

TCAN1044-Q1 Automotive Fault-Protected CAN FD Transceiver With 1.8V I/O Support

1 Features

- AEC-Q100: Qualified for automotive applications
 - Temperature grade 1: –40°C to 125°C T_△
- Meets the requirements of ISO 11898-2:2016 and ISO 11898-5:2007 physical layer standards
- **Functional Safety-Capable**
 - Documentation available to aid functional safety system design
- Support of classical CAN and optimized CAN FD performance at 2, 5, and 8Mbps
 - Short and symmetrical propagation delays for enhanced timing margin
 - Higher data rates in loaded CAN networks
- I/O voltage range supports 1.7V to 5.5V
 - Support for 1.8V, 2.5V, 3.3V, and 5V applications
- Protection features:
 - Bus fault protection: ±58V
 - Undervoltage protection
 - TXD dominant timeout (DTO)
 - Data rates down to 9.2kbps
 - Thermal-shutdown protection (TSD)
- Operating modes:
 - Normal mode
 - Low power standby mode supporting remote wake-up request
- Optimized behavior when unpowered
 - Bus and logic pins are high impedance (no load to operating bus or application)
 - Hot-plug capable: power up/down glitch free operation on bus and RXD output
- Junction temperatures from: -40°C to 150°C
- Receiver common mode input voltage: ±12V
- Available in SOIC (8), SOT23 (8) packages and leadless VSON (8) packages with improved automated optical inspection (AOI) capability

2 Applications

- Automotive and Transportation
 - Body control modules
 - Automotive gateway
 - Advanced driver assistance system (ADAS)
 - Infotainment

3 Description

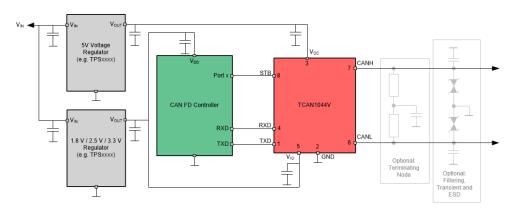
The TCAN1044-Q1 is a high speed controller area network (CAN) transceiver that meets the physical layer requirements of the ISO 11898-2:2016 highspeed CAN specification.

TCAN1044-Q1 transceiver supports classical CAN and CAN FD networks up to 8 megabits per second (Mbps). The TCAN1044V-Q1 includes internal logic level translation with the V_{IO} pin to allow for interfacing the transceiver I/O directly to 1.8V, 2.5V, 3.3V, or 5V logic levels. The transceiver supports a low-power standby mode and wake over CAN compliant to the ISO 11898-2:2016 defined wake-up pattern (WUP). The TCAN1044-Q1 transceiver also includes protection and diagnostic features supporting thermal-shutdown (TSD), TXDdominant time-out (DTO), supply undervoltage detection, and bus fault protection up to ±58V.

Package Information

PART NUMBER	PACKAGE ⁽¹⁾	PACKAGE SIZE ⁽²⁾
	SOT (DDF, 8)	2.9mm x 2.8mm
TCAN1044-Q1	VSON (DRB, 8)	3mm x 3mm
	SOIC (D, 8)	4.9mm x 6mm

- For more information, see Section 12. (1)
- The package size (length × width) is a nominal value and includes pins, where applicable.



Simplified Schematic



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4 Device Comparison

Table 4-1. Device Comparison Table

Part Number	Low Voltage I/O Logic Support on Pin 5	Pin 8 Mode Selection
TCAN1044-Q1	No	Low Power Standby Mode with Remote
TCAN1044V-Q1	Yes	Wake

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5 Pin Configuration and Functions



Figure 5-1. DDF Package, 8-Pin SOT (Top View)

Figure 5-2. D Package, 8-Pin SOIC (Top View)

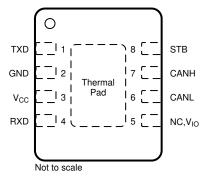


Figure 5-3. DRB Package, 8-Pin VSON (Top View)

Table 5-1. Pin Functions

	Pins		Description
Name	No.	Type Description	
TXD	1	Digital Input	CAN transmit data input
GND	2	GND	Ground connection
V _{CC}	3	Supply	5V supply voltage
RXD	4	Digital Output	CAN receive data output, tri-state when powered off
NC	5	_	No Connect (not internally connected); Devices without V _{IO}
V _{IO}	3	Supply	I/O supply voltage
CANL	6	Bus IO	Low-level CAN bus input/output line
CANH	7	Bus IO	High-level CAN bus input/output line
STB	8	Digital Input	Standby input for mode control, integrated pull up
Thermal Pad (VSON only)		_	Electrically connected to GND, connect the thermal pad to the printed circuit board (PCB) ground plane for thermal relief



6 Specifications

6.1 Absolute Maximum Ratings

over operating free-air temperature range (unless otherwise noted)(1) (2)

		MIN	MAX	UNIT
V _{CC}	Supply voltage	-0.3	6	V
V _{IO}	Supply voltage I/O level shifter	-0.3	6	V
V _{BUS}	CAN Bus IO voltage CANH and CANL	-58	58	V
V _{DIFF}	Max differential voltage between CANH and CANL	-45	45	V
V _{Logic_Input}	Logic input terminal voltage	-0.3	6	V
V _{RXD}	RXD output terminal voltage range	-0.3	6	V
I _{O(RXD)}	RXD output current	-8	8	mA
T _J	Operating virtual junction temperature range	-40	150	°C
T _{STG}	Storage temperature	-65	150	°C

- (1) Operation outside the Absolute Maximum Ratings may cause permanent device damage. Absolute Maximum Ratings do not imply functional operation of the device at these or any other conditions beyond those listed under Recommended Operating Conditions. If used outside the Recommended Operating Conditions but within the Absolute Maximum Ratings, the device may not be fully functional, and this may affect device reliability, functionality, performance, and shorten the device lifetime.
- (2) All voltage values, except differential IO bus voltages, are with respect to ground terminal.

6.2 ESD Ratings

				VALUE	UNIT
V _{ESD} E	Human-body model (HBM), per AEC Q100-002 ⁽¹⁾		HBM classification level 3A for all pins	±3000	V
		HBM classification level 3B for global pins CANH & CANL	±10000	V	
				±750	V

(1) AEC Q100-002 indicates that HBM stressing shall be in accordance with the ANSI/ESDA/JEDEC JS-001 specification.

6.3 ESD Ratings

				VALUE	UNIT
V Contain Local Floring Otatic Disch	System Loyal Floatro Statio Discharge (FSD)(3)	CAN bus terminals (CANH, CANL) to GND	SAE J2962-2 per ISO 10650 Powered Contact Discharge	±8000	V
V _{ESD}		SAE J2962-2 per ISO 10650 Powered Air Discharge		V	
	ISO 7637 ISO Pulse Transients ⁽¹⁾		Pulse 1	-100	V
			Pulse 2a	75	V
V _{Tran}	130 7037 130 Pulse Harislenis	CAN bus terminals (CANH, CANL)	Pulse 3a	-	V
			Pulse 3b	100	V
	ISO 7637 Slow transients pulse ⁽²⁾	CAN bus terminals (CANH, CANL) to GND	DCC slow transient pulse	±85	V

- (1) Tested according to IEC 62228-3:2019 CAN Transceivers, Section 6.3; standard pulses parameters defined in ISO 7637-2 (2011)
- (2) Tested according to ISO 7637-3 (2017); Electrical transient transmission by capacitive and inductive coupling via lines other than supply lines
- (3) Results given here are specific to the SAE J2962-2 Communication Transceivers Qualification Requirements CAN. Testing performed by OEM approved independent 3rd party, EMC report available upon request.

6.4 Recommended Operating Conditions

	<u> </u>				
		MIN	NOM	MAX	UNIT
V _{CC}	Supply voltage	4.5	5	5.5	V
V _{IO}	Supply voltage for I/O level shifter	1.7		5.5	V
I _{OH(RXD)}	RXD terminal high level output current	-2			mA
I _{OL(RXD)}	RXD terminal low level output current			2	mA
T _A	Operating ambient temperature	-40		125	℃

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6.5 Thermal Characteristics

THERMAL METRIC(1)			TCAN1044x-Q1			
	THERMAL METRIC	D (SOIC)	DDF (SOT)	DRB (VSON)	UNIT	
R _{⊙JA}	Junction-to-ambient thermal resistance	128.1	119.9	49.9	°C/W	
R _{OJC(top)}	Junction-to-case (top) thermal resistance	68.3	61.8	58.2	°C/W	
R _{⊝JB}	Junction-to-board thermal resistance	71.6	39.7	23.9	°C/W	
Ψ_{JT}	Junction-to-top characterization parameter	19.7	2.1	1.7	°C/W	
Ψ_{JB}	Junction-to-board characterization parameter	70.8	39.5	23.8	°C/W	
R _{OJC(bot)}	Junction-to-case (bottom) thermal resistance	-	-	6.4	°C/W	

⁽¹⁾ For more information about traditional and new thermal metrics, see the Semiconductor and IC Package Thermal Metrics application report.

6.6 Supply Characteristics

Over recommended operating conditions with $T_A = -40$ °C to 125°C (unless otherwise noted)

	PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT
		Dominant	TXD = 0 V, STB = 0 V, R _L = 60 Ω , C _L = open See Figure 7-1		45	70	mA
	Supply current	Dominant	TXD = 0 V, STB = 0 V, R _L = 50 Ω , C _L = open See Figure 7-1		49	80	mA
Icc	Normal mode	Recessive	TXD = V_{CC} , STB = 0 V, R_L = 50 Ω , C_L = open See Figure 7-1		4.5	7.5	mA
		Dominant with bus fault	TXD = 0 V, STB = 0 V, CANH = CANL = ±25 V, R _L = open, C _L = open See Figure 7-1			130	mA
Icc	Supply current Standby mode Devices with V _{IO}		TXD = STB = V_{IO} , R_L = 50 Ω , C_L = open See Figure 7-1		0.2	1	μΑ
I _{CC}	Supply current Standby mode Devices without V _{IO}		TXD = STB = V_{CC} , R_L = 50 Ω , C_L = open See Figure 7-1			14.5	μΑ
I _{IO}	I/O supply current Normal mode	Dominant	TXD = 0 V, STB= 0 V RXD floating		125	300	μΑ
I _{IO}	I/O supply current Normal mode	Recessive	TXD = 0 V, STB = 0 V RXD floating		25	48	μΑ
I _{IO}	I/O supply current Standby mode		TXD = 0 V, STB = V _{IO} RXD floating		8.5	13.5	μΑ
UV _{VCC}	Rising under voltage det	Itage detection on V _{CC} for protected mode			4.2	4.4	V
UV _{VCC}	Falling under voltage detection on V _{CC} for protected mode		3.5	4	4.25	V	
UV _{VIO}	Rising under voltage detection on V _{IO} (Devices with V _{IO})			1.56	1.65	V	
UV _{VIO}	Falling under voltage det	ection on V _{IO} (Device	es with V _{IO})	1.4	1.51	1.59	V



6.7 Dissipation Ratings

	PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
	Т	$V_{\rm CC}$ = 5 V, $V_{\rm IO}$ = 1.8 V, $T_{\rm J}$ = 27°C, $R_{\rm L}$ = 60 Ω , TXD input = 250 kHz 50% duty cycle square wave, $C_{\rm L_RXD}$ = 15 pF		110		mW
		$V_{\rm CC}$ = 5 V, $V_{\rm IO}$ = 3.3 V, $T_{\rm J}$ = 27°C, $R_{\rm L}$ = 60 Ω , TXD input = 250 kHz 50% duty cycle square wave, $C_{\rm L_RXD}$ = 15 pF		110		mW
D	Average power dissipation	V_{CC} = 5 V, V_{IO} = 5 V, T_J = 27°C, R_L = 60 Ω , TXD input = 250 kHz 50% duty cycle square wave, C_{L_RXD} = 15 pF		110		mW
P _D	Normal mode	V_{CC} = 5.5 V, V_{IO} = 1.8 V, T_A = 125°C, R_L = 60 Ω , TXD input = 2.5 MHz 50% duty cycle square wave, C_{L_RXD} = 15 pF		120		mW
		V_{CC} = 5.5 V, V_{IO} = 3.3 V, T_{A} = 125°C, R_{L} = 60 Ω , TXD input = 2.5 MHz 50% duty cycle square wave, C_{L_RXD} = 15 pF		120		mW
		V_{CC} = 5.5 V, V_{IO} = 5 V, T_{A} = 125°C, R_{L} = 60 Ω , TXD input = 2.5 MHz 50% duty cycle square wave, C_{L_RXD} = 15 pF		120		mW
T _{TSD}	Thermal shutdown temperature			192		°C
T _{TSD_HYS}	Thermal shutdown hysteresis			10		C



6.8 Electrical Characteristics

Over recommended operating conditions with T_A = -40°C to 125°C (unless otherwise noted)

	PARAMETER	Mill I A	TEST CONDITIONS	MIN	TYP MA	X UNIT
Driver Fleets			TEST CONDITIONS	IMIIA	IIF WA	A ONT
Driver Electr	ical Characteristics	CANH	TVD = 0.V CTD = 0.V 50.0 < D < 05.0	0.75		- V
$V_{O(DOM)}$	Dominant output voltage Normal mode	CANL	$ TXD = 0 \text{ V}, \text{ STB} = 0 \text{ V}, 50 \Omega \leq R_L \leq 65 \Omega,$ $ C_L = \text{ open, } R_{CM} = \text{ open}$ See Figure 7-2 and $ Figure 8-3 $,	0.5	2.2	
V _{O(REC)}	Recessive output voltage Normal mode	CANH and CANL	TXD = V _{IO} , STB = 0 V, R _L = open (no load), R _{CM} = open See Figure 7-2 and Figure 8-3	2		3 V
V _{SYM}	Driver symmetry (V _{O(CANL)})/V _{CC}		STB = 0 V, R_L = 60 Ω , C_{SPLIT} = 4.7 nF, C_L = open, R_{CM} = open, TXD = 250 kHz, 1 MHz, 2.5 MHz See Figure 7-2 and Figure 9-2	0.9	1.	1 V/V
V _{SYM_DC}	DC output symmetry (Vcc - Vo(CANH) - Vo(CANL))		STB = 0 V, R _L = 60 Ω, C _L = open See Figure 7-2 and Figure 8-3	-400	40	0 mV
			TXD = 0 V, STB = 0 V, 50 Ω \leq R _L \leq 65 Ω , C _L = open See Figure 7-2 and Figure 8-3	1.5		3 V
$V_{OD(DOM)}$	Differential output voltage Normal mode Dominant	CANH - CANL	TXD = 0 V, STB = 0 V, 45 Ω ≤ R _L ≤ 70 Ω , C _L = open See Figure 7-2 and Figure 8-3	1.4	3.	3 V
			TXD = 0 V, STB = 0 V, R _L = 2240 Ω , C _L = open See Figure 7-2 and Figure 8-3	1.5		5 V
$V_{OD(REC)}$	Differential output voltage Normal mode	CANH - CANL	TXD = V_{IO} , STB = 0 V, R_L = 60 Ω , C_L = open See Figure 7-2 and Figure 8-3	-120	1	2 mV
VOD(REC)	Recessive	OANT - OANE	$\begin{aligned} &TXD = V_{IO} \text{ , STB} = 0 \; V, \; R_{L} = open, \; C_{L} = \\ &open \\ &See \; Figure \; 7-2 \; and \; Figure \; 8-3 \end{aligned}$	-50	5	0 mV
		CANH		-0.1	0.	1 V
V _{O(STB)}	Bus output voltage Standby mode	CANL	STB = V _{IO} , R _L = open (no load) See Figure 7-2 and Figure 8-3	-0.1	0.	1 V
		CANH - CANL		-0.2	0.	2 V
laavaa aass	Short-circuit steady-state out	tput current,	STB = 0 V, V _(CANH) = -15 V to 40 V, CANL = open, TXD = 0 V See Figure 7-7 and Figure 8-3	-115		mA
Ios(ss_dom)	Normal mode		$ \begin{aligned} &STB = 0 \ V, \ V_{(CAN_L)} = -15 \ V \ to \ 40 \ V, \ CANH \\ &= open, \ TXD = 0 \ V \\ &See \ Figure \ 7-7 \ and \ Figure \ 8-3 \end{aligned} $		11	5 mA
I _{OS(SS_REC)}	Short-circuit steady-state our recessive Normal mode	tput current,	$\begin{split} &STB = 0 \; V, -27 \; V \leq V_{BUS} \leq 32 \; V, where \\ &V_{BUS} = CANH = CANL, \; TXD = V_{IO} \\ &See \; Figure \; 7-7 \; and \; Figure \; 8-3 \end{split}$	-5		5 mA
Receiver Ele	ctrical Characteristics					
V _{IT}	Input threshold voltage Normal mode		STB = 0 V, -12 V ≤ V _{CM} ≤ 12 V See Figure 7-3, Table 7-1, and Table 8-6	500	90	0 mV
V _{IT(STB)}	Input threshold Standby mode		STB = V_{IO} , -12 V \leq $V_{CM} \leq$ 12 V See Figure 7-3, Table 7-1, and Table 8-6	400	115	0 mV
V_{DOM}	Dominant state differential in Normal mode	put voltage range	STB = 0 V, -12 V ≤ V _{CM} ≤ 12 V See Figure 7-3, Table 7-1, and Table 8-6	0.9		9 V
V _{REC}	Recessive state differential input voltage range Normal mode		STB = 0 V, -12 V ≤ V _{CM} ≤ 12 V See Figure 7-3, Table 7-1, and Table 8-6	-4	0.	5 V
V _{DOM(STB)}	Dominant state differential input voltage range Standby mode		STB = V_{IO} , -12 V \leq $V_{CM} \leq$ 12 V See Figure 7-3, Table 7-1, and Table 8-6	1.15		9 V
V _{REC(STB)}	Recessive state differential in Standby mode	nput voltage range	STB = V_{IO} , -12 V \leq $V_{CM} \leq$ 12 V See Figure 7-3, Table 7-1, and Table 8-6	-4	0.	4 V
V _{HYS}	Hysteresis voltage for input t Normal mode	hreshold	STB = 0 V, -12 V ≤ V _{CM} ≤ 12 V See Figure 7-3, Table 7-1, and Table 8-6		100	mV
V _{CM}	Common mode range Normal and standby modes		See Figure 7-3 and Table 8-6 Table 8-6	-12	1	2 V
I _{LKG(IOFF)}	Unpowered bus input leakag	e current	CANH = CANL = 5 V, V _{CC} = V _{IO} = GND			5 µA



6.8 Electrical Characteristics (continued)

Over recommended operating conditions with $T_A = -40$ °C to 125°C (unless otherwise noted)

	PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
Cı	Input capacitance to ground (CANH or CANL)	TVD - V (1)			20	pF
C _{ID}	Differential input capacitance	$-TXD = V_{IO}^{(1)}$			10	pF
R _{ID}	Differential input resistance	TVD = V (1) CTD = 0.V 40.V 5V	40		90	kΩ
R _{IN}	Single ended input resistance (CANH or CANL)	$-$ TXD = V_{IO} (1), STB = 0 V, -12 V \leq V_{CM} \leq 12 V	20		45	kΩ
R _{IN(M)}	Input resistance matching [1 – (R _{IN(CANH)} / R _{IN(CANL)})] × 100 %	$V_{(CAN_H)} = V_{(CAN_L)} = 5 \text{ V}$	-1%		1%	
TXD Termin	nal (CAN Transmit Data Input)					
V _{IH}	High-level input voltage	Devices without V _{IO}	0.7 V _{CC}			V
V _{IH}	High-level input voltage	Devices with V _{IO}	0.7 V _{IO}			V
V _{IL}	Low-level input voltage	Devices without V _{IO}			0.3 V _{CC}	V
V _{IL}	Low-level input voltage	Devices with V _{IO}			0.3 V _{IO}	V
I _{IH}	High-level input leakage current	$TXD = V_{CC} = V_{IO} = 5.5 V$	-2.5	0	1	μA
I _{IL}	Low-level input leakage current	TXD = 0 V, V _{CC} = V _{IO} = 5.5 V	-200	-100	-20	μA
I _{LKG(OFF)}	Unpowered leakage current	TXD = 5.5 V, V _{CC} = V _{IO} = 0 V	-1	0	1	μA
Cı	Input Capacitance	$V_{IN} = 0.4 \times \sin(2 \times \pi \times 2 \times 10^6 \times t) + 2.5 \text{ V}$		5		pF
RXD Termin	nal (CAN Receive Data Output)					
V _{OH}	High-level output voltage	I _O = -2 mA, Devices without V _{IO} See Figure 7-3	0.8 V _{CC}			V
V _{OH}	High-level output voltage	I _O = -2 mA, Devices with V _{IO} See Figure 7-3	0.8 V _{IO}			V
V _{OL}	Low-level output voltage	I _O = 2 mA, Devices without V _{IO} See Figure 7-3			0.2 V _{CC}	V
V _{OL}	Low-level output voltage	I _O = -2 mA, Devices with V _{IO} See Figure 7-3			0.2 V _{IO}	V
I _{LKG(OFF)}	Unpowered leakage current	RXD = 5.5 V, V _{CC} = V _{IO} = 0 V	-1	0	1	μA
STB Termir	nal (Standby Mode Input)					
V _{IH}	High-level input voltage	Devices without V _{IO}	0.7 V _{CC}			V
V _{IH}	High-level input voltage	Devices with V _{IO}	0.7 V _{IO}			V
V _{IL}	Low-level input voltage	Devices without V _{IO}			0.3 V _{CC}	V
V _{IL}	Low-level input voltage	Devices with V _{IO}			0.3 V _{IO}	V
I _{LKG(OFF)}	Unpowered leakage current	STB = 5.5V, V _{CC} = V _{IO} = 0 V	-1	0	1	μA

⁽¹⁾ $V_{IO} = V_{CC}$ in non-V variants of device

6.9 Switching Characteristics

Over recommended operating conditions with $T_A = -40$ °C to 125°C (unless otherwise noted)

	PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
Device Switchin	ng Characteristics					
t _{PROP(LOOP1)}	Total loop delay, driver input (TXD) to receiver output (RXD), recessive to dominant	Normal mode, R_L = 60 Ω , C_L = 100 pF, $C_{L(RXD)}$ = 15 pF V_{IO} = 2.8 V to 5.5 V See Figure 7-4		125	210	ns
t _{PROP(LOOP1)}	Total loop delay, driver input (TXD) to receiver output (RXD), recessive to dominant	Normal mode, R _L = $60~\Omega$, C _L = $100~pF$, C _{L(RXD)} = $15~pF$ V _{IO} = $1.7~V$ See Figure 7-4		165	255	ns
t _{PROP(LOOP2)}	Total loop delay, driver input (TXD) to receiver output (RXD), dominant to recessive	Normal mode, R_L = 60 Ω , C_L = 100 pF, $C_{L(RXD)}$ = 15 pF V_{IO} = 2.8 V to 5.5 V See Figure 7-4		150	210	ns
t _{PROP(LOOP2)}	Total loop delay, driver input (TXD) to receiver output (RXD), dominant to recessive	Normal mode, R _L = 60 Ω , C _L = 100 pF, C _{L(RXD)} = 15 pF V _{IO} = 1.7 V See Figure 7-4		180	255	ns

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6.9 Switching Characteristics (continued)

Over recommended operating conditions with T_A = -40°C to 125°C (unless otherwise noted)

	PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
t _{MODE}	Mode change time, from normal to standby or from standby to normal	See Figure 7-5			20	μs
t _{WK_FILTER}	Filter time for a valid wake-up pattern	See Figure 8-5	0.5		1.8	μs
t _{WK_TIMEOUT}	Bus wake-up timeout	See Figure 8-5	0.8		6	ms
Driver Switchin	ng Characteristics		•		'	
t _{pHR}	Propagation delay time, high TXD to driver recessive (dominant to recessive)			80		ns
t _{pLD}	Propagation delay time, low TXD to driver dominant (recessive to dominant)			70		ns
t _{sk(p)}	Pulse skew (tpHR - tpLD)	STB = 0 V, R_L = 60 Ω , C_L = 100 pF See Figure 7-2 and Figure 7-6		20		ns
t _R	Differential output signal rise time			30		ns
t _F	Differential output signal fall time			50		ns
t _{TXD_DTO}	Dominant timeout		1.2		4.0	ms
Receiver Switch	ching Characteristics					
t _{pRH}	Propagation delay time, bus recessive input to high output (dominant to recessive)			90		ns
t _{pDL}	Propagation delay time, bus dominant input to low output (recessive to dominant)	STB = 0 V, C _{L(RXD)} = 15 pF See Figure 7-3		65		ns
t _R	RXD output signal rise time			10		ns
t _F	RXD output signal fall time	10			ns	
FD Timing Cha	racteristics					
t _{BIT(BUS)}	Bit time on CAN bus output pins $t_{BIT(TXD)} = 500 \text{ ns}$		450		530	ns
t _{BIT(BUS)}	Bit time on CAN bus output pins $t_{BIT(TXD)} = 200 \text{ ns}$		155		210	ns
t _{BIT(RXD)}	Bit time on RXD output pins t _{BIT(TXD)} = 500 ns	STB = 0 V, R_L = 60 Ω , C_L = 100 pF, $C_{L(RXD)}$ = 15 pF	400		550	ns
t _{BIT(RXD)}	Bit time on RXD output pins t _{BIT(TXD)} = 200 ns	Atrec = t _{BIT(RXD)} - t _{BIT(BUS)} See Figure 7-4 120			220	ns
t _{REC}	Receiver timing symmetry t _{BIT(TXD)} = 500 ns		-50		20	ns
t _{REC}	Receiver timing symmetry t _{BIT(TXD)} = 200 ns		-45		15	ns

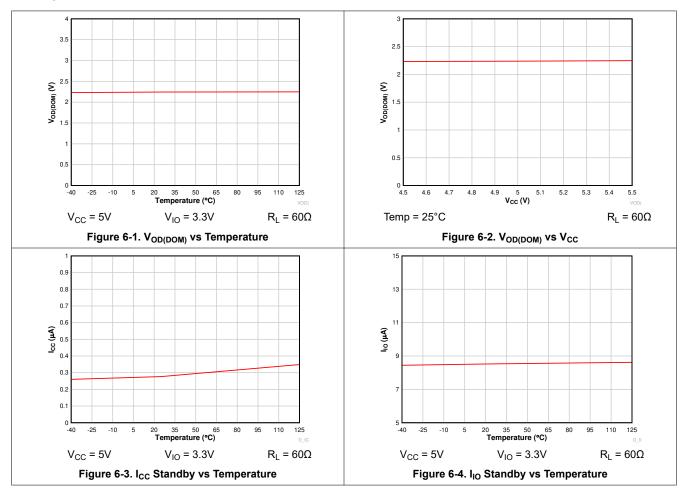
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6.10 Typical Characteristics



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7 Parameter Measurement Information

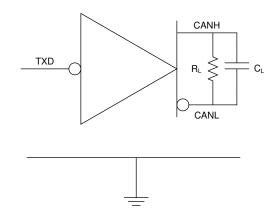


Figure 7-1. I_{CC} Test Circuit

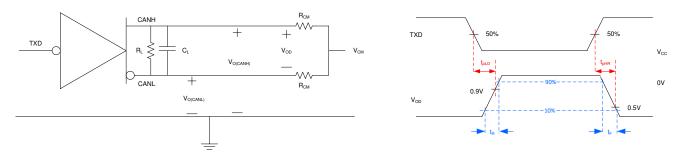


Figure 7-2. Driver Test Circuit and Measurement

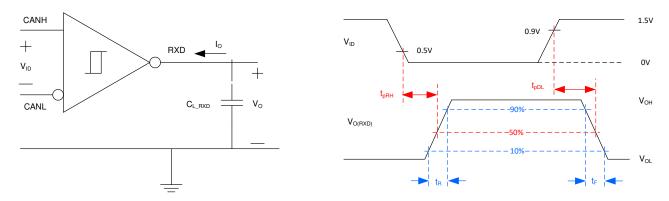


Figure 7-3. Receiver Test Circuit and Measurement



Table 7-1. Receiver Differential Input Voltage Threshold Test

	Input (See Figure 7-3)		Out	tput
V _{CANH}	V_{CANL}	V _{ID}	R	KD
-11.5V	-12.5V	1000mV		
12.5V	11.5V	1000mV	Low	V
-8.55V	-9.45V	900mV	Low	V _{OL}
9.45 V	8.55V	900mV		
-8.75V	-9.25V	500mV		
9.25V	8.75V	500mV		
-11.8 V	-12.2V	400mV	High	V _{OH}
12.2V	11.8V	400mV		
Open	Open	X		

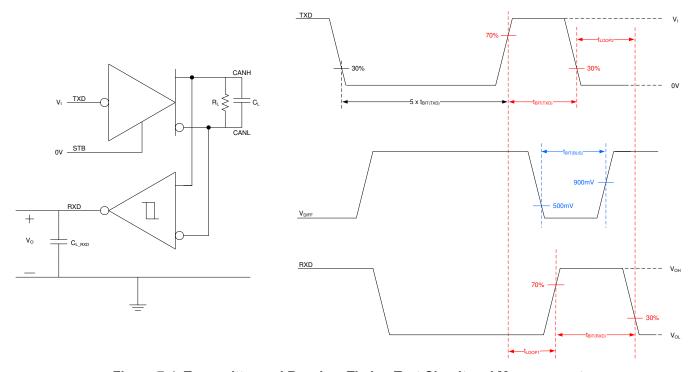


Figure 7-4. Transmitter and Receiver Timing Test Circuit and Measurement



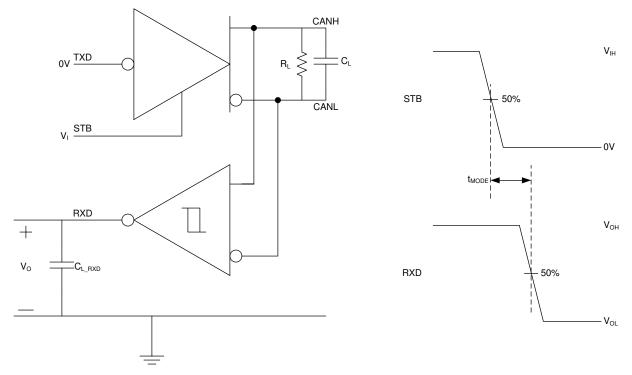


Figure 7-5. t_{MODE} Test Circuit and Measurement

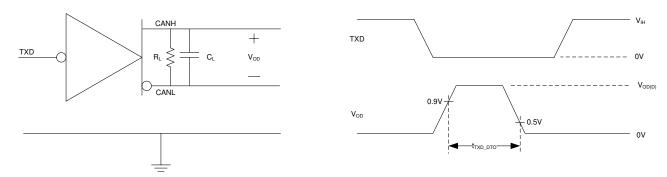


Figure 7-6. TXD Dominant Timeout Test Circuit and Measurement

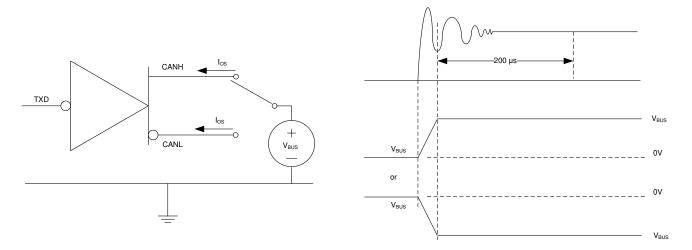


Figure 7-7. Driver Short-Circuit Current Test and Measurement

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8 Detailed Description

8.1 Overview

The TCAN1044-Q1 meets or exceeds the specifications of the ISO 11898-2:2016 high speed CAN (Controller Area Network) physical layer standard. The device has been certified to the requirements of ISO 11898-2:2016 and ISO 11898-5:2007 physical layer requirements according to the GIFT/ICT high speed CAN test specification. The transceiver provides a number of different protection features making it ideal for the stringent automotive system requirements while also supporting CAN FD data rates up to 8Mbps.

The TCAN1044-Q1 conforms to the following CAN standards:

- CAN transceiver physical layer standards:
 - ISO 11898-2:2016 High speed medium access unit
 - ISO 11898-5:2007 High speed medium access unit with low-power mode
 - SAE J2284-1: High Speed CAN (HSC) for Vehicle Applications at 125kbps
 - SAE J2284-2: High Speed CAN (HSC) for Vehicle Applications at 250kbps
 - SAE J2284-3: High Speed CAN (HSC) for Vehicle Applications at 500kbps
 - SAE J2284-4: High-Speed CAN (HSC) for Vehicle Applications at 500kbps with CAN FD Data at 2Mbps
 - SAE J2284-5: High-Speed CAN (HSC) for Vehicle Applications at 500kbps with CAN FD Data at 5Mbps
 - ARINC 825-4 General Standardization of CAN (Controller Area Network) Bus Protocol For Airborne Use
- EMC requirements:
 - VeLIO (Vehicle LAN Interoperability and Optimization) CAN and CAN-FD Transceiver Requirements
 - SAE J2962-2 Communication Transceivers Qualification Requirements CAN
- Conformance test requirements:
 - ISO 16845-2 Road vehicles Controller area network (CAN) conformance test plan Part 2: High-speed medium access unit conformance test plan



8.2 Functional Block Diagram

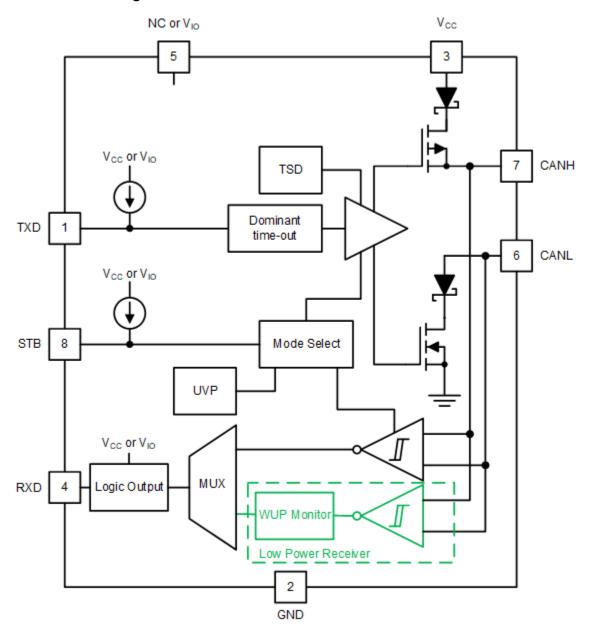


Figure 8-1. Block Diagram



8.3 Feature Description

8.3.1 Pin Description

8.3.1.1 TXD

The TXD input is a logic-level signal, referenced to either V_{CC} or V_{IO} from a CAN controller to the TCAN1044-Q1 transceivers

8.3.1.2 GND

GND is the ground pin of the transceiver, it must be connected to the PCB ground.

8.3.1.3 V_{CC}

V_{CC} provides the 5V power supply to the CAN transceiver.

8.3.1.4 RXD

The RXD output is a logic-level signal, referenced to either V_{CC} or V_{IO} , from the TCAN1044-Q1 transceivers to the CAN controller. RXD is only driven once V_{IO} is present.

When a wake event takes place RXD is driven low.

8.3.1.5 V_{IO}

The V_{IO} pin provides the digital I/O voltage to match the CAN controller voltage thus avoiding the requirement for a level shifter. It supports voltages from 1.7V to 5.5V providing the widest range of controller support.

8.3.1.6 CANH and CANL

These are the CAN high and CAN low differential bus pins. These pins are connected to the CAN transceiver and the low-voltage WUP CAN receiver.

8.3.1.7 STB (Standby)

The STB pin is an input pin used for mode control of the transceiver. The STB pin can be supplied from either the system processor or from a static system voltage source. If normal mode is the only intended mode of operation, then the STB pin can be tied directly to GND.

8.3.2 CAN Bus States

The CAN bus has two logical states during operation: recessive and dominant. See Figure 8-2 and Figure 8-3.

A dominant bus state occurs when the bus is driven differentially and corresponds to a logic low on the TXD and RXD pins. A recessive bus state occurs when the bus is biased to V_{CC}/2 via the high-resistance internal input resistors R_{IN}) of the receiver and corresponds to a logic high on the TXD and RXD pins.

A dominant state overwrites the recessive state during arbitration. Multiple CAN nodes may be transmitting a dominant bit at the same time during arbitration, and in this case the differential voltage of the bus is greater than the differential voltage of a single driver.

The TCAN1044-Q1 transceiver implements a low-power standby (STB) mode which enables a third bus state where the bus pins are weakly biased to ground via the high resistance internal resistors of the receiver. See Figure 8-2 and Figure 8-3.

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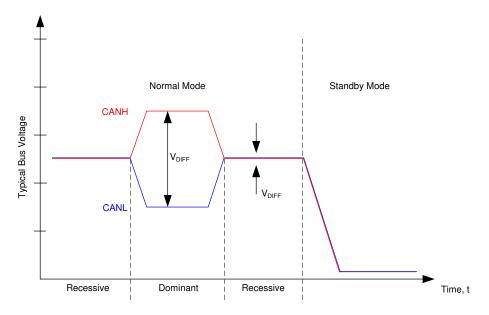
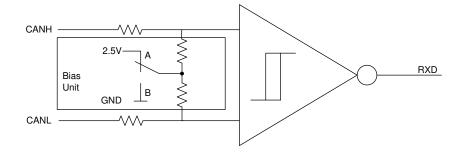


Figure 8-2. Bus States



- A. Normal Mode
- B. Standby Mode

Figure 8-3. Simplified Recessive Common Mode Bias Unit and Receiver

8.3.3 TXD Dominant Timeout (DTO)

During normal mode, the only mode where the CAN driver is active, the TXD DTO circuit prevents the local node from blocking network communication in the event of a hardware or software failure where TXD is held dominant longer than the timeout period t_{TXD_DTO} . The TXD DTO circuit is triggered by a falling edge on TXD. If no rising edge is seen before the timeout period of the circuit, t_{TXD_DTO} , the CAN driver is disabled. This frees the bus for communication between other nodes on the network. The CAN driver is reactivated when a recessive signal is seen on the TXD pin, thus clearing the dominant timeout. The receiver remains active and biased to $V_{CC}/2$ and the RXD output reflects the activity on the CAN bus during the TXD DTO fault.

The minimum dominant TXD time allowed by the TXD DTO circuit limits the minimum possible transmitted data rate of the device. The CAN protocol allows a maximum of eleven successive dominant bits (on TXD) for the worst case, where five successive dominant bits are followed immediately by an error frame. The minimum transmitted data rate may be calculated using Equation 1.

Minimum Data Rate = 11 bits /
$$t_{TXD}$$
 DTO = 11 bits / 1.2 ms = 9.2 kbps (1)



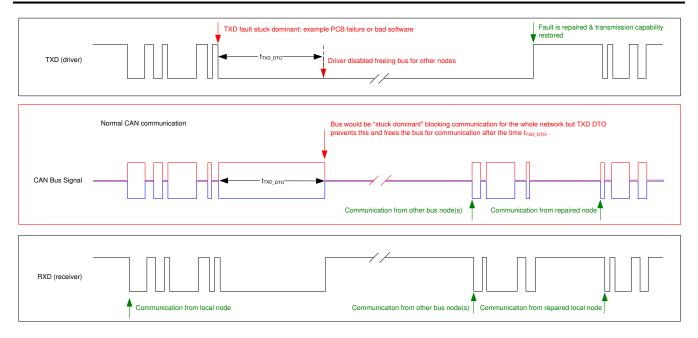


Figure 8-4. Example Timing Diagram for TXD Dominant Timeout

8.3.4 CAN Bus Short Circuit Current Limiting

The TCAN1044-Q1 has several protection features that limit the short circuit current when a CAN bus line is shorted. These include CAN driver current limiting in the dominant and recessive states and TXD dominant state timeout which prevents permanently having the higher short circuit current of a dominant state in case of a system fault. During CAN communication the bus switches between the dominant and recessive states, thus the short circuit current may be viewed as either the current during each bus state or as a DC average current. When selecting termination resistors or a common mode choke for the CAN design the average power rating, I_{OS(AVG)}, should be used. The percentage dominant is limited by the TXD DTO and the CAN protocol which has forced state changes and recessive bits due to bit stuffing, control fields, and interframe space. These ensure there is a minimum amount of recessive time on the bus even if the data field contains a high percentage of dominant bits.

The average short circuit current of the bus depends on the ratio of recessive to dominant bits and their respective short circuit currents. The average short circuit current may be calculated using Equation 2.

Where:

- I_{OS(AVG)} is the average short circuit current
- % Transmit is the percentage the node is transmitting CAN messages
- % Receive is the percentage the node is receiving CAN messages
- % REC_Bits is the percentage of recessive bits in the transmitted CAN messages
- % DOM_Bits is the percentage of dominant bits in the transmitted CAN messages
- I_{OS(SS) REC} is the recessive steady state short circuit current
- I_{OS(SS)} DOM is the dominant steady state short circuit current

This short circuit current and the possible fault cases of the network should be taken into consideration when sizing the power supply used to generate the transceivers V_{CC} supply.

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8.3.5 Thermal Shutdown (TSD)

If the junction temperature of the TCAN1044-Q1 exceeds the thermal shutdown threshold, T_{TSD} , the device turns off the CAN driver circuitry and blocks the TXD to bus transmission path. The shutdown condition is cleared when the junction temperature of the device drops below T_{TSD} . The CAN bus pins are biased to $V_{CC}/2$ during a TSD fault and the receiver to RXD path remains operational. The TCAN1044-Q1 TSD circuit includes hysteresis which prevents the CAN driver output from oscillating during a TSD fault.

8.3.6 Undervoltage Lockout

The supply pins, V_{CC} and V_{IO} , have undervoltage detection that places the device into a protected state. This protects the bus during an undervoltage event on either supply pin.

Table 8-1. Undervoltage Lockout - TCAN1044-Q1

V _{cc}	DEVICE STATE	BUS	RXD PIN
> UV _{VCC}	Normal	Per TXD	Mirrors bus
< UV _{VCC}	Protected	High impedance Weak pull-down to ground ⁽¹⁾	High impedance

(1) $V_{CC} = GND$, see $I_{LKG(OFF)}$

Table 8-2. Undervoltage Lockout - TCAN1044V-Q1

V _{cc}	V _{IO}	DEVICE STATE	BUS	RXD PIN
> UV _{VCC}	> UV _{VIO}	Normal	Per TXD	Mirrors bus
< 11V	> UV _{VIO}	STB = V _{IO} : standby mode		V _{IO} : Remote wake request ⁽²⁾
< UV _{VCC}	- O V VIO	STB = GND: Protected	High impedance Weak pull-down to	Recessive
> UV _{VCC}	< UV _{VIO}	Protected	ground ⁽¹⁾	High impedance
< UV _{VCC}	< UV _{VIO}	Protected	_	High impedance

V_{CC} = GND, see I_{LKG(OFF)}

Once the undervoltage condition is cleared and t_{MODE} has expired the TCAN1044-Q1 transitions to normal mode and the host controller can send and receive CAN traffic again.

8.3.7 Unpowered Device

The TCAN1044-Q1 is designed to be a suitable passive or no load to the CAN bus if the device is unpowered. The bus pins were designed to have low leakage currents when the device is unpowered, so the pins do not load the bus. This is critical if some nodes of the network are unpowered while the rest of the of network remains operational.

The logic pins also have low leakage currents when the device is unpowered, so the pins do not load other circuits which may remain powered.

8.3.8 Floating pins

The TCAN1044-Q1 has internal pull-ups on critical pins which place the device into known states if the pin floats. This internal bias should not be relied upon by design though, especially in noisy environments, but instead should be considered a failsafe protection feature.

When a CAN controller supporting open-drain outputs is used an adequate external pull-up resistor must be chosen. This makes sure the TXD output of the CAN controller maintains acceptable bit time to the input of the CAN transceiver. See Table 8-3 for details on pin bias conditions.

Table 8-3. Pin Bias

Pin	Pull-up or Pull-down	Comment
TXD	Pull-up	Weakly biases TXD towards recessive to prevent bus blockage or TXD DTO triggering
STB	Pull-up	Weakly biases STB towards low-power standby mode to prevent excessive system power

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⁽²⁾ See Section 8.4.3.1

8.4 Device Functional Modes

8.4.1 Operating Modes

The TCAN1044-Q1 has two main operating modes: normal mode and standby mode. Operating mode selection is made by applying a high or low level to the STB pin on the TCAN1044-Q1.

Table 8-4. Operating Modes

STB	Device Mode	Driver	Receiver	RXD Pin
High	Low current standby mode with bus wake-up	Disabled	Low-power receiver and bus monitor enable	High (recessive) until valid WUP is received See Section 8.4.3.1
Low	Normal Mode	Enabled	Enabled	Mirrors bus state

8.4.2 Normal Mode

This is the normal operating mode of the TCAN1044-Q1. The CAN driver and receiver are fully operational and CAN communication is bi-directional. The driver is translating a digital input on the TXD input to a differential output on the CANH and CANL bus pins. The receiver is translating the differential signal from CANH and CANL to a digital output on the RXD output.

8.4.3 Standby Mode

This is the low-power mode of the TCAN1044-Q1. The CAN driver and main receiver are switched off and bi-directional CAN communication is not possible. The low-power receiver and bus monitor circuits are enabled to allow for RXD wake-up requests via the CAN bus. A wake-up request is output to RXD as shown in Figure 8-5. The local CAN protocol controller should monitor RXD for transitions (high-to-low) and reactivate the device to normal mode by pulling the STB pin low. The CAN bus pins are weakly pulled to GND in this mode; see Figure 8-2 and Figure 8-3.

In standby mode, only the V_{IO} supply is required therefore the V_{CC} may be switched off for additional system level current savings.

8.4.3.1 Remote Wake Request via Wake-Up Pattern (WUP) in Standby Mode

The TCAN1044-Q1 supports a remote wake-up request that is used to indicate to the host controller that the bus is active and the node should return to normal operation.

The device uses the multiple filtered dominant wake-up pattern (WUP) from the ISO 11898-2:2016 standard to qualify bus activity. Once a valid WUP has been received, the wake request is indicated to the controller by a falling edge and low period corresponding to a filtered dominant on the RXD output of the TCAN1044-Q1.

The WUP consists of a filtered dominant pulse, followed by a filtered recessive pulse, and finally by a second filtered dominant pulse. The first filtered dominant initiates the WUP, and the bus monitor then waits on a filtered recessive; other bus traffic does not reset the bus monitor. Once a filtered recessive is received the bus monitor is waiting for a filtered dominant and again, other bus traffic does not reset the bus monitor. Immediately upon reception of the second filtered dominant the bus monitor recognizes the WUP and drives the RXD output low every time an additional filtered dominant signal is received from the bus.

For a dominant or recessive to be considered filtered, the bus must be in that state for more than the t_{WK_FILTER} time. Due to variability in t_{WK_FILTER} the following scenarios are applicable. Bus state times less than $t_{WK_FILTER(MIN)}$ are never detected as part of a WUP and thus no wake request is generated. Bus state times between $t_{WK_FILTER(MIN)}$ and $t_{WK_FILTER(MAX)}$ may be detected as part of a WUP and a wake-up request may be generated. Bus state times greater than $t_{WK_FILTER(MAX)}$ are always detected as part of a WUP, and thus a wake request is always generated. See Figure 8-5 for the timing diagram of the wake-up pattern.

The pattern and t_{WK_FILTER} time used for the WUP prevents noise and bus stuck dominant faults from causing false wake-up requests while allowing any valid message to initiate a wake-up request.

The ISO 11898-2:2016 standard has defined times for a short and long wake-up filter time. The t_{WK_FILTER} timing for the device has been picked to be within the minimum and maximum values of both filter ranges. This timing

has been chosen such that a single bit time at 500kbps, or two back-to-back bit times at 1Mbps triggers the filter in either bus state. Any CAN frame at 500kbps or less would contain a valid WUP.

For an additional layer of robustness and to prevent false wake-ups, the device implements a wake-up timeout feature. For a remote wake-up event to successfully occur, the entire WUP must be received within the timeout value $t \le t_{WK_TIMEOUT}$. If not, the internal logic is reset and the transceiver remains in its current state without waking up. The full pattern must then be transmitted again, conforming to the constraints mentioned in this section. See Figure 8-5 for the timing diagram of the wake-up pattern with wake timeout feature.

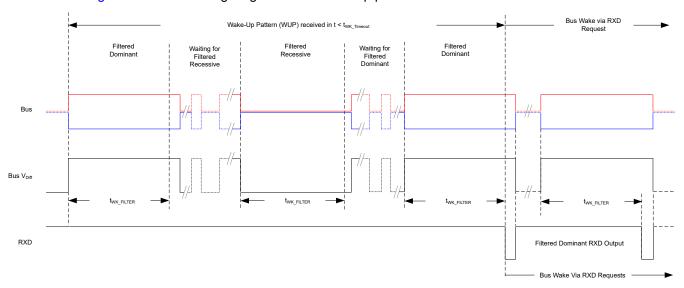


Figure 8-5. Wake-Up Pattern (WUP) with twk TIMEOUT



8.4.4 Driver and Receiver Function

The digital logic input and output levels for the TCAN1044-Q1 are CMOS levels with respect to either V_{CC} for 5V systems or V_{IO} for compatible with MCUs having 1.8V, 2.5V, 3.3V, or 5V systems.

Table 8-5. Driver Function Table

Device Mode	TXD Input ⁽¹⁾	Bus	Driven Bus State ⁽²⁾	
Device Mode	I AD IIIput	CANH	CANL	Driven bus State
Normal	Low	High	Low	Dominant
Nomai	High or open	High impedance	High impedance	Biased recessive
Standby	Х	High impedance	High impedance	Biased to ground

⁽¹⁾ X = irrelevant

Table 8-6. Receiver Function Table Normal and Standby Mode

Device Mode	CAN Differential Inputs V _{ID} = V _{CANH} - V _{CANL}	Bus State	RXD Pin
	V _{ID} ≥ 0.9V	Dominant	Low
Normal	0.5V < V _{ID} < 0.9V	Undefined	Undefined
	V _{ID} ≤ 0.5V	Recessive	High
	V _{ID} ≥ 1.15V	Dominant	High
Standby	0.4V < V _{ID} < 1.15V	Undefined	Low if a remote wake event occurred
	V _{ID} ≤ 0.4V	Recessive	See Figure 8-5
Any	Open (V _{ID} ≈ 0V)	Open	High

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⁽²⁾ For bus state and bias see Figure 8-2 and Figure 8-3

9 Application and Implementation

Note

Information in the following applications sections is not part of the TI component specification, and TI does not warrant its accuracy or completeness. TI's customers are responsible for determining suitability of components for their purposes, as well as validating and testing their design implementation to confirm system functionality.

9.1 Application Information

9.2 Typical Application

The TCAN1044-Q1 transceiver can be used in applications with a host controller or FPGA that includes the link layer portion of the CAN protocol. Figure 9-1 shows a typical configuration for 5V controller applications. The bus termination is shown for illustrative purposes.

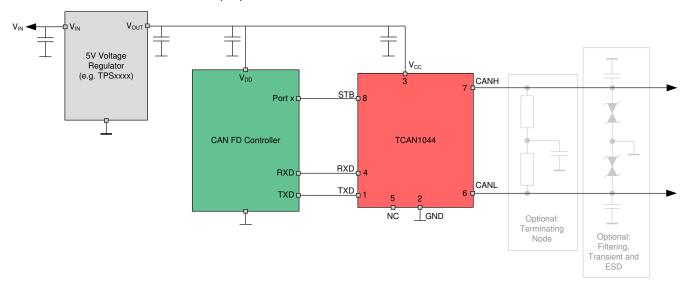


Figure 9-1. Transceiver Application Using 5V I/O Connections

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9.2.1 Design Requirements

9.2.1.1 CAN Termination

Termination may be a single 120Ω resistor at each end of the bus, either on the cable or in a terminating node. If filtering and stabilization of the common-mode voltage of the bus is desired then split termination may be used, see Figure 9-2. Split termination improves the electromagnetic emissions behavior of the network by filtering higher-frequency common-mode noise that may be present on the differential signal lines.

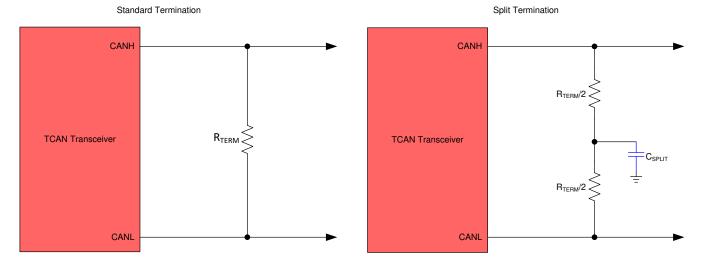


Figure 9-2. CAN Bus Termination Concepts

9.2.2 Detailed Design Procedures

9.2.2.1 Bus Loading, Length and Number of Nodes

A typical CAN application may have a maximum bus length of 40 meters and maximum stub length of 0.3m. However, with careful design, users can have longer cables, longer stub lengths, and many more nodes to a bus. A high number of nodes requires a transceiver with high input impedance such as the TCAN1044-Q1.

Many CAN organizations and standards have scaled the use of CAN for applications outside the original ISO 11898-2 standard. They made system level trade off decisions for data rate, cable length, and parasitic loading of the bus. Examples of these CAN systems level specifications are ARINC 825, CANopen, DeviceNet, SAE J2284, SAE J1939, and NMEA 2000.

A CAN network system design is a series of tradeoffs. In the ISO 11898-2:2016 specification the driver differential output is specified with a bus load that can range from 50Ω to 65Ω where the differential output must be greater than 1.5V. The TCAN1044-Q1 family is specified to meet the 1.5V requirement down to 50Ω and is specified to meet 1.4V differential output at 45Ω bus load. The differential input resistance of the TCAN1044-Q1 is a minimum of $40k\Omega$. If 100 TCAN1044-Q1 transceivers are in parallel on a bus, this is equivalent to a 400Ω differential load in parallel with the nominal 60Ω bus termination which gives a total bus load of approximately 52Ω . Therefore, the TCAN1044-Q1 family theoretically supports over 100 transceivers on a single bus segment. However, for a CAN network design margin must be given for signal loss across the system and cabling, parasitic loadings, timing, network imbalances, ground offsets and signal integrity; thus, a practical maximum number of nodes is often lower. Bus length may also be extended beyond 40 meters by careful system design and data rate tradeoffs. For example, CANopen network design guidelines allow the network to be up to 1 km with changes in the termination resistance, cabling, less than 64 nodes and significantly lowered data rate.

This flexibility in CAN network design is one of the key strengths of the various extensions and additional standards that have been built on the original ISO 11898-2 CAN standard. However, when using this flexibility, the CAN network system designer must take the responsibility of good network design for a robust network operation.

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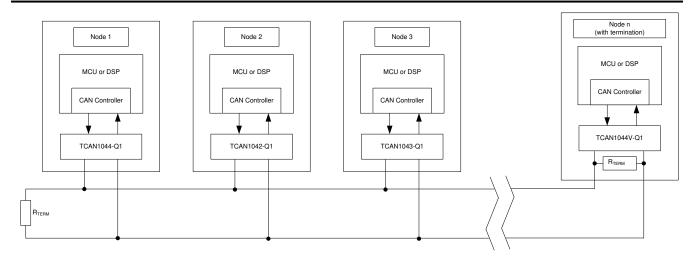
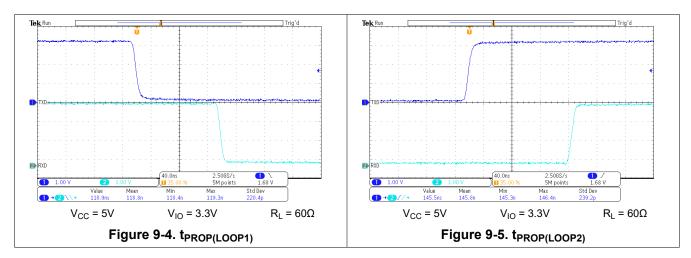


Figure 9-3. Typical CAN Bus

9.2.3 Application Curves





9.3 System Examples

The TCAN1044-Q1 CAN transceiver is typically used in applications with a host controller or FPGA that includes the link layer portion of the CAN protocol. A 1.8V, 2.5V, or 3.3V application is shown in Figure 9-6. The bus termination is shown for illustrative purposes.

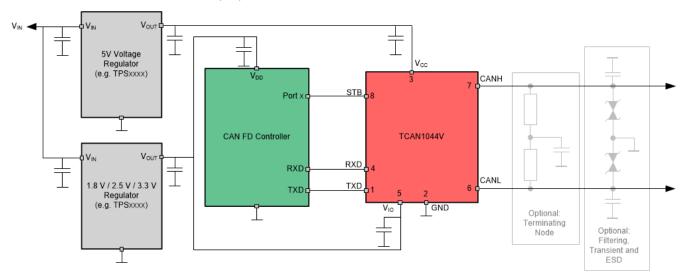


Figure 9-6. Typical Transceiver Application Using 1.8V, 2.5V, 3.3V IO Connections

9.4 Power Supply Recommendations

The TCAN1044-Q1 transceiver is designed to operate with a main V_{CC} input voltage supply range between 4.5V and 5.5V. The TCAN1044V-Q1 implements an IO level shifting supply input, V_{IO} , designed for a range between 1.8V and 5.5V. Both supply inputs must be well regulated. A decoupling capacitance, typically 100nF, should be placed near the CAN transceiver main V_{CC} supply pin in addition to bypass capacitors. A decoupling capacitor, typically 100nF, should be placed near the CAN transceiver V_{IO} supply pin in addition to bypass capacitors.

9.5 Layout

Robust and reliable CAN node design may require special layout techniques depending on the application and automotive design requirements. Since transient disturbances have high frequency content and a wide bandwidth, high-frequency layout techniques should be applied during PCB design.

9.5.1 Layout Guidelines

- Place the protection and filtering circuitry close to the bus connector, J1, to prevent transients, ESD, and
 noise from propagating onto the board. This layout example shows an optional transient voltage suppression
 (TVS) diode, D1, which may be implemented if the system-level requirements exceed the specified rating of
 the transceiver. This example also shows optional bus filter capacitors C4 and C5.
- Design the bus protection components in the direction of the signal path. Do not force the transient current to divert from the signal path to reach the protection device.
- Decoupling capacitors should be placed as close as possible to the supply pins V_{CC} and V_{IO} of transceiver.
- Use at least two vias for supply and ground connections of bypass capacitors and protection devices to minimize trace and via inductance.

Note

High frequency current follows the path of least impedance and not the path of least resistance.

• This layout example shows how split termination could be implemented on the CAN node. The termination is split into two resistors, R6 and R7, with the center or split tap of the termination connected to ground via capacitor C3. Split termination provides common mode filtering for the bus. See Section 9.2.1.1, Section

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- 8.3.4, and Equation 2 for information on termination concepts and power ratings needed for the termination resistor(s).
- To limit current, digital lines series resistors can be used. Examples are R2, R3 and R4.
- Pin 1 is shown for the TXD input of the device with R1 as an optional pull-up resistor. If an open drain host controller is used, making sure the bit timing into the device is met is mandatory.
- Pin 8 is shown with R4 assuming the mode pin STB, is used. If the device is used in normal mode only, R4 is not needed and the pads of C4 could be used for the pull down resistor R5 to GND.

9.5.2 Layout Example

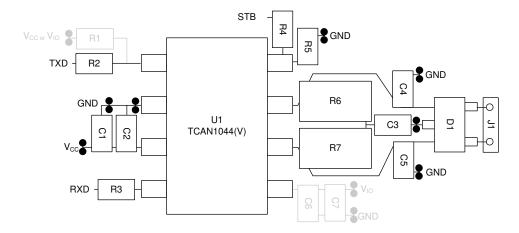


Figure 9-7. Layout Example



10 Device and Documentation Support

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This integrated circuit can be damaged by ESD. Texas Instruments recommends that all integrated circuits be handled with appropriate precautions. Failure to observe proper handling and installation procedures can cause damage.

ESD damage can range from subtle performance degradation to complete device failure. Precision integrated circuits may be more susceptible to damage because very small parametric changes could cause the device not to meet its published specifications.

10.5 Glossary

TI Glossary

This glossary lists and explains terms, acronyms, and definitions.

11 Revision History

NOTE: Page numbers for previous revisions may differ from page numbers in the current version.

Changes from Revision C (October 2024) to Revision D (March 2025)	
 Added new text to the second paragraph in the <i>Description</i> Added the <i>Device Comparison</i> table 	1 2
Changes from Revision B (October 2021) to Revision C (October 2024)	Page
Changed Feature: "Available in SOIC"	1
Deleted part number TCAN1044V-Q1 from the data sheet title and header information	
Changed the Device Information table to the Package Information table	1
Changes from Revision A (December 2019) to Revision B (October 2021)	Page
Added Feature "Functional Safety-Capable"	1
Changed the Simplified Schematic image	1
Changed Figure 9-2	24

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C	hanges from Revision * (August 2019) to Revision A (December 2019)	Page
•	First public release of the data sheet	1
	Added SAE j2962-2 ESD	
•	Changed footnote to Tested according to IEC 62228-3:2019 CAN Transceivers, Section 6.3; standard parameters defined in ISO 7637-2 (2011)	oulses

12 Mechanical, Packaging, and Orderable Information

The following pages include mechanical, packaging, and orderable information. This information is the most current data available for the designated devices. This data is subject to change without notice and revision of this document. For browser-based versions of this data sheet, refer to the left-hand navigation.

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PACKAGING INFORMATION

Orderable part number Status		Material type	Package Pins	Package qty Carrier	RoHS	Lead finish/ Ball material	MSL rating/ Peak reflow	Op temp (°C)	Part marking	
	(1)	(2)			(3)	(4)	(5)		(6)	
TCAN1044DRBRQ1	Active	Production	SON (DRB) 8	3000 LARGE T&R	Yes	NIPDAU	Level-2-260C-1 YEAR	-40 to 125	1044	
TCAN1044DRBRQ1.A	Active	Production	SON (DRB) 8	3000 LARGE T&R	Yes	NIPDAU	Level-2-260C-1 YEAR	-40 to 125	1044	
TCAN1044DRQ1	Active	Production	SOIC (D) 8	2500 LARGE T&R	Yes	NIPDAU	Level-1-260C-UNLIM	-40 to 125	1044	
TCAN1044DRQ1.A	Active	Production	SOIC (D) 8	2500 LARGE T&R	Yes	NIPDAU	Level-1-260C-UNLIM	-40 to 125	1044	
TCAN1044VDDFRQ1	Active	Production	SOT-23-THIN (DDF) 8	3000 LARGE T&R	Yes	NIPDAU	Level-1-260C-UNLIM	-40 to 125	26SF	
TCAN1044VDDFRQ1.A	Active	Production	SOT-23-THIN (DDF) 8	3000 LARGE T&R	Yes	NIPDAU	Level-1-260C-UNLIM	-40 to 125	26SF	
TCAN1044VDRBRQ1	Active	Production	SON (DRB) 8	3000 LARGE T&R	Yes	NIPDAU	Level-2-260C-1 YEAR	-40 to 125	1044V	
TCAN1044VDRBRQ1.A	Active	Production	SON (DRB) 8	3000 LARGE T&R	Yes	NIPDAU	Level-2-260C-1 YEAR	-40 to 125	1044V	
TCAN1044VDRQ1	Active	Production	SOIC (D) 8	2500 LARGE T&R	Yes	NIPDAU	Level-1-260C-UNLIM	-40 to 125	1044V	
TCAN1044VDRQ1.A	Active	Production	SOIC (D) 8	2500 LARGE T&R	Yes	NIPDAU	Level-1-260C-UNLIM	-40 to 125	1044V	

⁽¹⁾ Status: For more details on status, see our product life cycle.

Multiple part markings will be inside parentheses. Only one part marking contained in parentheses and separated by a "~" will appear on a part. If a line is indented then it is a continuation of the previous line and the two combined represent the entire part marking for that device.

Important Information and Disclaimer: The information provided on this page represents TI's knowledge and belief as of the date that it is provided. TI bases its knowledge and belief on information provided by third parties, and makes no representation or warranty as to the accuracy of such information. Efforts are underway to better integrate information from third parties. TI has taken and continues to take reasonable steps to provide representative

⁽²⁾ Material type: When designated, preproduction parts are prototypes/experimental devices, and are not yet approved or released for full production. Testing and final process, including without limitation quality assurance, reliability performance testing, and/or process qualification, may not yet be complete, and this item is subject to further changes or possible discontinuation. If available for ordering, purchases will be subject to an additional waiver at checkout, and are intended for early internal evaluation purposes only. These items are sold without warranties of any kind.

⁽³⁾ RoHS values: Yes, No, RoHS Exempt. See the TI RoHS Statement for additional information and value definition.

⁽⁴⁾ Lead finish/Ball material: Parts may have multiple material finish options. Finish options are separated by a vertical ruled line. Lead finish/Ball material values may wrap to two lines if the finish value exceeds the maximum column width.

⁽⁵⁾ MSL rating/Peak reflow: The moisture sensitivity level ratings and peak solder (reflow) temperatures. In the event that a part has multiple moisture sensitivity ratings, only the lowest level per JEDEC standards is shown. Refer to the shipping label for the actual reflow temperature that will be used to mount the part to the printed circuit board.

⁽⁶⁾ Part marking: There may be an additional marking, which relates to the logo, the lot trace code information, or the environmental category of the part.



PACKAGE OPTION ADDENDUM

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and accurate information but may not have conducted destructive testing or chemical analysis on incoming materials and chemicals. TI and TI suppliers consider certain information to be proprietary, and thus CAS numbers and other limited information may not be available for release.

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TAPE AND REEL INFORMATION



TAPE DIMENSIONS + K0 - P1 - B0 W Cavity - A0 -

	Dimension designed to accommodate the component width
В0	Dimension designed to accommodate the component length
K0	Dimension designed to accommodate the component thickness
W	Overall width of the carrier tape
P1	Pitch between successive cavity centers

QUADRANT ASSIGNMENTS FOR PIN 1 ORIENTATION IN TAPE



*All dimensions are nominal

Device	Package Type	Package Drawing		SPQ	Reel Diameter (mm)	Reel Width W1 (mm)	A0 (mm)	B0 (mm)	K0 (mm)	P1 (mm)	W (mm)	Pin1 Quadrant
TCAN1044DRBRQ1	SON	DRB	8	3000	330.0	12.4	3.3	3.3	1.1	8.0	12.0	Q1
TCAN1044DRQ1	SOIC	D	8	2500	330.0	12.4	6.4	5.2	2.1	8.0	12.0	Q1
TCAN1044VDDFRQ1	SOT-23- THIN	DDF	8	3000	180.0	8.4	3.2	3.2	1.4	4.0	8.0	Q3
TCAN1044VDRBRQ1	SON	DRB	8	3000	330.0	12.4	3.3	3.3	1.1	8.0	12.0	Q1
TCAN1044VDRQ1	SOIC	D	8	2500	330.0	12.4	6.4	5.2	2.1	8.0	12.0	Q1



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*All dimensions are nominal

Device Package Type		Package Drawing	Pins	SPQ	Length (mm)	Width (mm)	Height (mm)
TCAN1044DRBRQ1	TCAN1044DRBRQ1 SON		8	3000	367.0	367.0	35.0
TCAN1044DRQ1 SOIC		D	8	2500	356.0	356.0	35.0
TCAN1044VDDFRQ1	CAN1044VDDFRQ1 SOT-23-THIN		8	3000	210.0	185.0	35.0
TCAN1044VDRBRQ1	SON	DRB	8	3000	367.0	367.0	35.0
TCAN1044VDRQ1	SOIC	D	8	2500	356.0	356.0	35.0

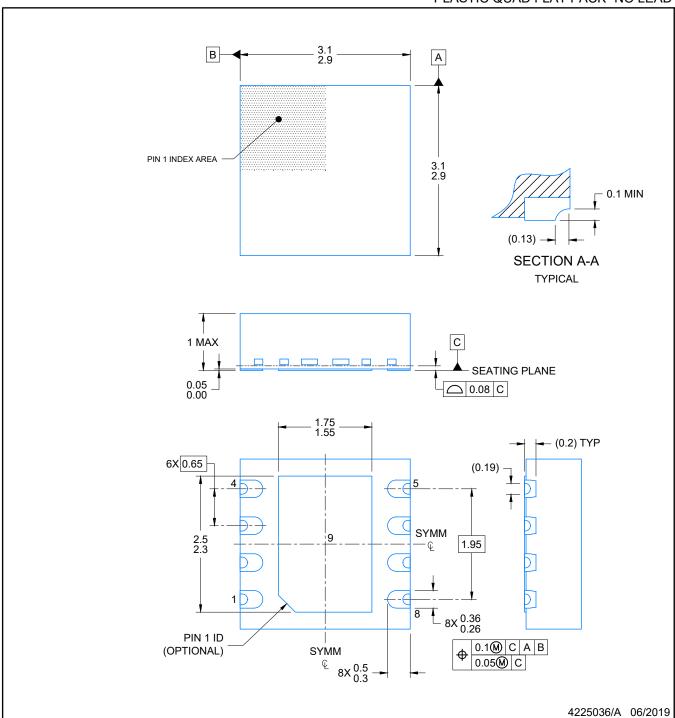


Images above are just a representation of the package family, actual package may vary. Refer to the product data sheet for package details.

4203482/L



PLASTIC QUAD FLAT PACK- NO LEAD

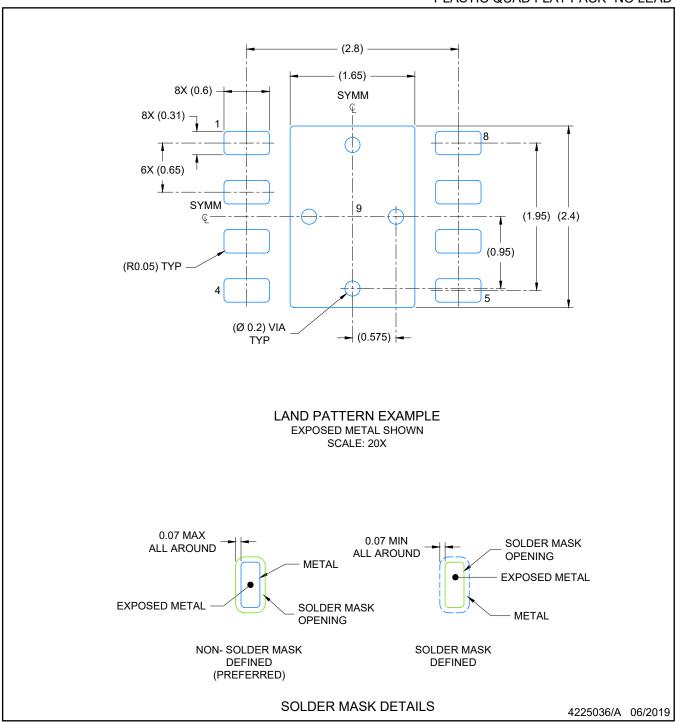


NOTES:

- All linear dimensions are in millimeters. Any dimensions in parenthesis are for reference only. Dimensioning and tolerancing per ASME Y14.5M.
- 2. This drawing is subject to change without notice.
- 3. The package thermal pad must be soldered to the printed circuit board for optimal thermal and mechanical performance.



PLASTIC QUAD FLAT PACK- NO LEAD

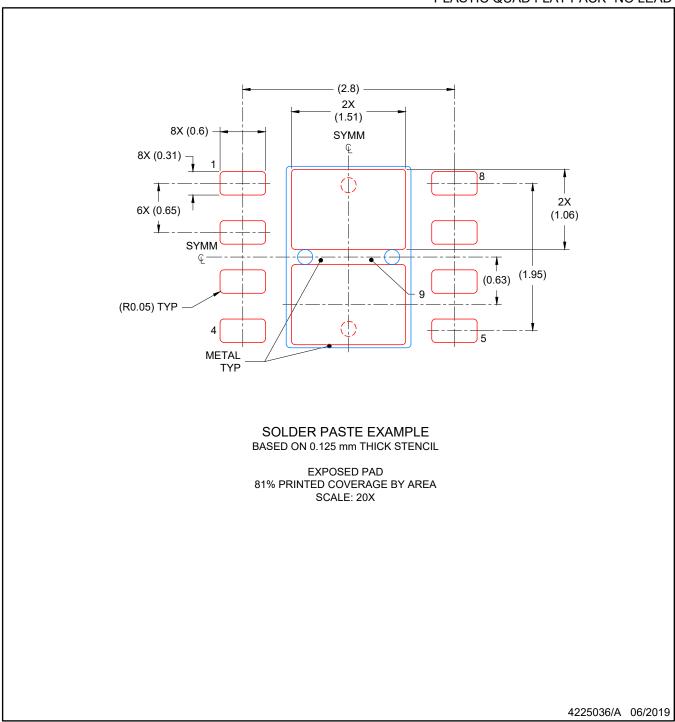


NOTES: (continued)

- 4. This package is designed to be soldered to a thermal pad on the board. For more information, see Texas Instruments literature number SLUA271 (www.ti.com/lit/slua271).
- 5. Vias are optional depending on application, refer to device data sheet. If any vias are implemented, refer to their locations shown on this view. It is recommended that vias under paste be filled, plugged or tented.



PLASTIC QUAD FLAT PACK- NO LEAD



NOTES: (continued)

6. Laser cutting apertures with trapezoidal walls and rounded corners may offer better paste release. IPC-7525 may have alternate design recommendations.



PLASTIC SMALL OUTLINE



NOTES:

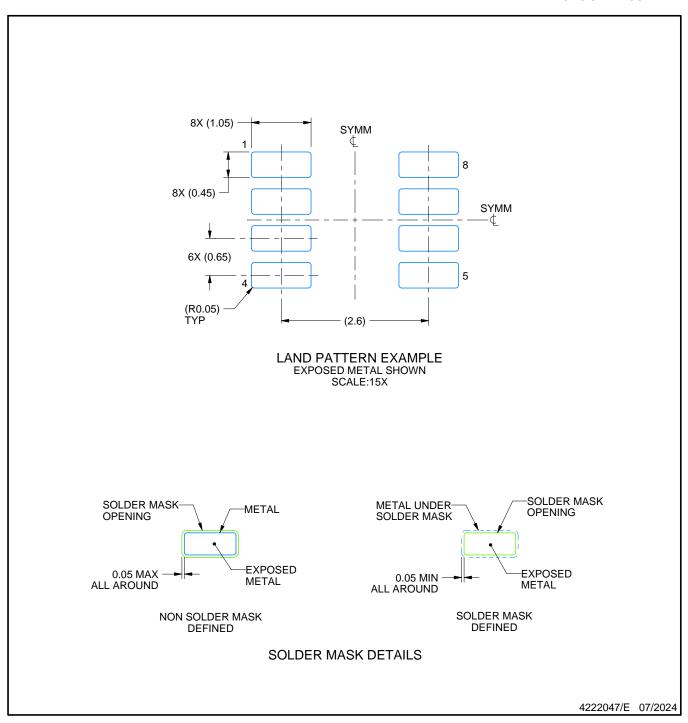
- 1. All linear dimensions are in millimeters. Any dimensions in parenthesis are for reference only. Dimensioning and tolerancing per ASME Y14.5M.

 2. This drawing is subject to change without notice.

 3. This dimension does not include mold flash, protrusions, or gate burrs. Mold flash, protrusions, or gate burrs shall not
- exceed 0.15 mm per side.



PLASTIC SMALL OUTLINE

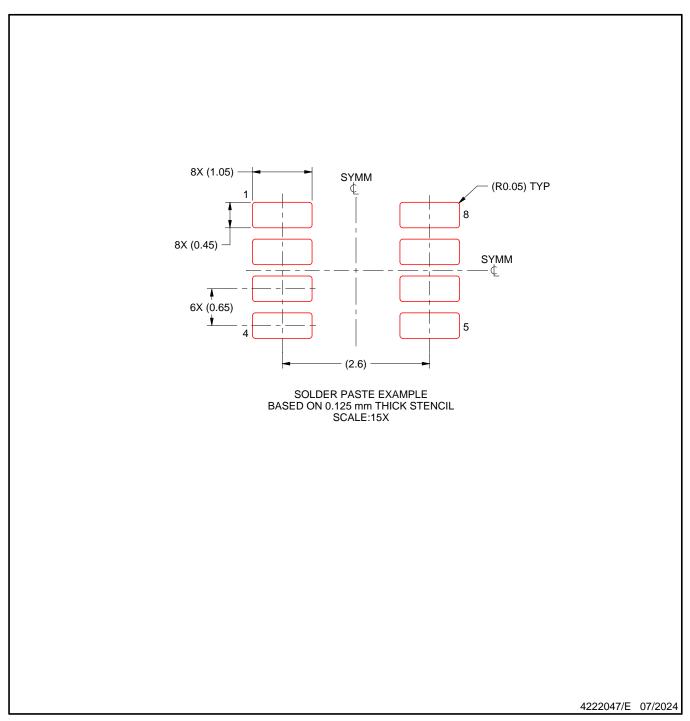


NOTES: (continued)

- 4. Publication IPC-7351 may have alternate designs.
- 5. Solder mask tolerances between and around signal pads can vary based on board fabrication site.



PLASTIC SMALL OUTLINE



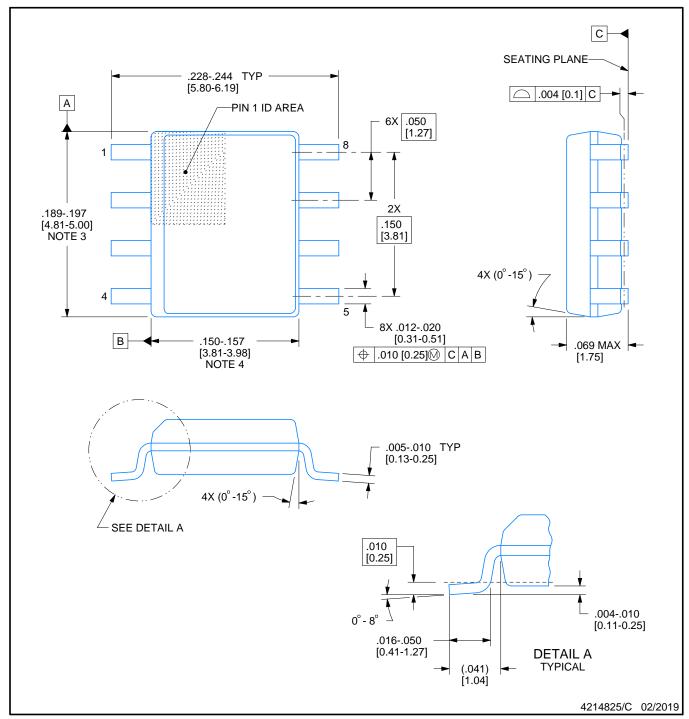
NOTES: (continued)

- 6. Laser cutting apertures with trapezoidal walls and rounded corners may offer better paste release. IPC-7525 may have alternate design recommendations.
- 7. Board assembly site may have different recommendations for stencil design.





SMALL OUTLINE INTEGRATED CIRCUIT

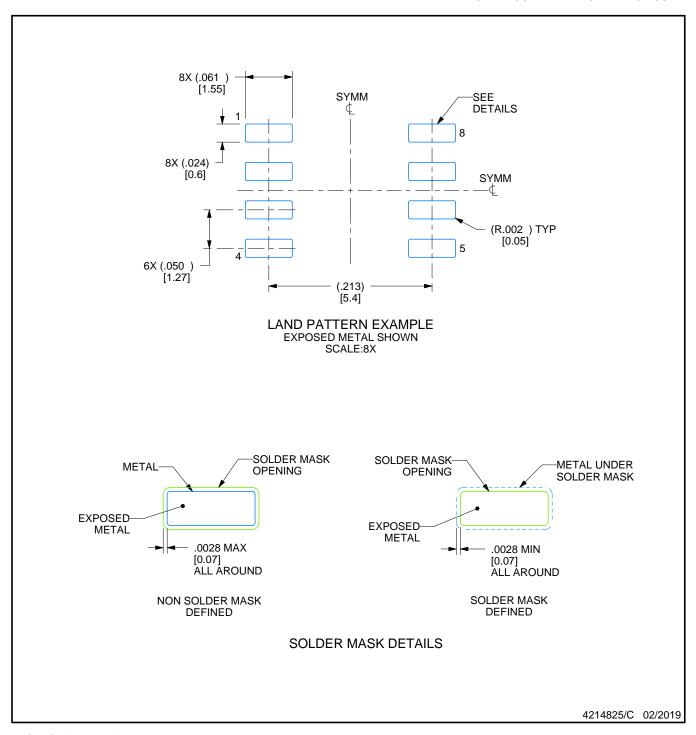


NOTES:

- 1. Linear dimensions are in inches [millimeters]. Dimensions in parenthesis are for reference only. Controlling dimensions are in inches. Dimensioning and tolerancing per ASME Y14.5M.
- 2. This drawing is subject to change without notice.
- 3. This dimension does not include mold flash, protrusions, or gate burrs. Mold flash, protrusions, or gate burrs shall not exceed .006 [0.15] per side.
- 4. This dimension does not include interlead flash.
- 5. Reference JEDEC registration MS-012, variation AA.



SMALL OUTLINE INTEGRATED CIRCUIT

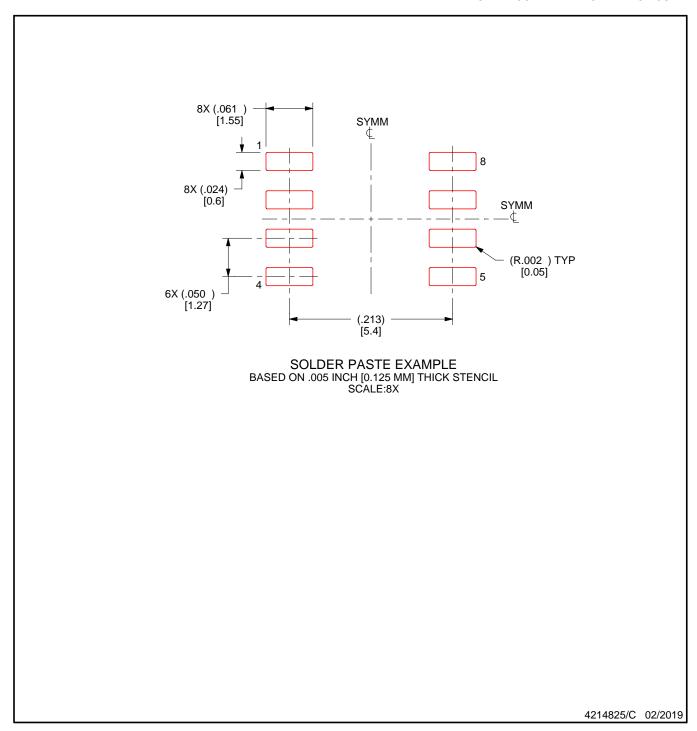


NOTES: (continued)

6. Publication IPC-7351 may have alternate designs.

7. Solder mask tolerances between and around signal pads can vary based on board fabrication site.

SMALL OUTLINE INTEGRATED CIRCUIT



NOTES: (continued)

- 8. Laser cutting apertures with trapezoidal walls and rounded corners may offer better paste release. IPC-7525 may have alternate design recommendations.
- 9. Board assembly site may have different recommendations for stencil design.



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