4. DRV
   This block drives the gate of the external power FET by the PWM switching Duty.

5. VREF
   This block outputs the internal reference voltage of 2.5V (TYP).
   This circuit's reference voltage is controlled (ON / OFF) by the ENB terminal.

6. Protection Circuits (UVLO / TSD)
   UVLO (low-voltage Lock Out circuit) shuts down the circuits when the voltage is below 3.5V (MIN).
   TSD (temperature protection circuit) shuts down the IC when the temperature reaches 175°C (TYP).

7. Soft Start Circuit
   The Soft Start Circuit limits the current when the output voltage is slowly increasing during start-up.
   Through this, the overshoot of output voltage and current sinking can be prevented.

8. Timer Latch
   It is an output short protection circuit that detects if the output of error amplifier (COMP voltage) is more than 1.7V (TYP). If the COMP voltage becomes more than 1.7V, the counter begins to operate. The LATCH is locked when the counter counts to 2200 and the GD output shuts down. The frequency of counter is determined by RT and CT.
   Once the LATCH was locked, the GD output will not operate until it is restarted by ENB or VCC. If the output short is removed while the Timer latch is counting, the counter will be reset.

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Absolute Maximum Ratings (Ta = 25°C)

Parameter	Symbol	Limit	Unit
Power Supply Voltage (Note 2)	V_{CC}	20	V
Power Dissipation	Pd	0.58 (Note 1)	W
Operating Temperature Range	Topr	-40 to +85	°C
Storage Temperature Range	Tstg	-55 to +150	°C
Maximum Junction Temperature	Tjmax	150	°C

(Note 1) When mounted on a glass epoxy 4-layer board (70 mm x 70 mm x 1.6 mm). Derate by 4.7 mW/°C for Ta over 25°C.
(Note 2) Must not exceed Pd.
Caution: Operating the IC over the absolute maximum ratings may damage the IC. The damage can either be a short circuit between pins or an open circuit between pins. Therefore, it is important to consider circuit protection measures, such as adding a fuse, in case the IC is operated over the absolute maximum ratings.

Recommended Operating Conditions (Ta=-40°C to +85°C)

Parameter	Symbol	Limit			Unit
		Min	Typ	Max	
Power Supply Voltage	V_{CC}	4.2	12	18	V
Control Voltage	V_{ENB}	-	-	V_{CC}	V
Timing Capacitance	C_{CT}	100	-	1000	pF
Timing Resistance	R_{RT}	5	-	50	kΩ
Oscillating Frequency	f_{OSC}	100	-	800	kHz

Electrical Characteristics (Unless otherwise specified Ta=25°C, V_{CC}=12V, C_{CT}=200pF, R_{RT}=20kΩ)

Parameter	Symbol	Limit			Unit	Conditions
		Min	Typ	Max		
【Triangular Waveform Oscillator Block】						
Oscillating Frequency	f_{OSC}	165	220	275	kHz	V_{CC}=5V
Charge Threshold Voltage	V_{OSC+}	0.80	0.85	0.90	V	
Discharge Threshold Voltage	V_{OSC-}	0.20	0.25	0.30	V	
【Under-voltage Lockout Protection Circuit】						
Threshold Voltage	V_{UT}	3.5	-	4.2	V	
【Error Amp Block】						
Feed Back Voltage	V_{FB}	1.230	1.250	1.270	V	
Input Bias Current	I_{IB}	-	0.05	1	μA	V_{FB}=1.5V
COMP Sink Current	I_{OI}	35	50	65	μA	V_{FB}=1.5V V_{COMP}=1.25V
COMP Source Current	I_{OO}	35	50	65	μA	V_{FB}=1.0V V_{COMP}=1.25V
【Gate Drive Block】						
ON-Resistance	R_{ON}	-	5	-	Ω	
Gate Drive Voltage L	V_{GDL}	-	0	0.5	V	No Load
Gate Drive Voltage H	V_{GDH}	V_{CC}-0.5	V_{CC}	-	V	No Load
MAX Duty	MDT	—	83	—	%	V_{CC}=5V
【Control Block】						
ON Voltage	V_{ON}	2	-	-	V	
OFF Voltage	V_{OFF}	-	-	0.3	V	
ENB Sink Current	I_{ENB}	40	60	90	μA	V_{ENB}=5V
【Soft Start Block】						
Soft Start Time	t_S	-	10	-	ms	
【Timer Latch Protection Circuit】						
Latch Detection COMP Voltage	V_{LC}	1.5	1.7	1.9	V	
Latch Delay OSC Count Number	CNT	-	2200	-	COUNT	
Latch Delay Time	DLY	-	10	-	ms	
【Overall】						
Standby Current	I_{STBY}	-	0	10	μA	ENB=OFF
Average Consumption Current	I_{CC}	1.0	1.5	2.5	mA	No Switching

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Typical Performance Curves

(Unless otherwise specified, V_{CC}=12V, Ta=25°C)

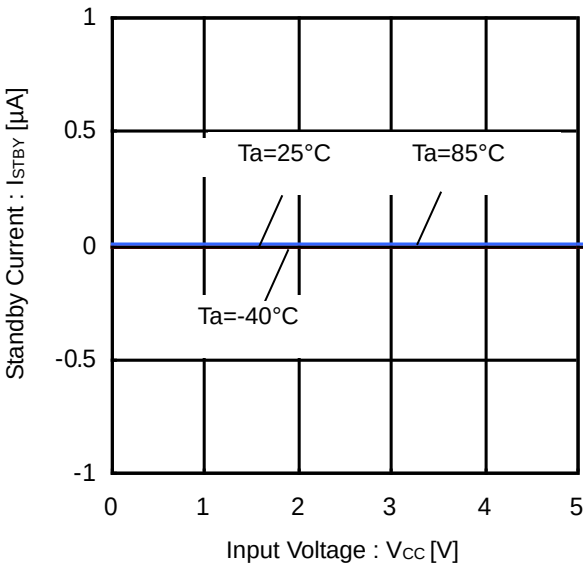


Figure 2. Standby Current vs Input Voltage

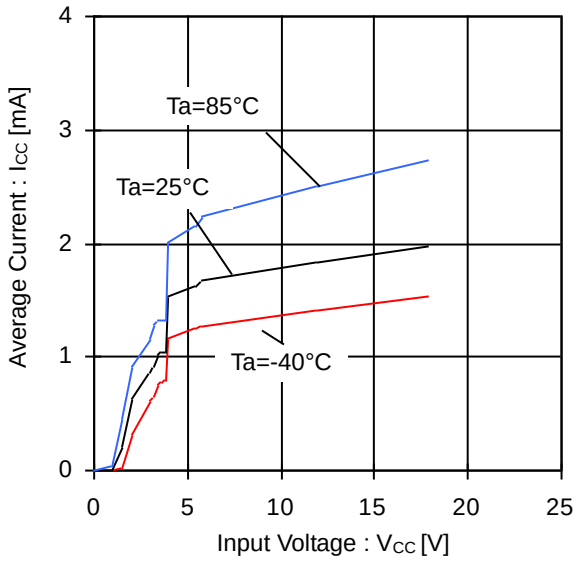


Figure 3. Average Consumption Current vs Input Voltage

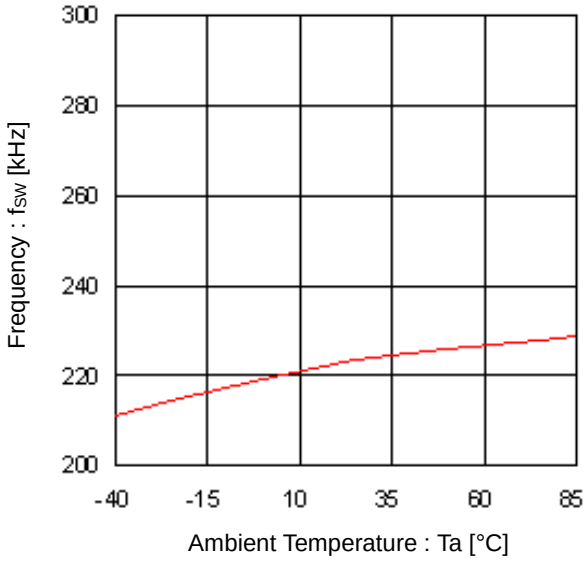


Figure 4. Frequency vs Ambient Temperature

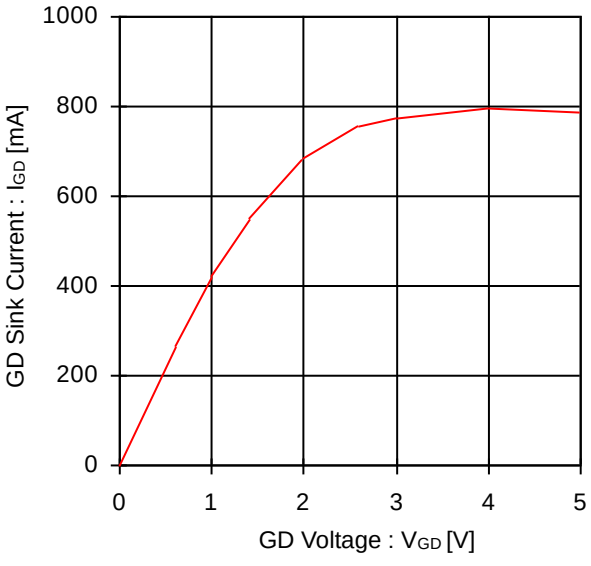


Figure 5. GD Sink Current vs GD Voltage

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Typical Performance Curves - continued

(Unless otherwise specified, V_{CC}=12V, Ta=25°C)

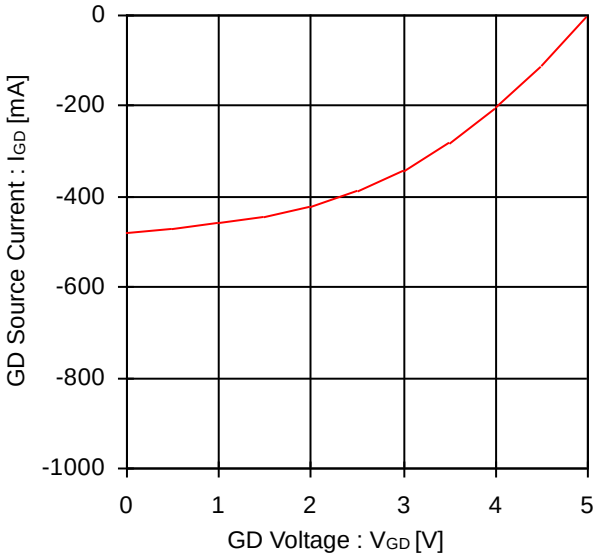


Figure 6. GD Source Current vs GD Voltage

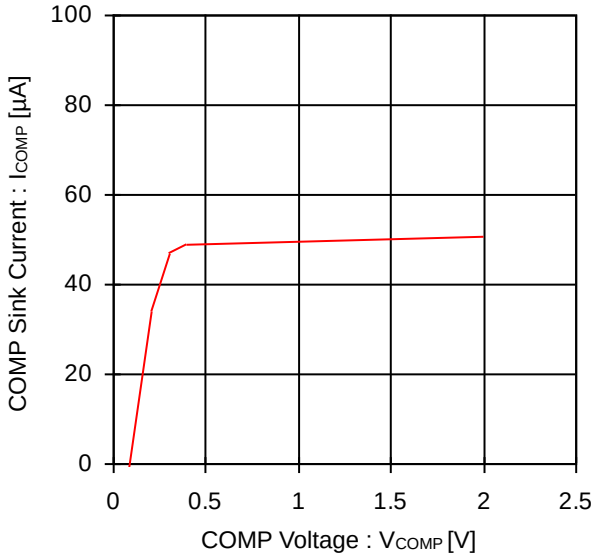


Figure 7. COMP Sink Current vs COMP Voltage

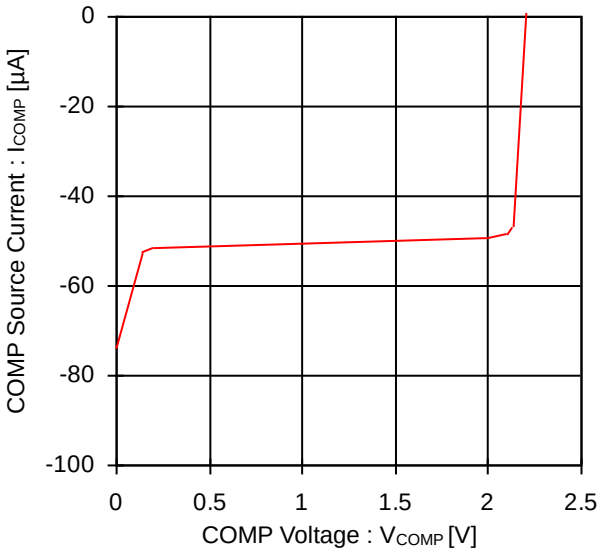


Figure 8. COMP Source Current vs COMP Voltage

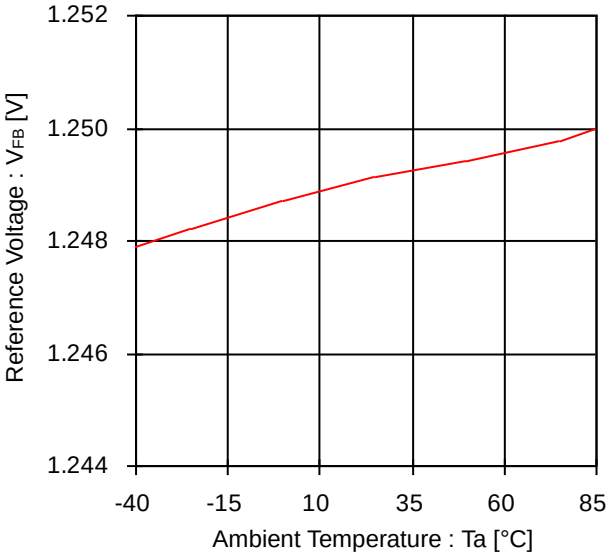


Figure 9. Reference Voltage vs Ambient Temperature

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Typical Performance Curves - continued
(Unless otherwise specified, V_{CC}=12V, Ta=25°C)

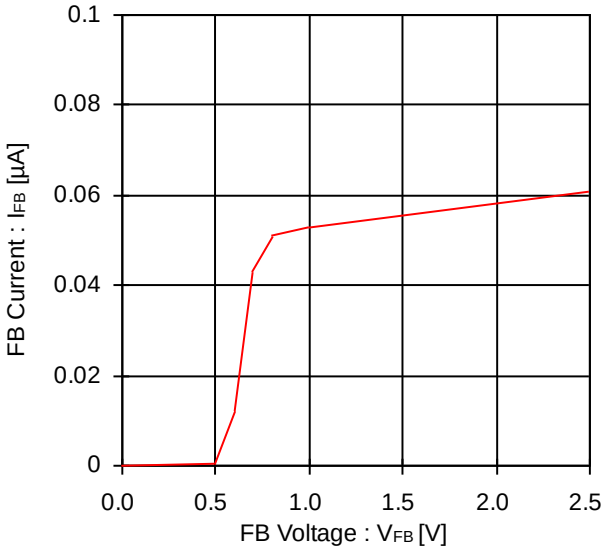


Figure 10. FB Input Bias Current vs FB Voltage

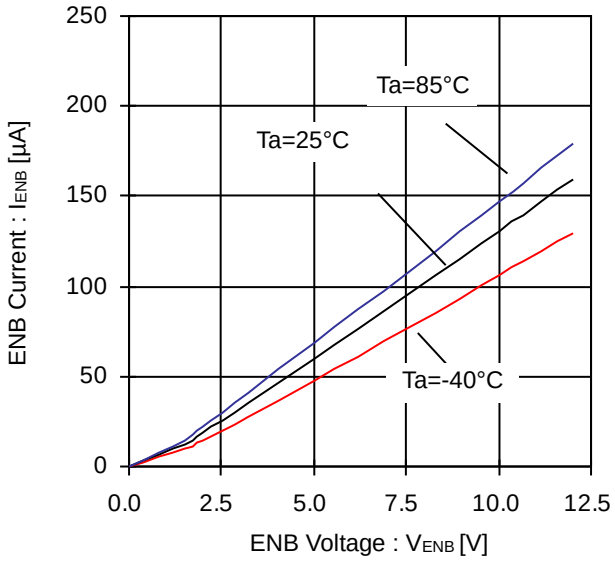


Figure 11. ENB Input Current vs ENB Voltage

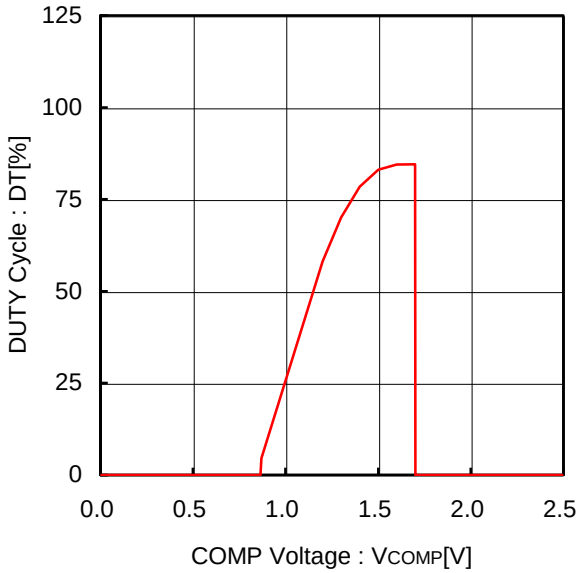


Figure 12. DUTY Cycle vs COMP Voltage

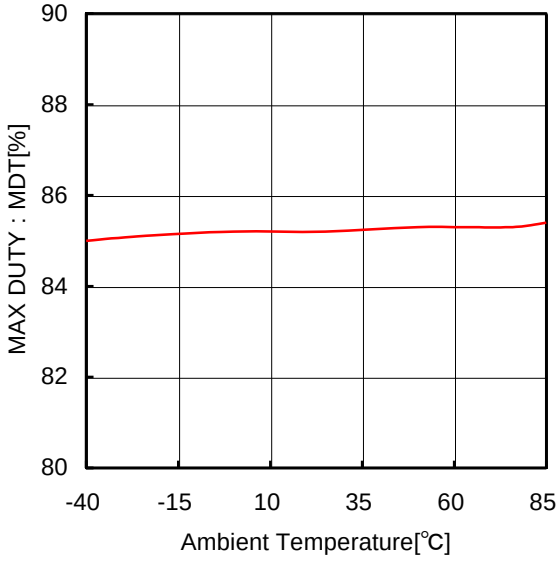


Figure 13. MAX DUTY vs Ambient Temperature

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Typical Performance Curves - continued
(Unless otherwise specified, V_{CC}=12V, Ta=25°C)

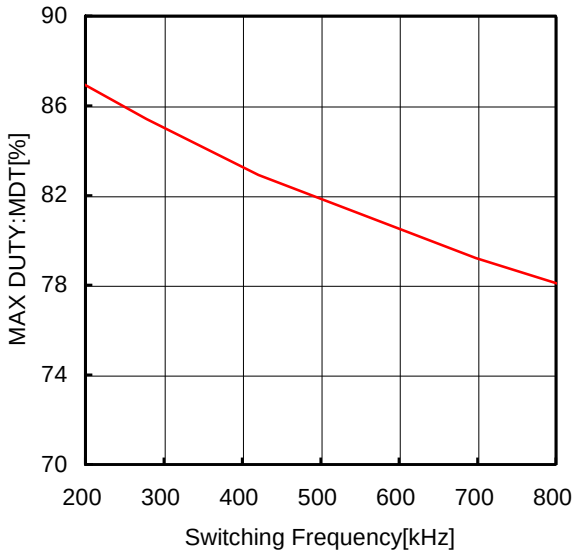


Figure 14. MAX DUTY vs Switching Frequency

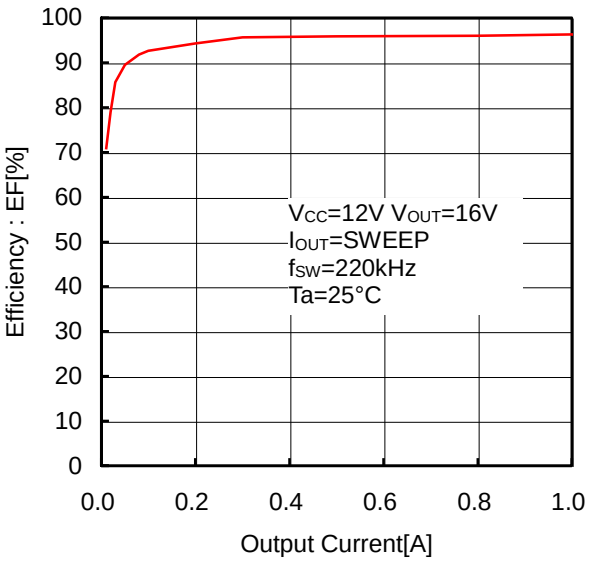


Figure 15. Efficiency vs Output Current

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Typical Waveforms

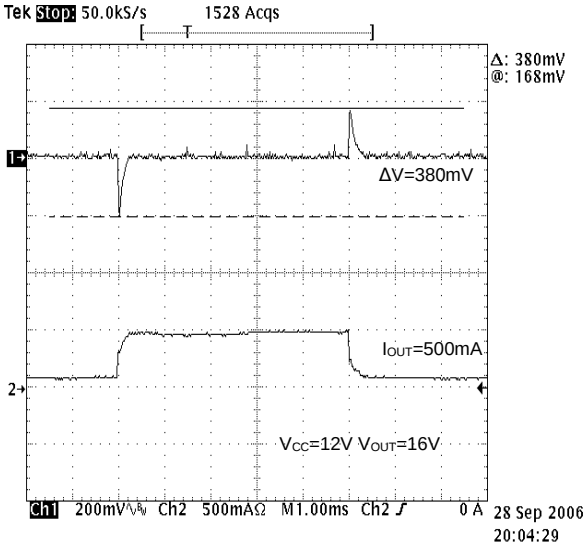


Figure 16. Load Response

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Application Information

1.Selecting Application Components

(1) Setting the Output L Constant (Step Down DC/DC)
The inductance L used for output was decided by the rated current I_{LR} and input current maximum value I_{OMAX} of the inductance.

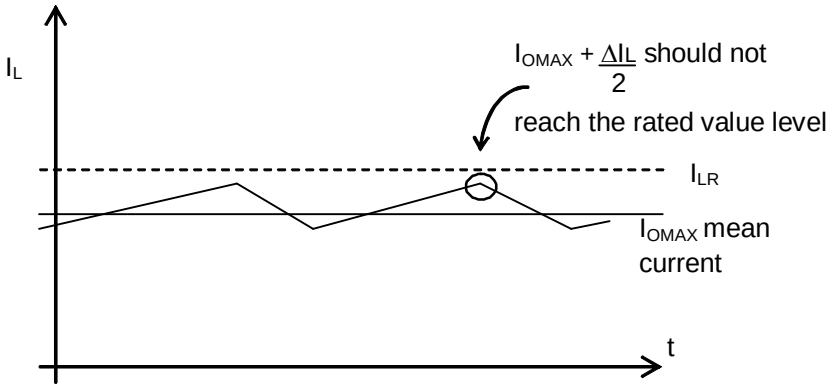


Figure 17. Coil Current Waveform (Step Down DC/DC)

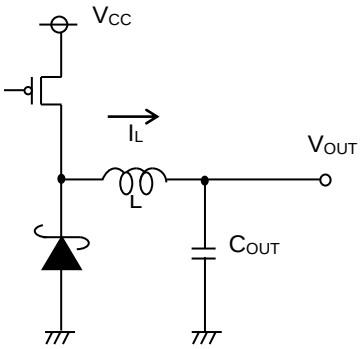


Figure 18. Output Application Circuit (Step Down DC/DC)

Adjust so that $I_{OMAX}+\Delta I_L/2$ does not reach the rated current value I_{LR}.
At this time, ΔI_L can be obtained by the following equation.

$$\Delta I_L = \frac{1}{L}\times(V_{CC}-V_{OUT})\times\frac{V_{OUT}}{V_{CC}}\times\frac{1}{f}\ [A]$$

Set a sufficient margin because the inductance L value may have ± 30% dispersion.
If the coil current exceeds the rating current I_{LR} of the coil, it can cause damage to the IC internal elements.

(2) Setting the Output L Constant (Step Up DC/DC)
The inductance L to use for output is decided by the rated current I_{LR} and input current maximum value I_{INMAX} of the inductance.

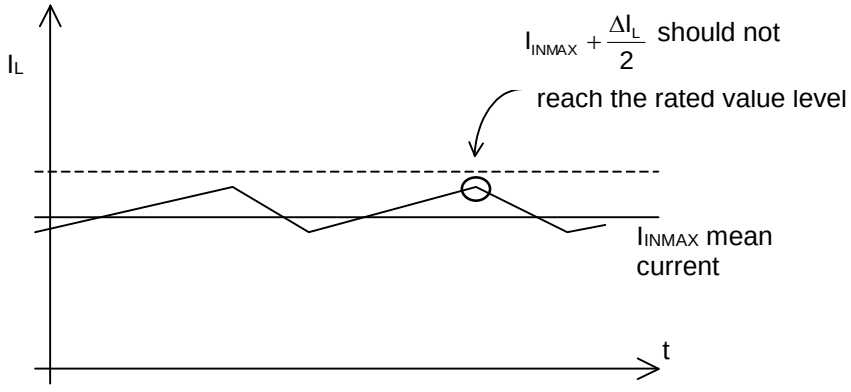


Figure 19. Coil Current Waveform (Step Up DC/DC)

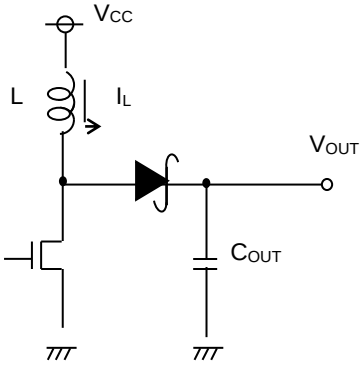


Figure 20. Output Application Circuit (Step Up DC/DC)

Adjust so that $I_{INMAX}+\Delta I_L/2$ does not reach the rated current value I_{LR}.
At this time, ΔI_L can be obtained by the following equation.

$$\Delta I_L = \frac{1}{L}V_{CC}\times\frac{V_{OUT}-V_{CC}}{V_{OUT}}\times\frac{1}{f}\ [A]$$

where:
f is the switching frequency

Set a sufficient margin because the inductance L value may have ± 30% dispersion.
If the coil current exceeds the rating current I_{LR} of the coil, it can cause damage to the IC internal elements.

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(3) Setting the Output Capacitor

For the output capacitor C, select a capacitor which has a larger value at the ripple voltage V_{PP} allowance value and the drop voltage allowance value when there's a sudden load change.

Output ripple voltage is determined by the following equation.

$$\Delta Vpp = \Delta I_L \times R_{ESR} + \frac{\Delta I_L}{2C_{OUT}} \times \frac{V_{OUT}}{V_{CC}} \times \frac{1}{f} \ [V] \qquad \text{(Step Down DC/DC)}$$

$$\Delta Vpp = I_{LMAX} \times R_{ESR} + \frac{1}{fC_{OUT}} \times \frac{V_{CC}}{V_{OUT}} \times \left(I_{LMAX} - \frac{\Delta I_L}{2} \right) \ [V] \qquad \text{(Step Up DC/DC)}$$

Apply the setting so that the voltage is within the allowable ripple voltage range.
For the drop voltage during the sudden load change (V_{DR}), perform a rough calculation by the following equation.

$$V_{DR} = \frac{\Delta I}{C_{OUT}} \times 10\mu \sec \ [V]$$

However, 10 µs is the rough calculation value of the DC/DC response speed. Set the capacitance while considering a sufficient margin so that these two values are within the standard value range.

(4) Setting of Feedback Resistance Constant

Refer to the following formula for setting of feedback resistance.

$$V_{OUT} = \frac{R_1 + R_2}{R_2} \times 1.25 \ [V]$$

It is recommend to use 10kΩ to 330kΩ setting range. If a resistance below 10kΩ was set, voltage efficiency will be dropped. If a resistance of more than 330kΩ was set, the offset voltage becomes large because of the internal error amplifier's input bias current of 0.05µA(Typ). Please set the maximum setting voltage of BD9306AFVM (step up) in such a way that Duty : (V_{OUT} - V_{CC}) / V_{OUT} is less than 70%.

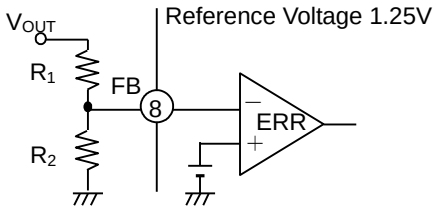


Figure 21. Feedback Resistance Setting

(5) Setting of Oscillating Frequency

The angular wave oscillation frequency can be set by connecting a resistor and a capacitor to RT (Pin 1) and CT (Pin 2) respectively. The charge and discharge currents at the capacitor of CT will be determined by the RT resistor.
Refer to the configuration below for setting the RT's resistor and the CT's capacitor.
R_{RT}: 5kΩ to 50kΩ, C_{CT}: 100pF to 1000pF. The frequency range of 100kHz to 800kHz are recommended.
Remember that the switching will stop if your setting is off this range.

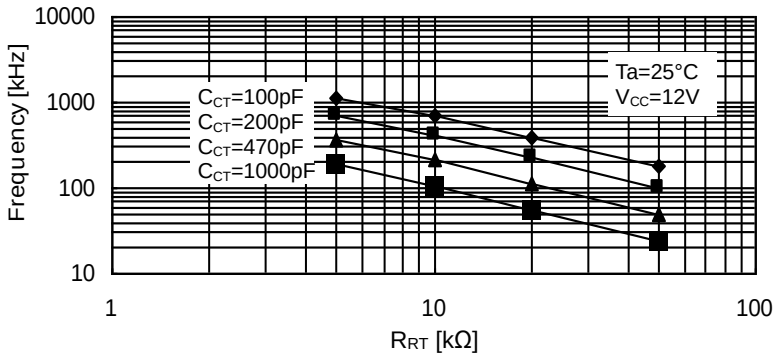


Figure 22. Frequency Setting

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(6) Selection of Input Capacitor

For DC/DC converter, the capacitor at the input side is also necessary because maximum current will be flowing between input and output. Therefore, it is recommended that an input capacitor with over 10μF and low ESR below 100mΩ. If a selected capacitor is outside this range, excessive large ripple voltage will overlap with the input voltage which may cause IC malfunction. However, this condition varies with negative overcurrent, input voltage, output voltage, inductor's value, and switching frequency so make sure to have a margin check with actual devices.

(7) Selection of Output Rectifier Diode

Schottky barrier diode is recommended as the diode for rectification at the output stage of DC/DC converter. Refer below for choosing the maximum inductor current, the maximum output voltage, and the power supply voltage.

<step-down DC/DC>

Maximum inductor current	$I_{OMAX} + \frac{\Delta I_L}{2}$	<	Diode's rated current
Power supply voltage	V_{CC}	<	Diode's rated voltage

<step-up DC/DC>

Maximum inductor current	$I_{INMAX} + \frac{\Delta I_L}{2}$	<	Diode's rated current
Maximum output voltage	V_{OMAX}	<	Diode's rated voltage

Furthermore, each parameter has a deviation of 30% to 40%, so design in such a way that you have provided enough margin for the deviation in your design.

(8) Setting of Power FET

If step-up DC/DC is configured by BD9306AFVM, Nch FET is necessary.
Consider the following conditions when you choose:

<step-down DC/DC>

Maximum inductor current	$I_{OMAX} + \frac{\Delta I_L}{2}$	<	FET's rated current
Power supply voltage	V_{CC}	<	FET's rated voltage
Power supply voltage	V_{CC}	>	FET's gate ON voltage
Gate capacitance (Note 1)	C_{GATE}	<	2000pF

<step-up DC/DC>

Maximum inductor current	$I_{INMAX} + \frac{\Delta I_L}{2}$	<	FET's rated current
Maximum output voltage	V_{OMAX}	<	FET's rated voltage
Power supply voltage	V_{CC}	>	FET's gate ON voltage
Gate capacitance (Note 1)	C_{GATE}	<	2000pF

Furthermore, each parameter has a deviation of 30% to 40%, so design in such a way that you have provided enough margin for the deviation in your design.

(Note 1) If the Gate capacity becomes large, the switching speed will be slower, which may cause heat generation and breakdown, so check thoroughly the actual devices.

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(9) Phase Compensation

Phase Setting Method

The following conditions are required to ensure the stability of the negative feedback circuit.

Phase lag should be 150° or lower when gain is 1 (0 dB) (phase margin of 30° or higher).

Because DC/DC converter applications are sampled using the switching frequency, the overall GBW should be set to 1/10 the switching frequency or lower. The target application characteristics can be summarized as follows:

- Phase lag should be 150° or lower when gain is 1 (0 dB) (phase margin of 30° or higher).
- The GBW at that time (i.e., the frequency of a 0-dB gain) is 1/10 of the switching frequency or below.

In other words, because the response is limited by the GBW, it is necessary to use higher switching frequencies to raise response.

One way to maintain stability through phase compensation involves cancellation of the secondary phase lag (-180°) caused by LC resonance with a secondary phase advance (by inserting 2 phase advances).
The GBW (i.e., the frequency with the gain set to 1) is determined by the phase compensation capacitor connected to the error amp. Increase the capacitance if a GBW reduction is required.

(a) Standard integrator (low-pass filter)

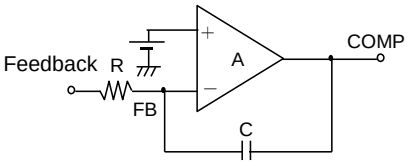


Figure 23

(b) Open loop characteristics of integrator

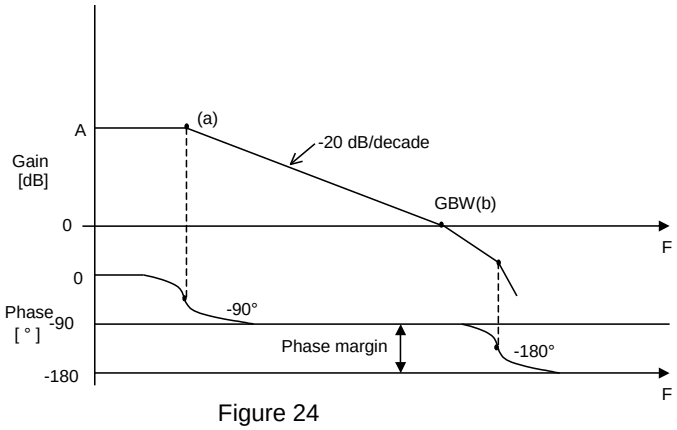
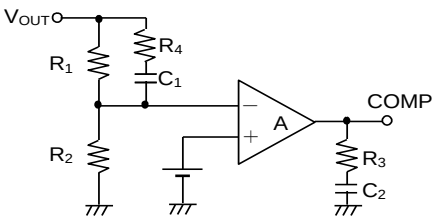


Figure 24

Point (a) $fa = \dfrac{1}{2\pi RCA}$ $[Hz]$

Point (b) $fb = GBW = \dfrac{1}{2\pi RC}$ $[Hz]$

The error amp performs the phase compensation at points (a) and (b) and it acts as a low-pass filter.
For DC/DC converter applications, R refers to feedback resistors connected in parallel.
From the LC resonance of output, the number of phase advances to be inserted is two.



LC resonant frequency $fp = \dfrac{1}{2\pi\sqrt{LC}}$ $[Hz]$

Phase advance $fz1 = \dfrac{1}{2\pi C_1 R_1}$ $[Hz]$

Phase advance $fz2 = \dfrac{1}{2\pi C_2 R_3}$ $[Hz]$

Figure 25

Set a phase advance frequency close to the LC resonant frequency for the purpose of canceling the LC resonance.

(Note) If high-frequency noise is generated in the output, FB is affected through capacitor C_1.
Therefore, insert the resistor R_4=1kΩ or so, which is in series with capacitor C_1.

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2.Example of Application

(Note) We strongly recommend the following application circuit examples but check thoroughly the characteristics before putting using them.
When you made changes at the external circuit, design a sufficient margin after considering the deviation, etc. of the external components and ROHM IC in terms of not only the static characteristic but also the transient characteristic.
Moreover, understand that our company can not confirm fully with regard to the patent right.

<Master Slave Function>

The master slave function is built-in. Synchronous switching is possible by the multi-connection of BD9305AFVM/ BD9306AFVM ICs. The following drawing shows an example of circuit connection in which BD9305AFVM is connected on the master side and BD9306AFVM is connected on the slave side.

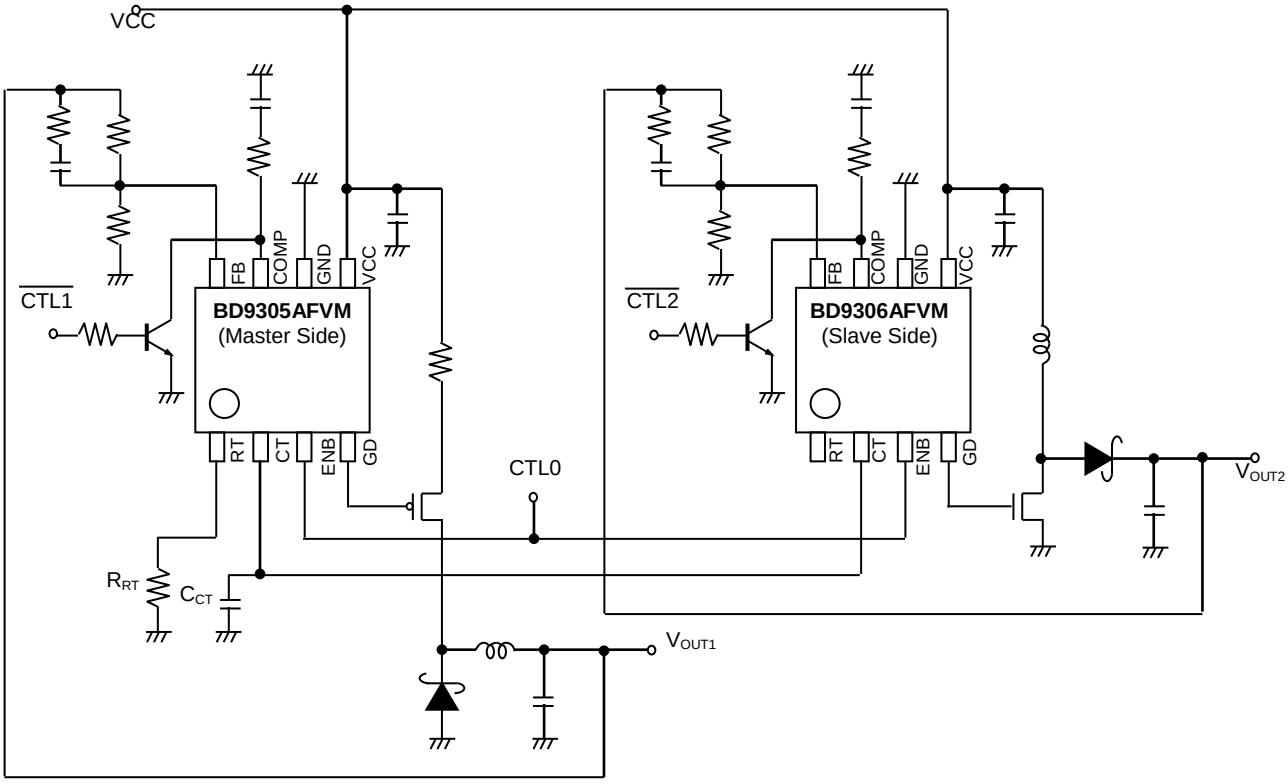


Figure 26. Master Slave Application Circuit

In the circuit above, BD9306AFVM is synchronized with the switching frequency which is determined by RT and CT of BD9305AFVM (master). In addition, the ON/OFF of output can be controlled by connecting the switch to the COMP terminal. (Refer to the following table)

Control signal correspondence table

Output state		Control signal		
V_{OUT1}	V_{OUT2}	CTL0	$\overline{CTL1}$	$\overline{CTL2}$
OFF	OFF	Low	(Note)	(Note)
OFF	ON	High	High	Low
ON	OFF	High	Low	High
ON	ON	High	Low	Low

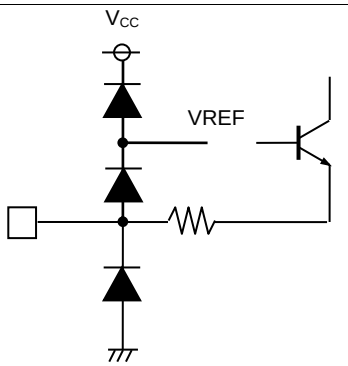
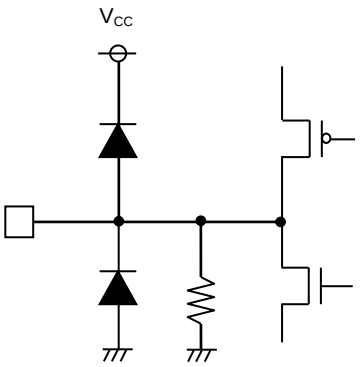
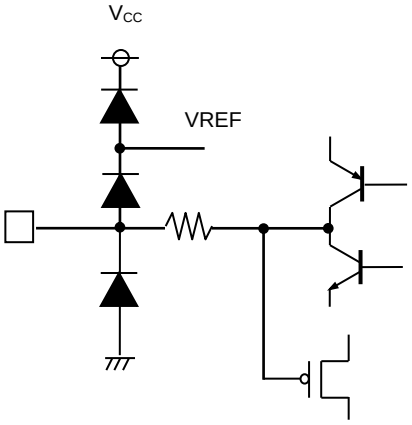
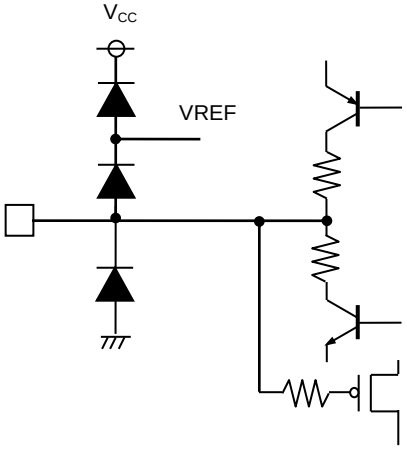
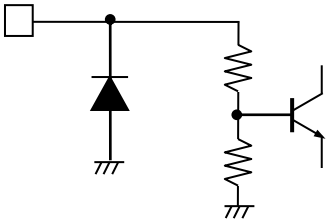
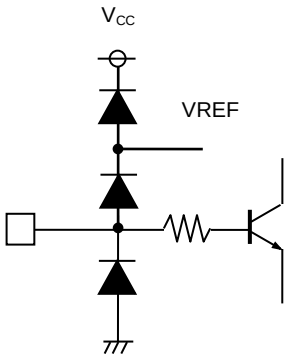
(Note) The same in either case of High / Low.

Similarly in the case of connecting three or more than three, synchronization is still possible by connecting the CT terminal of Master and the CT terminal of Slave.

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I/O Equivalent Circuits

1.RT	4.GD
	
2.CT	**7.COMP**
	
3.ENB	**8.FB**
	

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Operational Notes

1. **Reverse Connection of Power Supply**
   Connecting the power supply in reverse polarity can damage the IC. Take precautions against reverse polarity when connecting the power supply, such as mounting an external diode between the power supply and the IC's power supply pins.

2. **Power Supply Lines**
   Design the PCB layout pattern to provide low impedance supply lines. Separate the ground and supply lines of the digital and analog blocks to prevent noise in the ground and supply lines of the digital block from affecting the analog block. Furthermore, connect a capacitor to ground at all power supply pins. Consider the effect of temperature and aging on the capacitance value when using electrolytic capacitors.

3. **Ground Voltage**
   Ensure that no pins are at a voltage below that of the ground pin at any time, even during transient condition.

4. **Ground Wiring Pattern**
   When using both small-signal and large-current ground traces, the two ground traces should be routed separately but connected to a single ground at the reference point of the application board to avoid fluctuations in the small-signal ground caused by large currents. Also ensure that the ground traces of external components do not cause variations on the ground voltage. The ground lines must be as short and thick as possible to reduce line impedance.

5. **Thermal Consideration**
   Should by any chance the maximum junction temperature rating be exceeded the rise in temperature of the chip may result in deterioration of the properties of the chip. In case of exceeding this absolute maximum rating, increase the board size and copper area to prevent exceeding the maximum junction temperature rating.

6. **Recommended Operating Conditions**
   These conditions represent a range within which the expected characteristics of the IC can be approximately obtained. The electrical characteristics are guaranteed under the conditions of each parameter.

7. **Inrush Current**
   When power is first supplied to the IC, it is possible that the internal logic may be unstable and inrush current may flow instantaneously due to the internal powering sequence and delays, especially if the IC has more than one power supply. Therefore, give special consideration to power coupling capacitance, power wiring, width of ground wiring, and routing of connections.

8. **Operation Under Strong Electromagnetic Field**
   Operating the IC in the presence of a strong electromagnetic field may cause the IC to malfunction.

9. **Testing on Application Boards**
   When testing the IC on an application board, connecting a capacitor directly to a low-impedance output pin may subject the IC to stress. Always discharge capacitors completely after each process or step. The IC's power supply should always be turned off completely before connecting or removing it from the test setup during the inspection process. To prevent damage from static discharge, ground the IC during assembly and use similar precautions during transport and storage.

10. **Inter-pin Short and Mounting Errors**
    Ensure that the direction and position are correct when mounting the IC on the PCB. Incorrect mounting may result in damaging the IC. Avoid nearby pins being shorted to each other especially to ground, power supply and output pin. Inter-pin shorts could be due to many reasons such as metal particles, water droplets (in very humid environment) and unintentional solder bridge deposited in between pins during assembly to name a few.

11. **Unused Input Pins**
    Input pins of an IC are often connected to the gate of a MOS transistor. The gate has extremely high impedance and extremely low capacitance. If left unconnected, the electric field from the outside can easily charge it. The small charge acquired in this way is enough to produce a significant effect on the conduction through the transistor and cause unexpected operation of the IC. So unless otherwise specified, unused input pins should be connected to the power supply or ground line.

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Operational Notes – continued

12. **Regarding the Input Pin of the IC**

This monolithic IC contains P+ isolation and P substrate layers between adjacent elements in order to keep them isolated. P-N junctions are formed at the intersection of the P layers with the N layers of other elements, creating a parasitic diode or transistor. For example (refer to figure below):

When GND > Pin A and GND > Pin B, the P-N junction operates as a parasitic diode.
When GND > Pin B, the P-N junction operates as a parasitic transistor.

Parasitic diodes inevitably occur in the structure of the IC. The operation of parasitic diodes can result in mutual interference among circuits, operational faults, or physical damage. Therefore, conditions that cause these diodes to operate, such as applying a voltage lower than the GND voltage to an input pin (and thus to the P substrate) should be avoided.

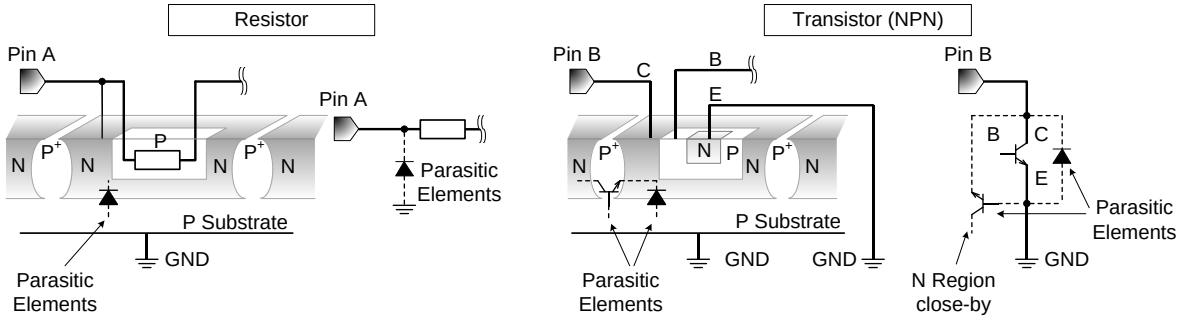


Figure 27. Example of monolithic IC structure

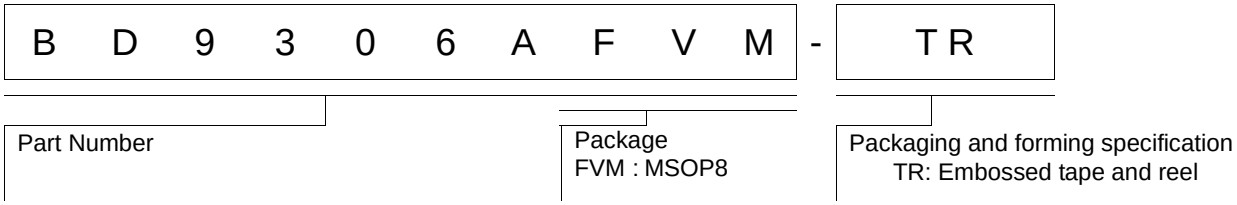
13. **Thermal Shutdown Circuit(TSD)**

This IC has a built-in thermal shutdown circuit that prevents heat damage to the IC. Normal operation should always be within the IC's power dissipation rating. If however the rating is exceeded for a continued period, the junction temperature (Tj) will rise which will activate the TSD circuit that will turn OFF all output pins. When the Tj falls below the TSD threshold, the circuits are automatically restored to normal operation.
Note that the TSD circuit operates in a situation that exceeds the absolute maximum ratings and therefore, under no circumstances, should the TSD circuit be used in a set design or for any purpose other than protecting the IC from heat damage.

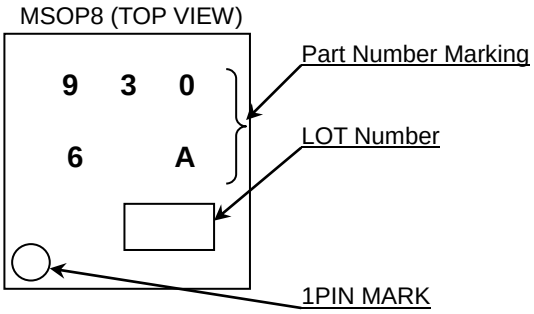
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Ordering Information



Marking Diagram

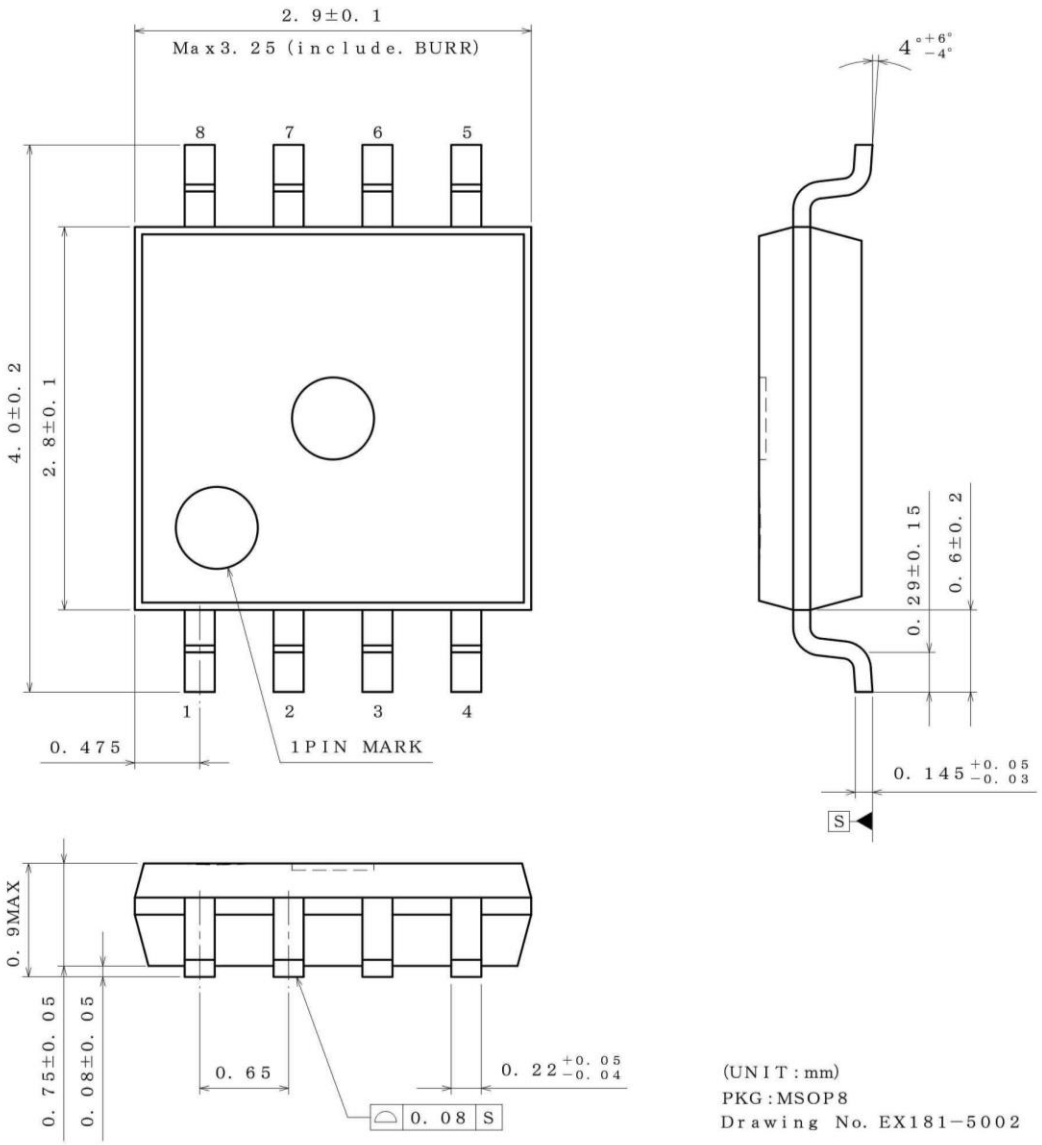


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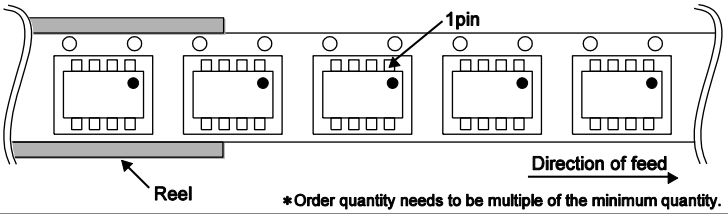
Physical Dimension, Tape and Reel Information

Package Name	MSOP8



<Tape and Reel information>

Tape	Embossed carrier tape
Quantity	3000pcs
Direction of feed	TR (The direction is the 1pin of product is at the upper right when you hold reel on the left hand and you pull out the tape on the right hand)



*Order quantity needs to be multiple of the minimum quantity.

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Revision History

Date	Revision	Changes
13.Nov.2015	001	New Release

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JAPAN	USA	EU	CHINA
CLASSⅢ	CLASSⅢ	CLASSⅡb	CLASSⅢ
CLASSⅣ		CLASSⅢ	

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   [b] Use of our Products outdoors or in places where the Products are exposed to direct sunlight or dust
   [c] Use of our Products in places where the Products are exposed to sea wind or corrosive gases, including Cl_2, H_2S, NH_3, SO_2, and NO_2
   [d] Use of our Products in places where the Products are exposed to static electricity or electromagnetic waves
   [e] Use of our Products in proximity to heat-producing components, plastic cords, or other flammable items
   [f] Sealing or coating our Products with resin or other coating materials
   [g] Use of our Products without cleaning residue of flux (even if you use no-clean type fluxes, cleaning residue of flux is recommended); or Washing our Products by using water or water-soluble cleaning agents for cleaning residue after soldering
   [h] Use of the Products in places subject to dew condensation

4. The Products are not subject to radiation-proof design.

5. Please verify and confirm characteristics of the final or mounted products in using the Products.

6. In particular, if a transient load (a large amount of load applied in a short period of time, such as pulse. is applied, confirmation of performance characteristics after on-board mounting is strongly recommended. Avoid applying power exceeding normal rated power; exceeding the power rating under steady-state loading condition may negatively affect product performance and reliability.

7. De-rate Power Dissipation depending on ambient temperature. When used in sealed area, confirm that it is the use in the range that does not exceed the maximum junction temperature.

8. Confirm that operation temperature is within the specified range described in the product specification.

9. ROHM shall not be in any way responsible or liable for failure induced under deviant condition from what is defined in this document.

Precaution for Mounting / Circuit board design

1. When a highly active halogenous (chlorine, bromine, etc.) flux is used, the residue of flux may negatively affect product performance and reliability.

2. In principle, the reflow soldering method must be used on a surface-mount products, the flow soldering method must be used on a through hole mount products. If the flow soldering method is preferred on a surface-mount products, please consult with the ROHM representative in advance.

For details, please refer to ROHM Mounting specification

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Precautions Regarding Application Examples and External Circuits

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2. You agree that application notes, reference designs, and associated data and information contained in this document are presented only as guidance for Products use. Therefore, in case you use such information, you are solely responsible for it and you must exercise your own independent verification and judgment in the use of such information contained in this document. ROHM shall not be in any way responsible or liable for any damages, expenses or losses incurred by you or third parties arising from the use of such information.

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This Product is electrostatic sensitive product, which may be damaged due to electrostatic discharge. Please take proper caution in your manufacturing process and storage so that voltage exceeding the Products maximum rating will not be applied to Products. Please take special care under dry condition (e.g. Grounding of human body / equipment / solder iron, isolation from charged objects, setting of Ionizer, friction prevention and temperature / humidity control).

Precaution for Storage / Transportation

1. Product performance and soldered connections may deteriorate if the Products are stored in the places where:
   - [a] the Products are exposed to sea winds or corrosive gases, including Cl2, H2S, NH3, SO2, and NO2
   - [b] the temperature or humidity exceeds those recommended by ROHM
   - [c] the Products are exposed to direct sunshine or condensation
   - [d] the Products are exposed to high Electrostatic

2. Even under ROHM recommended storage condition, solderability of products out of recommended storage time period may be degraded. It is strongly recommended to confirm solderability before using Products of which storage time is exceeding the recommended storage time period.

3. Store / transport cartons in the correct direction, which is indicated on a carton with a symbol. Otherwise bent leads may occur due to excessive stress applied when dropping of a carton.

4. Use Products within the specified time after opening a humidity barrier bag. Baking is required before using Products of which storage time is exceeding the recommended storage time period.

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