

XR8051, XR8052, XR8054 Low Cost, High Speed Rail-to-Rail Amplifiers

General Description

The XR8051 (single), XR8052 (dual) and XR8054 (quad) are low cost, voltage feedback amplifiers. These amplifiers are designed to operate on +3V to +5V, or \pm 5V supplies. The input voltage range extends 300mV below the negative rail and 0.9V below the positive rail.

The XR8051, XR8052, and XR8054 offer superior dynamic performance with a 260MHz small signal bandwidth and 190V/µs slew rate. The combination of low power, high output current drive, and rail-to-rail performance make these amplifiers well suited for battery-powered systems and video applications.

The combination of low cost and high performance make the XR8051, XR8052, and XR8054 suitable for high volume applications in both consumer and industrial applications such as video surveillance and distribution systems, professional and IPC cameras, active filter circuits, coaxial cable drivers, and electronic white boards.

FEATURES

- 260MHz bandwidth
- Fully specified at +3V, +5V and ±5V supplies
- Output voltage range:
 0.03V to 4.95V; V_S = +5; R_L = 2kΩ
- Input voltage range:
- -0.3V to +4.1V; $V_{S} = +5$
- 190V/µs slew rate
- 2.6mA supply current per amplifier
- ±100mA linear output current
- ±125mA short circuit current
- XR8051 directly replaces AD8051, AD8091
- XR8052 directly replaces AD8052, AD8092
- XR8054 directly replaces AD8054

APPLICATIONS

- Video driver
- Video surveillance and distribution
- A/D driver
- Active filters
- CCD imaging systems
- CD/DVD ROM
- Coaxial cable drivers
- High capacitive load driver
- Portable/battery-powered applications
- Twisted pair driver
- Telecom and optical terminals

Ordering Information - page 26



Large Signal Frequency Response

Output Voltage Swing vs Competition



Absolute Maximum Ratings

Stresses beyond the limits listed below may cause permanent damage to the device. Exposure to any Absolute Maximum Rating condition for extended periods may affect device reliability and lifetime.

V _S	0V to +14V
V _{IN} V _S - 0.5V to	+V _S +0.5V

Operating Conditions

Supply Voltage Range	2.7 to 12.6V
Operating Temperature Range	40°C to 125°C
Junction Temperature	150°C
Storage Temperature Range	65°C to 170°C
Lead Temperature (Soldering, 10s)	260°C

Package Thermal Resistance

θ _{JA} (TSOT-5)	215°C/W
θ _{JA} (SOIC-8)	150°C/W
θ _{JA} (MSOP-8)	200°C/W
θ _{JA} (SOIC-14)	90°C/W
θ _{JA} (TSSOP-14)	100°C/W
Package thermal resistance (θ_{JA}), JEDEC st test boards, still air.	andard, multi-layer

ESD Protection

XR8051, XR8052, XR8054 (HBM)1kV ESD Rating for HBM (Human Body Model).

Electrical Characteristics at +3V

 T_A = 25°C, V_S = +3V, R_f = 1.5k\Omega, R_L = 2k\Omega to $V_S/2;$ G = 2; unless otherwise noted.

Symbol	Parameter	Conditions	Min	Тур	Max	Units
Frequency	Domain Response					
GBWP	-3dB Gain Bandwidth Product	$G = +11, V_{OUT} = 0.2V_{pp}$		90		MHz
UGBW	Unity Gain Bandwidth	$V_{OUT} = 0.2V_{pp}, R_F = 0$		245		MHz
BW _{SS}	-3dB Bandwidth	$V_{OUT} = 0.2V_{pp}$		85		MHz
f _{0.1dB}	0.1dB Gain Flatness	$V_{OUT} = 0.2V_{pp}, R_L = 150\Omega$		16		MHz
BW _{LS}	Large Signal Bandwidth	$V_{OUT} = 2V_{pp}$		40		MHz
50	Differential Only	DC-coupled Output		0.03		%
DG	Differential Gain	AC-coupled Output		0.04		%
55	D'fferential Disease	DC-coupled Output		0.03		0
DP	Differential Phase	AC-coupled Output		0.06		•
Time Doma	in					
t _R , t _F	Rise and Fall Time	V _{OUT} = 0.2V step; (10% to 90%)		5		ns
t _S	Settling Time to 0.1%	V _{OUT} = 1V step		25		ns
OS	Overshoot	V _{OUT} = 0.2V step		8		%
SR	Slew Rate	G = -1, 2V step		165		V/µs
Distortion/N	loise Response					
THD	Total Harmonic Distortion	$1MHz, V_{OUT} = 1V_{pp}$		75		dBc
e _n	Input Voltage Noise	>50kHz		16		nV/√Hz
X _{TALK}	Crosstalk	f = 5MHz		58		dB
DC Perform	nance		· · · ·			
V _{IO}	Input Offset Voltage			0.5		mV
d _{VIO}	Average Drift			5		μV/°C
I _B	Input Bias Current			1.4		μA
dl _B	Average Drift			2		nA/°C
I _{OS}	Input Offset Current			0.05		μA
PSRR	Power Supply Rejection Ratio	DC		102		dB
A _{OL}	Open Loop Gain	$R_L = 2k\Omega$		92		dB
I _S	Supply Current	per channel		2.6		mA
Input Chara	acteristics					
C _{IN}	Input Capacitance			0.5		pF
CMIR	Common Mode Input Range			-0.3 to 2.1		V
CMRR	Common Mode Rejection Ratio	DC, V _{CM} = 0 to 1.5V		100		dB
Output Cha	racteristics			1		
		D 1500		0.3 to		v
V _{OUT}	Output Swing	$R_L = 150\Omega$		2.75		v
V001	Cuputowing	$R_L = 2k\Omega$		0.02 to 2.96		V
I _{OUT}	Output Current			±100		mA
I _{SC}	Short Circuit Current	$V_{OUT} = V_S / 2$		±125		V
V _S	Power Supply Operating Range			2.7 to 12.6		V

Electrical Characteristics at +5V

 T_A = 25°C, V_S = +5V, R_f = 1.5k\Omega, R_L = 2k\Omega to $V_S/2;$ G = 2; unless otherwise noted.

Symbol	Parameter	Conditions	Min	Тур	Max	Units
Frequency	Domain Response					
GBWP	-3dB Gain Bandwidth Product	$G = +11, V_{OUT} = 0.2V_{pp}$		95		MHz
UGBW	Unity Gain Bandwidth	$V_{OUT} = 0.2V_{pp}, R_F = 0$		250		MHz
BW _{SS}	-3dB Bandwidth	$V_{OUT} = 0.2V_{pp}$		85		MHz
f _{0.1dB}	0.1dB Gain Flatness	$V_{OUT} = 0.2V_{pp}, R_L = 150\Omega$		35		MHz
BW _{LS}	Large Signal Bandwidth	$V_{OUT} = 2V_{pp}$		45		MHz
D 0	Differential Only	DC-coupled Output		0.03		%
DG	Differential Gain	AC-coupled Output		0.04		%
		DC-coupled Output		0.03		0
DP	Differential Phase	AC-coupled Output		0.06		0
Time Doma	in					
t _R , t _F	Rise and Fall Time	V _{OUT} = 0.2V step		5		ns
t _S	Settling Time to 0.1%	V _{OUT} = 2V step		25		ns
OS	Overshoot	V _{OUT} = 0.2V step		5		%
SR	Slew Rate	G = -1, 4V step		185		V/µs
Distortion/N	loise Response					
THD	Total Harmonic Distortion	1MHz, V _{OUT} = 2 V _{pp}		-75		dBc
e _n	Input Voltage Noise	>50kHz		16		nV/√Hz
X _{TALK}	Crosstalk	f = 5MHz		58		dB
DC Perform	ance					
V _{IO}	Input Offset Voltage		-7	0.5	7	mV
d _{VIO}	Average Drift			5		µV/°C
I _B	Input Bias Current		-2	1.4	2	μA
dl _B	Average Drift			2		nA/°C
I _{OS}	Input Offset Current		-0.75	0.05	0.75	μA
PSRR	Power Supply Rejection Ratio	DC	80	102		dB
A _{OL}	Open Loop Gain	$R_L = 2k\Omega$	80	92		dB
I _S	Supply Current	per channel		2.6	4	mA
Input Chara	acteristics					
C _{IN}	Input Capacitance			0.5		pF
CMIR	Common Mode Input Range			-0.3 to 4.1		V
CMRR	Common Mode Rejection Ratio	DC, V _{CM} = 0 to 3.5V	75	100		dB
Output Cha	racteristics		I	1		1
	0.4	R _L = 150Ω	0.35	0.1 to 4.9	4.65	V
V _{OUT}	Output Swing	$R_L = 2k\Omega$		0.03 to 4.95		v
I _{OUT}	Output Current			±100		mA
I _{SC}	Short Circuit Current	$V_{OUT} = V_S / 2$		±125		V
V _S	Power Supply Operating Range			2.7 to 12.6		V

Electrical Characteristics at ±5V

 T_A = 25°C, V_S = ±5V, R_f = 1.5k $\Omega,~R_L$ = 2k Ω to GND; G = 2; unless otherwise noted.

Symbol	Parameter	Conditions	Min	Тур	Max	Units
Frequency	Domain Response					
GBWP -3dB Gain Bandwidth Product		$G = +11, V_{OUT} = 0.2V_{pp}$		90		MHz
UGBW	Unity Gain Bandwidth	$V_{OUT} = 0.2V_{pp}, R_F = 0$		260		MHz
BW _{SS}	-3dB Bandwidth	$V_{OUT} = 0.2V_{pp}$		85		MHz
f _{0.1dB}	0.1dB Gain Flatness	$V_{OUT} = 0.2V_{pp}, R_L = 150\Omega$		22		MHz
BW _{LS}	Large Signal Bandwidth	$V_{OUT} = 2V_{pp}$		50		MHz
5.0		DC-coupled Output		0.03		%
DG	Differential Gain	AC-coupled Output		0.04		%
		DC-coupled Output		0.03		0
DP	Differential Phase	AC-coupled Output		0.06		0
Time Doma	in		K			
t _R , t _F	Rise and Fall Time	V _{OUT} = 0.2V step		5		ns
t _S	Settling Time to 0.1%	$V_{OUT} = 2V$ step, $R_L = 100\Omega$		25		ns
OS	Overshoot	V _{OUT} = 0.2V step		5		%
SR	Slew Rate	G = -1, 5V step		190		V/µs
Distortion/N	loise Response		· · · ·			
THD	Total Harmonic Distortion	1MHz, V _{OUT} = 2V _{pp}		76		dBc
e _n	Input Voltage Noise	>50kHz		16		nV/√H
X _{TALK}	Crosstalk	f = 5MHz		58		dB
DC Perform	lance		· · · ·	,		
V _{IO}	Input Offset Voltage			0.5		mV
d _{VIO}	Average Drift			5		μV/°C
IB	Input Bias Current			1.3		μA
dl _B	Average Drift			2		nA/°C
I _{OS}	Input Offset Current			0.04		μA
PSRR	Power Supply Rejection Ratio	DC		102		dB
A _{OL}	Open Loop Gain	$R_L = 2k\Omega$		92		dB
Is	Supply Current	per channel		2.6		mA
Input Chara	interistics			1		1
C _{IN}	Input Capacitance			0.5		pF
CMIR	Common Mode Input Range			-5.3 to 4.1		v
CMRR	Common Mode Rejection Ratio	DC, V _{CM} = -5 to 3.5V		100		dB
Output Cha			H	1		1
·		D 1500		-4.8 to		
Max.	Output Swing	$R_L = 150\Omega$		4.8		V
V _{OUT}	Output Swing	$R_L = 2k\Omega$		-4.95 to 4.93		v
I _{OUT}	Output Current			±100		mA
I _{SC}	Short Circuit Current	$V_{OUT} = V_S / 2$		±125		V
V _S	Power Supply Operating Range			2.7 to 12.6		V

XR8051 Pin Configurations TSOT-5



SOIC-8



XR8051 Pin Assignments

TSOT-5

Pin No.	Pin Name	Description
1	OUT	Output
2	-V _S	Negative supply
3	+IN	Positive input
4	-IN	Negative input
5	+V _S	Positive supply

SOIC-8

Pin No.	Pin Name	Description
1	NC	No Connect
2	-IN	Negative input
3	+IN	Positive input
4	-V _S	Negative supply
5	NC	No Connect
6	OUT	Output
7	+V _S	Positive supply
8	NC	No Connect

XR8052 Pin Configuration SOIC-8 / MSOP-8



XR8052 Pin Assignments SOIC-8 / MSOP-8

Pin No.	Pin Name	Description
1	OUT1	Output, channel 1
2	-IN1	Negative input, channel 1
3	+IN1	Positive input, channel 1
4	-V _S	Negative supply
5	+IN2	Positive input, channel 2
6	-IN2	Negative input, channel 2
7	OUT2	Output, channel 2
8	+V _S	Positive supply

XR8054 Pin Configuration SOIC-14/TSSOP-14



XR8054 Pin Assignments

SOIC-14 / TSSOP-14

Pin No.	Pin Name	Description
1	OUT1	Output, channel 1
2	-IN1	Negative input, channel 1
3	+IN1	Positive input, channel 1
4	+V _S	Positive supply
5	+IN2	Positive input, channel 2
6	-IN2	Negative input, channel 2
7	OUT2	Output, channel 2
8	OUT3	Output, channel 3
9	-IN3	Negative input, channel 3
10	+IN3	Positive input, channel 3
11	-V _S	Negative supply
12	+IN4	Positive input, channel 4
13	-IN4	Negative input, channel 4
14	OUT4	Output, channel 4

 $T_A = 25^{\circ}C$, $V_S = +3V$, $R_L = 2k\Omega$ to $V_S/2$, G = +2, $R_F = 1.5k\Omega$; unless otherwise noted.

Non-Inverting Freq. Resp.



Freq. Resp. vs CL



Large Signal Freq. Resp.



Inverting Freq. Resp.



Freq. Resp. vs R_L







 T_A = 25°C, V_S = +3V, R_L = 2k Ω to $V_S/2,~G$ = +2, R_F = 1.5k $\Omega;$ unless otherwise noted.

Input Voltage Noise vs Freq.



3rd Harmonic Distortion vs R_L over Freq.



3rd Harmonic Distortion vs VO over Freq.



2nd Harmonic Distortion vs R_L over Freq.



2nd Harmonic Distortion vs VO over Freq.



Non-Inverting Small Signal Pulse Response



 $T_A = 25^{\circ}C$, $V_S = +3V$, $R_L = 2k\Omega$ to $V_S/2$, G = +2, $R_F = 1.5k\Omega$; unless otherwise noted.

Non-Inverting Large Signal Pulse Response



Differential Gain & Phase_DC Coupled



Crosstalk vs Frequency (XR8052)



Differential Gain & Phase_AC Coupled



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 T_A = 25°C, V_S = +5V, R_L = 2k Ω to $V_S/2,~G$ = +2, R_F = 1.5k $\Omega;$ unless otherwise noted.

Non-Inverting Freq. Resp.



Freq. Resp. vs CL



Large Signal Freq. Resp.



Inverting Freq. Resp.



Freq. Resp. vs R_L







 T_A = 25°C, V_S = +5V, R_L = 2k Ω to $V_S/2,~G$ = +2, R_F = 1.5k $\Omega;$ unless otherwise noted.

Input Voltage Noise vs Freq.



3rd Harmonic Distortion vs R_L over Freq.



3rd Harmonic Distortion vs VO over Freq.



2nd Harmonic Distortion vs RL over Freq.



2nd Harmonic Distortion vs VO over Freq.



Non-Inverting Small Signal Pulse Response



 $T_A = 25^{\circ}C$, $V_S = +5V$, $R_L = 2k\Omega$ to $V_S/2$, G = +2, $R_F = 1.5k\Omega$; unless otherwise noted.

Non-Inverting Large Signal Pulse Response



Differential Gain & Phase_DC Coupled



Crosstalk vs Frequency (XR8052)



Differential Gain & Phase_AC Coupled



 $T_A = 25^{\circ}C$, $V_S = \pm 5V$, $R_L = 2k\Omega$ to GND, G = +2, $R_F = 1.5k\Omega$; unless otherwise noted.

Non-Inverting Freq. Resp.



Freq. Resp. vs CL



Large Signal Freq. Resp.



Inverting Freq. Resp.



Freq. Resp. vs R_L







 T_A = 25°C, V_S = ±5V, R_L = 2k Ω to GND, G = +2, R_F = 1.5k Ω ; unless otherwise noted.

Input Voltage Noise vs Freq.



3rd Harmonic Distortion vs R_L over Freq.



3rd Harmonic Distortion vs VO over Freq.



2nd Harmonic Distortion vs RL over Freq.



2nd Harmonic Distortion vs VO over Freq.



Non-Inverting Small Signal Pulse Response



 $T_A = 25^{\circ}C$, $V_S = \pm 5V$, $R_L = 2k\Omega$ to GND, G = +2, $R_F = 1.5k\Omega$; unless otherwise noted.

Non-Inverting Large Signal Pulse Response



Differential Gain & Phase_DC Coupled



Crosstalk vs Frequency (XR8052)



Differential Gain & Phase_AC Coupled



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Application Information

General Description

The XR8051, XR8052, and XR8054 are single supply, general purpose, voltage-feedback amplifiers fabricated on a complementary bipolar process using a patent pending topography. They feature a rail-to-rail output stage and is unity gain stable.

The common mode input range extends to 300mV below ground and to 0.9V below V_s. Exceeding these values will not cause phase reversal. However, if the input voltage exceeds the rails by more than 0.5V, the input ESD devices will begin to conduct. The output will stay at the rail during this overdrive condition.

The design is short circuit protected and offers "soft" saturation protection that improves recovery time.

Figures 1, 2, and 3 illustrate typical circuit configurations for non-inverting, inverting, and unity gain topologies for dual supply applications. They show the recommended bypass capacitor values and overall closed loop gain equations. Figure 4 shows the typical non-inverting gain circuit for single supply applications.



Figure 1: Typical Non-Inverting Gain Circuit



Figure 2: Typical Inverting Gain Circuit



Figure 3: Unity Gain Circuit



Figure 4: Single Supply Non-Inverting Gain Circuit

Overdrive Recovery

For an amplifier, an overdrive condition occurs when the output and/or input ranges are exceeded. The recovery time varies based on whether the input or output is overdriven and by how much the ranges are exceeded. The XR8051, XR8052, and XR8054 will typically recover in less than 20ns from an overdrive condition. Figure 5 shows the XR8052 in an overdriven condition.



Figure 5: Overdrive Recovery

Power Dissipation

Power dissipation should not be a factor when operating under the stated $2k\Omega$ load condition. However, applications with low impedance, DC coupled loads should be analyzed to ensure that maximum allowed junction temperature is not exceeded. Guidelines listed below can be used to verify that the particular application will not cause the device to operate beyond it's intended operating range.

Maximum power levels are set by the absolute maximum junction rating of 170°C. To calculate the junction temperature, the package thermal resistance value Theta_{JA} (θ_{JA}) is used along with the total die power dissipation.

$$T_{Junction} = T_{Ambient} + (\theta_{JA} \times P_D)$$

Where $\mathsf{T}_{\mathsf{Ambient}}$ is the temperature of the working environment.

In order to determine P_D , the power dissipated in the load needs to be subtracted from the total power delivered by the supplies.

$$P_D = P_{supply} - P_{load}$$

Supply power is calculated by the standard power equation.

$$P_{supply} = V_{supply} \times I_{RMSsupply}$$
$$V_{supply} = V_{S+} - V_{S-}$$

Power delivered to a purely resistive load is:

$$P_{load} = ((V_{load})_{RMS^2})/Rload_{eff}$$

The effective load resistor ($Rload_{eff}$) will need to include the effect of the feedback network. For instance,

Rload_{eff} in Figure 3 would be calculated as:

$$R_L \parallel (R_f + R_g)$$

These measurements are basic and are relatively easy to perform with standard lab equipment. For design purposes however, prior knowledge of actual signal levels and load impedance is needed to determine the dissipated power. Here, P_D can be found from

$$P_D = P_{Quiescent} + P_{Dynamic} - P_{load}$$

Quiescent power can be derived from the specified I_S values along with known supply voltage, V_{supply} . Load power can be calculated as above with the desired signal amplitudes using:

$$(V_{load})_{RMS} = V_{peak} / \sqrt{2}$$

 $(I_{load})_{RMS} = (V_{load})_{RMS} / Rload_{eff}$

The dynamic power is focused primarily within the output stage driving the load. This value can be calculated as:

 $P_{Dynamic} = (V_{S+} - V_{load})_{RMS} \times (I_{load})_{RMS}$

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Assuming the load is referenced in the middle of the power rails or $V_{supply}/2$.

The XR8051 is short circuit protected. However, this may not guarantee that the maximum junction temperature (+150°C) is not exceeded under all conditions. Figure 6 shows the maximum safe power dissipation in the package vs. the ambient temperature for the packages available.



Figure 6. Maximum Power Derating

Driving Capacitive Loads

Increased phase delay at the output due to capacitive loading can cause ringing, peaking in the frequency response, and possible unstable behavior. Use a series resistance, R_S , between the amplifier and the load to help improve stability and settling performance. Refer to Figure 7.



Figure 7. Addition of R_S for Driving Capacitive Loads

Table 1 provides the recommended R_S for various capacitive loads. The recommended R_S values result in approximately <1dB peaking in the frequency response.

C _L (pF)	R _S (Ω)	-3dB BW (MHz)
22pF	0	120
47pF	15	80
100pF	15	65
492pF	6.5	40

Table 1: Recommended R_S vs. C_L

For a given load capacitance, adjust R_S to optimize the tradeoff between settling time and bandwidth. In general, reducing R_S will increase bandwidth at the expense of additional overshoot and ringing.

Layout Considerations

General layout and supply bypassing play major roles in high frequency performance. Exar has evaluation boards to use as a guide for high frequency layout and as an aid in device testing and characterization. Follow the steps below as a basis for high frequency layout:

- Include 6.8µF and 0.1µF ceramic capacitors for power supply decoupling
- Place the 6.8µF capacitor within 0.75 inches of the power pin
- Place the 0.1µF capacitor within 0.1 inches of the power pin
- Remove the ground plane under and around the part, especially near the input and output pins to reduce parasitic capacitance
- Minimize all trace lengths to reduce series inductances

Refer to the evaluation board layouts below for more information.

Evaluation Board Information

The following evaluation boards are available to aid in the testing and layout of these devices:

Evaluation Board #	Products
CEB002	XR8051 in TSOT
CEB003	XR8051 in SOIC
CEB006	XR8052 in SOIC
CEB010	XR8052 in MSOP
CEB018	XR8054 in TSSOP

Evaluation Board Schematics

Evaluation board schematics and layouts are shown in Figures 8-14 These evaluation boards are built for dualsupply operation. Follow these steps to use the board in a single-supply application:

- 1. Short -V $_{\rm S}$ to ground.
- 2. Use C3 and C4, if the -V_S pin of the amplifier is not directly connected to the ground plane.



Figure 8. CEB002 & CEB003 Schematic



Figure 9. CEB002 Top View



Figure 10. CEB002 Bottom View



Figure 11. CEB003 Top View



Figure 12. CEB003 Bottom View



Figure 13. CEB006 & CEB010 Schematic



Figure 14. CEB006 Top View



Figure 15. CEB006 Bottom View



Figure 16. CEB010 Top View



Figure 17. CEB010 Bottom View



Figure 18. CEB018 Schematic



Figure 19. CEB018 Top View

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Figure 20. CEB018 Bottom View

Mechanical Dimensions

TSOT-5 Package



5 Pin TSOT (OPTION 2)						
SYMBOLS		ISION II ntrol U		DIMENSION IN INCH (Reference Unit)		
	MIN	N NOM		MIN	NOM	MAX
A	0.75	—	0.80	0.030 —		0.031
A1	0.00	—	0.05	0.000	—	0.002
A2	0.70	0.75	0.78	0.028	0.030	0.031
b	0.35	_	0.50	0.012	—	0.020
с	0.10	_	0.20	0.003	—	0.008
D	2	2.90 BS	C	0.114 BSC		
E	2	2.80 BS	iC	0.110 BSC		
E1	1	.60 BS	iC	0.063 BSC		
е	0).95 BS	iC	0.038 BSC		
e1	1	.90 BS	iC	0.075 BSC		
L	0.37	0.45	0.60	0.012	0.018	0.024
L1	0).60 RE	F	0.024 REF		
L2	0.25 BSC			0.010 BSC		
R	0.10	—	_	0.004	—	_
R1	0.10	_	0.25	0.004	—	0.010
θ	0*	4'	8'	0,	4*	8'
θ1	4.	10'	12	4.	10*	12*
N		5			5	

Side View

Front View

MSOP-8 Package







Side View

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SOIC-8 Package



RECOMMENDED PCB LAND PATTERN





Side View

A

Front View

8 Pin	SOICN	JEDE	EC MS-	-012	Variatio	n AA	
SYMBOLS		NSIONS II ontrol Un		DIMENSIONS IN INCH (Reference Unit)			
	MIN	NOM	MAX	MIN	NOM	MAX	
A	1.35		1.75	0.053	—	0.069	
A1	0.10		0.25	0.004	—	0.010	
A2	1.25	—	1.65	0.049	—	0.065	
b	0.31		0.51	0.012		0.020	
с	0.17		0.25	0.007	—	0.010	
E		5.00 BSC)	0.236 BSC			
E1		3.90 BSC	2	0.154 BSC			
е		1.27 BSC)	0.050 BSC			
h	0.25	—	0.50	0.010	—	0.020	
L	0.40	—	1.27	0.016	—	0.050	
L1		1.04 REF		0.041 REF			
L2		0.25 BSC	2	0.010 BSC			
R	0.07		—	0.003	—	—	
R1	0.07		—	0.003	_	_	
θ	0*		8*	0°	_	8'	
θ1	5°		15°	5°	_	15°	
θ2	0.	_	_	0°	_	_	
D	4	4.90 BSC	;	0.193 BSC			
N		8		8			

SOIC-14 Package



A2

Top View



Side View



	PACKAGE OUTLINE NSOIC .150" BODY JEDEC MS-012									
SYMBO	OLS	COMMON DIMENSIONS IN MM (Control Unit)				COMMON DIMENSIONS IN INCH (Reference Unit)				
		MIN	4	NOM		IAX	MIN		NOM	MAX
Α		1.3	5	_	1	.75	0.05	3	—	0.069
A1		0.1	0	—	C).25	0.00	4	—	0.010
A2		1.2	5	_	1	.65	0.04	9	—	0.065
b		0.3	51	_	0).51	0.01	2	_	0.020
с		0.1	7	—	C	.25	0.00	7	—	0.010
Е			6	.00 BS0	C			0	.236 BS	SC
E1			3	5.90 BS	С			0	.154 BS	SC
e				.27 BS	С		0.050 BSC			
h		0.2	5	_	0.50		0.01	0	_	0.020
L		0.40		_		.27	0.01	6	_	0.050
L1		1.04 REF				0.041 REF				
L2				0.25 BSC			0.010 BSC			
R		0.0		_		_	0.00		_	-
R1		0.0)7	_	—		0.00	3	—	-
θ		0	_	_		8'	0.		_	8'
θ1		5		_		15'	5'		_	15
θ2		0	•	_		—	0.		—	-
D		SEE VARIATIONS								
N		SEE VARIATIONS								
VARIATION D										
VARIATIONS	(DIMENSIONS IN MM (Control Unit)			DIMENSIONS IN INCH (Reference Unit)			N		
SNC	MI	N	NO	MA NA	хΤ	MIN	NC	м	MAX	
AA		4.90 BSC				0.193 BSC			8	
AB		8.65 BSC				0.341 BSC			14	
AC		9.90 BSC					0.390	E	SC	16

Front View

TSSOP-14 Package





Side View

14 Pin TSSOP JEDEC MO-153 Variation AB-1						
SYMBOLS		NSIONS ontrol U		DIMENSIONS IN INCH (Reference Unit)		
	MIN	NOM	MAX	MIN	NOM	MAX
А	—	—	1.20	—	—	0.047
A1	0.05		0.15	0.002	—	0.006
A2	0.80 1.00		1.05	0.031	0.039	0.041
b	0.19 — 0.3		0.30	0.007	—	0.012
с	0.09	_	0.20	0.004	—	0.008
E	6.40 BSC			0.252 BSC		
E1	4.30	4.40	4.50	0.169	0.173	0.177
е	0.65 BSC			0.026 BSC		
L	0.45	0.60	0.75	0.018	0.024	0.030
L1	1	.00 RE	F	0.039 REF		
L2	0.25 BSC			0.010 BSC		
R	0.09	—	—	0.035	—	—
R1	0.09	—	_	0.035	—	—
θ	12" REF			12" REF		
θ1	0*	_	8'	0.	_	8"
D	4.90	5.00	5.10	0.193	0.197	0.200
N		14		14		

А

Ordering Information

Part Number	Package	Green	Operating Temperature Range	Packaging	
XR8051 Ordering Information					
XR8051AST5X	TSOT-5	Yes	-40°C to +125°C	Tape & Reel	
XR8051AST5MTR	TSOT-5	Yes	-40°C to +125°C	Tape & Reel	
XR8051AST5EVB	Evaluation Board	N/A	N/A	N/A	
XR8051ASO8X	SOIC-8	Yes	-40°C to +125°C	Tape & Reel	
XR8051ASO8MTR	SOIC-8	Yes	-40°C to +125°C	Tape & Reel	
XR8051ASO8EVB	Evaluation Board	N/A	N/A	N/A	
XR8052 Ordering Information			· · · · · · · · · · · · · · · · · · ·		
XR8052ASO8X	SOIC-8	Yes	-40°C to +125°C	Tape & Reel	
XR8052ASO8MTR	SOIC-8	Yes	-40°C to +125°C	Tape & Reel	
XR8052ASO8EVB	Evaluation Board	N/A	N/A	N/A	
XR8052AMP8X	MSOP-8	Yes	-40°C to +125°C	Tape & Reel	
XR8052AMP8MTR	MSOP-8	Yes	-40°C to +125°C	Tape & Reel	
XR8052AMP8EVB	Evaluation Board	N/A	N/A	N/A	
XR8054 Ordering Information		•			
XR8054ATP14X	TSSOP-14	Yes	-40°C to +125°C	Tape & Reel	
XR8054ATP14MTR	TSSOP-14	Yes	-40°C to +125°C	Tape & Reel	
XR8054ATP14EVB	Evaluation Board	N/A	N/A	N/A	
XR8054ASO14X	SOIC-14	Yes	-40°C to +125°C	Tape & Reel	
XR8054ASO14MTR	SOIC-14	Yes	-40°C to +125°C	Tape & Reel	
XR8054ASO14EVB	Evaluation Board	N/A	N/A	N/A	

Moisture sensitivity level for all parts is MSL-1. Mini tape and reel quantity is 250.

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Revision History

Revision	Date	Description
1B (ECN 1451-05)	December 2014	Reformat into Exar data sheet template. Updated ordering information table to include MTR and EVB part numbers. Updated thermal resistance numbers and package outline drawings. Updated typical small signal bandwidth specifications and plots.

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