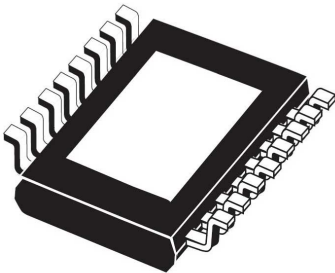


Double channel high-side driver with current sense analog feedback for automotive applications


PowerSSO-16
Product status link
[VND9008AJ](#)
Product summary

Order code	VND9008AJTR
Package	PowerSSO-16
Packing	Tape and reel

Features

Max. transient supply voltage	V_{CC}	36 V
Operating voltage range	V_{CC}	4 to 28 V
Typ. on-state resistance (per channel)	R_{ON}	9.4 m Ω
Current limitation (typ.)	I_{LIMH}	67 A
Standby current (max.)	I_{STBY}	0.5 μ A



- AEC-Q100 qualified
- Extreme low voltage operation for deep cold cranking applications (compliant with LV124, revision 2013)
- General
 - Double channel smart high-side driver with current sense analog feedback
 - Very low standby current
 - Compatible with 3 V and 5 V CMOS outputs
- Current sense diagnostic functions
 - Multiplexed analog feedback of load current with high precision proportional current mirror
 - Overload and short to ground (power limitation) indication
 - Thermal shutdown indication
 - OFF-state open-load detection
 - Output short to V_{CC} detection
 - Sense enable/disable
- Protections
 - Undervoltage shutdown
 - Overvoltage clamp
 - Load current limitation
 - Self limiting of fast thermal transients
 - Configurable latch-off on overtemperature or power limitation with dedicated fault reset pin
 - Loss of ground and loss of V_{CC}
 - Reverse battery through self turn-on
 - Electrostatic discharge protection

Applications

- Automotive resistive, inductive and capacitive loads
- Protected supply for ADAS systems: radars and sensors
- Automotive headlamps (up to H7)

Description

The device is a double channel high-side driver manufactured using ST proprietary VIPower M0-9 technology and housed in PowerSSO-16 package. The device is designed to drive 12 V automotive grounded loads through a 3 V and 5 V CMOS-compatible interface, providing protection and diagnostics.

The device integrates advanced protective functions such as load current limitation, overload active management by power limitation and overtemperature shutdown with configurable latch-off.

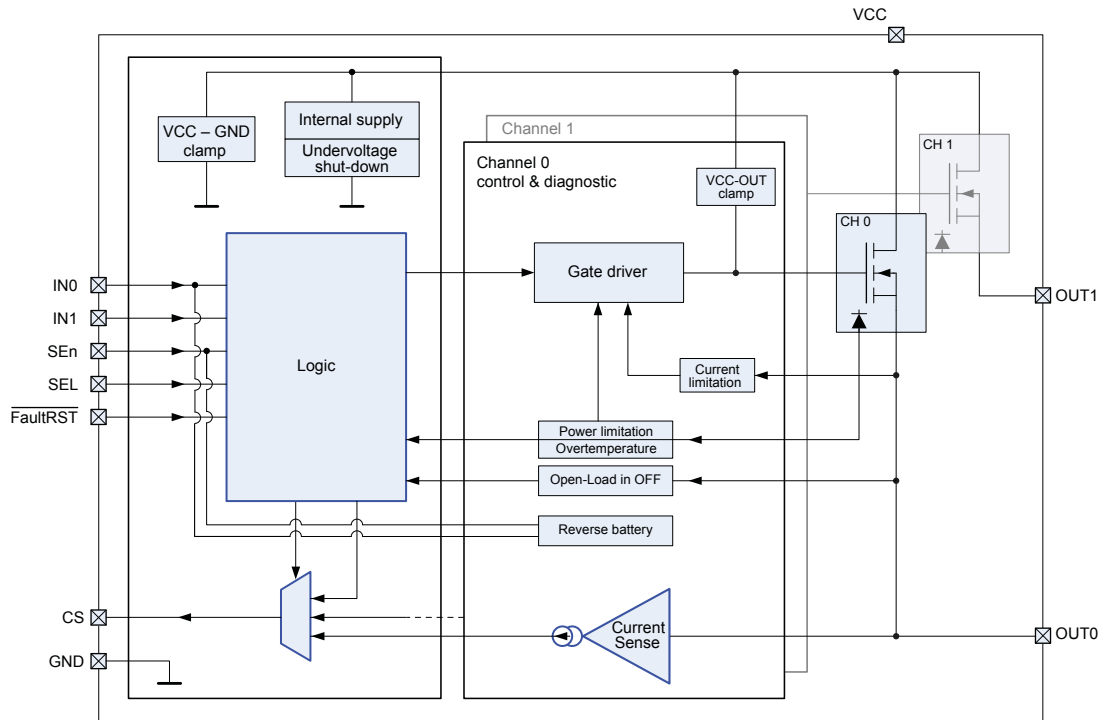
A $\overline{\text{FaultRST}}$ pin unlatches the output in case of fault or disables the latch-off functionality.

A dedicated multifunction multiplexed analog output pin delivers diagnostic functions including high precision proportional load current sense, in addition to the detection of overload and short circuit to ground, short to V_{CC} and OFF-state open-load.

A sense enable pin allows OFF-state diagnosis to be disabled during the module low-power mode as well as external sense resistor sharing among similar devices.

1 Block diagram and pin description

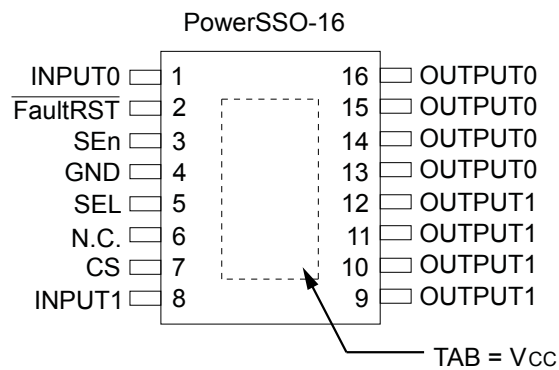
Figure 1. Block diagram



GADG0507181254PS

Table 1. Pin functions

Name	Function
V _{CC}	Battery connection.
OUTPUT _{0,1}	Power output.
GND	Ground connection. Must be reverse battery protected by an external diode / resistor network.
INPUT _{0,1}	Voltage controlled input pin with hysteresis, compatible with 3 V and 5 V CMOS outputs. It controls output switch state.
CS	Multiplexed analog sense output pin; it delivers a current proportional to the selected load current.
SEn	Active high compatible with 3 V and 5 V CMOS outputs pin; it enables the CS diagnostic pin.
SEL	Active high compatible with 3 V and 5 V CMOS outputs pin; it addresses the CS multiplexer.
FaultRST	Active low compatible with 3 V and 5 V CMOS outputs pin; it unlatches the output in case of fault; If kept low, sets the outputs in auto-restart mode.

Figure 2. Configuration diagram (top view)


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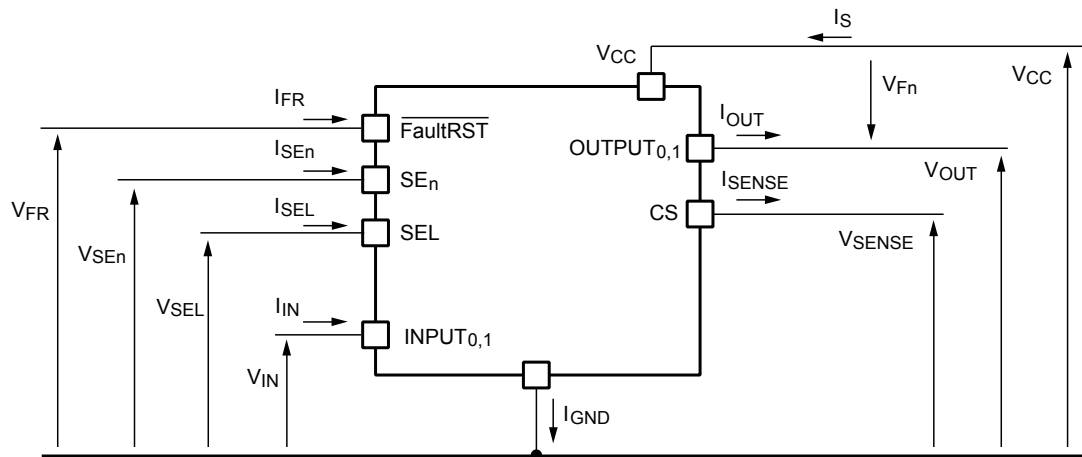
Table 2. Suggested connections for unused and not connected pins

Connection / pin	CS	N.C.	Output	Input	SEn, SELx, FaultRST
Floating	Not allowed	X ⁽¹⁾	X	X	X
To ground	Through 1 kΩ resistor	X	Not allowed	Through 15 kΩ resistor	Through 15 kΩ resistor

1. X: do not care.

2 Electrical specifications

Figure 3. Current and voltage conventions



GADG1605171029PS

Note: $V_{Fn} = V_{OUTn} - V_{CC}$ during reverse battery condition.

2.1 Absolute maximum ratings

Stressing the device above the rating listed in [Table 3. Absolute maximum ratings](#) may cause permanent damage to the device. These are stress ratings only and operation of the device at these or any other conditions above those indicated in the operating sections of this specification is not implied. Exposure to the conditions in the table below for extended periods may affect the device reliability.

Table 3. Absolute maximum ratings

Symbol	Parameter		Value	Unit
V _{CC}	DC supply voltage		36	V
-V _{CC}	Reverse DC supply voltage		0.3	
V _{CCJS}	Maximum jump start voltage for single pulse short-circuit protection		28	V
-I _{GND}	DC reverse ground pin current		200	mA
I _{OUT}	OUTPUT _{0,1} DC output current		Internally limited	A
-I _{OUT}	Reverse DC output current		15	
I _{SEn}	SEn DC input current		-1 to 3	mA
I _{IN}	INPUT _{0,1} DC input current		-1 to 10	
I _{SEL}	SEL DC input current			
I _{FR}	FaultRST DC input current			
I _{SENSE}	CS pin DC output current (V _{GND} = V _{CC} and V _{SENSE} < 0 V)		10	mA
	CS pin DC output current in reverse (V _{CC} < 0 V)		-20	
E _{MAX}	Maximum switching energy (single pulse) (T _{Jstart} = 150 °C)		26	mJ
V _{ESD}	Electrostatic discharge (JEDEC 22A-114F)	INPUT _{0,1}	2000	V
		CS	2000	
		SEn, SEL, FaultRST	2000	

Symbol	Parameter		Value	Unit
V _{ESD}	Electrostatic discharge (JEDEC 22A-114F)	OUTPUT _{0,1}	4000	V
		V _{CC}	4000	
	Charge device model (CDM-AEC-Q100-011)		750	
T _J	Operating junction temperature range		-40 to 150	°C
T _{stg}	Storage temperature range		-55 to 150	°C

2.2 Thermal data

Table 4. Thermal data

Symbol	Parameter	Typ. value	Unit
R _{thJB}	Thermal resistance, junction-to-board (JEDEC JESD 51-5 / 51-8) ⁽¹⁾⁽²⁾	6.3	°C/W
R _{thJA}	Thermal resistance, junction-to-ambient (JEDEC JESD 51-5) ⁽¹⁾⁽³⁾	53.8	°C/W
	Thermal resistance, junction-to-ambient (JEDEC JESD 51-7) ⁽¹⁾⁽²⁾	22	

1. One channel ON.
2. Device mounted on a four-layer 2s2p PCB.
3. Device mounted on a two-layer 2s0p PCB with 2 cm² heatsink copper trace.

2.3 Main electrical characteristics

7 V < V_{CC} < 28 V; -40 °C < T_J < 150 °C, unless otherwise specified.

All typical values refer to V_{CC} = 13 V; T_J = 25 °C, unless otherwise specified.

Table 5. Power section

Symbol	Parameter	Test conditions	Min.	Typ.	Max.	Unit
V _{CC}	Operating supply voltage		4	13	28	V
V _{USD}	Undervoltage shutdown			2.1	2.7	V
V _{USDReset}	Undervoltage shutdown reset				4.5	V
V _{USDhyst}	Undervoltage shutdown hysteresis			0.15		V
R _{ON}	On-state resistance ⁽¹⁾	I _{OUT} = 4.6 A; T _J = 25 °C		9.4		mΩ
		I _{OUT} = 4.6 A; T _J = 150 °C			20.7	
		I _{OUT} = 4.6 A; V _{CC} = 4 V; T _J = 25 °C			16	
		I _{OUT} = 0.46 A; V _{CC} = 2.7 V; V _{CC} decreasing			56.4	
R _{ON_Rev}	R _{DS(on)} in reverse battery condition	V _{CC} = -13 V, I _{OUT} = -4.6 A; T _J = 25 °C		9.4		mΩ
V _{clamp}	Clamp voltage	I _S = 20 mA; 25 °C < T _J < 150 °C	36	38	45	V
		I _S = 20 mA; T _J = -40 °C	36			V
I _{STBY}	Supply current in standby at V _{CC} = 13 V ⁽²⁾	V _{CC} = 13 V; V _{IN0,1} = V _{OUT0,1} = V _{FR} = V _{SEn} = 0 V; V _{SEL} = 0 V; T _J = 25 °C			0.5	μA
		V _{CC} = 13 V; V _{IN0,1} = V _{OUT0,1} = V _{FR} = V _{SEn} = 0 V; V _{SEL} = 0 V; T _J = 85 °C ⁽³⁾			0.5	
		V _{CC} = 13 V; V _{IN0,1} = V _{OUT0,1} = V _{FR} = V _{SEn} = 0 V; V _{SEL} = 0 V; T _J = 125 °C			3	

Symbol	Parameter	Test conditions	Min.	Typ.	Max.	Unit
t_{D_STBY}	Standby mode blanking time	$V_{CC} = 13\text{ V}$; $V_{IN0,1} = V_{OUT0,1} = V_{FR} = V_{SEL} = 0\text{ V}$; $V_{SEn} = 5\text{ V to } 0\text{ V}$	60	260	550	μs
$I_{S(ON)}$	Supply current	$V_{CC} = 13\text{ V}$; $V_{SEn} = V_{FR} = V_{SEL} = 0\text{ V}$; $V_{IN0} = 5\text{ V}$; $V_{IN1} = 5\text{ V}$; $I_{OUT0} = 0\text{ A}$; $I_{OUT1} = 0\text{ A}$		2.9	4	mA
$I_{GND(ON)}$	Control stage current consumption in ON state. All channels active.	$V_{CC} = 13\text{ V}$; $V_{SEn} = 5\text{ V}$; $V_{FR} = V_{SEL} = 0\text{ V}$; $V_{IN0} = 5\text{ V}$; $V_{IN1} = 5\text{ V}$; $I_{OUT0} = 4.6\text{ A}$; $I_{OUT1} = 4.6\text{ A}$			4.5	mA
$I_{L(off)}$	Off-state output current at $V_{CC} = 13\text{ V}$ ⁽²⁾	$V_{IN0,1} = V_{OUT0,1} = 0\text{ V}$; $V_{CC} = 13\text{ V}$; $T_J = 25^\circ\text{C}$	0	0.01	0.5	μA
		$V_{IN0,1} = V_{OUT0,1} = 0\text{ V}$; $V_{CC} = 13\text{ V}$; $T_J = 125^\circ\text{C}$	0		3	
V_F	Output - V_{CC} diode voltage at $T_J = 150^\circ\text{C}$	$I_{OUT} = -4.6\text{ A}$; $T_J = 150^\circ\text{C}$			0.7	V

1. For each channel.
2. Power MOSFET leakage included.
3. Parameter specified by design; not tested in production.

Table 6. Switching

$V_{CC} = 13\text{ V}$; $-40^\circ\text{C} < T_J < 150^\circ\text{C}$, unless otherwise specified						
Symbol	Parameter	Test conditions	Min.	Typ.	Max.	Unit
$t_{d(on)}^{(1)}$	Turn-on delay time at $T_J = 25^\circ\text{C}$	$R_L = 2.8\ \Omega$	10	46	120	μs
$t_{d(off)}^{(1)}$	Turn-off delay time at $T_J = 25^\circ\text{C}$		10	30	100	
$(dV_{OUT}/dt)_{on}^{(1)}$	Turn-on voltage slope at $T_J = 25^\circ\text{C}$	$R_L = 2.8\ \Omega$	0.2	0.4	0.7	V/ μs
$(dV_{OUT}/dt)_{off}^{(1)}$	Turn-off voltage slope at $T_J = 25^\circ\text{C}$		0.35	0.6	0.7	
$W_{ON}^{(2)}$	Switching energy losses at turn-on (t_{won})	$R_L = 2.8\ \Omega$		0.55	1.38	mJ
$W_{OFF}^{(2)}$	Switching energy losses at turn-off (t_{woff})	$R_L = 2.8\ \Omega$		0.37	0.74	mJ
$t_{SKEW}^{(1)}$	Differential Pulse skew ($t_{PHL} - t_{PLH}$)	$R_L = 2.8\ \Omega$	-75	-25	25	μs

1. See Figure 4. Switching time and pulse skew.
2. Parameter specified by design and evaluated by characterization; not tested in production.

Table 7. Logic inputs

7 V < V _{CC} < 28 V; -40 °C < T _J < 150 °C						
Symbol	Parameter	Test conditions	Min.	Typ.	Max.	Unit
INPUT _{0,1} characteristics						
V _{IL}	Input low level voltage				0.9	V
I _{IL}	Low level input current	V _{IN} = 0.9 V	1			μA
V _{IH}	Input high level voltage		2.1			V
I _{IH}	High level input current	V _{IN} = 2.1 V			10	μA
V _{I(hyst)}	Input hysteresis voltage		0.2			V
V _{ICL}	Input clamp voltage	I _{IN} = 1 mA	6		8.5	V
		I _{IN} = -1 mA		-0.7		
FaultRST characteristics						
V _{FRL}	Input low level voltage				0.9	V
I _{FRL}	Low level input current	V _{IN} = 0.9 V	1			μA
V _{FRH}	Input high level voltage		2.1			V
I _{FRH}	High level input current	V _{IN} = 2.1 V			10	μA
V _{FR(hyst)}	Input hysteresis voltage		0.2			V
V _{FRCL}	Input clamp voltage	I _{IN} = 1 mA	6		8.5	V
		I _{IN} = -1 mA		-0.7		
SEL characteristics (7 V < V _{CC} < 18 V)						
V _{SELL}	Input low level voltage				0.9	V
I _{SELL}	Low level input current	V _{IN} = 0.9 V	1			μA
V _{SELH}	Input high level voltage		2.1			V
I _{SELH}	High level input current	V _{IN} = 2.1 V			10	μA
V _{SEL(hyst)}	Input hysteresis voltage		0.2			V
V _{SELCL}	Input clamp voltage	I _{IN} = 1 mA	6		8.5	V
		I _{IN} = -1 mA		-0.7		
SEn characteristics (7 V < V _{CC} < 18 V)						
V _{SEnL}	Input low level voltage				0.9	V
I _{SEnL}	Low level input current	V _{IN} = 0.9 V	1			μA
V _{SEnH}	Input high level voltage		2.1			V
I _{SEnH}	High level input current	V _{IN} = 2.1 V			10	μA
V _{SEn(hyst)}	Input hysteresis voltage		0.2			V
V _{SEnCL}	Input clamp voltage	I _{IN} = 1 mA	9		12	V
		I _{IN} = -1 mA		-0.7		

Table 8. Protections

7 V < V _{CC} < 18 V; -40 °C < T _J < 150 °C						
Symbol	Parameter	Test conditions	Min.	Typ.	Max.	Unit
I _{LIMH} ⁽¹⁾	DC short circuit current	V _{CC} = 16 V, T _J = -40 °C	-15%	74.5	+15%	A
		V _{CC} = 16 V, T _J = 150 °C	-15%	56	+15%	A
I _{LIMH2} ⁽²⁾	DC short circuit current	V _{CC} = 19 V, T _J = -40 °C	-15%	56.5	+15%	A
		V _{CC} = 19 V, T _J = 150 °C	-15%	42.8	+15%	A
I _{LIMH} at 22 V	DC short circuit current	V _{CC} = 22 V, T _J = 25 °C		18.9		A
T _{TSD}	Shutdown temperature		150	175	210	°C
	Shutdown temperature (V _{CC} decreasing)	V _{CC} = 2.7 V	140			
T _{RS}	Thermal reset of fault diagnostic indication	V _{FR} = 0 V; V _{SEn} = 5 V	135			°C
T _{HYST}	Thermal hysteresis (T _{TSD} - T _{RS}) ⁽³⁾			7		°C
ΔT _{J_SD}	Dynamic temperature	V _{CC} = 16 V		80		K
		V _{CC} = 19 V		55		K
t _{LATCH_RST}	Fault reset time for output unlatch ⁽³⁾	V _{FR} = 5 V to 0 V; V _{SEn} = 5 V; • for example Ch ₀ : V _{IN0} = 5 V; V _{SEL} = 0 V	3	10	20	μs
t _{D_Restart}	Latch-OFF delay time before automatic restart			50	75	ms
V _{DEMAG}	Turn-off output voltage clamp	I _{OUT} = 1 A; L = 6 mH; T _J = -40 °C	V _{CC} - 36			V
		I _{OUT} = 1 A; L = 6 mH; T _J = 25 °C to 150 °C	V _{CC} - 36	V _{CC} - 38	V _{CC} - 45	V

1. I_{LIMH}, guaranteed between 7 V and 16 V, -40 °C < T_J < 150 °C.
2. I_{LIMH2}, guaranteed between 16 V and 19 V, -40 °C < T_J < 150 °C.
3. Parameter specified by design and evaluated by characterization; not tested in production.

Table 9. Current sense

7 V < V _{CC} < 18 V; -40°C < T _J < 150°C						
Symbol	Parameter	Test conditions	Min.	Typ.	Max.	Unit
V _{SENSE_CL}	CS clamp voltage	V _{SEn} = 0 V; I _{SENSE} = 1 mA	-9	-8	-7	V
		V _{SEn} = 0 V; I _{SENSE} = -1 mA		7		
Current sense characteristics						
K _{OL}	I _{OUT} /I _{SENSE}	I _{OUT} = 100 mA; V _{SENSE} = 0.5 V; V _{SEn} = 5 V	-35%	9150	+35%	
dK _{OL} /K _{OL} ⁽¹⁾⁽²⁾	Current sense ratio drift at calibration point	I _{OUT} = 100 mA; V _{SENSE} = 0.5 V; V _{SEn} = 5 V	-25		25	%
K ₀	I _{OUT} /I _{SENSE}	I _{OUT} = 0.5 A; V _{SENSE} = 0.5 V; V _{SEn} = 5 V	-25%	9150	+25%	
dK ₀ /K ₀ ⁽¹⁾⁽²⁾	Current sense ratio drift	I _{OUT} = 0.5 A; V _{SENSE} = 0.5 V; V _{SEn} = 5 V	-20		20	%
K ₁	I _{OUT} /I _{SENSE}	I _{OUT} = 0.9 A; V _{SENSE} = 3.5 V; V _{SEn} = 5 V	-15%	9150	+15%	
dK ₁ /K ₁ ⁽¹⁾⁽²⁾	Current sense ratio drift	I _{OUT} = 0.9 A; V _{SENSE} = 3.5 V; V _{SEn} = 5 V	-10		10	%
K ₂	I _{OUT} /I _{SENSE}	I _{OUT} = 4.6 A; V _{SENSE} = 3.5 V; V _{SEn} = 5 V	-7%	9150	+7%	

7 V < V _{CC} < 18 V; -40°C < T _J < 150°C						
Symbol	Parameter	Test conditions	Min.	Typ.	Max.	Unit
dK ₂ /K ₂ ⁽¹⁾⁽²⁾	Current sense ratio drift	I _{OUT} = 4.6 A; V _{SENSE} = 3.5 V; V _{SEn} = 5 V	-7		7	%
K ₃	I _{OUT} /I _{SENSE}	I _{OUT} = 13.8 A; V _{SENSE} = 3.5 V; V _{SEn} = 5 V	-5%	9150	+5%	
dK ₃ /K ₃ ⁽¹⁾⁽²⁾	Current sense ratio drift	I _{OUT} = 13.8 A; V _{SENSE} = 3.5 V; V _{SEn} = 5 V	-5		5	%
I _{SENSE0}	Current sense leakage current	CS disabled: V _{SEn} = 0 V	0		0.5	μA
		CS disabled: -1 V < V _{SENSE} < 5 V ⁽¹⁾	-1		1	
		CS enabled: V _{SEn} = 5 V; All channels ON; I _{OUTX} = 0 A; Ch _X diagnostic selected; • for example Ch ₀ : V _{IN0} = 5 V; V _{IN1} = 5 V; V _{SEL} = 0 V; I _{OUT0} = 0 A; I _{OUT1} = 4.6 A	0		3	
		CS enabled: V _{SEn} = 5 V; Ch _X OFF; Ch _X diagnostic selected; • for example Ch ₀ : V _{IN0} = 0 V; V _{IN1} = 5 V; V _{SEL} = 0 V; I _{OUT0} = 0 A; I _{OUT1} = 4.6 A	0		1	
V _{OUT_MSD} ⁽¹⁾	Output Voltage for CS shutdown	V _{SEn} = 5 V; R _{SENSE} = 2.7 kΩ; • for example Ch ₀ : V _{IN0} = 5 V; V _{SEL} = 0 V; I _{OUT0} = 4.6 A		5		V
V _{SENSE_SAT}	CS saturation voltage	V _{CC} = 7 V; R _{SENSE} = 10 kΩ; V _{SEn} = 5 V; V _{IN0} = 5 V; V _{SEL} = 0 V; I _{OUT0} = 13.8 A; T _J = -40 °C	4.8			V
I _{SENSE_SAT} ⁽¹⁾	CS saturation current	V _{CC} = 7 V; V _{SENSE} = 3.5 V; V _{IN0} = 5 V; V _{SEn} = 5 V; V _{SEL} = 0 V; T _J = 150 °C	2			mA
I _{OUT_SAT} ⁽¹⁾	Output saturation current	V _{CC} = 7 V; V _{SENSE} = 3.5 V; V _{IN0} = 5 V; V _{SEn} = 5 V; V _{SEL} = 0 V; T _J = 150 °C	30			A
OFF-state diagnostic						
V _{OL}	OFF-state open-load voltage detection threshold	V _{SEn} = 5 V; Ch _X OFF; Ch _X diagnostic selected • for example: Ch ₀ V _{IN0} = 0 V; V _{SEL} = 0 V;	2	3	4	V
I _{L(off2)}	OFF-state output sink current	V _{IN} = 0 V; V _{OUT} = V _{OL} ; T _J = -40°C to 125°C	-150	-40	-5	μA
t _{DSTKON}	OFF-state diagnostic delay time from falling edge of INPUT (see Figure 6. T _{DSTKON})	V _{SEn} = 5 V; Ch _X ON to OFF transition; Ch _X diagnostic selected • for example: Ch ₀ V _{IN0} = 5 V to 0 V; V _{SEL} = 0 V; I _{OUT0} = 0 A; V _{OUT0} = 4 V	150	260	400	μs
t _{D_OL_V}	Settling time for valid OFF-state open load diagnostic indication from rising edge of SE _n	V _{IN0} = 0 V; V _{IN1} = 0 V; V _{FR} = 0 V; V _{SEL} = 0 V; V _{OUT0} = 4 V; V _{SEn} = 0 V to 5 V			60	μs

7 V < V _{CC} < 18 V; -40°C < T _J < 150°C						
Symbol	Parameter	Test conditions	Min.	Typ.	Max.	Unit
t _{D_VOL}	OFF-state diagnostic delay time from rising edge of V _{OUT}	V _{SEn} = 5 V; Ch _X OFF; Ch _X diagnostic selected • for example: Ch ₀ V _{IN0} = 0 V; V _{SEL} = 0 V; V _{OUT} = 0 V to 4 V		5	30	μs
Fault diagnostic feedback (see Table 10. Truth table)						
V _{SENSEH}	Current sense output voltage in fault condition	13 V < V _{CC} < 18 V; Ch0 in open load; R _{SENSE} = 0.7 kΩ; V _{IN0} = 0 V; V _{SEn} = 5 V; I _{OUT0} = 0 A; V _{OUT0} = 4 V	5		7.5	V
		V _{CC} = 7 V; Ch0 in open load; R _{SENSE} = 0.7 kΩ; V _{IN0} = 0 V; V _{SEn} = 5 V; I _{OUT0} = 0 A; V _{OUT0} = 4 V	4.3			
I _{SENSEH}	Current sense output current in fault condition	13 V < V _{CC} < 18 V; V _{SENSE} = 5 V; Ch0 in open load; V _{IN0} = 0 V; V _{SEn} = 5 V; I _{OUT0} = 0 A; V _{OUT0} = 4 V	7	8.6	12	mA
		V _{CC} = 7 V; V _{SENSE} = 5 V; Ch0 in open load; V _{IN0} = 0 V; V _{SEn} = 5 V; I _{OUT0} = 0 A; V _{OUT0} = 4 V	4.4			
Current sense timings (current sense mode - see Figure 5. Current sense timings (current sense mode)) ⁽³⁾						
t _{DSENSE1H}	Current sense settling time from rising edge of SEn	V _{IN} = 5 V; V _{SEn} = 0 V to 5 V; R _{SENSE} = 1 kΩ; R _L = 2.8 Ω			60	μs
t _{DSENSE1L}	Current sense disable delay time from falling edge of SEn	V _{IN} = 5 V; V _{SEn} = 5 V to 0 V; R _{SENSE} = 1 kΩ; R _L = 2.8 Ω		5	20	μs
t _{DSENSE2H}	Current sense settling time from rising edge of INPUT	V _{IN} = 0 V to 5 V; V _{SEn} = 5 V; R _{SENSE} = 1 kΩ; R _L = 2.8 Ω		100	200	μs
Δt _{DSENSE2H}	Current sense settling time from rising edge of I _{OUT} (dynamic response to a step change of I _{OUT})	V _{IN} = 5 V; V _{SEn} = 5 V; R _{SENSE} = 1 kΩ; I _{SENSE} = 90% of I _{SENSEMAX} ; R _L = 2.8 Ω			100	μs
t _{DSENSE2L}	Current sense turn-off delay time from falling edge of INPUT	V _{IN} = 5 V to 0 V; V _{SEn} = 5 V; R _{SENSE} = 1 kΩ; R _L = 2.8 Ω		70	250	μs
t _{DSENSE3H}	Current sense latch-OFF filtering time		1.4	2.0	2.6	ms
Current sense timings (multiplexer transition times) ⁽³⁾						
t _{D_XtoY}	Current sense transition delay from Ch _X to Ch _Y	V _{IN0} = 5 V; V _{IN1} = 5 V; V _{SEn} = 5 V; V _{SEL} = 0 V to 5 V; I _{OUT0} = 0 A; I _{OUT1} = 4.6 A; R _{SENSE} = 1 kΩ			30	μs
t _{D_CStoVSENSEH}	Current sense transition delay from stable current sense on Ch _X to V _{SENSEH} on Ch _Y	V _{IN0} = 5 V; V _{IN1} = 0 V; V _{SEn} = 5 V; V _{SEL} = 0 V to 5 V; I _{OUT0} = 4.6 A; V _{OUT1} = 4 V; R _{SENSE} = 1 kΩ			20	μs

1. Parameter specified by design and evaluated by characterization; not tested in production.

2. All values refer to $V_{CC} = 13\text{ V}$; $T_J = 25^\circ\text{C}$, unless otherwise specified.
3. Transition delay is measured up to $\pm 10\%$ of final conditions.

Figure 4. Switching time and pulse skew

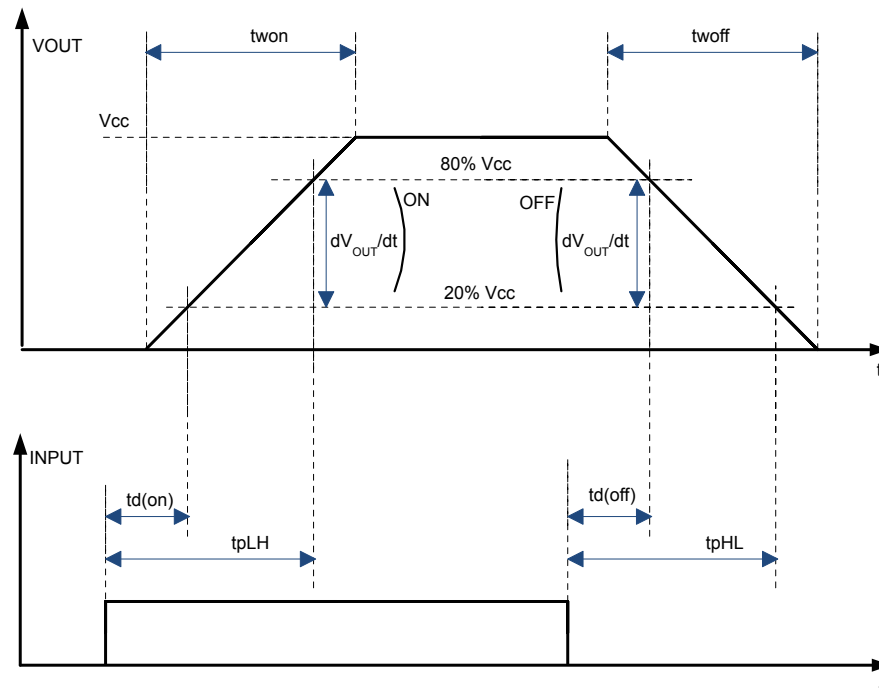
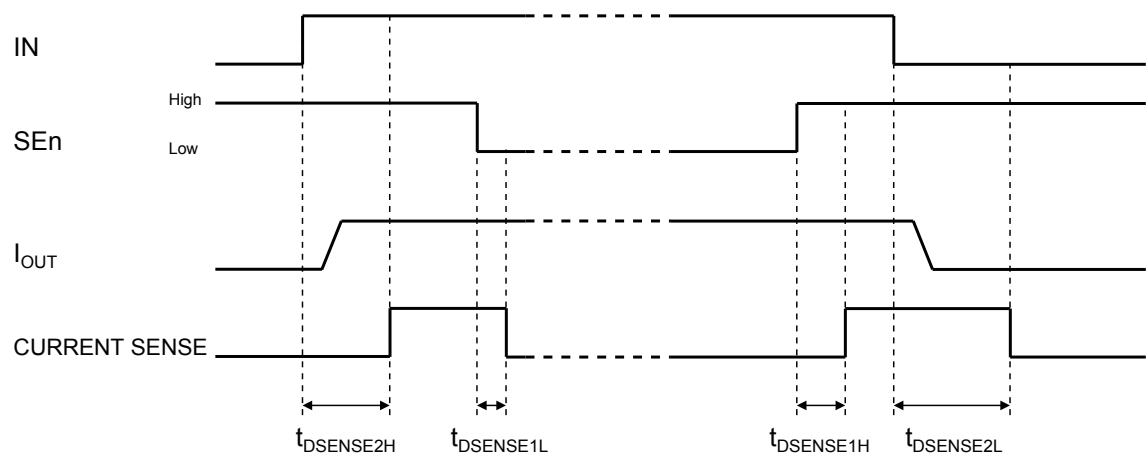
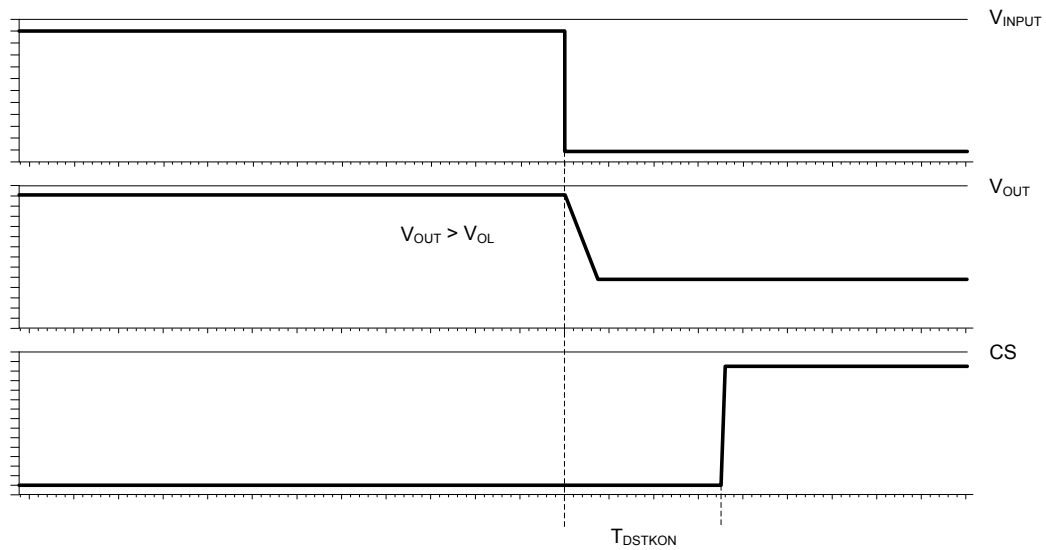


Figure 5. Current sense timings (current sense mode)



GADG1007181523PS

Figure 6. T_{DSTKON}


GADG0112170745PS

Table 10. Truth table

Mode	Conditions	IN _x	FR	SEn	SEL	OUT _x	CS	Comments
Standby	All logic inputs low	L	L	L	L	L	Hi-Z	Low quiescent current consumption
Normal	Nominal load connected; $T_J < 150\text{ }^{\circ}\text{C}$	L	X	See ⁽¹⁾		L	See ⁽¹⁾	
		H	L			H	See ⁽¹⁾	Outputs configured for auto-restart
		H	H			H	See ⁽¹⁾	Outputs configured for Latch-off
Overload	Overload or short to GND causing: $T_J > T_{TSD}$ or $\Delta T_J > \Delta T_{J_SD}$	L	X	See ⁽¹⁾		L	See ⁽¹⁾	
		H	L			H	See ⁽¹⁾	Output cycles with temperature hysteresis
		H	H			L	See ⁽¹⁾	Output latches-off
Undervoltage	$V_{CC} < V_{USD}$ (falling)	X	X	X	X	L L	Hi-Z Hi-Z	Re-start when $V_{CC} > V_{USD} + V_{USDhyst}$ (rising)
OFF-state diagnostics	Short to V_{CC}	L	X	See ⁽¹⁾		H	See ⁽¹⁾	
	Open-load	L	X			H	See ⁽¹⁾	External pull-up
Negative output voltage	Inductive loads turn-off	L	X	See ⁽¹⁾		< 0 V	See ⁽¹⁾	

1. Refer to Table 11. Current sense multiplexer addressing

Table 11. Current sense multiplexer addressing

SEn	SEL	MUX channel	CS output			
			Normal mode	Overload	OFF-state diag.	Negative output
L	X		Hi-Z			
H	L	Channel 0 diagnostic	$I_{\text{SENSE}} = 1/K * I_{\text{OUT0}}$	$V_{\text{SENSE}} = V_{\text{SENSEH}}$	$V_{\text{SENSE}} = V_{\text{SENSEH}}$	Hi-Z
H	H	Channel 1 diagnostic	$I_{\text{SENSE}} = 1/K * I_{\text{OUT1}}$	$V_{\text{SENSE}} = V_{\text{SENSEH}}$	$V_{\text{SENSE}} = V_{\text{SENSEH}}$	Hi-Z

2.4 Electrical characteristics (curves)

Figure 7. OFF-state output current

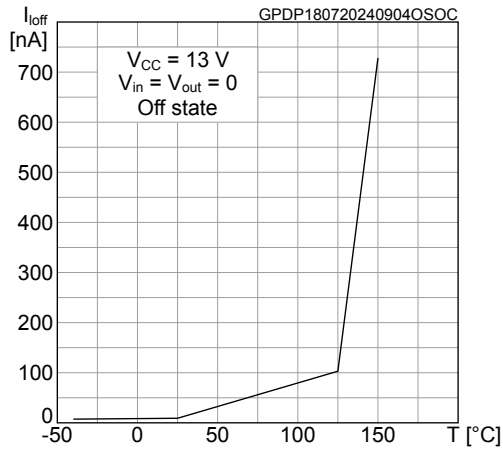


Figure 8. Standby current

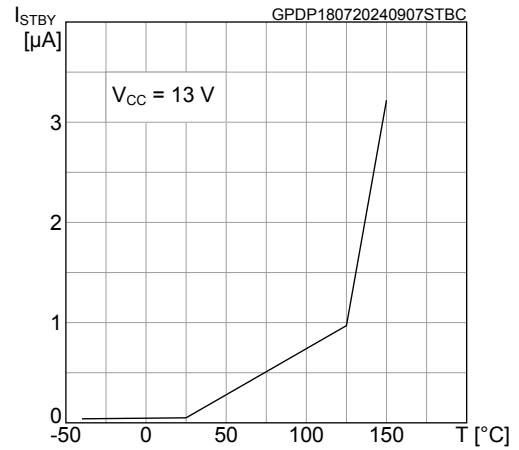


Figure 9. $I_{GND(ON)}$ vs. I_{out}

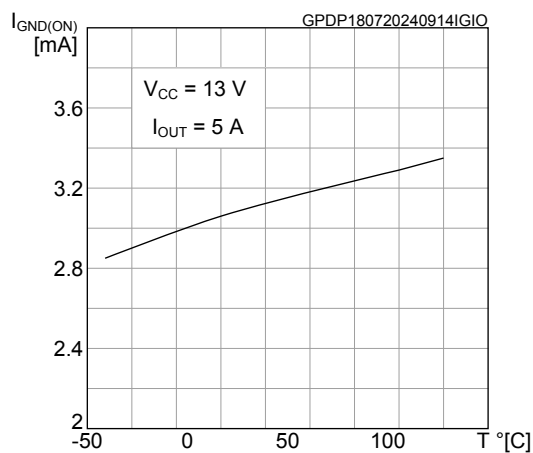


Figure 10. Logic input high level voltage

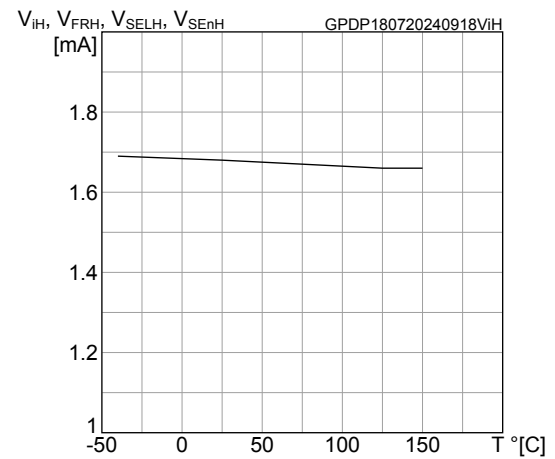


Figure 11. Logic input low level voltage

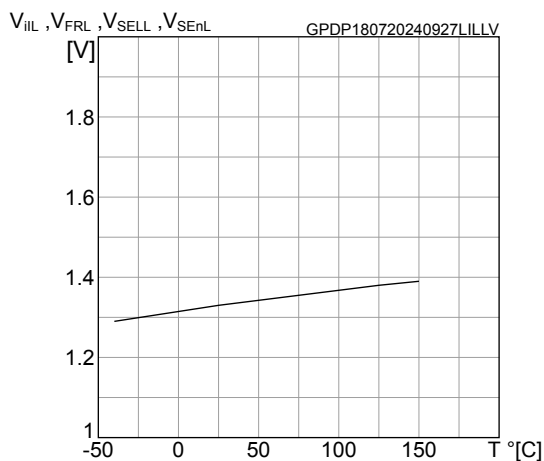


Figure 12. High level logic input current

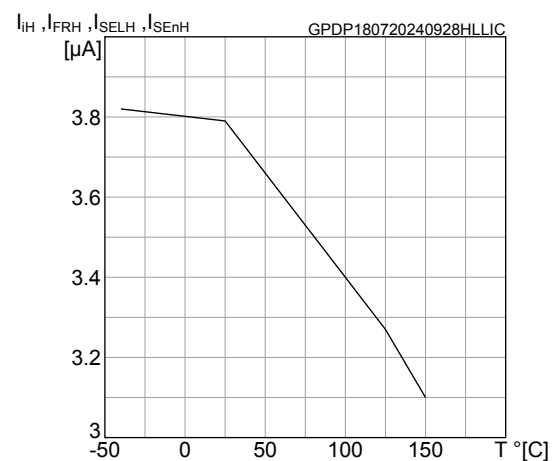


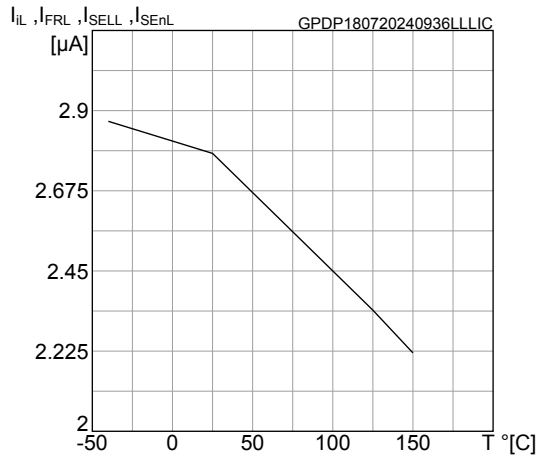
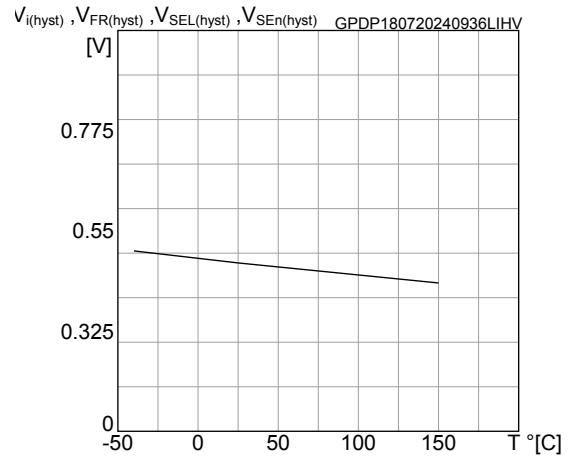
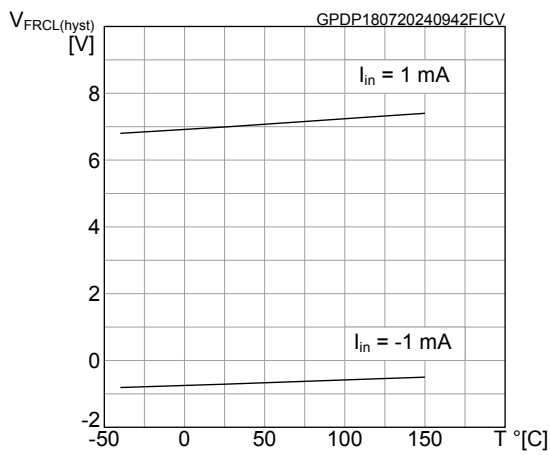
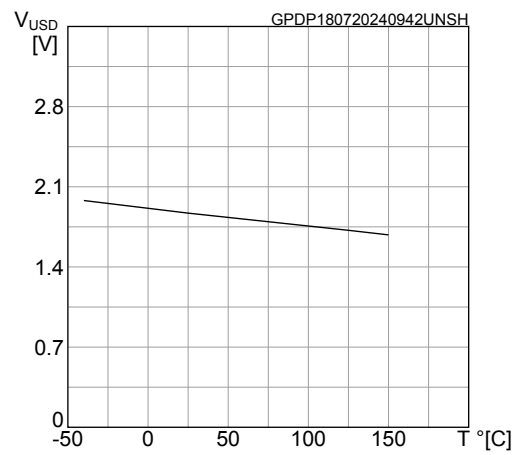
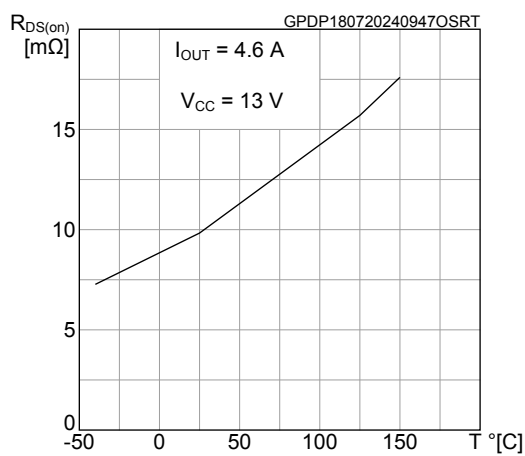
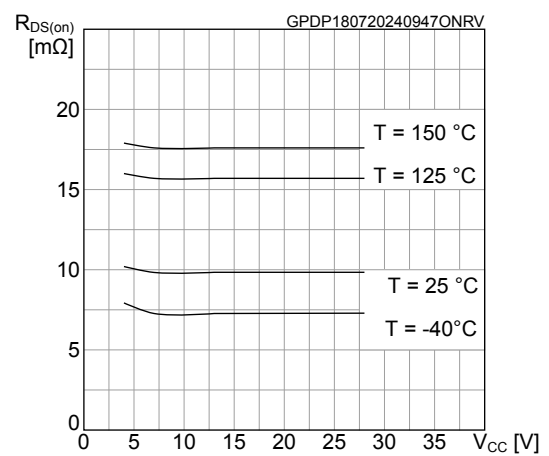
Figure 13. Low level logic input current

Figure 14. Logic input hysteresis voltage

Figure 15. FaultRST input clamp voltage

Figure 16. Undervoltage shutdown

Figure 17. On-state resistance vs. T_C

Figure 18. On-state resistance vs. V_{CC}


Figure 19. Turn-on voltage slope

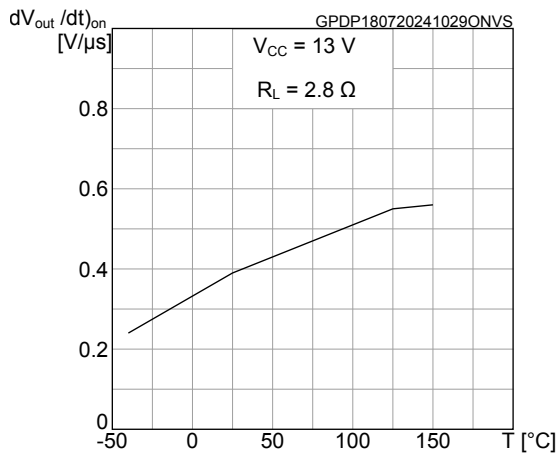


Figure 20. Turn-off voltage slope

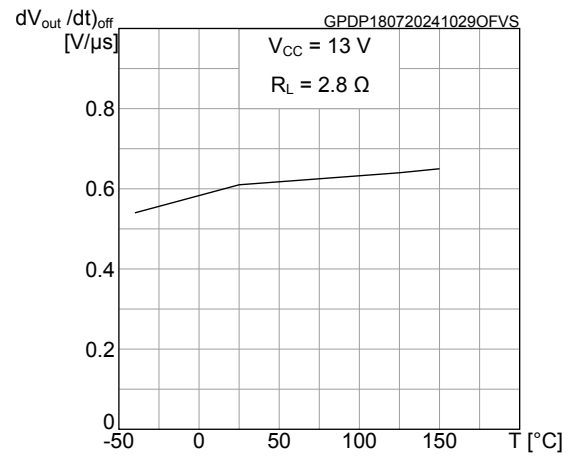


Figure 21. W_{on} vs T_C

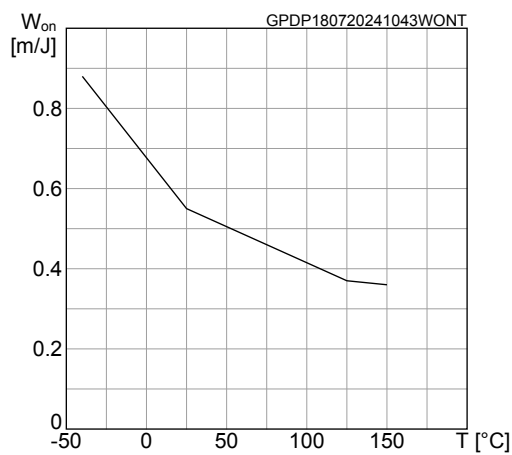


Figure 22. W_{off} vs T_C

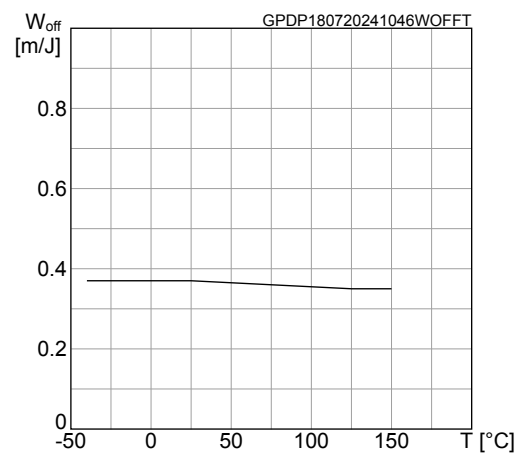


Figure 23. I_{limH} vs T_C

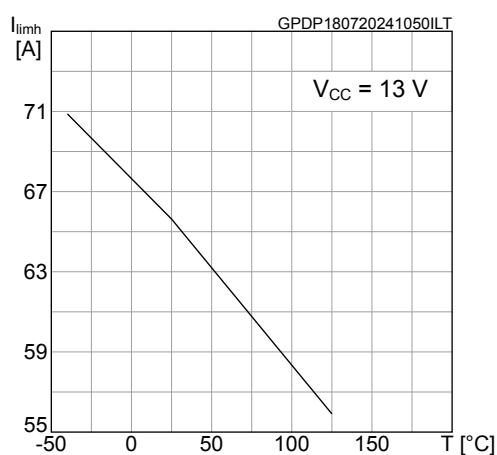


Figure 24. OFF-state open-load voltage detection threshold

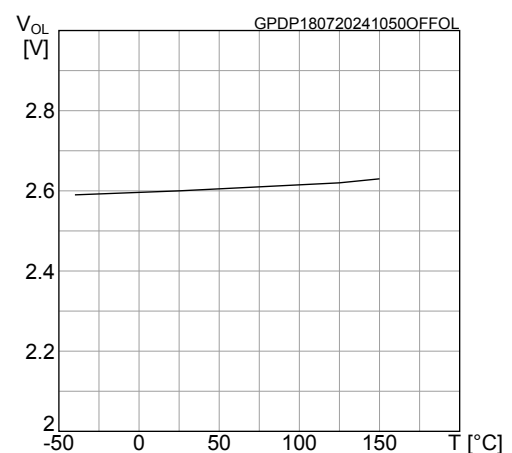


Figure 25. $V_{\text{Sense clamp}}$ vs T_C

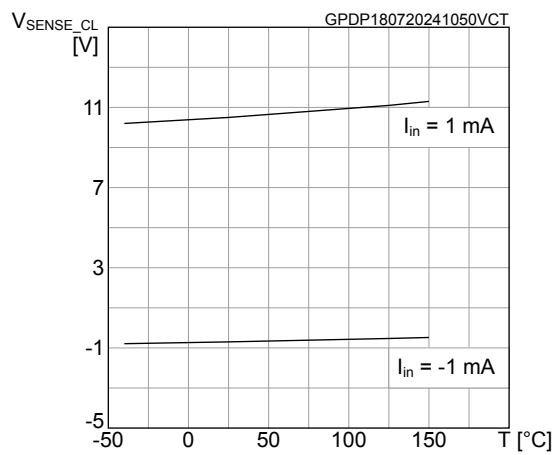
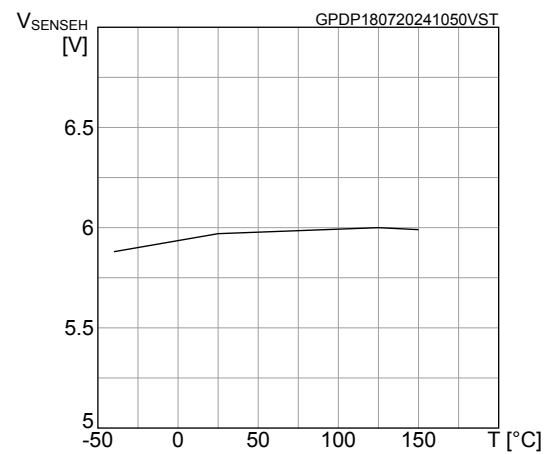


Figure 26. V_{senseh} vs T_C



2.5 Waveforms

Figure 27. Latch-off mode - Intermittent short circuit

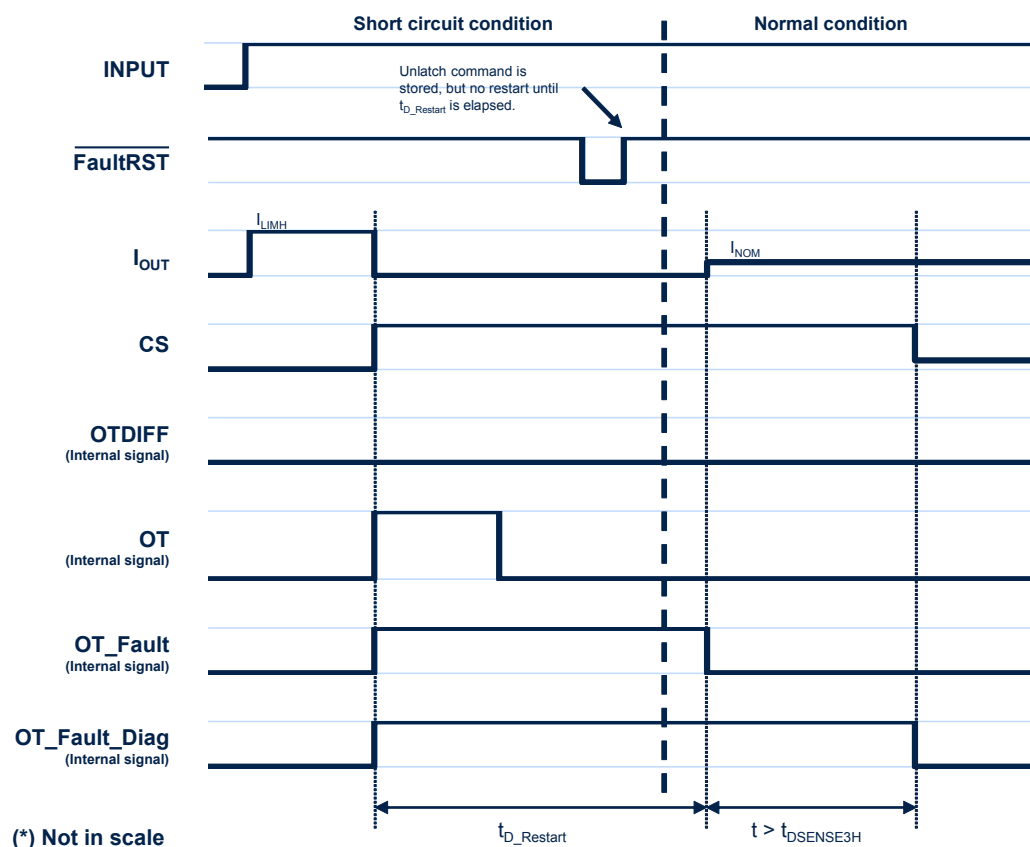


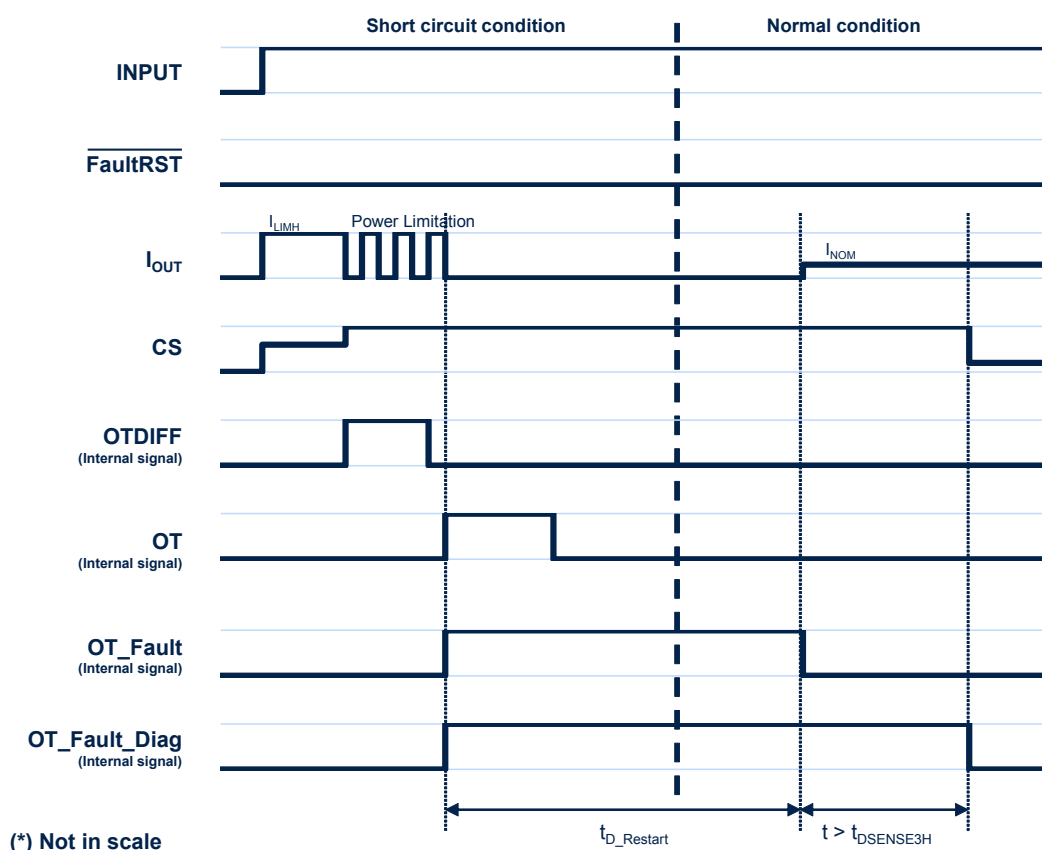
Figure 28. Auto-restart mode - Intermittent short circuit


Figure 29. Auto-restart mode - Permanent short circuit

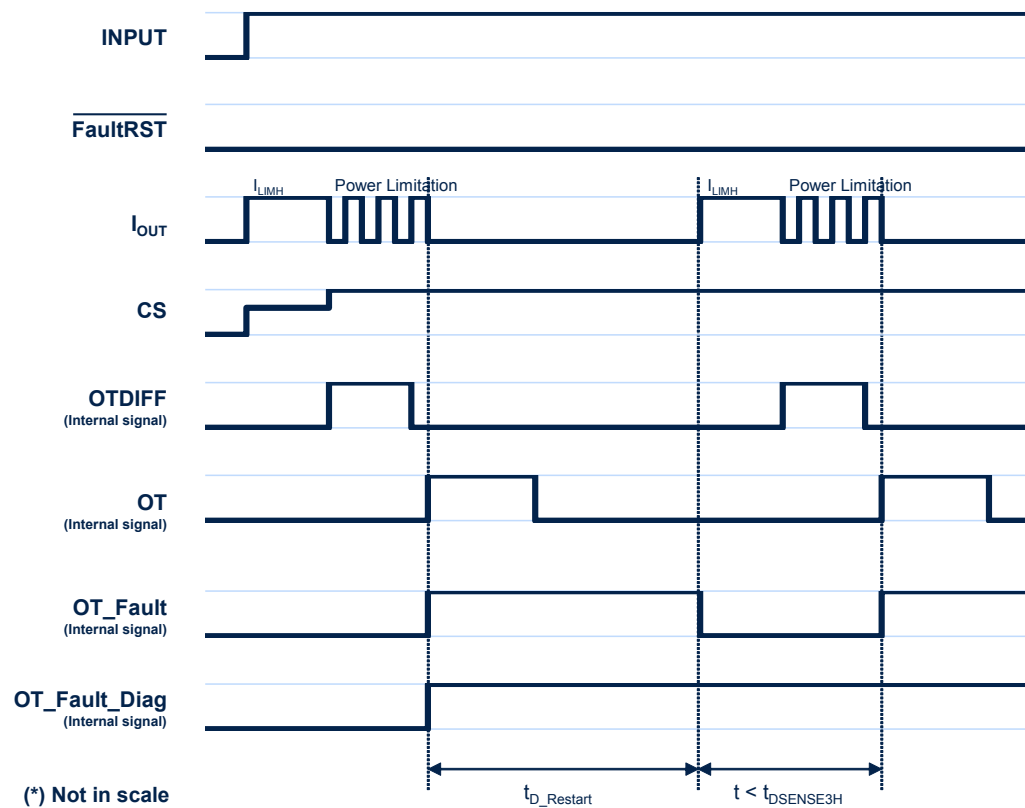


Figure 30. Standby mode activation

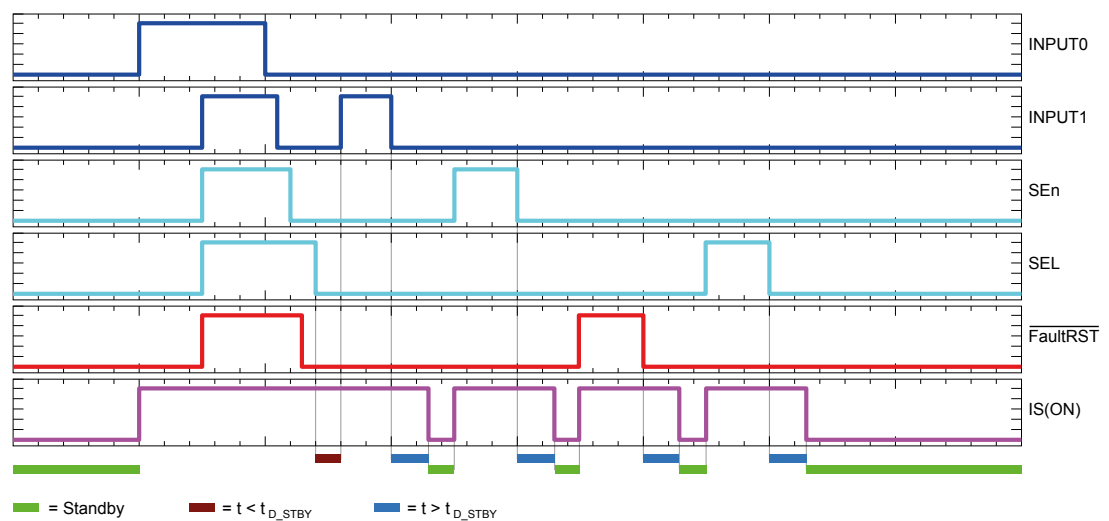
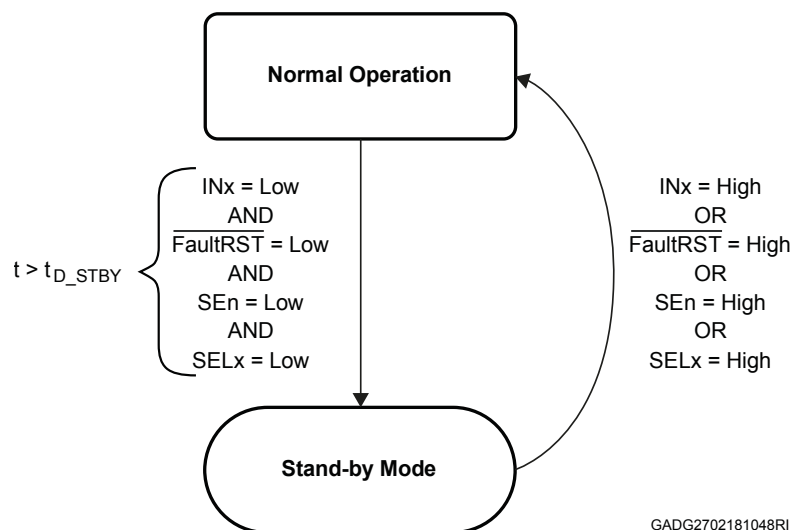


Figure 31. Standby state diagram



3 Protections

3.1 Power limitation

The basic working principle of this protection consists of an indirect measurement of the junction temperature swing ΔT_J through the direct measurement of the spatial temperature gradient on the device surface in order to automatically shut off the output MOSFET as soon as ΔT_J exceeds the safety level of ΔT_{J_SD} . According to the voltage level on the $\overline{\text{FaultRST}}$ pin, the output MOSFET switches on and cycles with a thermal hysteresis according to the maximum instantaneous power which can be handled ($\overline{\text{FaultRST}} = \text{Low}$) or remain off ($\overline{\text{FaultRST}} = \text{High}$). The protection prevents fast thermal transient effects and, consequently, reduces thermo-mechanical fatigue.

3.2 Thermal shutdown

In case the junction temperature of the device exceeds the maximum allowed threshold (typically 175 °C), it automatically switches off and the diagnostic indication is triggered. According to the voltage level on the $\overline{\text{FaultRST}}$ pin, the device switches on again as soon as its junction temperature drops to T_R ($\overline{\text{FaultRST}} = \text{Low}$) or remains off ($\overline{\text{FaultRST}} = \text{High}$).

3.3 Current limitation

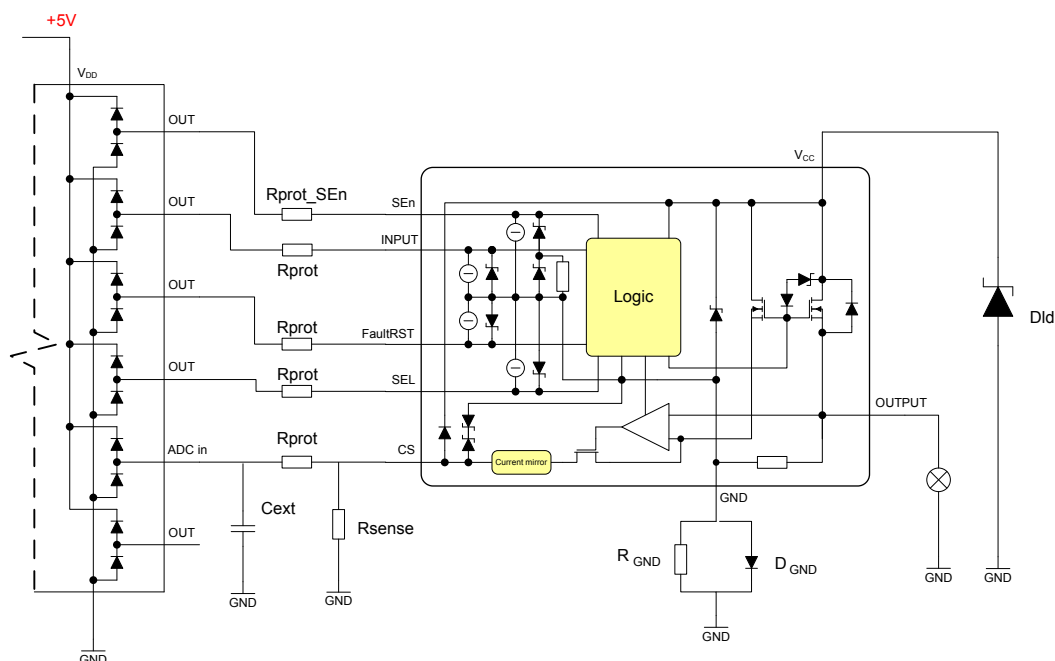
The device is equipped with an output current limiter in order to protect the silicon as well as the other components of the system (for example bonding wires, wiring harness, connectors, loads, etc.) from excessive current flow. Consequently, in case of short-circuit, overload or during load power-up, the output current is clamped to a safety level, I_{LIMH} , by operating the output power MOSFET in the active region.

3.4 Negative voltage clamp

In case the device drives inductive load, the output voltage reaches a negative value during turn off. A negative voltage clamp structure limits the maximum negative voltage to a certain value, V_{DEMG} , allowing the inductor energy to be dissipated without damaging the device.

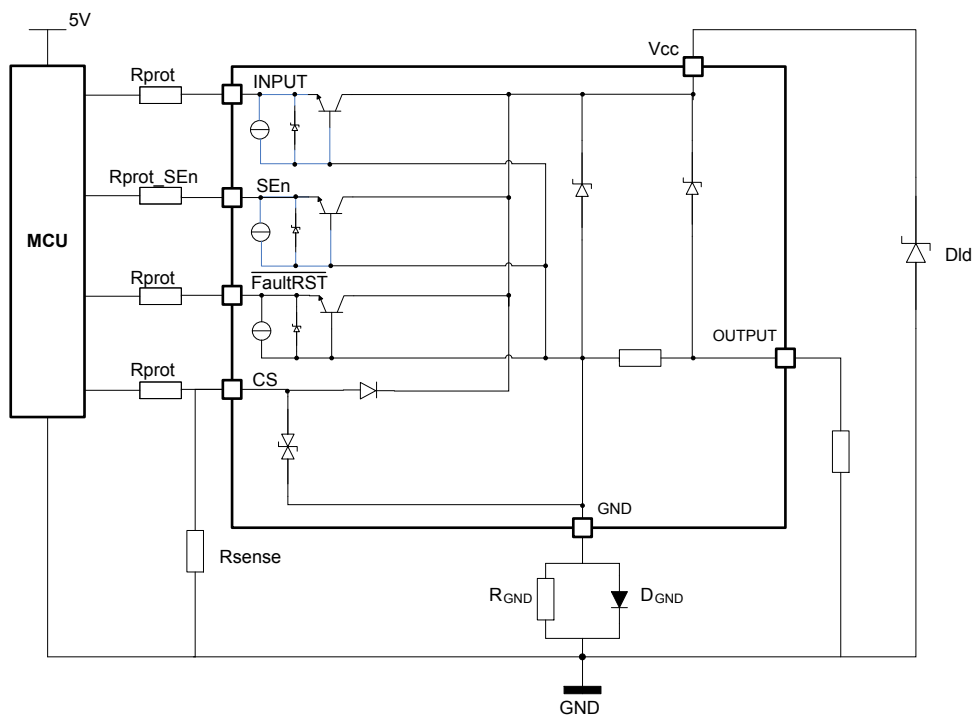
4 Application information

Figure 32. Application diagram



4.1 GND protection network against reverse battery

Figure 33. Simplified internal structure



4.1.1 Diode (DGND) in the ground line

A resistor (typ. $R_{GND} = 4.7 \text{ k}\Omega$) should be inserted in parallel to D_{GND} if the device drives an inductive load.

This small signal diode can be safely shared amongst several different HSDs. Also in this case, the presence of the ground network produces a shift ($\approx 600 \text{ mV}$) in the input threshold and in the status output values if the microprocessor ground is not common to the device ground. This shift does not vary if more than one HSD shares the same diode/resistor network.

4.2 Immunity against transient electrical disturbances

The immunity of the device against transient electrical emissions, conducted along the supply lines and injected into the V_{CC} pin, is tested in accordance with ISO7637-2:2011 (E) and ISO 16750-2:2010.

The related function performance status classification is shown in [Table 12. ISO 7637-2 - electrical transient conduction along supply line](#).

Test pulses are applied directly to DUT (device under test) both in ON and OFF state and in accordance to ISO 7637-2:2011(E), chapter 4.

The DUT is intended as the current device only, with external components as shown in [Figure 34. M0-9 application schematic](#).

Status II is defined in ISO 7637-1 Function Performance Status Classification (FPSC) as follows: "The function does not perform as designed during the test but returns automatically to normal operation after the test".

Table 12. ISO 7637-2 - electrical transient conduction along supply line

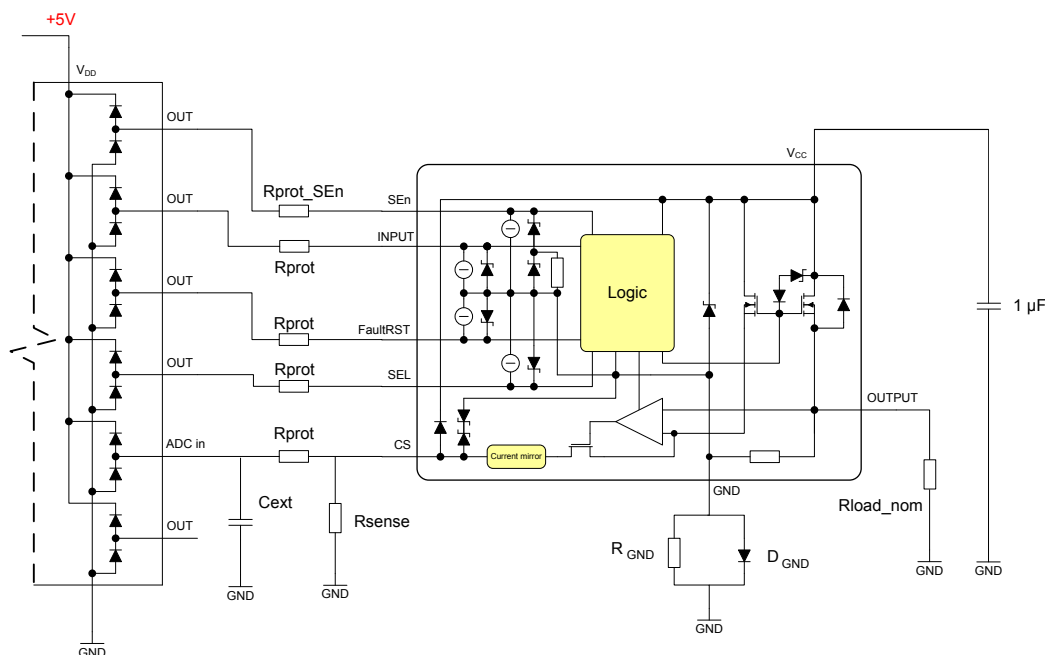
Test Pulse 2011(E)	Test pulse severity level with Status II functional performance status		Minimum number of pulses or test time	Burst cycle / pulse repetition time		Pulse duration and pulse generator internal impedance
	Level	$U_S^{(1)}$		min	max	
1	III	-112 V	500 pulses	0.5 s		2 ms, 10 Ω
2a	III	+55 V	500 pulses	0.2 s	5 s	50 μs , 2 Ω
3a	IV	-220 V	1h	90 ms	100 ms	0.1 μs , 50 Ω
3b	IV	+150 V	1h	90 ms	100 ms	0.1 μs , 50 Ω
4 ⁽²⁾	IV	-7 V	1 pulse			100 ms, 0.01 Ω
Load dump according to ISO 16750-2:2010						
Test B ⁽³⁾		35 V	5 pulses	1 min		400 ms, 2 Ω

1. U_S is the peak amplitude as defined for each test pulse in ISO 7637-2:2011(E), chapter 5.6.

2. Test pulse from ISO 7637-2:2004(E).

3. With 35 V external suppressor referred to ground ($-40^\circ\text{C} < T_j < 150^\circ\text{C}$).

Figure 34. M0-9 application schematic



4.3 MCU I/Os protection

If a ground protection network is used and negative transients are present on the V_{CC} line, the control pins will be pulled negative. ST suggests to insert a resistor (R_{prot}) in line both to prevent the microcontroller I/O pins from latching-up and to protect the HSD inputs.

The value of these resistors is a compromise between the leakage current of microcontroller and the current required by the HSD I/Os (Input levels compatibility) with the latch-up limit of microcontroller I/Os.

Equation

$$V_{CCpeak}/I_{latchup} \leq R_{prot} \leq (V_{OH\mu C} - V_{IH} - V_{GND}) / I_{IHmax}$$

Calculation example:

For $V_{CCpeak} = -150\text{ V}$; $I_{latchup} \geq 20\text{ mA}$; $V_{OHuic} \geq 4.5\text{ V}$

$$7.5 \text{ k}\Omega \leq R_{\text{prot}} \leq 140 \text{ k}\Omega.$$

Recommended values: $R_{\text{prot}} = 15 \text{ k}\Omega$

A different value of the resistor has to be used for SEn pin, Rprot_SEn, as reported in [Figure 32. Application diagram](#).

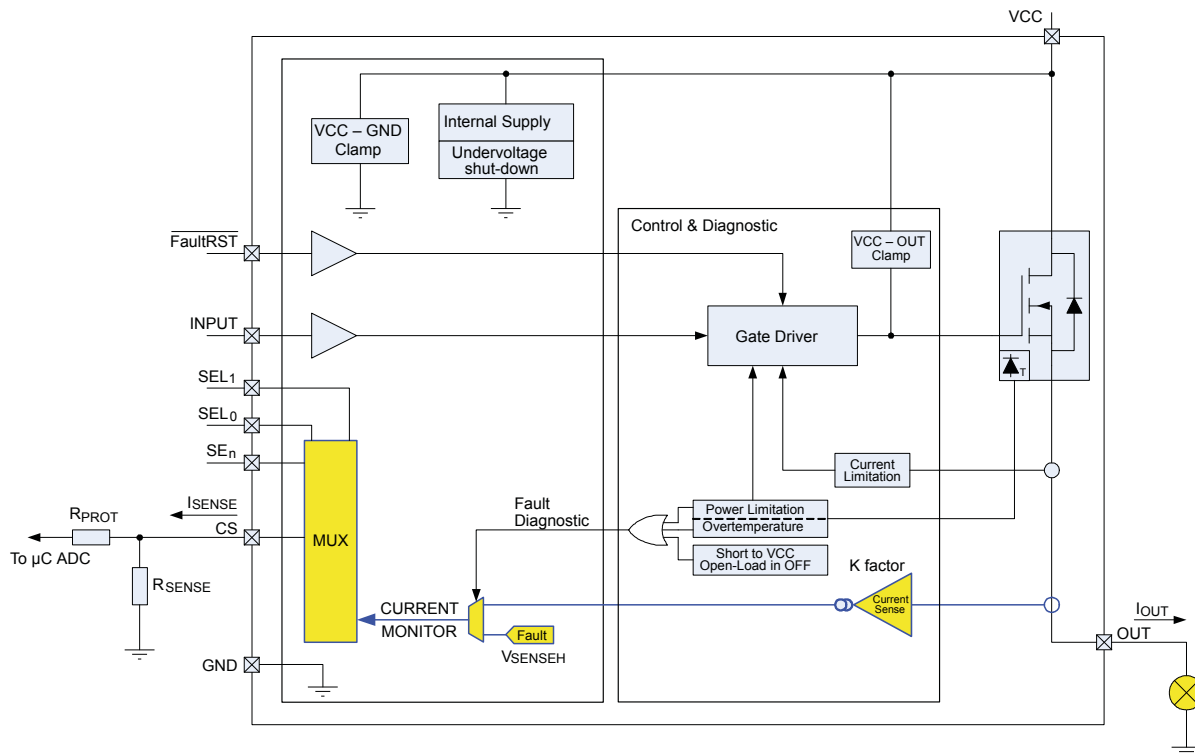
4.4 CS - analog current sense

Diagnostic information on device and load status are provided by an analog output pin (CS) delivering the following signals:

- Current monitor: current mirror of channel output current

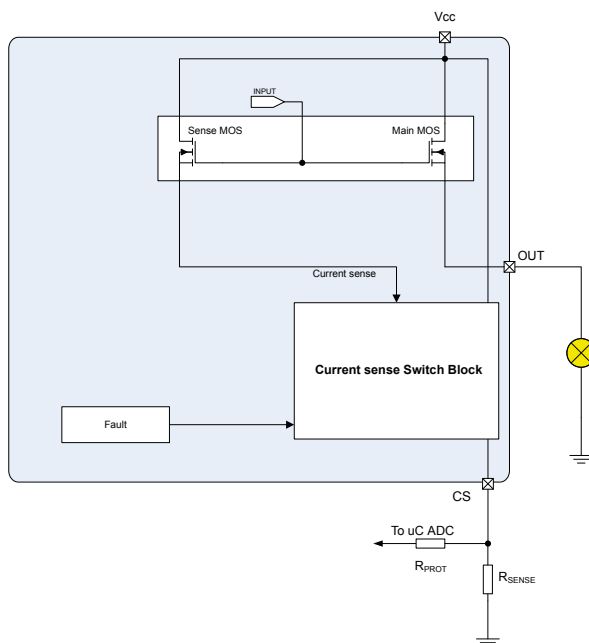
Those signals are routed through an analog multiplexer which is configured and controlled by means of SELx and SEn pins, according to the address map in [Table 7. Logic inputs.](#)

Figure 35. Current Sense and diagnostic – block diagram



4.4.1 Principle of current sense signal generation

Figure 36. CurrentSense block diagram



GAPG2307131200CFT

Current sense

The output is capable of providing:

- Current mirror proportional to the load current in normal operation, delivering current proportional to the load according to known ratio named K

- Diagnostics flag in fault conditions delivering fixed voltage V_{SENSEH}

The current delivered by the current sense circuit, I_{SENSE} , can be easily converted into a voltage V_{SENSE} by using an external sense resistor, R_{SENSE} , allowing continuous load monitoring and abnormal condition detection.

Normal operation (channel ON, no fault, SEn active)

While device is operating in normal conditions (no fault intervention), V_{SENSE} calculation can be done using simple equations

Current provided by MultiSense output: $I_{SENSE} = I_{OUT}/K$

Voltage on R_{SENSE} : $V_{\text{SENSE}} = R_{\text{SENSE}} \cdot I_{\text{SENSE}} = R_{\text{SENSE}} \cdot I_{\text{OUT}}/K$

Where:

- V_{SENSE} is voltage measurable on R_{SENSE} resistor
- I_{SENSE} is current provided from CurrentSense pin in current output mode
- I_{OUT} is current flowing through output
- K factor represents the ratio between PowerMOS cells and SenseMOS cells; its spread includes geometric factor spread, current sense amplifier offset and process parameters spread of overall circuitry specifying the ratio between I_{OUT} and I_{SENSE} .

Failure flag indication

In case of power limitation/overtemperature, the fault is indicated by the CurrentSense pin which is switched to a “current limited” voltage source, V_{SENSEH} .

In any case, the current sourced by the CurrentSense in this condition is limited to I_{SENSEH} .

The typical behavior in case of overload or hard short circuit is shown in Waveforms.

Figure 37. Analog HSD – open-load detection in off-state

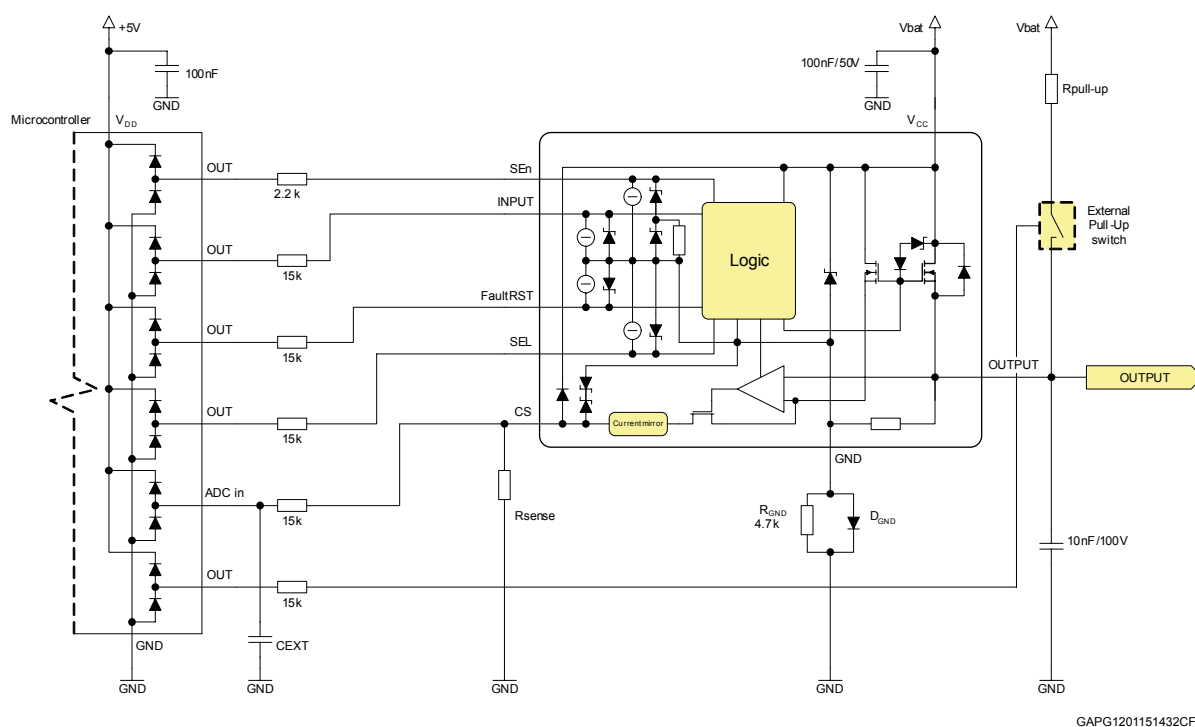
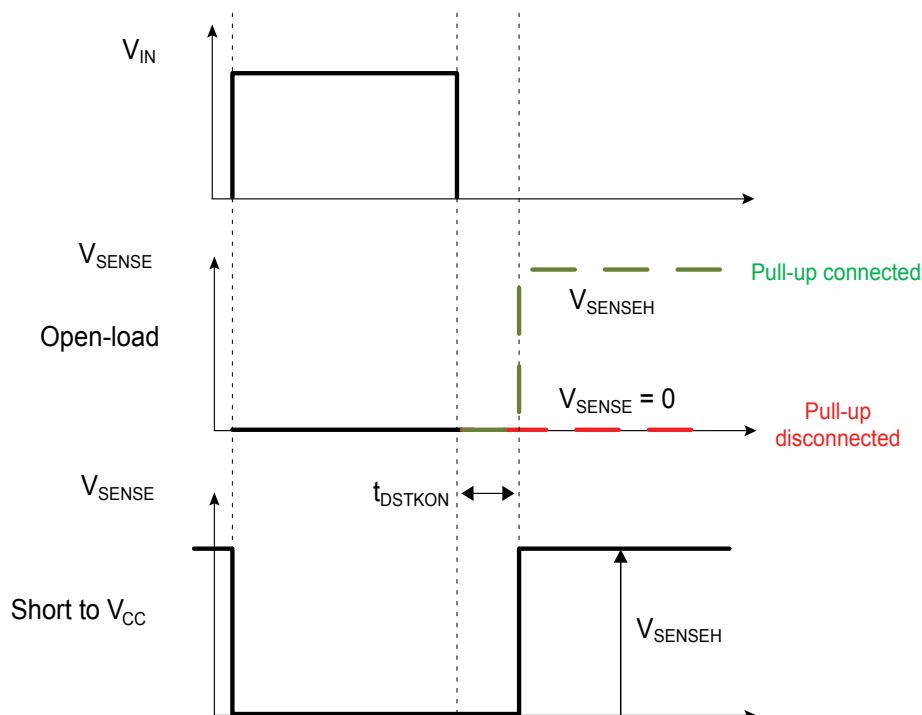


Figure 38. Open-load / short to V_{CC} condition


GADG2702181218R1

Table 13. CS pin levels in off-state

Condition	Output	CS	SEn
Open-load	$V_{OUT} > V_{OL}$	Hi-Z	L
		V_{SENSEH}	H
	$V_{OUT} < V_{OL}$	Hi-Z	L
		0	H
Short to V_{CC}	$V_{OUT} > V_{OL}$	Hi-Z	L
		V_{SENSEH}	H
Nominal	$V_{OUT} < V_{OL}$	Hi-Z	L
		0	H

4.4.2 Short to V_{CC} and OFF-state open-load detection

Short to V_{CC}

A short circuit between V_{CC} and output is indicated by the relevant current sense pin set to V_{SENSEH} during the device off-state. Small or no current is delivered by the current sense during the on-state depending on the nature of the short-circuit.

OFF-state open-load with external circuitry

Detection of an open-load in off mode requires an external pull-up resistor R_{PU} connecting the output to a positive supply voltage V_{PU} .

It is preferable that V_{PU} is switched off during the module standby mode in order to avoid the overall standby current consumption to increase in normal conditions, that is when load is connected.

R_{PU} must be selected in order to ensure V_{OUT} > V_{OLmax} in accordance with the following equation:

$$R_{PU} < \frac{V_{PU} - 4}{I_{L(off2)min @ 4V}}$$

5 Maximum demagnetization energy ($V_{CC} = 16\text{ V}$)

Figure 39. Maximum turn off current versus inductance

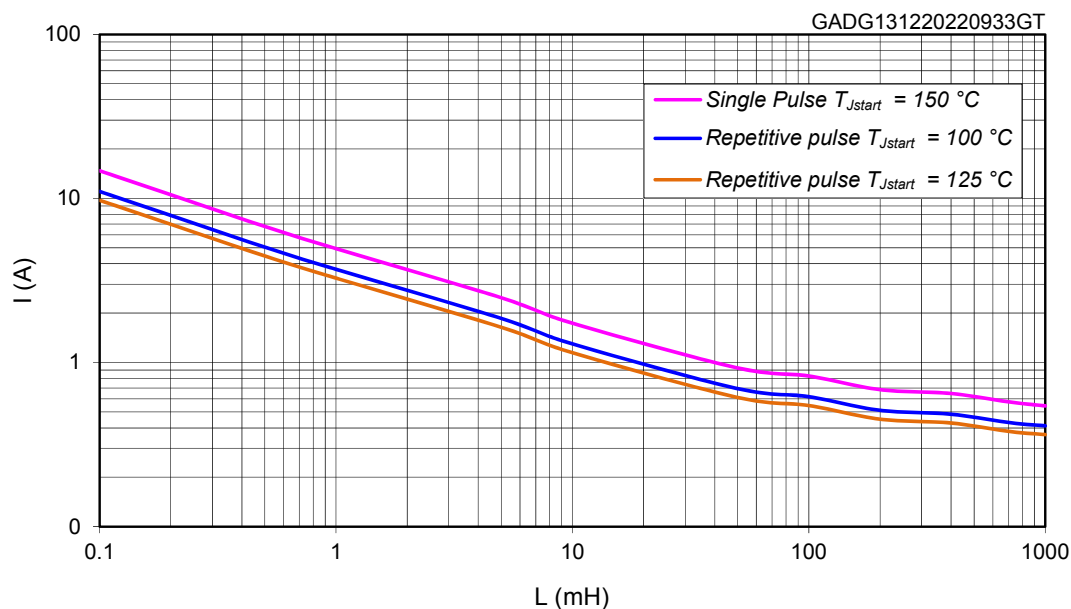
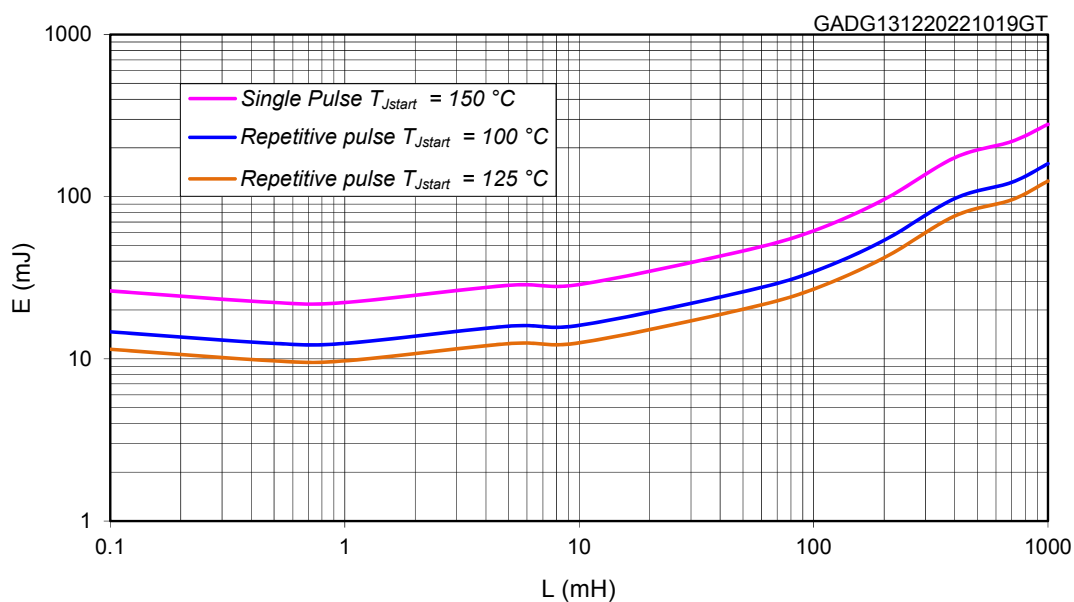


Figure 40. Maximum turn off energy versus inductance



6 Package and PCB thermal data

6.1 PowerSSO-16 thermal data

Figure 41. PowerSSO-16 on two-layer PCB (2s0p to JEDEC JESD 51-5)

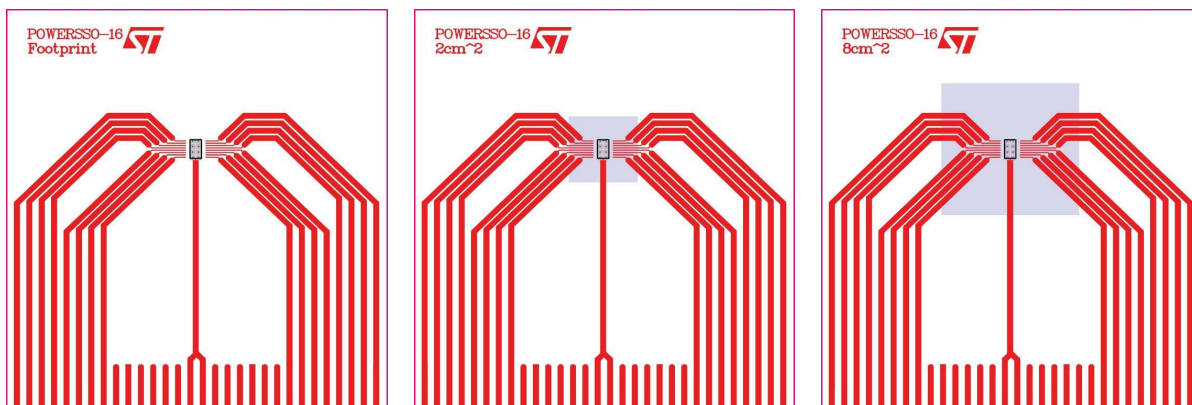


Figure 42. PowerSSO-16 on four-layer PCB (2s2p to JEDEC JESD 51-7)

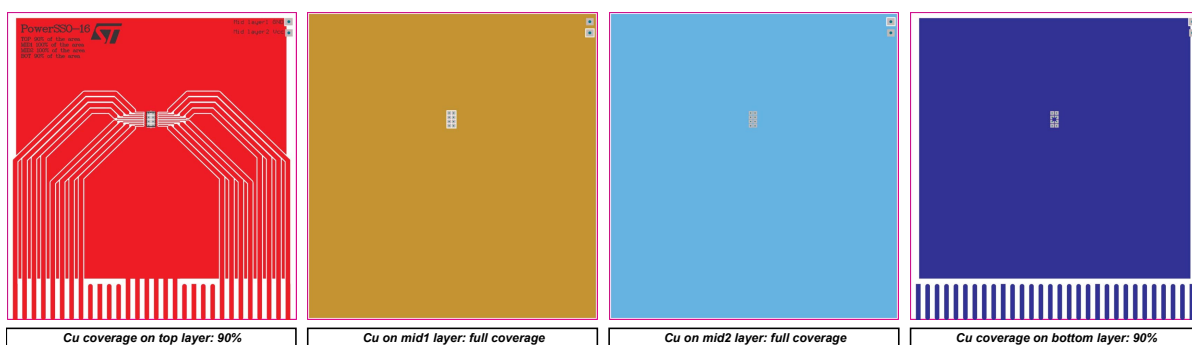


Table 14. PCB properties

Dimension	Value
Board finish thickness	1.6 mm $\pm 10\%$
Board dimension	77 mm x 86 mm
Board Material	FR4
Copper thickness (top and bottom layers)	0.070 mm
Copper thickness (inner layers)	0.035 mm
Thermal vias separation	1.2 mm
Thermal via diameter	0.3 mm ± 0.08 mm
Copper thickness on vias	0.025 mm
Footprint dimension (top layer)	2.2 mm x 3.9 mm
Heatsink copper area dimension (bottom layer)	Footprint, 2 cm ² or 8 cm ²

Figure 43. R_{thJA} vs PCB copper area in open box free air condition (one channel on)

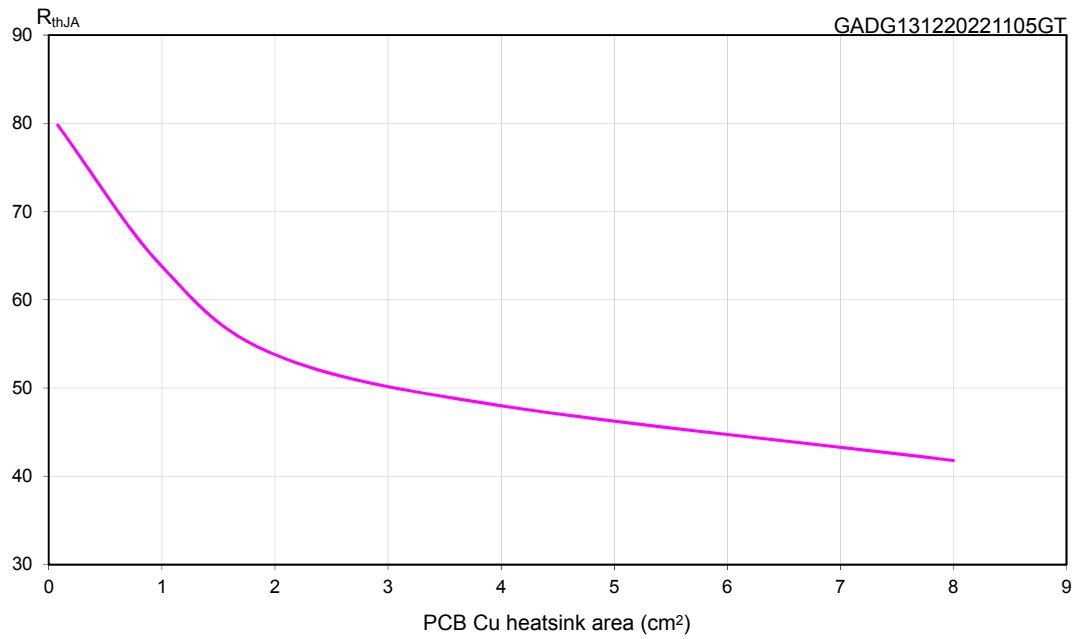
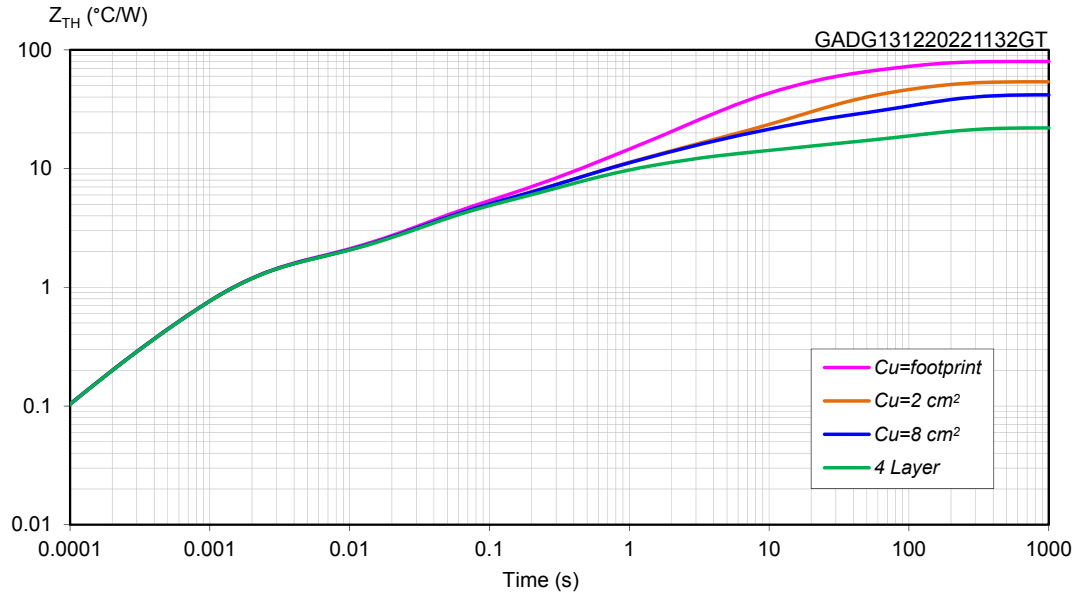


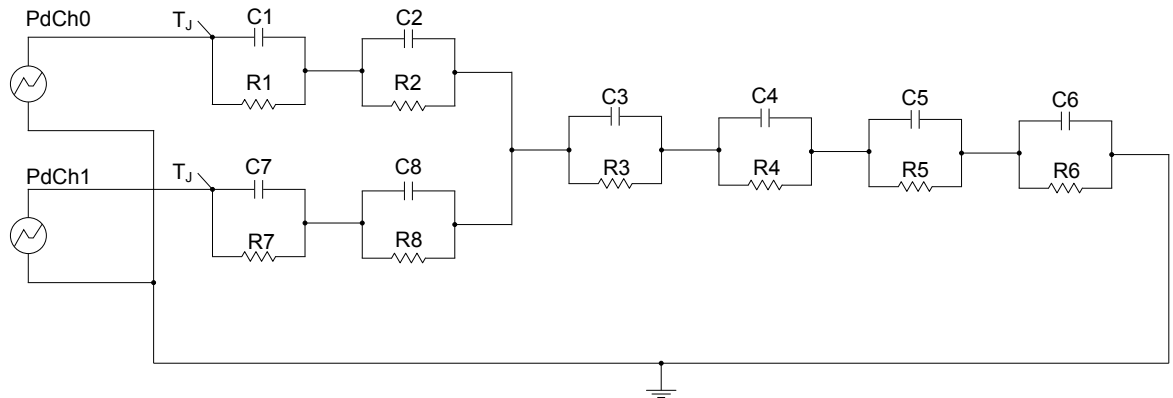
Figure 44. PowerSSO-16 thermal impedance junction ambient single pulse (one channel on)



Equation: pulse calculation formula

$$Z_{th\delta} = R_{th} \cdot \delta + Z_{thp} (1 - \delta)$$

where $\delta = t_p/T$

Figure 45. Thermal fitting model of a double-channel HSD in PowerSSO-16


GPDG280620241410FSR

Note: The fitting model is a simplified thermal tool and is valid for transient evolutions where the embedded protections (power limitation or thermal cycling during thermal shutdown) are not triggered.

Table 15. Thermal parameters

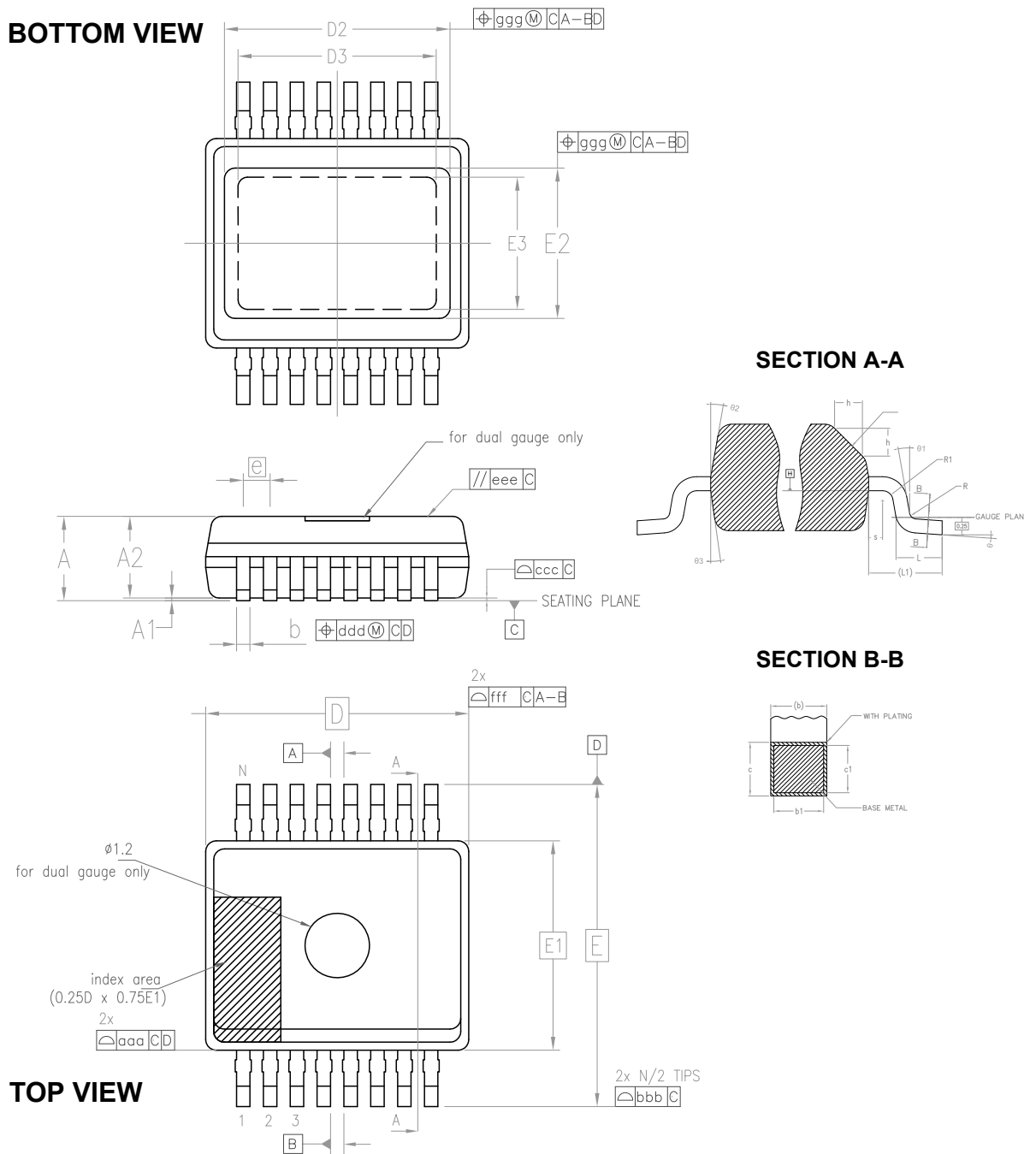
Area/island (cm ²)	FP	2	8	4L
R1 = R7 (°C/W)	1.35			
R2 = R8 (°C/W)	2.45			
R3 (°C/W)	4	4	4	4
R4 (°C/W)	16	6	6	4.2
R5 (°C/W)	30	20	10	3
R6 (°C/W)	26	20	18	7
C1 = C7 (W·s/°C)	0.001			
C2 = C8 (W·s/°C)	0.0152			
C3 (W·s/°C)	0.09			
C4 (W·s/°C)	0.2	0.3	0.3	0.4
C5 (W·s/°C)	0.4	1	1	4
C6 (W·s/°C)	3	5	7	18

7 Package information

To meet environmental requirements, ST offers these devices in different grades of **ECOPACK** packages, depending on their level of environmental compliance. ECOPACK specifications, grade definitions, and product status are available at: www.st.com. ECOPACK is an ST trademark.

7.1 PowerSSO-16 package information

Figure 46. PowerSSO-16 package outline



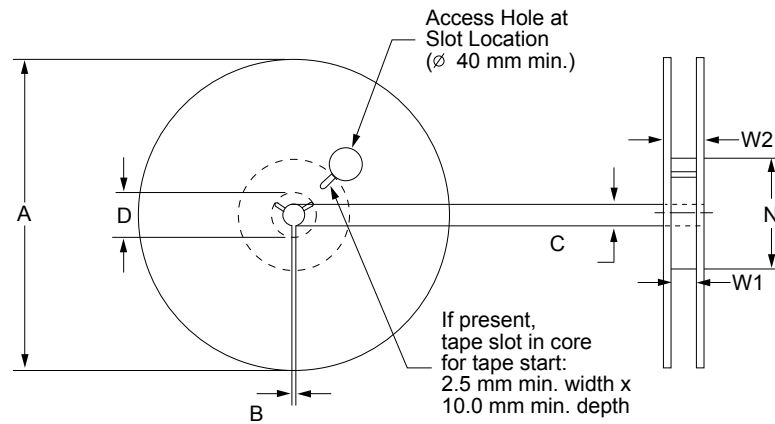
8017965_10_D

Table 16. PowerSSO-16 package mechanical data

Symbol	Millimeters		
	Min.	Typ.	Max.
Θ	0°		8°
$\Theta 1$	0°		
$\Theta 2$	5°		15°
$\Theta 3$	5°		15°
A			1.70
A1	0.00		0.10
A2	1.10		1.60
b	0.20		0.30
b1	0.20	0.25	0.28
c	0.19		0.25
c1	0.19	0.20	0.23
D	4.9 BSC		
D2	3.60		4.20
D3	2.90		
e	0.50 BSC		
E	6.00 BSC		
E1	3.90 BSC		
E2	2.50		3.10
E3	1.80		
h	0.25		0.50
L	0.40	0.60	0.85
L1	1.00 REF		
N	16		
R	0.07		
R1	0.07		
S	0.20		
Tolerance of form and position			
aaa	0.10		
bbb	0.10		
ccc	0.08		
ddd	0.08		
eee	0.10		
fff	0.10		
ggg	0.15		

7.2 PowerSSO-16 packing information

Figure 47. PowerSSO-16 reel 13"



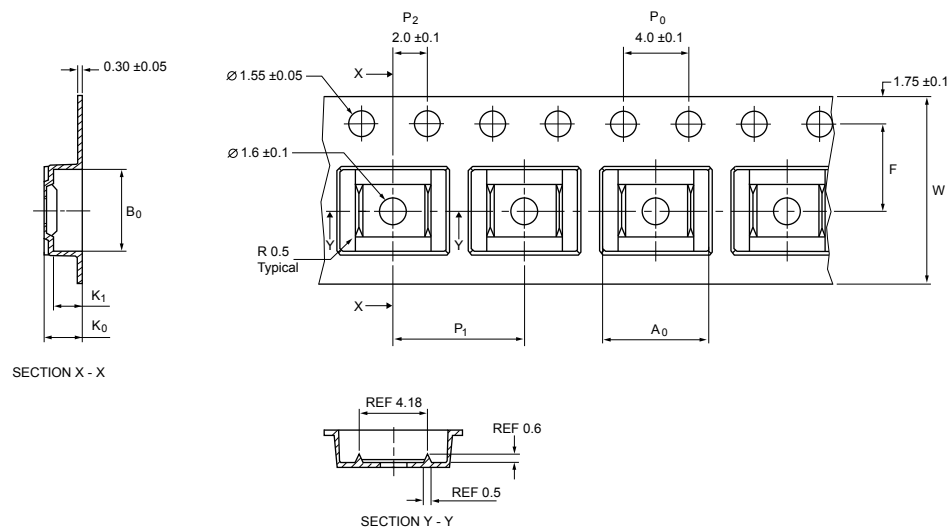
TAPG2004151655CFT

Table 17. Reel dimensions

Description	Value ⁽¹⁾
Base quantity	2500
Bulk quantity	2500
A (max)	330
B (min)	1.5
C (+0.5, -0.2)	13
D (min)	20.2
N	100
W1 (+2 /-0)	12.4
W2 (max)	18.4

1. All dimensions are in mm.

Figure 48. PowerSSO-16 carrier tape

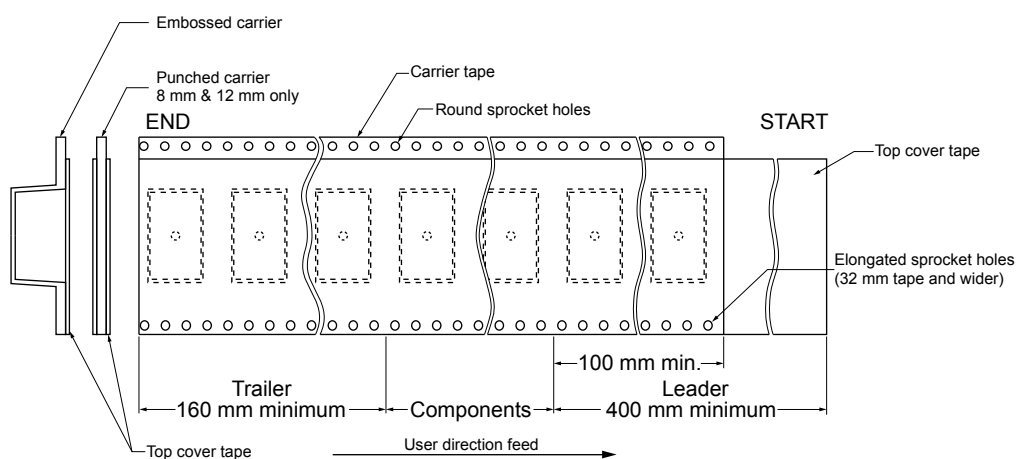


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Table 18. PowerSSO-16 carrier tape dimensions

Description	Value ⁽¹⁾
A ₀	6.50 ± 0.1
B ₀	5.25 ± 0.1
K ₀	2.10 ± 0.1
K ₁	1.80 ± 0.1
F	5.50 ± 0.1
P ₁	8.00 ± 0.1
W	12.00 ± 0.3

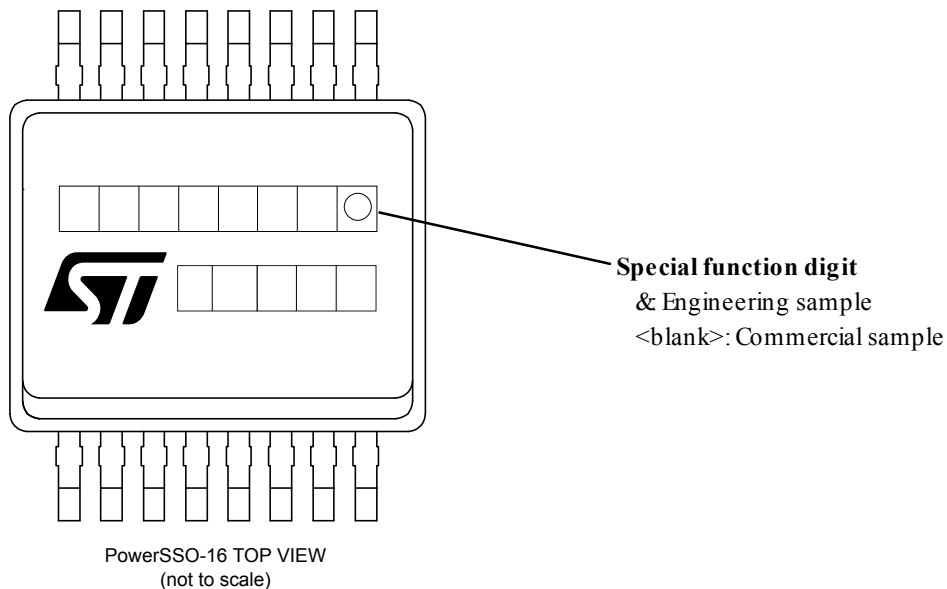
1. All dimensions are in mm.

Figure 49. PowerSSO-16 schematic drawing of leader and trailer tape


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7.3 PowerSSO-16 marking information

Figure 50. PowerSSO-16 marking information



GADG0310161234SMD

Parts marked as '&' are not yet qualified and therefore not approved for use in production. ST is not responsible for any consequences resulting from such use. In no event will ST be liable for the customer using any of these engineering samples in production. ST's Quality department must be contacted prior to any decision to use these engineering samples to run a qualification activity.

Revision history

Table 19. Document revision history

Date	Revision	Changes
06-Dec-2017	1	Initial release.
13-Dec-2017	2	No content changed, removed the watermark "Restricted".
25-Jun-2021	3	Added: - Section 5 Package information. Updated: - Figure 1. Block diagram; - Table 5. Power section; - Table 6. Switching; - Table 8. Protections; - Table 9. Current sense; - Section Features; - Section Applications; - Section 2.4 Waveforms. Minor text changes in: - Table 7. Logic inputs; - Table 3. Absolute maximum ratings.
23-Dec-2022	4	Update Section 2.1 Absolute maximum ratings and Section 2.2 Thermal data. Updated Table 5. Power section, Table 6. Switching, Table 8. Protections and Table 9. Current sense. Updated Section 4 Application information. Added Section 5 Maximum demagnetization energy ($V_{CC} = 16\text{ V}$). Added Section 6 Package and PCB thermal data. Minor text changes.
17-Jan-2025	5	Updated Table 3. Absolute maximum ratings, Table 9. Current sense and Figure 45. Thermal fitting model of a double-channel HSD in PowerSSO-16. Added Section 2.4: Electrical characteristics (curves).

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