



CSD87333Q3D Synchronous Buck NexFET™ Power Block

1 Features

- Half-Bridge Power Block
- Optimized for High-Duty Cycle
- Up to 24 V_{IN}
- 94.7% System Efficiency at 8 A
- 1.5 W P_{LOSS} at 8 A
- Up to 15-A Operation
- High-Frequency Operation (up to 1.5 MHz)
- High-Density SON 3.3-mm x 3.3-mm Footprint
- Optimized for 5-V Gate Drive
- Low-Switching Losses
- Ultra-Low Inductance Package
- RoHS Compliant
- Halogen Free
- Lead-Free Terminal Plating

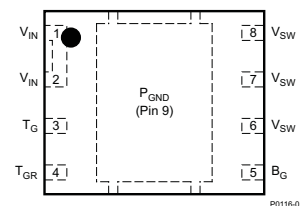
2 Applications

- Synchronous Buck Converters
 - High-Frequency Applications
 - High-Duty Cycle Applications
- Synchronous Boost Converters
- POL DC-DC Converters

3 Description

The CSD87333Q3D NexFET™ power block is an optimized design for synchronous buck and boost applications offering high-current, high-efficiency, and high-frequency capability in a small 3.3-mm x 3.3-mm outline. Optimized for 5-V gate drive applications, this product offers a flexible solution in high-duty cycle applications when paired with an external controller or driver.

Top View

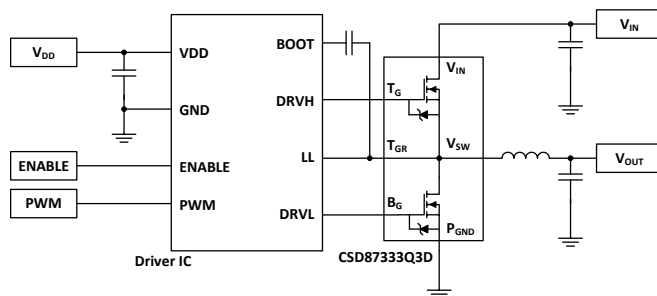


Device Information⁽¹⁾

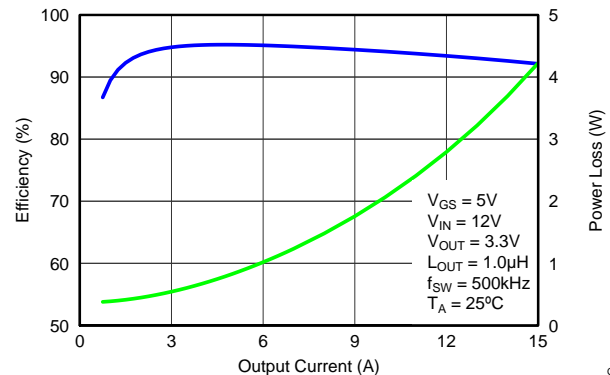
DEVICE	MEDIA	QTY	PACKAGE	SHIP
CSD87333Q3D	13-Inch Reel	2500	SON	Tape and Reel
CSD87333Q3DT	7-Inch Reel	250	3.30-mm x 3.30-mm Plastic Package	

(1) For all available packages, see the orderable addendum at the end of the data sheet.

Typical Circuit



Typical Power Block Efficiency and Power Loss



G001



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4 Revision History

Changes from Original (February 2014) to Revision A

Page

• Added pulse duration note to I_{DM} in the <i>Absolute Maximum Ratings</i> table.....	3
• Added <i>Receiving Notification of Documentation Updates</i> section and <i>Community Resources</i> section to the <i>Device and Documentation Support</i> section	13

5 Specifications

5.1 Absolute Maximum Ratings⁽¹⁾

 $T_A = 25^{\circ}\text{C}$ (unless otherwise noted)

PARAMETER	CONDITIONS	MIN	MAX	UNIT
Voltage	V_{IN} to P_{GND}	–0.8	30	V
	V_{SW} to P_{GND}		30	V
	V_{SW} to P_{GND} (10 ns)		32	V
	T_G to T_{GR}	–0.3	10	V
	B_G to P_{GND}	–0.3	10	V
Pulsed current rating, I_{DM} ⁽²⁾			40	A
Power dissipation, P_D			6	W
Avalanche energy, E_{AS}	Sync FET, $I_D = 19$, $L = 0.1$ mH		18	mJ
	Control FET, $I_D = 19$, $L = 0.1$ mH		18	
Operating junction temperature, T_J		–55	150	$^{\circ}\text{C}$
Storage temperature, T_{stg}		–55	150	$^{\circ}\text{C}$

- (1) Stresses beyond those listed under *Absolute Maximum Ratings* may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated in the *Recommended Operating Conditions* is not implied. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.
- (2) Pulse duration ≤ 50 μs . Duty cycle $\leq 0.01\%$.

5.2 Recommended Operating Conditions

 $T_A = 25^{\circ}\text{C}$ (unless otherwise noted)

PARAMETER	CONDITIONS	MIN	MAX	UNIT
V_{GS} Gate drive voltage		3.3	8	V
V_{IN} Input supply voltage			24	V
f_{SW} Switching frequency	$C_{BST} = 0.1$ μF (min)		1500	kHz
Operating current			15	A
T_J Operating temperature			125	$^{\circ}\text{C}$

5.3 Power Block Performance⁽¹⁾

 $T_A = 25^{\circ}\text{C}$ (unless otherwise noted)

PARAMETER	CONDITIONS	MIN	TYP	MAX	UNIT
P_{LOSS} Power loss ⁽¹⁾	$V_{IN} = 12$ V, $V_{GS} = 5$ V, $V_{OUT} = 3.3$ V, $I_{OUT} = 8$ A, $f_{SW} = 500$ kHz, $L_{OUT} = 1$ μH , $T_J = 25^{\circ}\text{C}$		1.5		W
I_{QVIN} V_{IN} quiescent current	T_G to $T_{GR} = 0$ V B_G to $P_{GND} = 0$ V		10		μA

- (1) Measurement made with six 10- μF (TDK C3216X5R1C106KT or equivalent) ceramic capacitors placed across V_{IN} to P_{GND} pins and using a high current 5-V driver IC.

5.4 Thermal Information

 $T_A = 25^{\circ}\text{C}$ (unless otherwise stated)

THERMAL METRIC		MIN	TYP	MAX	UNIT
$R_{\theta JA}$	Junction-to-ambient thermal resistance (min Cu) ⁽¹⁾			150	$^{\circ}\text{C/W}$
	Junction-to-ambient thermal resistance (max Cu) ⁽¹⁾⁽²⁾			80	
$R_{\theta JC}$	Junction-to-case thermal resistance (top of package) ⁽¹⁾			36	$^{\circ}\text{C/W}$
	Junction-to-case thermal resistance (P_{GND} pin) ⁽¹⁾			3.7	

- (1) $R_{\theta JC}$ is determined with the device mounted on a 1-in² (6.45-cm²), 2-oz (0.071-mm) thick Cu pad on a 1.5-in \times 1.5-in (3.81-cm \times 3.81-cm), 0.06-in (1.52-mm) thick FR4 board. $R_{\theta JC}$ is specified by design while $R_{\theta JA}$ is determined by the user's board design.
- (2) Device mounted on FR4 material with 1-in² (6.45-cm²) Cu.

CSD87333Q3D

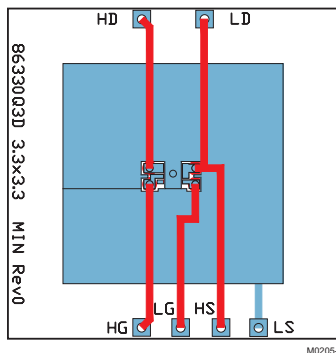
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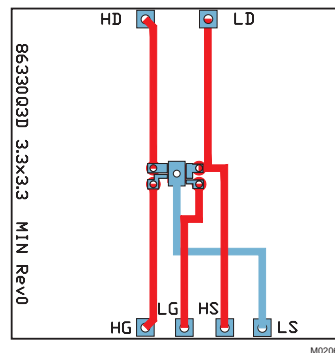
5.5 Electrical Characteristics

 $T_A = 25^\circ\text{C}$ (unless otherwise stated)

PARAMETER		TEST CONDITIONS	Q1 Control FET			Q2 Sync FET			UNIT
			MIN	TYP	MAX	MIN	TYP	MAX	
STATIC CHARACTERISTICS									
BV _{DSS}	Drain-to-source voltage	V _{GS} = 0 V, I _{DS} = 250 μA	30			30			V
I _{DSS}	Drain-to-source leakage current	V _{GS} = 0 V, V _{DS} = 20 V				1			μA
I _{GSS}	Gate-to-source leakage current	V _{DS} = 0 V, V _{GS} = +10 / −8 V				100			nA
V _{GS(th)}	Gate-to-source threshold voltage	V _{DS} = V _{GS} , I _{DS} = 250 μA	0.75	0.95	1.20	0.75	0.95	1.20	V
R _{DS(on)}	Drain-to-source on resistance	V _{GS} = 3.5 V, I _{DS} = 4 A	14.7			17.7			mΩ
		V _{GS} = 4.5 V, I _{DS} = 4 A	13.4			16.1			
		V _{GS} = 8 V, I _{DS} = 4 A	11.9			14.3			
g _{fs}	Transconductance	V _{DS} = 15 V, I _{DS} = 4 A	43			43			S
DYNAMIC CHARACTERISTICS									
C _{ISS}	Input capacitance	V _{GS} = 0 V, V _{DS} = 15 V, f = 1 MHz	509			662			pF
C _{OSS}	Output capacitance		222			289			pF
C _{RSS}	Reverse transfer capacitance		8.2			10.7			pF
R _G	Series gate resistance		3.4			6.8			Ω
Q _g	Gate charge total (4.5 V)	V _{DS} = 15 V, I _{DS} = 4 A	3.5			4.6			nC
Q _{gd}	Gate charge gate-to-drain		0.3			0.3			nC
Q _{gs}	Gate charge gate-to-source		1.6			1.6			nC
Q _{g(th)}	Gate charge at V _{th}		0.6			0.6			nC
Q _{OSS}	Output charge	V _{DS} = 15 V, V _{GS} = 0 V	5.3			5.3			nC
t _{d(on)}	Turnon delay time	V _{DS} = 15 V, V _{GS} = 4.5 V, I _{DS} = 4 A, R _G = 2 Ω	2.1			2.1			ns
t _r	Rise time		3.9			3.9			ns
t _{d(off)}	Turnoff delay time		9.4			9.4			ns
t _f	Fall time		2.2			2.2			ns
DIODE CHARACTERISTICS									
V _{SD}	Diode forward voltage	I _{DS} = 4 A, V _{GS} = 0 V	0.80			1.0			V
Q _{rr}	Reverse recovery charge	V _{DS} = 15 V, I _F = 4 A, di/dt = 300 A/μs	10			10			nC
t _{rr}	Reverse recovery time		11			11			ns



Max $R_{\theta JA} = 80^\circ\text{C/W}$
when mounted on 1 in²
(6.45 cm²) of 2-oz
(0.071-mm) thick Cu.



Max $R_{\theta JA} = 150^\circ\text{C/W}$
when mounted on
minimum pad area of
2-oz (0.071-mm) thick
Cu.

5.6 Typical Power Block Device Characteristics

The typical power block system characteristic curves (Figure 1 through Figure 9) are based on measurements made on a PCB design with dimensions of 4 in (W) × 3.5 in (L) × 0.062 in (H) and 6 copper layers of 1-oz copper thickness. See Applications for detailed explanation. $T_A = 125^\circ\text{C}$, unless stated otherwise.

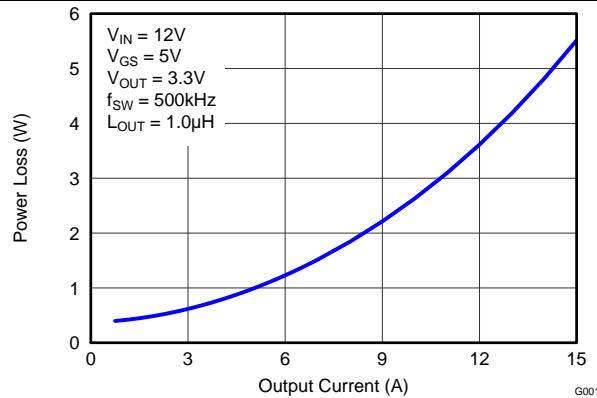


Figure 1. Power Loss vs Output Current

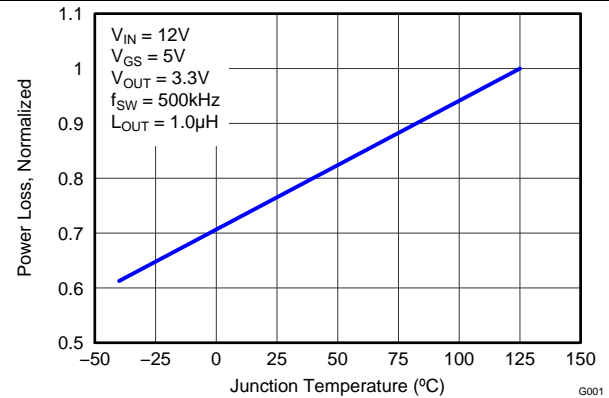


Figure 2. Power Loss vs Temperature

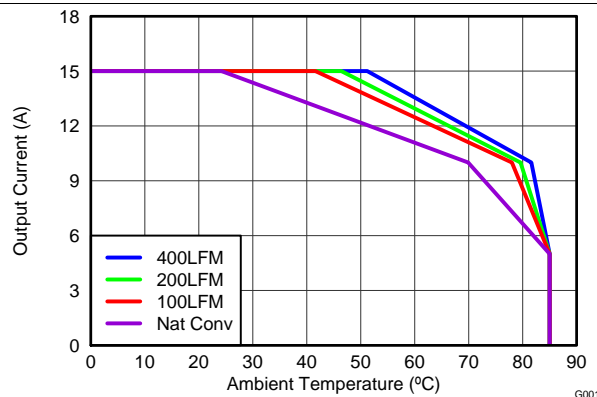


Figure 3. Safe Operating Area – PCB Horizontal Mount

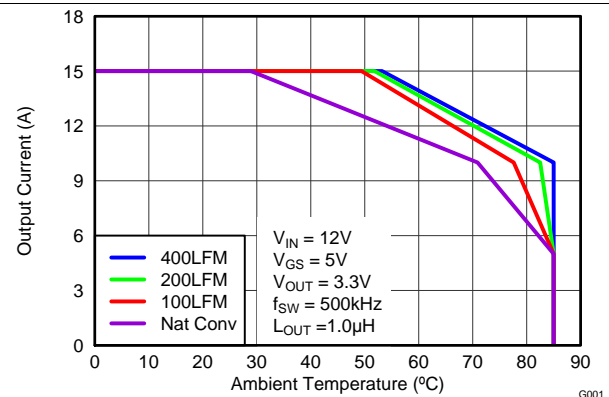


Figure 4. Safe Operating Area – PCB Vertical Mount

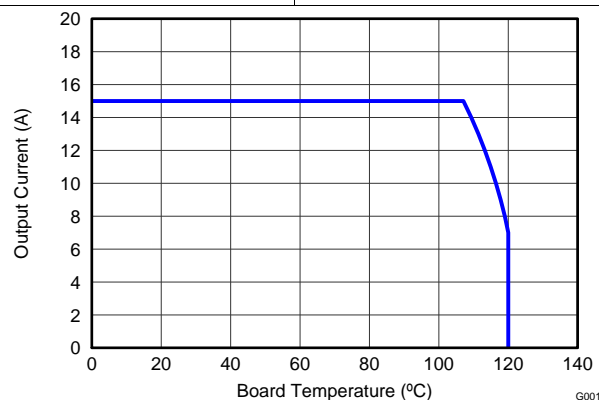
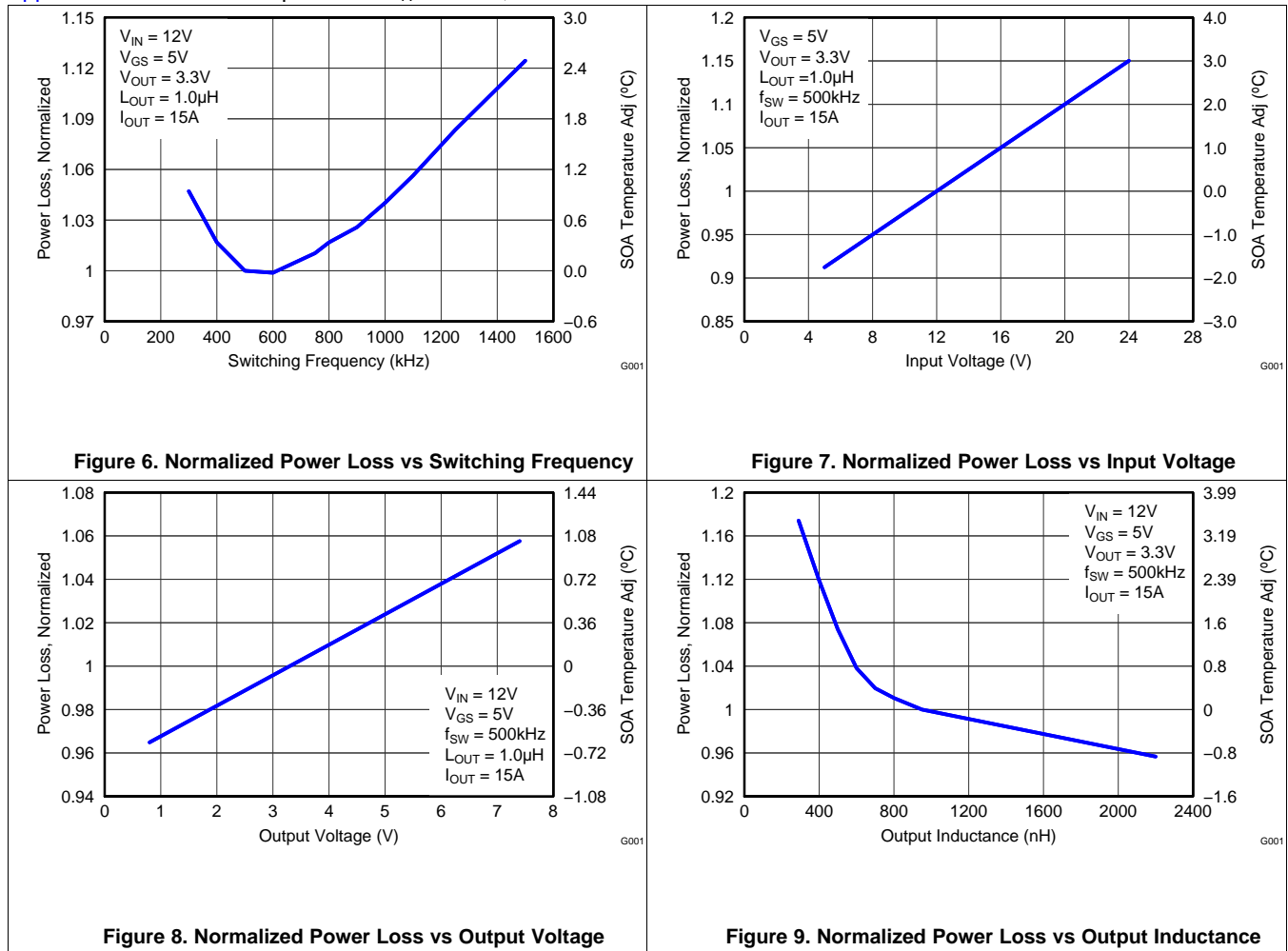


Figure 5. Typical Safe Operating Area

Typical Power Block Device Characteristics (continued)

The typical power block system characteristic curves (Figure 1 through Figure 9) are based on measurements made on a PCB design with dimensions of 4 in (W) × 3.5 in (L) × 0.062 in (H) and 6 copper layers of 1-oz copper thickness. See [Applications](#) for detailed explanation. $T_A = 125^\circ\text{C}$, unless stated otherwise.



5.7 Typical Power Block MOSFET Characteristics

$T_A = 25^\circ\text{C}$, unless stated otherwise.

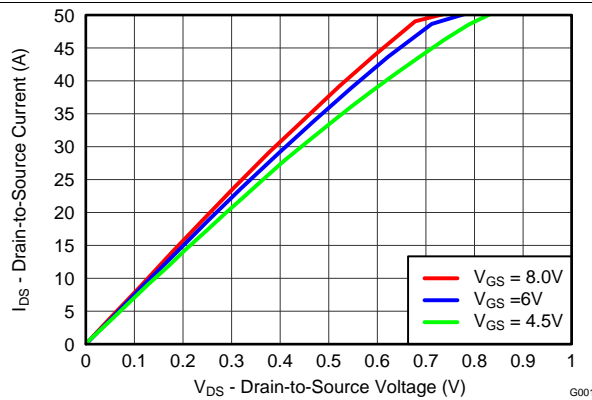


Figure 10. MOSFET Saturation Characteristics

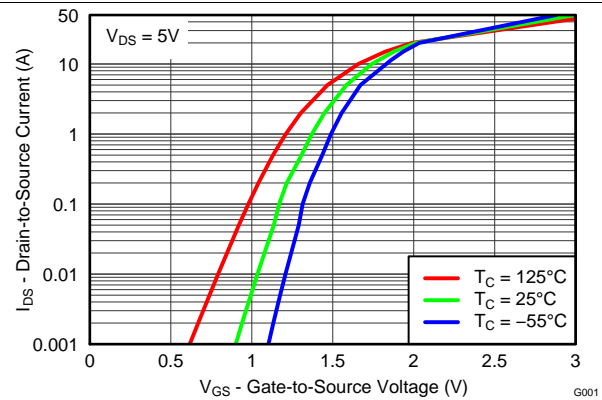


Figure 11. MOSFET Transfer Characteristics

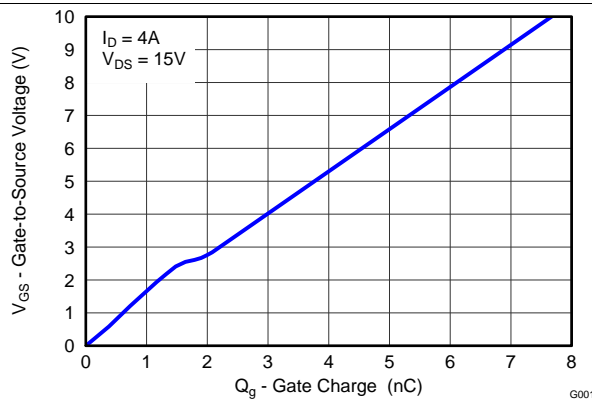


Figure 12. MOSFET Gate Charge

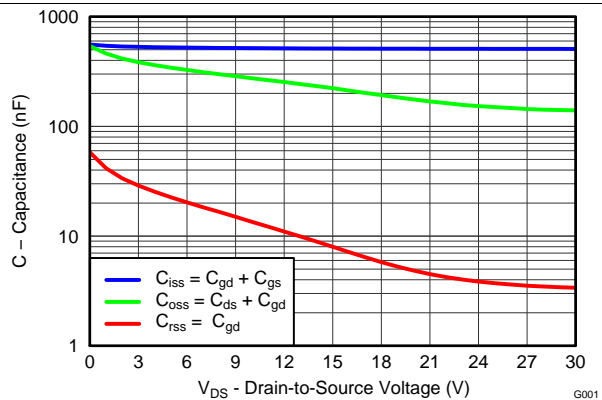


Figure 13. MOSFET Capacitance

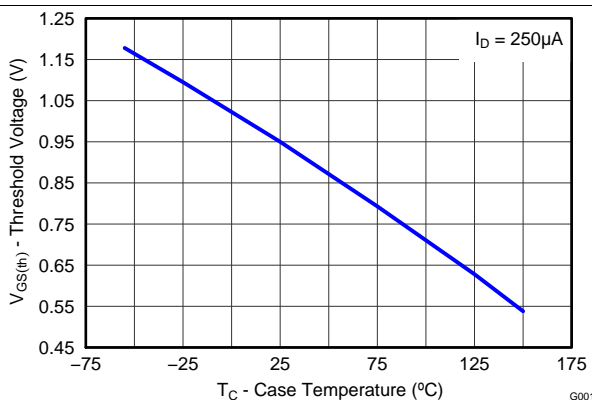


Figure 14. MOSFET $V_{GS(th)}$

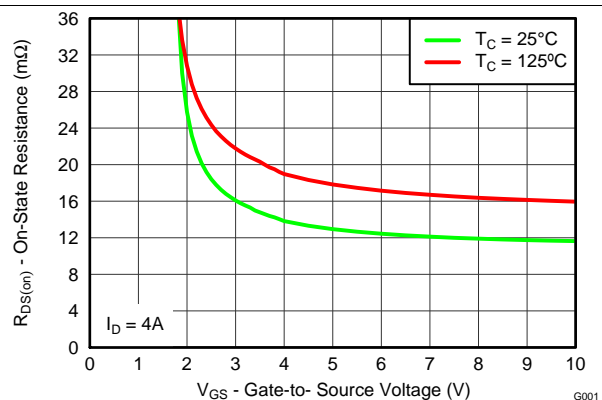


Figure 15. MOSFET $R_{DS(on)}$ vs V_{GS}

Typical Power Block MOSFET Characteristics (continued)

$T_A = 25^\circ\text{C}$, unless stated otherwise.

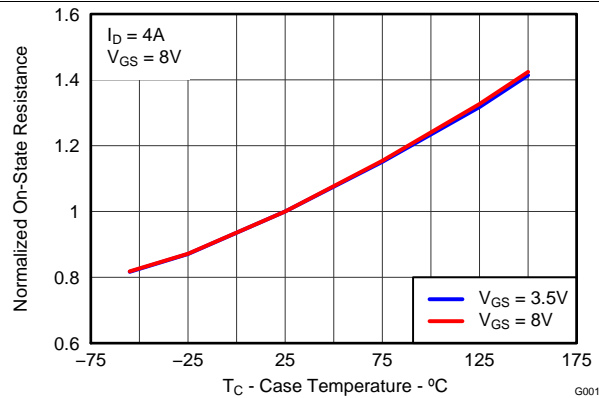


Figure 16. MOSFET Normalized $R_{DS(on)}$

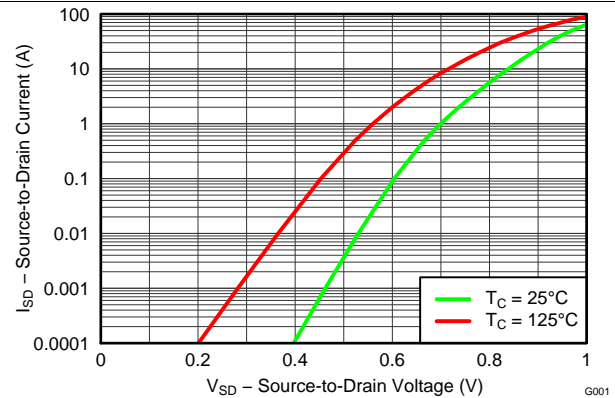


Figure 17. MOSFET Body Diode

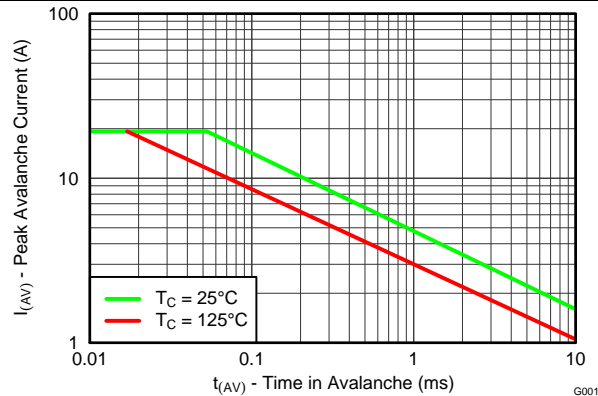


Figure 18. MOSFET Unclamped Inductive Switching

6 Applications

The CSD87333Q3D NexFET power block is an optimized design for synchronous buck applications using 5-V gate drive. The control FET and sync FET silicon are parametrically tuned to yield the lowest power loss and highest system efficiency. As a result, a new rating method is needed which is tailored towards a more systems-centric environment. System-level performance curves such as power loss, SOA, and normalized graphs allow engineers to predict the product performance in the actual application.

6.1 Power Loss Curves

MOSFET centric parameters such as $R_{DS(ON)}$ and Q_{gd} are needed to estimate the loss generated by the devices. In an effort to simplify the design process for engineers, Texas Instruments has provided measured power loss performance curves. Figure 1 plots the power loss of the CSD87333Q3D as a function of load current. This curve is measured by configuring and running the CSD87333Q3D as it would be in the final application (see Figure 19). The measured power loss is the CSD87333Q3D loss and consists of both input conversion loss and gate drive loss. Equation 1 is used to generate the power loss curve.

$$\text{Power loss} = (V_{IN} \times I_{IN}) + (V_{DD} \times I_{DD}) - (V_{SW_AVG} \times I_{OUT}) \quad (1)$$

The power loss curve in Figure 1 is measured at the maximum recommended junction temperatures of 125°C under isothermal test conditions.

6.2 Safe Operating Area (SOA) Curves

The SOA curves in the CSD87333Q3D data sheet provides guidance on the temperature boundaries within an operating system by incorporating the thermal resistance and system power loss. Figure 3 to Figure 5 outline the temperature and airflow conditions required for a given load current. The area under the curve dictates the safe operating area. All the curves are based on measurements made on a PCB design with dimensions of 4 in (W) × 3.5 in (L) × 0.062 in (T) and 6 copper layers of 1-oz copper thickness.

6.3 Normalized Curves

The normalized curves in the CSD87333Q3D data sheet provides guidance on the power loss and SOA adjustments based on their application specific needs. These curves show how the power loss and SOA boundaries adjust for a given set of system conditions. The primary Y-axis is the normalized change in power loss, and the secondary Y-axis is the change in system temperature required in order to comply with the SOA curve. The change in power loss is a multiplier for the power loss curve and the change in temperature is subtracted from the SOA curve.

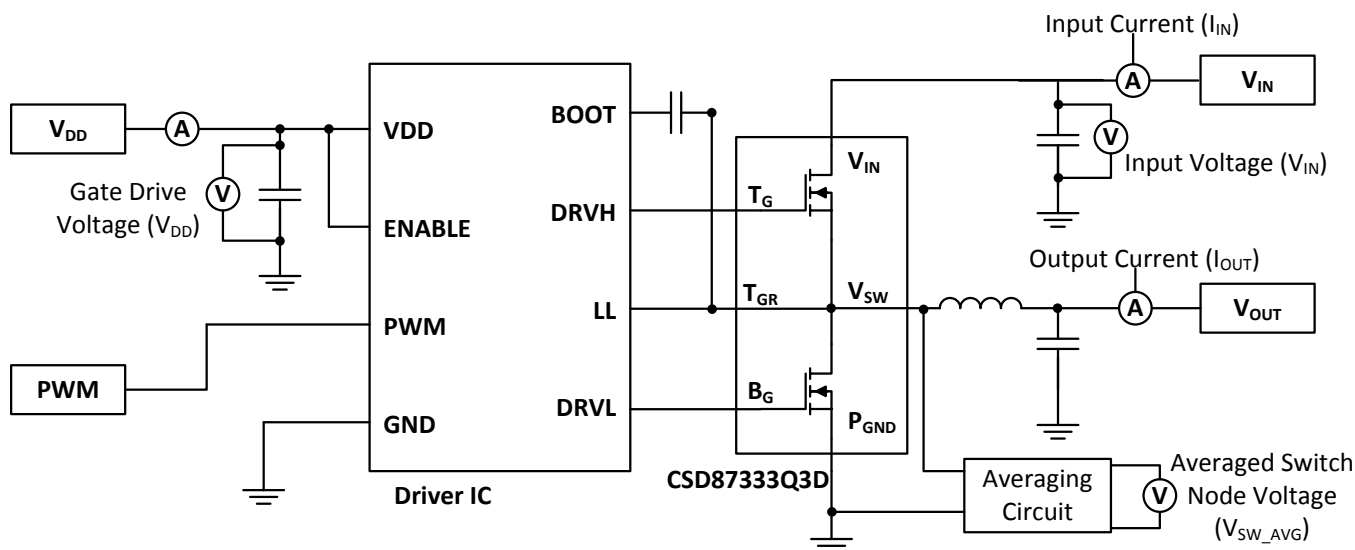


Figure 19. Typical Application

6.4 Calculating Power Loss and SOA

The user can estimate product loss and SOA boundaries by arithmetic means (see [Design Example](#)). Though the power loss and SOA curves in this data sheet are taken for a specific set of test conditions, the following procedure outlines the steps the user should take to predict product performance for any set of system conditions.

6.4.1 Design Example

Operating conditions:

- Output current = 10 A
- Input voltage = 20 V
- Output voltage = 1 V
- Switching frequency = 1000 kHz
- Inductor = 0.6 μ H

6.4.2 Calculating Power Loss

- Power loss at 10 A = 2.6 W ([Figure 1](#))
- Normalized power loss for input voltage ≈ 1.1 ([Figure 7](#))
- Normalized power loss for output voltage ≈ 0.96 ([Figure 8](#))
- Normalized power loss for switching frequency ≈ 1.04 ([Figure 6](#))
- Normalized power loss for output inductor ≈ 1.03 ([Figure 9](#))
- **Final calculated power loss = $2.6 \text{ W} \times 1.1 \times 0.96 \times 1.04 \times 1.03 \approx 2.9 \text{ W}$**

6.4.3 Calculating SOA Adjustments

- SOA adjustment for input voltage $\approx 2^\circ\text{C}$ ([Figure 7](#))
- SOA adjustment for output voltage $\approx -0.2^\circ\text{C}$ ([Figure 8](#))
- SOA adjustment for switching frequency $\approx 0.8^\circ\text{C}$ ([Figure 6](#))
- SOA adjustment for output inductor $\approx 0.8^\circ\text{C}$ ([Figure 9](#))
- **Final calculated SOA adjustment = $2 + (-0.2) + 0.8 + 0.8 \approx 3.4^\circ\text{C}$**

In the [Design Example](#), the estimated power loss of the CSD87333Q3D would increase to 2.9 W. In addition, the maximum allowable board or ambient temperature, or both, would have to decrease by 3.4°C . [Figure 20](#) graphically shows how the SOA curve would be adjusted accordingly.

1. Start by drawing a horizontal line from the application current to the SOA curve.
2. Draw a vertical line from the SOA curve intercept down to the board or ambient temperature.
3. Adjust the SOA board or ambient temperature by subtracting the temperature adjustment value.

In the design example, the SOA temperature adjustment yields a reduction in allowable board/ambient temperature of 3.4°C . In the event the adjustment value is a negative number, subtracting the negative number would yield an increase in allowable board or ambient temperature.

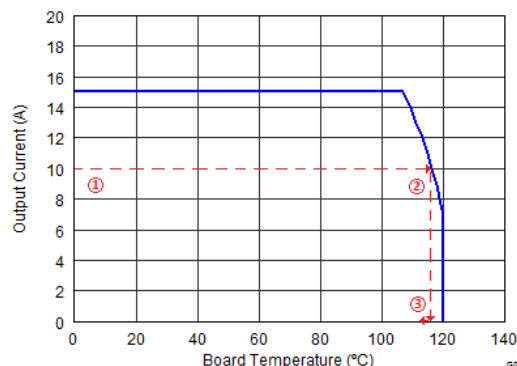


Figure 20. Power Block SOA

7 Recommended PCB Design Overview

There are two key system-level parameters that can be addressed with a proper PCB design: electrical and thermal performance. Properly optimizing the PCB layout yields maximum performance in both areas. A brief description on how to address each parameter is provided.

7.1 Electrical Performance

The power block has the ability to switch voltages at rates greater than 10 kV/μs. Special care must be then taken with the PCB layout design and placement of the input capacitors, driver IC, and output inductor.

- The placement of the input capacitors relative to the power block's VIN and PGND pins should have the highest priority during the component placement routine. It is critical to minimize these node lengths. As such, ceramic input capacitors need to be placed as close as possible to the VIN and PGND pins (see [Figure 21](#)). The example in [Figure 21](#) uses 6 × 10-μF ceramic capacitors (TDK part number C3216X5R1C106KT or equivalent). Notice there are ceramic capacitors on both sides of the board with an appropriate amount of vias interconnecting both layers. In terms of priority of placement next to the power block, C5, C7, C19, and C8 should follow in order.
 - The driver IC should be placed relatively close to the power block gate pins. T_G and B_G should connect to the outputs of the driver IC. The T_{GR} pin serves as the return path of the high-side gate drive circuitry and should be connected to the phase pin of the IC (sometimes called LX, LL, SW, PH, and so forth). The bootstrap capacitor for the driver IC will also connect to this pin.
 - The switching node of the output inductor should be placed relatively close to the power block VSW pins. Minimizing the node length between these two components will reduce the PCB conduction losses and actually reduce the switching noise level.⁽¹⁾ In the event the switch node waveform exhibits ringing that reaches undesirable levels, the use of a boost resistor or RC snubber can be an effective way to easily reduce the peak ring level. The recommended boost resistor value will range between 1 Ω to 4.7 Ω depending on the output characteristics of driver IC used in conjunction with the power block. The RC snubber values can range from 0.5 Ω to 2.2 Ω for the R and 330 pF to 2200 pF for the C. Please refer to [Snubber Circuits: Theory, Design and Application](#) (SLUP100) for more details on how to properly tune the RC snubber values. The RC snubber should be placed as close as possible to the Vsw node and PGND (see [Figure 21](#)).⁽¹⁾
- (1) Keong W. Kam, David Pommerenke, "EMI Analysis Methods for Synchronous Buck Converter EMI Root Cause Analysis", University of Missouri – Rolla

7.1 Thermal Performance

The power block has the ability to utilize the GND planes as the primary thermal path. As such, the use of thermal vias is an effective way to pull away heat from the device and into the system board. Concerns of solder voids and manufacturability problems can be addressed by the use of three basic tactics to minimize the amount of solder attach that will wick down the via barrel:

- Intentionally space out the vias from each other to avoid a cluster of holes in a given area.
- Use the smallest drill size allowed in your design. The example in [Figure 21](#) uses vias with a 10-mil drill hole and a 16-mil capture pad.
- Tent the opposite side of the via with solder-mask.

The number and drill size of the thermal vias should align with the end user's PCB design rules and manufacturing capabilities.

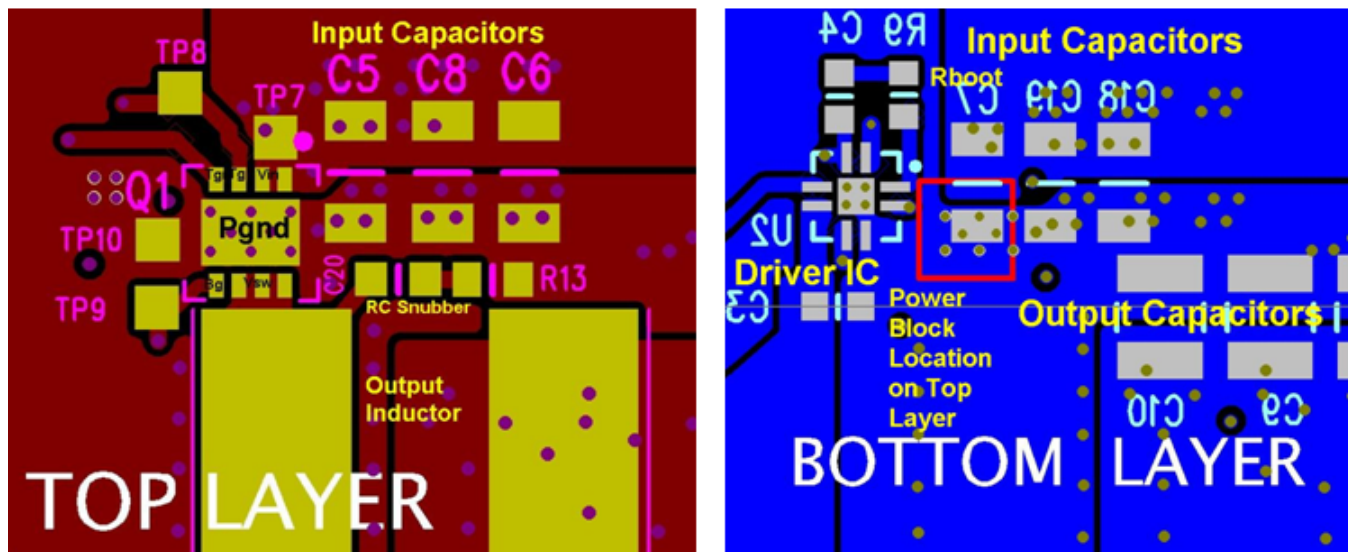


Figure 21. Recommended PCB Layout (Top Down)

8 Device and Documentation Support

8.1 Receiving Notification of Documentation Updates

To receive notification of documentation updates, navigate to the device product folder on ti.com. In the upper right corner, click on *Alert me* to register and receive a weekly digest of any product information that has changed. For change details, review the revision history included in any revised document.

8.2 Community Resources

The following links connect to TI community resources. Linked contents are provided "AS IS" by the respective contributors. They do not constitute TI specifications and do not necessarily reflect TI's views; see TI's [Terms of Use](#).

TI E2E™ Online Community *TI's Engineer-to-Engineer (E2E) Community*. Created to foster collaboration among engineers. At e2e.ti.com, you can ask questions, share knowledge, explore ideas and help solve problems with fellow engineers.

Design Support *TI's Design Support* Quickly find helpful E2E forums along with design support tools and contact information for technical support.

8.3 Trademarks

NexFET, E2E are trademarks of Texas Instruments.
All other trademarks are the property of their respective owners.

8.4 Electrostatic Discharge Caution



These devices have limited built-in ESD protection. The leads should be shorted together or the device placed in conductive foam during storage or handling to prevent electrostatic damage to the MOS gates.

8.5 Glossary

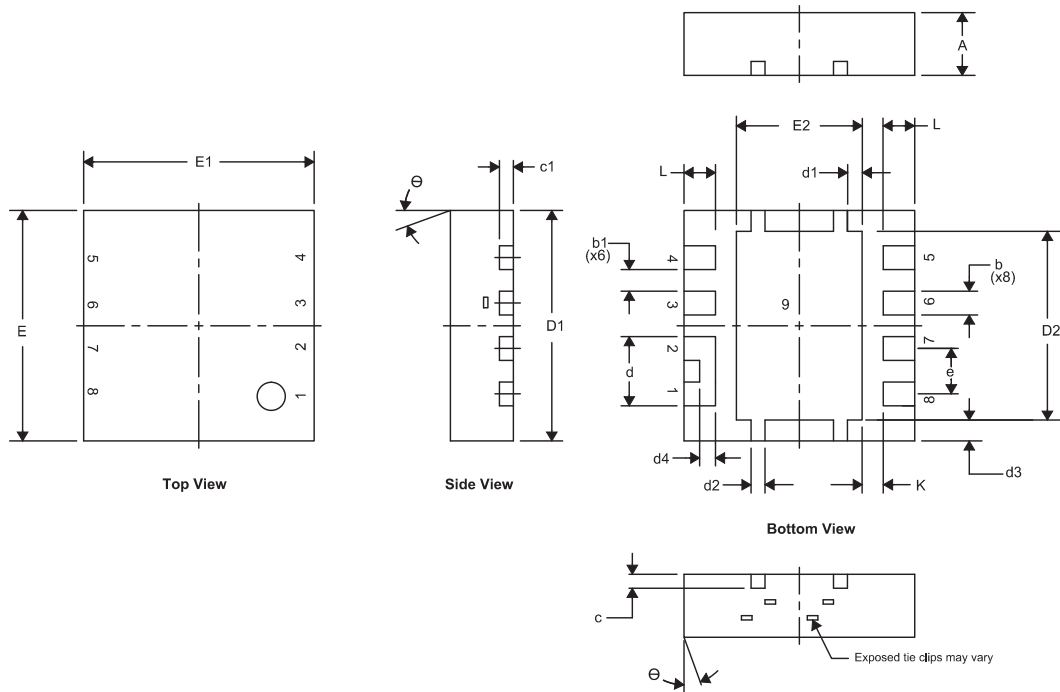
[SLYZ022](#) — *TI Glossary*.

This glossary lists and explains terms, acronyms, and definitions.

9 Mechanical, Packaging, and Orderable Information

The following pages include mechanical packaging and orderable information. This information is the most current data available for the designated devices. This data is subject to change without notice and revision of this document. For browser-based versions of this data sheet, refer to the left-hand navigation.

9.1 Q3D Package Dimensions



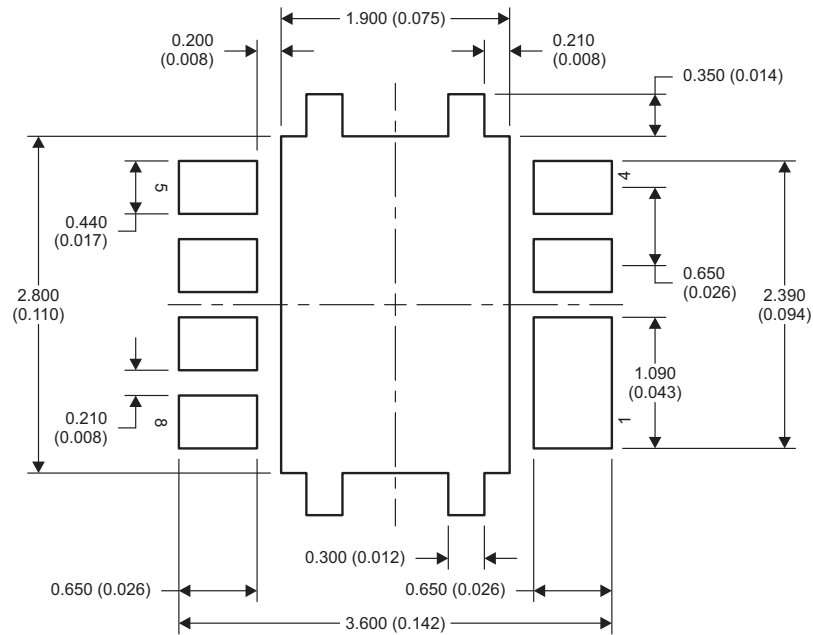
DIM	MILLIMETERS		INCHES	
	MIN	MAX	MIN	MAX
A	0.850	1.050	0.033	0.041
b	0.280	0.400	0.011	0.016
b1	0.310 NOM		0.012 NOM	
c	0.150	0.250	0.006	0.010
c1	0.150	0.250	0.006	0.010
d	0.940	1.040	0.037	0.041
d1	0.160	0.260	0.006	0.010
d2	0.150	0.250	0.006	0.010
d3	0.250	0.350	0.010	0.014
d4	0.175	0.275	0.007	0.011
D1	3.200	3.400	0.126	0.134
D2	2.650	2.750	0.104	0.108
E	3.200	3.400	0.126	0.134
E1	3.200	3.400	0.126	0.134
E2	1.750	1.850	0.069	0.073
e	0.650 TYP		0.026 TYP	
L	0.400	0.500	0.016	0.020
θ	0.000	—	—	—
K	0.300 TYP		0.012 TYP	

9.2 Pinout Configuration

Table 1. Pinout Configuration

POSITION	DESIGNATION
Pin 1	V_{IN}
Pin 2	V_{IN}
Pin 3	T_G
Pin 4	T_{GR}
Pin 5	B_G
Pin 6	V_{SW}
Pin 7	V_{SW}
Pin 8	V_{SW}
Pin 9	P_{GND}

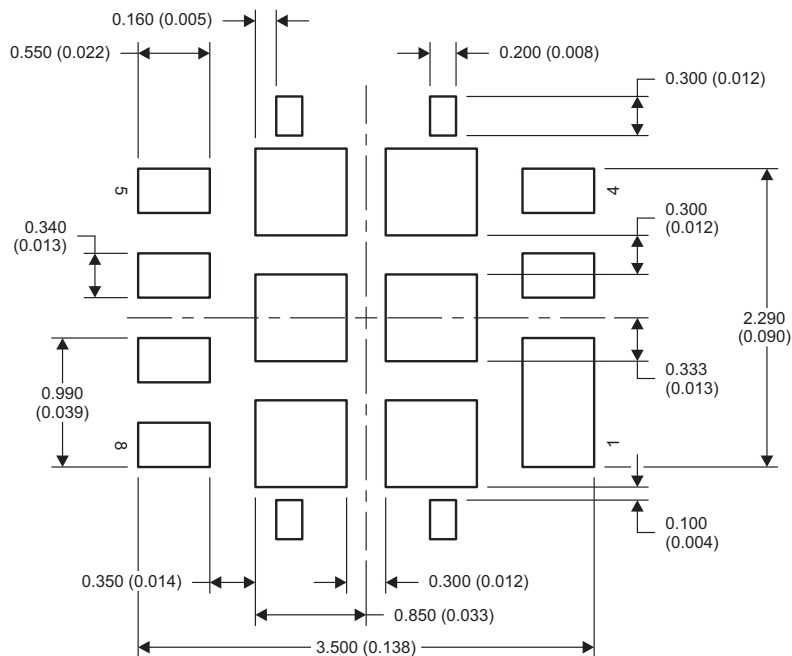
9.3 Land Pattern Recommendation



M0193-01

NOTE: Dimensions are in mm (in).

9.4 Stencil Recommendation

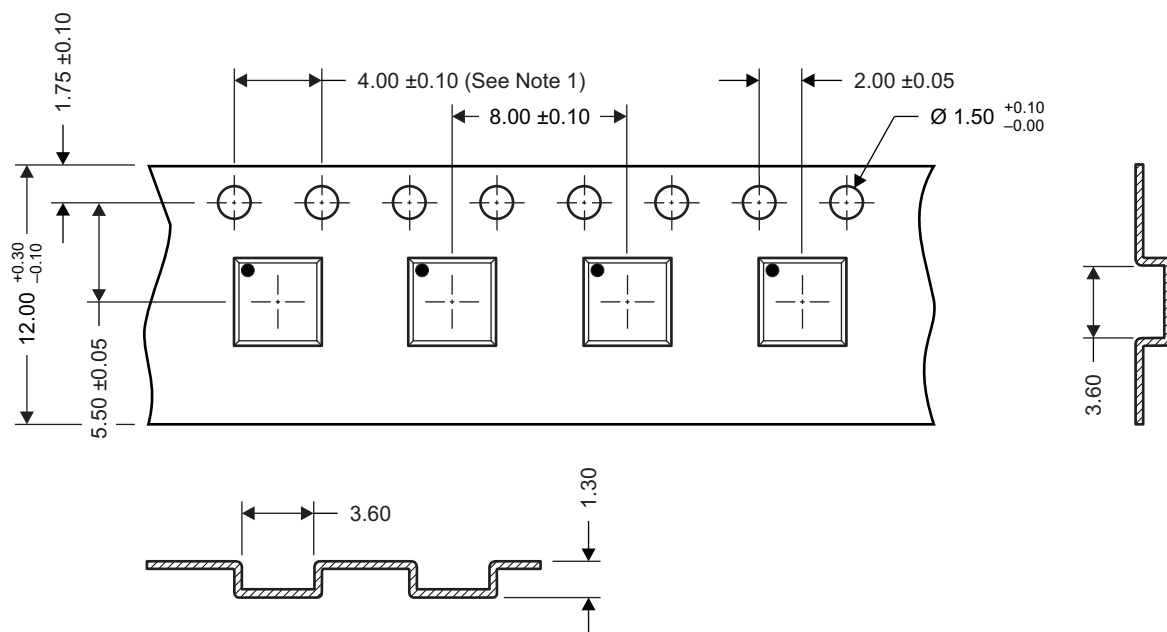


M0207-01

NOTE: Dimensions are in mm (in).

For recommended circuit layout for PCB designs, see [Reducing Ringing Through PCB Layout Techniques](#) (SLPA005).

9.5 Q3D Tape and Reel Information



M0144-01

- NOTES: 1. 10-sprocket hole-pitch cumulative tolerance ± 0.2 .
 2. Camber not to exceed 1 mm in 100 mm, noncumulative over 250 mm.
 3. Material: black static-dissipative polystyrene.
 4. All dimensions are in mm, unless otherwise specified.
 5. Thickness: 0.3 ± 0.05 mm.
 6. MSL1 260°C (IR and convection) PbF reflow compatible.

PACKAGING INFORMATION

Orderable part number	Status (1)	Material type (2)	Package Pins	Package qty Carrier	RoHS (3)	Lead finish/ Ball material (4)	MSL rating/ Peak reflow (5)	Op temp (°C)	Part marking (6)
CSD87333Q3D	Active	Production	VSON-CLIP (DPB) 8	2500 LARGE T&R	ROHS Exempt	NIPDAU SN	Level-1-260C-UNLIM	-55 to 150	87333D
CSD87333Q3DT	Active	Production	VSON-CLIP (DPB) 8	250 SMALL T&R	ROHS Exempt	NIPDAU SN	Level-1-260C-UNLIM	-55 to 150	87333D

⁽¹⁾ **Status:** For more details on status, see our [product life cycle](#).

⁽²⁾ **Material type:** When designated, preproduction parts are prototypes/experimental devices, and are not yet approved or released for full production. Testing and final process, including without limitation quality assurance, reliability performance testing, and/or process qualification, may not yet be complete, and this item is subject to further changes or possible discontinuation. If available for ordering, purchases will be subject to an additional waiver at checkout, and are intended for early internal evaluation purposes only. These items are sold without warranties of any kind.

⁽³⁾ **RoHS values:** Yes, No, RoHS Exempt. See the [TI RoHS Statement](#) for additional information and value definition.

⁽⁴⁾ **Lead finish/Ball material:** Parts may have multiple material finish options. Finish options are separated by a vertical ruled line. Lead finish/Ball material values may wrap to two lines if the finish value exceeds the maximum column width.

⁽⁵⁾ **MSL rating/Peak reflow:** The moisture sensitivity level ratings and peak solder (reflow) temperatures. In the event that a part has multiple moisture sensitivity ratings, only the lowest level per JEDEC standards is shown. Refer to the shipping label for the actual reflow temperature that will be used to mount the part to the printed circuit board.

⁽⁶⁾ **Part marking:** There may be an additional marking, which relates to the logo, the lot trace code information, or the environmental category of the part.

Multiple part markings will be inside parentheses. Only one part marking contained in parentheses and separated by a "~" will appear on a part. If a line is indented then it is a continuation of the previous line and the two combined represent the entire part marking for that device.

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TAPE AND REEL INFORMATION



*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Reel Diameter (mm)	Reel Width W1 (mm)	A0 (mm)	B0 (mm)	K0 (mm)	P1 (mm)	W (mm)	Pin1 Quadrant
CSD87333Q3D	VSON-CLIP	DPB	8	2500	330.0	12.4	3.6	3.6	1.2	8.0	12.0	Q1
CSD87333Q3DT	VSON-CLIP	DPB	8	250	180.0	12.4	3.6	3.6	1.2	8.0	12.0	Q1

TAPE AND REEL BOX DIMENSIONS



*All dimensions are nominal

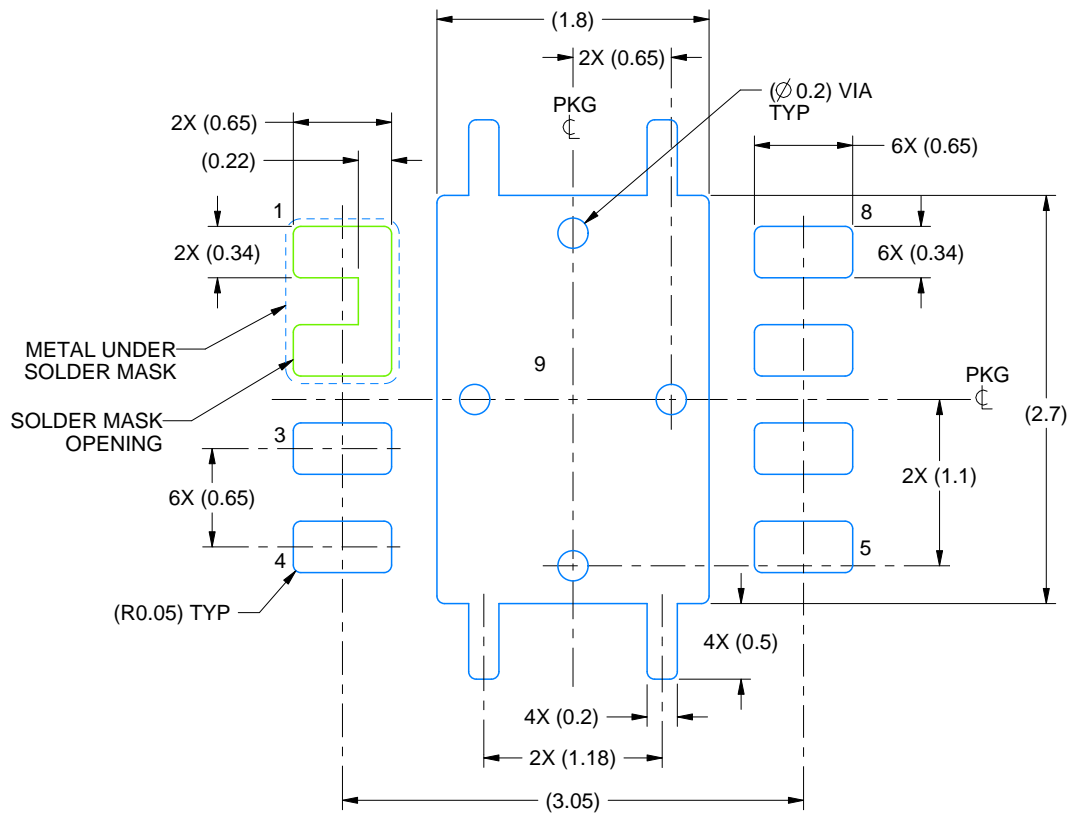
Device	Package Type	Package Drawing	Pins	SPQ	Length (mm)	Width (mm)	Height (mm)
CSD87333Q3D	VSON-CLIP	DPB	8	2500	346.0	346.0	33.0
CSD87333Q3DT	VSON-CLIP	DPB	8	250	182.0	182.0	20.0

EXAMPLE BOARD LAYOUT

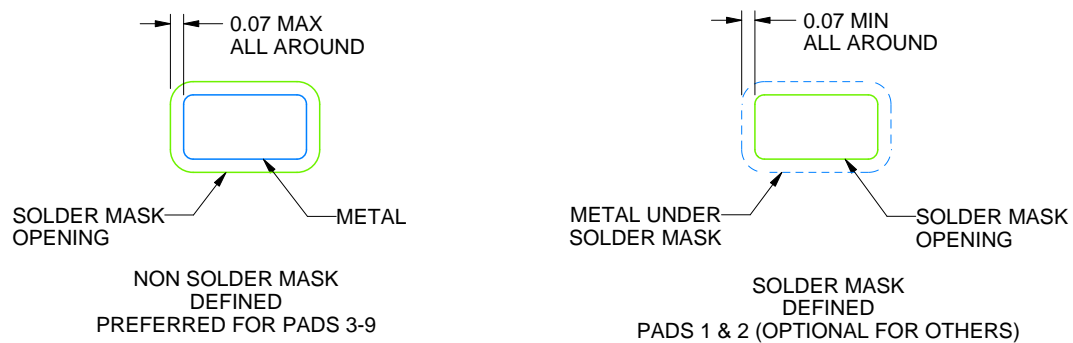
DPB0008A

VSON - 1.05 mm max height

PLASTIC SMALL OUTLINE - NO LEAD



LAND PATTERN EXAMPLE
SCALE:20X



SOLDER MASK DETAILS

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NOTES: (continued)

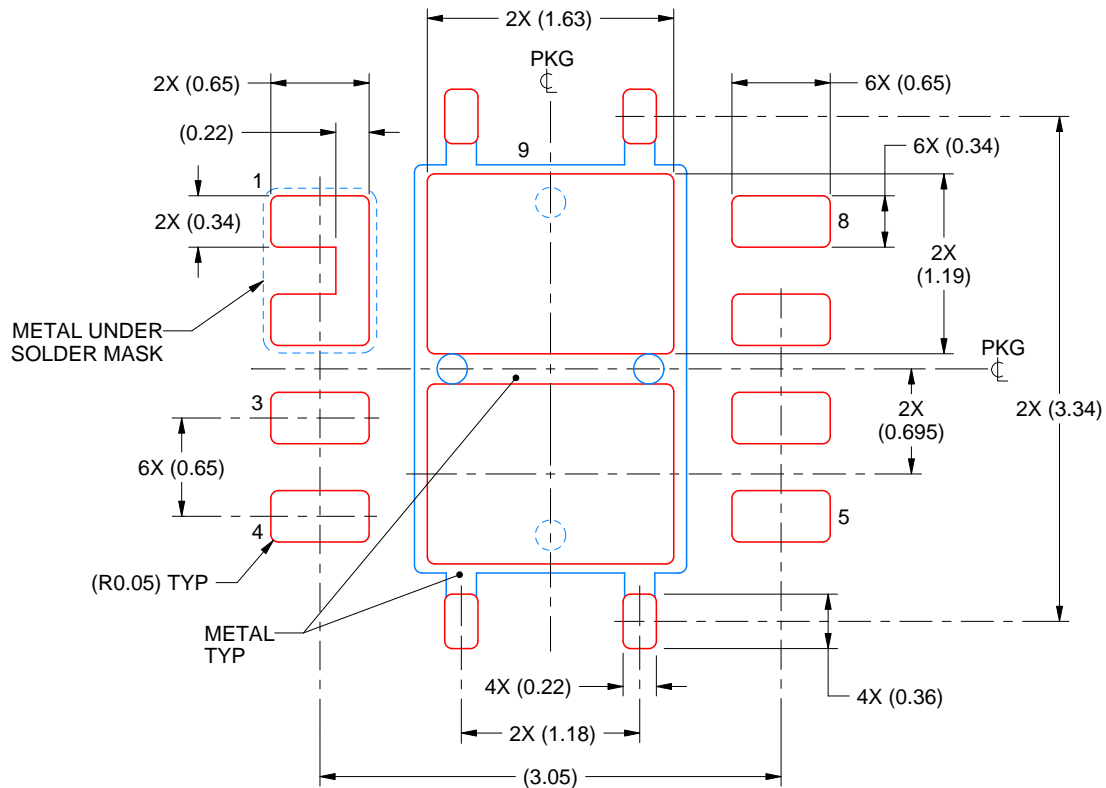
5. This package is designed to be soldered to a thermal pad on the board. For more information, see Texas Instruments literature number SLUA271 (www.ti.com/lit/sluea271).
6. Vias are optional depending on application, refer to device data sheet. If some or all are implemented, recommended via locations are shown.

EXAMPLE STENCIL DESIGN

DPB0008A

VSON - 1.05 mm max height

PLASTIC SMALL OUTLINE - NO LEAD



SOLDER PASTE EXAMPLE
BASED ON 0.125 mm THICK STENCIL

EXPOSED PAD 9
76% PRINTED SOLDER COVERAGE BY AREA UNDER PACKAGE
SCALE:20X

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NOTES: (continued)

7. Laser cutting apertures with trapezoidal walls and rounded corners may offer better paste release. IPC-7525 may have alternate design recommendations.

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