

Technical documentation



Support & training



TUSB217A-Q1 SLLSFK6 – SEPTEMBER 2021

# TUSB217A-Q1 USB 2.0 High-Speed Signal Conditioner With DCP and CDP Controllers

# 1 Features

- AEC-Q100 qualified for automotive applications
   Device temperature grade 2: -40°C to 105°C T<sub>A</sub>
- Wide supply voltage range: 2.3 6.5 V
- Ultra-low USB disconnect and shutdown power consumption
- Provides USB 2.0 high-speed signal conditioning
- Compatible with USB 2.0, OTG 2.0 and BC 1.2
- Support for low-speed, full-speed, high-speed signaling
- Integrated BC 1.2 Charging Downstream Port (CDP) and Dedicated Charging Port (DCP) controllers that dynamically changes per DCP/CDP pin
- Host or device agnostic
- Supports up to 5-m cable length
  - Four selectable signal boost (edge boost along with DC boost) settings through the external pull-down resistor values
  - Three selectable RX equalization settings through the pull-up-or-down to compensate ISI jitter for high-loss applications
- Supports up to 10-m cable length with two TUSB217A-Q1 devices
- Scalable solution devices can be daisy chained for high loss applications
- RWB is pin compatible with TUSB211/212/214/216

# 2 Applications

- · Automotive infotainment and cluster
- Automotive head unit
- Active cable, cable extenders, backplane

# **3 Description**

The TUSB217A-Q1 is a third-generation USB 2.0 high-speed signal conditioner designed to compensate both AC loss (due to capacitive load) and DC loss (due to resistive loss) in the transmission channel.

The TUSB217A-Q1 leverages a patented design to speed-up transition edges of USB 2.0 high-speed signal with an edge booster and increases static levels with a DC boost function. In addition, the TUSB217A-Q1 includes a pre-equalization function to improve the receiver sensitivity and compensate the inter-symbol interference (ISI) jitter in application with longer cable length. USB low-speed and fullspeed signal characteristics are unaffected by the TUSB217A-Q1.

The TUSB217A-Q1 improves signal quality without altering packet timing or adding propagation delay or latency.

The TUSB217A-Q1 helps a system to pass the USB 2.0 high-speed near end eye compliance with a cable as long as 5 meters.

The TUSB217A-Q1 is compatible with the USB On-The-Go (OTG) and battery charging (BC 1.2) protocols. The Integrated BC 1.2 battery charging controller can be enabled through a control pin.

#### Device Information<sup>(1)</sup>

PART NUMBER	PACKAGE	BODY SIZE (NOM)
TUSB217A-Q1	X2QFN (12RWB)	1.60 mm × 1.60 mm

 For all available packages, see the orderable addendum at the end of the data sheet.



#### **Simplified Schematic**

An IMPORTANT NOTICE at the end of this data sheet addresses availability, warranty, changes, use in safety-critical applications, intellectual property matters and other important disclaimers. PRODUCTION DATA.



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# **4 Revision History**

DATE	REVISION	NOTES
September 2021	*	Initial Release



# **5 Device Comparison**

	TUSB211	TUSB212	TUSB214	TUSB216I	TUSB217A
Supply (V)	3.3	3.3	3.3	2.3 to 6.5	2.3 to 6.5
DC Boost		3 levels	3 levels	Tandem with AC Boost	Tandem with AC Boost
RX pre-equalization for ISI compensation				3 levels	3 levels
Charging Downstream Port (CDP) controller			Always ON	Pin Controlled	Always ON. Dynamically selected by DCP/CDP pin
Dedicated Charging Port (DCP) controller					Always ON. Dynamically selected by DCP/CDP pin
Cable length compensation for near-end high-speed eye mask Compliance (pre-channel before redriver/post-channel after redriver) (meter - gauge)	2/1 - 28AWG	4/2 - 28AWG	4/2 - 28AWG	6/3 - 28AWG (10 - 24AWG with one redriver on each end)	6/3 - 28AWG (10 - 24AWG with one redriver on each end)
Cable length compensation for far- end high-speed eye mask Compliance (pre-channel before redriver/post-channel after redriver) (meter - gauge)	5/3 - 28AWG	8/6 - 28AWG	8/6 - 28AWG	10/8 - 26AWG (10 - 28AWG with one redriver on each end)	10/8 - 26AWG (10 - 28AWG with one redriver on each end)



# **6** Pin Configuration and Functions



# Figure 6-1. TUSB217A-Q1 RWB 12-Pin X2QFN Top View

PIN (RWB)		I/O	INTERNAL	DESCRIPTION
NAME	NO. (RWB)	1/0	PULLUP/PULLDOWN	DESCRIPTION
BOOST	6	I	N/A	USB High-speed boost select through the external pull down resistor. Both edge boost and DC boost are controlled by a single pin in non- I2C mode. In I2C mode edge boost and DC boost can be individually controlled. Sampled upon power up. Does not recognize real time adjustments. Auto selects BOOST LEVEL = 3 when left floating.
DCP/CDP	11	I	500 kΩ PU	DCP or CDP mode selection. Low=DCP and High=CDP TUSB217A-Q1RWB BC1.2 controller is always enabled.
RX_SEN <sup>(2)</sup> /ENA_HS	9	I/O	N/A	In I2C mode:         Reserved for TI test purpose.         In non-I2C mode:         At reset: 3-level input signal RX_SEN. USB High-speed RX         Equalization Setting to Compensate ISI Jitter         H (pin is pulled high) – high RX equalization (high loss channel)         M (pin is left floating) – medium RX equalization (medium loss channel)         L (pin is pulled low) – low RX equalization (low loss channel)         After reset: Output signal ENA_HS. Flag indicating that channel is in High-speed mode. Asserted upon:         1. Detection of USB-IF High-speed test fixture from an unconnected state followed by transmission of USB TEST_PACKET pattern.         2. Squelch detection following USB reset with a successful HS handshake [HS handshake is declared to be successful after single chirp J chirp K pair where each chirp is within 18 µs – 128 µs].
D2P	7	I/O	N/A	USB High-speed positive port.
D2M	8	I/O	N/A	USB High-speed negative port.
GND	10	Р	N/A	Ground
D1M	1	I/O	N/A	USB High-speed negative port
D1P	2	I/O	N/A	USB High-speed positive port.
SDA <sup>(1)</sup>	3	I/O	500 kΩ PU 1.8 MΩ PD	I2C Mode: Bidirectional I2C data pin [7-bit I2C slave address = 0x2C]. In non I2C mode: Reserved for TI test purpose.
VCC	12	Р	N/A	Supply power

#### Table 6-1. Pin Functions



#### Table 6-1. Pin Functions (continued)

PIN (RWB)		I/O	INTERNAL	DESCRIPTION		
NAME	NO. (RWB)	1/0	PULLUP/PULLDOWN	DESCRIPTION		
RSTN	5	I	500 kΩ PU 1.8 MΩ PD	Device disable/enable. Low – Device is at reset and in shutdown, and High - Normal operation. Recommend 0.1-µF external capacitor to GND to ensure clean power on reset if not driven. If the pin is driven, it must be held low until the supply voltage for the device reaches within specifications.		
SCL <sup>(1)</sup> /CD	4	I/O	When RSTN asserted there is a 500 kΩ PD	In I2C mode: I2C clock pin [I2C address = 0x2C]. Non I2C mode: After reset: Output CD. Flag indicating that a USB device is attached (connection detected). Asserted from an unconnected state upon detection of DP or DM pull-up resistor. De-asserted upon detection of disconnect.		

(1) Pull-up resistors for SDA and SCL pins in I<sup>2</sup>C mode should be R<sub>Pull-up</sub> (depending on I2C bus voltage). If both SDA and SCL are pulled up at power-up the device enters into I<sup>2</sup>C mode.

(2) Pull-down and pull-up resistors for RX\_SEN pin must follow R<sub>RXSEN1</sub> and R<sub>RXSEN2</sub> resistor recommendations in non I<sup>2</sup>C mode.



# 7 Specifications

## 7.1 Absolute Maximum Ratings

Over operating free-air temperature range (unless otherwise noted) (1)

		MIN	MAX	UNIT
Supply voltage range	VCC	-0.3	7	V
Voltage range USB data	DxP, DxM	-0.3	5.5	V
Voltage range on BOOST pin	BOOST	-0.3	1.98	V
Voltage range other pins	RX_SEN, DCP/CDP,SDA,SCL, RSTN	-0.3	5.5	V
Storage temperature, T <sub>stg</sub>		-65	150	°C
Maximum junction temperature,	Г <sub>Ј (max)</sub>		125	°C

(1) Operation outside the Absolute Maximum Ratings may cause permanent device damage. Absolute Maximum Ratings do not imply functional operation of the device at these or any other conditions beyond those listed under Recommended Operating Conditions. If used outside the Recommended Operating Conditions but within the Absolute Maximum Ratings, the device may not be fully functional, and this may affect device reliability, functionality, performance, and shorten the device lifetime.

#### 7.2 ESD Ratings

			VALUE	UNIT
V	Electrostatic discharge	Human-body model (HBM), per AEC Q100-002 <sup>(1)</sup>	±2000	V
V <sub>(ESD)</sub>		Charged-device model (CDM), per AEC Q100-011 <sup>(2)</sup>	±750	v

(1) AEC Q100-002 HBM ESD Classification Level 2

(2) AEC Q100- 011 CDM ESD Classification Level C4A

## 7.3 Recommended Operating Conditions

over operating free-air temperature range (unless otherwise noted)

		MIN	NOM	MAX	UNIT
V <sub>CC</sub>	Supply voltage	2.3	5	6.5	V
T <sub>A</sub>	Operating free-air temperature (AEC-Q100)	-40		105	°C
TJ	Junction temperature (AEC-Q100)			115	°C
V <sub>I2C_BUS</sub>	I2C Bus Voltage	1.62		3.6	V
DxP, DxM	Voltage range USB data	0		3.6	V
BOOST	Voltage range BOOST pin	0		1.98	V
DIGITAL	Voltage range other pins (SCL, SDA, RSTN, DCP/CDP)	0		3.6	V
RX_SEN	Voltage range RX_SEN pin	0		5.0	V

## 7.4 Thermal Information

			UNIT
		12 PINS	
R <sub>θJA</sub>	Junction-to-ambient thermal resistance	137.4	°C/W
R <sub>0JC(top)</sub>	Junction-to-case (top) thermal resistance	62	°C/W
R <sub>θJB</sub>	Junction-to-board thermal resistance	67.2	°C/W
Ψ <sub>JT</sub>	Junction-to-top characterization parameter	1.9	°C/W
Ψ <sub>JB</sub>	Junction-to-board characterization parameter	67.3	°C/W
R <sub>0JC(bot)</sub>	Junction-to-case (bottom) thermal resistance	N/A	°C/W

(1) For more information about traditional and new thermal metrics, see the *Semiconductor and IC Package Thermal Metrics* application report, SPRA953.



# 7.5 Electrical Characteristics

Over operating free-air temperature range (unless otherwise noted)

	PARAMETER	TEST CONDITIONS	MIN TY	P <sup>(1)</sup> MAX	UNIT
POWER					
ACTIVE_HS	High Speed Active Current	USB channel = HS mode. 480 Mbps traffic. V <sub>CC</sub> supply stable, with Boost = Max		22 36	mA
UDLE_HS	High Speed Idle Current	USB channel = HS mode, no traffic. V <sub>CC</sub> supply stable, Boost = Max		22 36	mA
HS_SUPSPEND	High Speed Suspend Current	USB channel = HS Suspend mode. V <sub>CC</sub> supply stable	(	).75 1.4	mA
FS	Full-Speed Current	USB channel = FS mode, 12 Mbps traffic, $V_{cc}$ supply stable	(	).75 1.4	mA
DISCONN	Disconnect Power	Host side application. No device attachment.	(	0.80 1.4	mA
I <sub>SHUTDN</sub>	Shutdown Power	RSTN driven low, V <sub>CC</sub> supply stable		60 115	μA
CONTROL PIN LE	AKAGE		·		
LKG_FS	Pin failsafe leakage current for SDA, RSTN	V <sub>CC</sub> = 0 V, pin at V <sub>IH, max</sub>		10 15	μA
LKG_FS	Pin failsafe leakage current for RX_SEN	V <sub>CC</sub> = 0 V, pin at V <sub>IH, max</sub>		6 15	μA
I <sub>LKG_FS</sub>	Pin failsafe leakage current for SCL	V <sub>CC</sub> = 0 V, pin at V <sub>IH, max</sub>		70	nA
INPUT RSTN		·		I	
V <sub>IH</sub>	High level input voltage		1.5	3.6	V
VIL	Low-level input voltage		0	0.5	V
. <u>.</u> I <sub>IH</sub>	High level input current	V <sub>IH</sub> = 3.6 V, R <sub>PU</sub> enabled		±15	μA
 I <sub>IL</sub>	Low level input current	$V_{IL} = 0V, R_{PU}$ enabled		±20	μA
INPUT DIGITAL	I			I	
V <sub>IH</sub>	High level input voltage (DCP/ CDP)		1.5	3.6	V
V <sub>IL</sub>	Low-level input voltage (DCP/ CDP)		0	0.5	V
IIL	Low level input current	V <sub>IL</sub> = 0V		±20	μA
IIH	High level input current	V <sub>IH</sub> = 3.6 V		±15	μA
NPUT RX_SEN (3	-level input, for mid level leave pin f	loating)		I	
V <sub>IH(Max)</sub>	Maximum High level input voltage	VCC = 2.3V to 6.5V		5.0	v
,	Minimum High level input voltage	VCC > 4.5V	3.3		V
V <sub>IH(Min)</sub>		VCC = 2.3V to 4.5V (% of VCC)	75		%
	Low level input voltage	VCC > 4.5V		0.75	V
V <sub>IL</sub>		VCC = 2.3V to 4.5V (% of VCC)		15	%
INPUT BOOST				I	
R <sub>BOOST_LVL0</sub>	External pulldown resistor for BOOST Level 0			160	Ω
R <sub>BOOST_LVL1</sub>	External pulldown resistor for BOOST Level 1		1.5	1.8 2	kΩ
R <sub>BOOST_LVL2</sub>	External pulldown resistor for BOOST Level 2		3.4	3.6 3.96	kΩ
RBOOST_LVL3	External pulldown resistor for BOOST Level 3 to remove upper limit for resistor value, can be left open		7.5		kΩ
OUTPUTS CD, EN	A_HS				
V <sub>OH</sub>	High level output voltage for CD and ENA_HS	I <sub>O</sub> = -50 μA, VCC >= 3.0V	2.5		V
V <sub>OH</sub>	High level output voltage for CD	I <sub>O</sub> = –25 μA, VCC = 2.3V	1.7		V

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# 7.5 Electrical Characteristics (continued)

Over operating free-air temperature range (unless otherwise noted)

	PARAMETER	TEST CONDITIONS	MIN	TYP <sup>(1)</sup>	MAX	UNIT
V <sub>OH</sub>	High level output voltage for ENA_HS	I <sub>O</sub> = -25 μA, VCC = 2.3V	1.8			V
V <sub>OL</sub>	Low level output voltage for CD and ENA_HS	I <sub>O</sub> = 50 μA			0.3	V
I2C	-					
C <sub>I2C_BUS</sub>	I <sup>2</sup> C Bus Capacitance		4		150	pF
I <sub>OL</sub>	I <sup>2</sup> C open drain output current	V <sub>OL</sub> = 0.4V	<sub>DL</sub> = 0.4V 1.5			mA
V <sub>IL</sub> 2.3V<= VCC<= 4.3V, V <sub>I2C_BUS</sub> = 1.8V +/-10%		$R_{Pull-up}$ =1.6k $\Omega$ to 2.5k $\Omega$ , % of $V_{I2C_BUS}$			25	%
VIL	V <sub>I2C_BUS</sub> = 3.3V +/-10%	$R_{Pull-up}$ =2.8k $\Omega$ to 7k $\Omega$ , % of $V_{I2C\_BUS}$			25	%
V <sub>IH</sub>	2.3V<= VCC<= 4.3V, V <sub>I2C_BUS</sub> = 1.8V +/-10%	$R_{Pull-up}$ =1.6k $\Omega$ to 2.5k $\Omega$ , % of $V_{I2C_BUS}$	80			%
V <sub>IH</sub>	V <sub>I2C_BUS</sub> = 3.3V +/-10%	$R_{Pull-up}$ =2.8k $\Omega$ to 7k $\Omega$ , % of $V_{I2C\_BUS}$	75			%
R <sub>Pull-up</sub>	V <sub>I2C_BUS</sub> = 1.8V +/-10%		1.6	2	2.5	kΩ
R <sub>Pull-up</sub>	V <sub>I2C_BUS</sub> = 3.3V +/-10%		2.8	4.7	7	kΩ
SCL Frequency					100	kHz
DxP, DxM	·				I	
C <sub>IO_DXX</sub>	Capacitance to GND	Measured with VNA at 240 MHz, V <sub>CC</sub> supply stable, Redriver off		2.5		pF

(1) All typical values are at V<sub>CC</sub> = 5 V, and T<sub>A</sub> = 25°C.

# 7.6 Switching Characteristics

Over operating free-air temperature range (unless otherwise noted)

	<u> </u>					
	PARAMETER	TEST CONDITIONS	MIN	TYP <sup>(1)</sup>	MAX	UNIT
DxP, DxM US	SB Signals					
F <sub>BR_DXX</sub>	Bit Rate	USB channel = HS mode. 480 Mbps traffic. V <sub>CC</sub> supply stable			480	Mbps
t <sub>R/F_DXX</sub>	Rise/Fall time		100			ps

(1) All typical values are at  $V_{CC}$  = 5 V, and  $T_A$  = 25°C.



# 7.7 Timing Requirements

		MIN	NOM MAX	UNIT		
POWER U	P TIMING					
T <sub>RSTN_PW</sub>	Minimum width to detect a valid RSTN signal assert when the pin is actively driven low	100		μs		
T <sub>STABLE</sub>	VCC must be stable before RSTN de-assertion	300		μs		
T <sub>READY</sub>	Maximum time needed for the device to be ready after RSTN is de- asserted.					
T <sub>RAMP</sub>	V <sub>CC</sub> ramp time		100	ms		
T <sub>RAMP</sub>	V <sub>CC</sub> ramp time	0.2		ms		
I2C (STD)						
t <sub>susto</sub>	Stop setup time, SCL (T <sub>r</sub> =600ns-1000ns), SDA (T <sub>f</sub> =6.5ns-106.5ns), 100kHz STD	4		μs		
t <sub>HDSTA</sub>	Start hold time, SCL (Tr=600ns-1000ns), SDA (Tf=6.5ns-106.5ns), 100kHz STD	4		μs		
t <sub>SUSTA</sub>	Start setup time, SCL (T <sub>r</sub> =600ns-1000ns), SDA (T <sub>f</sub> =6.5ns-106.5ns), 100kHz STD	4.7		μs		
t <sub>SUDAT</sub>	Data input or False start/stop, setup time, SCL (T_r=600ns-1000ns), SDA (T_f=6.5ns-106.5ns), 100kHz STD	250		ns		
t <sub>HDDAT</sub>	Data input or False start/stop, hold time, SCL (T_r=600ns-1000ns), SDA (T_f=6.5ns-106.5ns), 100kHz STD	5		μs		
t <sub>BUF</sub>	Bus free time between START and STOP conditions	4.7		μs		
t <sub>LOW</sub>	Low period of the I <sub>2C</sub> clock	4.7		μs		
t <sub>HIGH</sub>	High period of the I <sub>2C</sub> clock	4		μs		
t <sub>F</sub>	Fall time of both SDA and SCL signals		300	ns		
t <sub>R</sub>	Rise time of both SDA and SCL signals		1000	ns		



# 8 Detailed Description

## 8.1 Overview

The TUSB217A-Q1 is a USB High-Speed (HS) signal conditioner designed to compensate for ISI signal loss in a transmission channel. TUSB217A-Q1 has a patented design for USB Low Speed (LS) and Full Speed (FS) signals. It does not alter the signal characteristics. HS signals are compensated. The design is compatible with USB On-The-Go (OTG) and Battery Charging (BC) specifications.

Programmable signal gain through an external resistor permits fine tuning device performance to optimize signals. This helps pass USB HS electrical compliance tests at the connector. Additional RX sensitivity, tuned by external pull-up resistor and pull-down resistor, allows to overcome attenuation in cables. The TUSB217A-Q1 allows application in series to cover longer distances, or high loss transmission paths. A maximum of 4 devices can be daisy-chained.

## 8.2 Functional Block Diagram



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## 8.3 Feature Description

#### 8.3.1 High-Speed Boost

The high-speed booster (combination of edge boost and DC boost) improves the eye width for USB2.0 highspeed signals. It is direction independent and by that is compatible to OTG systems. The BOOST pin is configuring the booster strength with different values of pull down resistors to set 4 levels of boosts, alternatively the boost level can be set through the I2C register according to Section 8.4.6. Internal circuitry of the signal conditioner reduces possible overshoot.

#### 8.3.2 RX Sensitivity

The RX\_SEN pin is a tri-level pin. It is used to set the equalization gain of the device according to system channel inter-symbol interference (ISI) loss. RX equalization can be increased to compensate for the higher ISI loss of the channel for example due to a long cable.

## 8.4 Device Functional Modes

#### 8.4.1 Low-Speed (LS) Mode

TUSB217A-Q1 automatically detects a LS connection and does not enable signal compensation. CD pin is asserted high but ENA\_HS will be low.

#### 8.4.2 Full-Speed (FS) Mode

TUSB217A-Q1 automatically detects a FS connection and does not enable signal compensation. CD pin is asserted high but ENA\_HS will be low

#### 8.4.3 High-Speed (HS) Mode

TUSB217A-Q1 automatically detects a HS connection and will enable signal compensation as determined by the configuration of the RX\_SEN pin and the external pull down resistance on its BOOST pin.

CD pin and ENA\_HS pin are asserted high when high-speed boost is active.



#### 8.4.4 High-Speed Downstream Port Electrical Compliance Test Mode

TUSB217A-Q1 will detect HS compliance test fixture and enter downstream port high-speed eye diagram test mode. CD pin will be low and ENA\_HS pin is asserted high when TUSB217A-Q1 is in HS eye compliance test mode.

If RSTN pin is asserted low and de-asserted high while TUSB217A-Q1 is operating in HS functional mode, TUSB217A-Q1 may transition to HS eye compliance test mode and CD asserts low and ENA\_HS remains high. When this occurs signal compensation is enabled.

#### 8.4.5 Shutdown Mode

TUSB217A-Q1 can be disabled when its RSTN pin is asserted low. DP, DM traces are continuous through the device in shutdown mode. The USB channel is still fully operational, but there is neither signal compensation, nor any indication from the CD pin as to the status of the channel.

MODE	CD	ENA_HS					
Low-speed	HIGH	LOW					
Full-speed	HIGH	LOW					
High-speed	HIGH	HIGH					
High-speed downstream port electrical test	LOW	HIGH					
Shutdown	LOW	LOW					

#### Table 8-1. CD and ENA\_HS Pins in Different Modes

#### 8.4.6 I<sup>2</sup>C Mode

TUSB217A-Q1 supports 100 KHz I2C for device configuration, status read back and test purposes. For detail electrical and functional specifications refer to I2C Bus Specification 2.1, 2001 – STANDARD MODE. This controller is enabled after SCL and SDA pins are sampled high shortly after return from shutdown. In this mode, the CSR can be accessed by I2C read/write transaction to 7-bit slave address 0x2C. It is advised to set CFG\_ACTIVE bit before changing values. This halts the FSM, and reset it after all changes are made. This ensure proper startup into high-speed mode.

#### 8.4.7 BC 1.2 Battery Charging Controller

Battery charging controller feature is always enabled in TUSB217A-Q1 RWB and supports both CDP charging downstream port functionality and DCP dedicated charging port functionality depending on DCP/CDP pin. When DCP/CDP pin is high the BC 1.2 controller supports CDP mode and when DCP/CDP pin is low BC 1.2 controller supports DCP mode. DCP/CDP pin can be dynamically controlled. When host or hub is disabled DCP/CDP pin can be set low to support DCP mode and when host or hub is enabled DCP/CDP pin can be set to high to support CDP. Downstream VBUS should be toggled after the DCP/CDP pin change so BC 1.2 handshake starts over to indicate charging mode change.

DCP/CDP pin has an internal pull up resistor. When DCP/CDP pin is left unconnected the BC 1.2 controller will be in CDP mode.

Pin 11 (DCP/CDP)	CDP	DCP						
Low	NO	YES						
High	YES	NO						

#### Table 8-2. TUSB217A-Q1 RWB Battery Charging Controller Modes



# 8.5 TUSB217A-Q1 Registers

Table 8-3 lists the memory-mapped registers for the TUSB217A-Q1 registers. All register offset addresses not listed in Table 8-3 should be considered as reserved locations and the register contents should not be modified.

Table 8-3. TUSB21/A-Q1 Registers						
Offset	Acronym	Register Name	Section			
0x1	EDGE_BOOST	This register is setting EDGE BOOST level.	Go			
0x3	CONFIGURATION	This register is selecting device mode.	Go			
0xE	DC_BOOST	This register is setting DC BOOST level.	Go			
0x25	RX_SEN	This register is setting RX Sensitivity level.	Go			

## Table 8-3. TUSB217A-Q1 Registers

Complex bit access types are encoded to fit into small table cells. Table 8-4 shows the codes that are used for access types in this section.

Access Type	Code	Description			
Read Type					
RH	H R	Set or cleared by hardware Read			
Write Type					
W	W	Write			
Reset or Default	Value				
-n		Value after reset or the default value			

#### Table 8-4. TUSB217A-Q1 Access Type Codes

## 8.5.1 EDGE\_BOOST Register (Offset = 0x1) [reset = X]

EDGE\_BOOST is shown in Figure 8-1 and described in Table 8-5.

Return to Summary Table.

This register is setting EDGE BOOST level.

Figure 8-1. EDGE_BOOST Register
---------------------------------

7	6	5	4	3	2	1	0
	ACB	_LVL			RESE	RVED	
	RH	W-X			RH/	W-X	

#### Table 8-5. EDGE BOOST Register Field Descriptions

Bit	Field	Туре	Reset	Description
7-4	ACB_LVL	RH/W	Х	XXXXb (sampled at startup from BOOST pin) 0000b to 1111b range
				0x0 = BOOST PIN LEVEL 0 (lowest edge boost setting)
				0x3 = BOOST PIN LEVEL 1
				0x6 = BOOST PIN LEVEL 2
				0xA = BOOST PIN LEVEL 3
				0xF = (highest edge boost setting)
3-0	RESERVED	RH/W	x	These bits are reserved bits and set by hardware at reset. When this register is modified the software should first read these reserved bits and rewrite with the same values

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#### 8.5.2 CONFIGURATION Register (Offset = 0x3) [reset = X]

CONFIGURATION is shown in Figure 8-2 and described in Table 8-6.

#### Return to Summary Table.

This register is selecting device mode.

#### Figure 8-2. CONFIGURATION Register

7	6	5	4	3	2	1	0
RESERVED							CFG_ACTIVE
RH/W-X							RH/W-0x1

#### Table 8-6. CONFIGURATION Register Field Descriptions

Bit	Field	Туре	Reset	Description
7-1	RESERVED	RH/W	X	These bits are reserved bits and set by hardware at reset. When this register is modified the software should first read these reserved bits and rewrite with the same values
0	CFG_ACTIVE	RH/W	0x1	Configuration mode After reset, if I2C mode is true (SCL and SDA are both pulled high) set the bit to get into configuration mode and clear to return to normal mode. 0x0 = NORMAL MODE 0x1 = CONFIGURATION MODE

## 8.5.3 DC\_BOOST Register (Offset = 0xE) [reset = X]

DC\_BOOST is shown in Figure 8-3 and described in Table 8-7.

Return to Summary Table.

This register is setting DC BOOST level.

#### Figure 8-3. DC\_BOOST Register

7	6	5	4	3	2	1	0
	RESE	RVED			DCB	_LVL	
	RH/	W-X			RH/	W-X	

#### Table 8-7. DC\_BOOST Register Field Descriptions

Bit	Field	Туре	Reset	Description
7-4	RESERVED	RH/W	X	These bits are reserved bits and set by hardware at reset. When this register is modified the software should first read these reserved bits and rewrite with the same values
3-0	DCB_LVL	RH/W	x	XXXXb (sampled at startup from BOOST pin) 0000b to 1111b range 0x0 = BOOST PIN LEVEL 0 (lowest dc boost setting) 0x2 = BOOST PIN LEVEL 1 and 2 0x6 = BOOST PIN LEVEL 3 0xF = (highest dc boost setting)



# 8.5.4 RX\_SEN Register (Offset = 0x25) [reset = X]

RX\_SEN is shown in Figure 8-4 and described in Table 8-8.

Return to Summary Table.

This register is setting RX Sensitivity level.

Figure	8-4	RX	SEN	Register
rigure	0-4.		JEN	Register

7	6	5	4	3	2	1	0
			RX_	SEN			
			RH/	W-X			

# Bit Field Type Reset Description 7-0 RX\_SEN RH/W X XXXXb (sampled at startup from RX\_SEN pin) 00000000b to 1111111b range 0x0 = RX\_SEN LEVEL LOW 0x3 = RX\_SEN LEVEL LOW 0x33 = RX\_SEN LEVEL MID 0x66 = RX\_SEN LEVEL HIGH 0xFF = (highest setting)

#### Table 8-8. RX\_SEN Register Field Descriptions



# 9 Application and Implementation

#### Note

Information in the following applications sections is not part of the TI component specification, and TI does not warrant its accuracy or completeness. TI's customers are responsible for determining suitability of components for their purposes, as well as validating and testing their design implementation to confirm system functionality.

## 9.1 Application Information

The purpose of the TUSB217A-Q1 is to re-store the signal integrity of a USB High-speed channel up to the USB connector. The loss in signal quality stems from reduced channel bandwidth due to high loss PCB trace and other components that contribute a capacitive load. This can cause the channel to fail the USB near end eye mask. Proper use of the TUSB217A-Q1 can help to pass this eye mask.

A secondary purpose is to use the CD pin of the TUSB217A-Q1 to control other blocks on the customer platform, if so desired.

#### 9.2 Typical Application

A typical application for TUSB217A-Q1 with dynamic mode change between DCP and CDP is shown in Figure 9-1. BC 1.2 controller mode will be based on host/hub active state in this application. When host/hub is not active the controller will be in DCP mode and when the host/hub is active the controller will be in CDP mode. Downstream VBUS needs to be toggled by the power controller to change advertisement and for portable device to re-detect the BC 1.2 controller charging mode. In this setup, D2P and D2M face the USB connector while D1P and D1M face the USB host. The orientation may be reversed (that is, D2 faces transceiver and D1 faces connector).





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#### Figure 9-1. TUSB217A-Q1: A Reference Schematic (Design Example with DCP/CDP Dynamic Switching). Downstream VBUS Needs to be Toggled if Upstream VBUS State Changes for BC 1.2 Controller to Change DCP/CDP Advertisement.



#### 9.2.1 Design Requirements

TUSB217A-Q1 requires a valid reset signal as described in the *Power Supply Recommendations* section. The capacitor at RSTN pin is not required if a micro controller drives the RSTN pin according to recommendations.

For this design example, use the parameters shown in Table 9-1, Table 9-2 and Table 9-3.

#### Table 9-1. Design Parameters for 5-V Supply With High Loss System

		VALUE <sup>(1)</sup>		
V <sub>CC</sub>		5 V ±10%		
I <sup>2</sup> C support require	No			
		0-Ω	0	
Edge and DC Boost		1.8 kΩ ±1% 1		Boost Level 1: R <sub>BOOST</sub> = 1.8 kΩ
		3.6 kΩ ± 1%	2	
		Do Not Install (DNI)	3	
	R <sub>RXSEN1</sub>	R <sub>RXSEN2</sub>	RX_SEN Level	High RX
	22 kΩ - 40 kΩ (27 kΩ typical)	Do Not Install (DNI)	Low	Sensitivity Level: R <sub>RXSEN1</sub> = 37.5
RX Sensitivity	Do Not Install (DNI)	Do Not Install (DNI)	Medium	kΩ
	37.5 kΩ <sup>(2)</sup>	12.5 kΩ	High	R <sub>RXSEN2</sub> = 12.5 kΩ

(1) These parameters are starting values for a high loss system. Further tuning might be required based on specific host or device as well as cable length and loss profile. These settings are not specific to a 5 V supply system could be applicable to 3.3 V supply system as well.

(2) This resistor is needed for a 5 V supply to divide the voltage down so the RX\_SEN pin voltage does not exceed 5.0 V.

#### Table 9-2. Design Parameters for 3.3-V Supply With Low to Medium Loss System

PARAMETER							
V <sub>CC</sub>	3.3 V ±10%						
I <sup>2</sup> C support require	No						
		R <sub>BOOST</sub>	BOOST Level				
		0-Ω	0	1			
Edge and DC Boo	st	1.8 kΩ ±1%	1	Boost Level 0: R <sub>BOOST</sub> = 0-Ω			
		3.6 kΩ ±1% 2					
		Do Not Install (DNI)	3				
	R <sub>RXSEN1</sub>	R <sub>RXSEN2</sub>	RX_SEN Level	Medium RX			
22 kΩ – 40 kΩ (27 kΩ typi		Do Not Install (DNI)	Low	Sensitivity Level:			
RX Sensitivity	Do Not Install (DNI)	Do Not Install (DNI)	Medium	R <sub>RXSEN1</sub> = DNI			
	Do Not Install (DNI)	22 kΩ – 40 kΩ (27 kΩ typical)	High	R <sub>RXSEN2</sub> = DNI			

(1) These parameters are starting values for a low to medium loss system. Further tuning might be required based on specific host or device as well as cable length and loss profile. These settings are not specific to a 3.3 V supply system could be applicable to 5 V supply system as well.



#### Table 9-3. Design Parameters for 2.3-V to 4.3-V VBAT Supply With Low to Medium Loss System

	PAR	AMETER		VALUE <sup>(1)</sup>			
V <sub>cc</sub>							
I <sup>2</sup> C support require	No						
		R <sub>BOOST</sub>	BOOST Level				
	0-Ω		0				
Edge and DC Boo	ost	1.8 kΩ ±1%	1	Boost Level 0: R <sub>BOOST</sub> = 0-Ω			
		3.6 kΩ ±1%	2				
		Do Not Install (DNI)	3				
	R <sub>RXSEN1</sub>	R <sub>RXSEN2</sub>	RX_SEN Level	Medium RX			
DV Sensitivity	22 kΩ – 40 kΩ (27 kΩ typical)	Do Not Install (DNI)	Low	Sensitivity Level:			
RX Sensitivity	Do Not Install (DNI)	Do Not Install (DNI)	Medium	R <sub>RXSEN1</sub> = DNI			
	37.5 kΩ <sup>(2)</sup>	12.5 kΩ	High	- R <sub>RXSEN2</sub> = DNI			

(1) These parameters are starting values for a low to medium loss system. Further tuning might be required based on specific host or device as well as cable length and loss profile. These settings are not specific to a 2.3 V – 4.3 V supply system could be applicable to 5 V supply system as well.

(2) This resistor is needed for a VBAT supply (2.3 V – 4.3 V) to divide the voltage down so the RX\_SEN pin voltage does not exceed 5.0 V.

#### 9.2.2 Detailed Design Procedure

The ideal BOOST setting is dependent upon the signal chain loss characteristics of the target platform. The recommendation is to start with BOOST level 0, and then increment to BOOST level 1, and so on. Same applies to the RX sensitivity setting where it is recommended to plan for the required pads or connections to change boost settings, but to start with RX sensitivity level Low.

In order for the TUSB217A-Q1 to recognize any change to the BOOST setting, the RSTN pin must be toggled. This is because the BOOST pin is latched on power up and the pin is ignored thereafter.

**Note** The TUSB217A-Q1 compensates for extra attenuation in the signal path according to the configuration of the RX\_SEN pin. This maximum recommended voltage for this pin is 5 V when selecting the highest RX sensitivity level.

Placement of the device is also dependent on the application goal. Table 9-4 summarizes our recommendations.

PLATFORM GOAL	SUGGESTED TUSB217A-Q1 PLACEMENT					
Pass USB Near End Mask at the receptacle	Close to measurement point (connector)					
Pass USB Far End Eye Mask at the plug	Close to USB PHY					
Cascade multiple TUSB217A-Q1s to improve device enumeration	Midway between each USB interconnect					

#### Table 9-4. Platform Placement Guideline

## Table 9-5. Table of Recommended Settings

BOO	BOOST and RX_SEN settings <sup>(1)</sup> for channel loss						
Pre-channel cable length (Between USB PHY and TUSB217A-Q1)	BOOST	RX_SEN					
0-3 meter	Level 0	Medium or High					
2-5 meter	Level 1	Medium or High					
Post-channel cable length (Between TUSB217A-Q1 and inter-connect)	BOOST	RX_SEN					
0-2 meter	Level 0	Medium or High					
1-4 meter	Level 1	Medium or High					

(1) These parameters are starting values for different cable lengths. Further tuning might be required based on specific host or device as well as cable length and loss profile.

#### 9.2.2.1 Test Procedure to Construct USB High-speed Eye Diagram

#### Note

USB-IF certification tests for High-speed eye masks require the *mandated use* of the USB-IF developed test fixtures. These test fixtures do not require the use of oscilloscope probes. Instead they use SMA cables. More information can be found at the USB-IF Compliance Updates Page. It is located under the *Electrical Specifications* section, ID 86 dated March 2013.

The following procedure must be followed before using any oscilloscope compliance software to construct a USB High-speed Eye Mask:

#### 9.2.2.1.1 For a Host Side Application

- 1. Configure the TUSB217A-Q1 to the desired BOOST setting.
- 2. Power on (or toggle the RSTN pin if already powered on) the TUSB217A-Q1.
- 3. Using SMA cables, connect the oscilloscope and the USB-IF host-side test fixture to the TUSB217A-Q1.
- 4. Enable the host to transmit USB TEST\_PACKET.
- 5. Execute the oscilloscope USB compliance software.
- 6. Repeat the above steps in order to retest TUSB217A-Q1 with a different BOOST setting (must reset to change).

#### 9.2.2.1.2 For a Device Side Application

- 1. Configure the TUSB217A-Q1 to the desired BOOST setting.
- 2. Power on (or toggle the RSTN pin if already powered on) the TUSB217A-Q1.
- 3. Connect a USB host, the USB-IF device-side test fixture, and USB device to the TUSB217A-Q1. Ensure that the USB-IF device test fixture is configured to the 'INIT' position.
- 4. Allow the host to enumerate the device.
- 5. Enable the device to transmit USB TEST\_PACKET.
- 6. Using SMA cables, connect the oscilloscope to the USB-IF device-side test fixture and ensure that the device-side test fixture is configured to the 'TEST' position.
- 7. Execute the oscilloscope USB compliance software.
- 8. Repeat the above steps in order to re-test TUSB217A-Q1 with a different BOOST setting (must reset to change).



#### 9.2.3 Application Curves



#### Figure 9-2. Near End Eye Measurement Set-Up With Pre-Channel Cable





#### 9.2.3 Application Curves (continued)



## 9.2.3 Application Curves



#### Figure 9-11. Near End Eye Measurement Set-Up With Post-Channel Cable





#### 9.2.3 Application Curves (continued)





# **10 Power Supply Recommendations**

On power up, the interaction of the RSTN pin and power on ramp could result in digital circuits not being set correctly. The device should not be enabled until the power on ramp has settled to minimum recommended supply voltage or higher to ensure a correct power on reset of the digital circuitry. If RSTN cannot be held low by microcontroller or other circuitry until the power on ramp has settled, then an external capacitor from the RSTN pin to GND is required to hold the device in the low power reset state.

The RC time constant should be larger than five times of the power on ramp time (0 to  $V_{CC}$ ). With a typical internal pullup resistance of 500 k $\Omega$ , the recommended minimum external capacitance is calculated as:

[Ramp Time x 5]  $\div$  [500 k $\Omega$ ]

(1)

## 11 Layout

#### **11.1 Layout Guidelines**

Although the land pattern has matched trace width to pad width, optimal impedance control is based on the user's own PCB stack-up. The recommendation is to maintain 90  $\Omega$  differential routing underneath the device.

#### 11.2 Layout Example



Figure 11-1. Layout Example



# 12 Device and Documentation Support

#### 12.1 Receiving Notification of Documentation Updates

To receive notification of documentation updates, navigate to the device product folder on ti.com. Click on *Subscribe to updates* to register and receive a weekly digest of any product information that has changed. For change details, review the revision history included in any revised document.

#### **12.2 Support Resources**

TI E2E<sup>™</sup> support forums are an engineer's go-to source for fast, verified answers and design help — straight from the experts. Search existing answers or ask your own question to get the quick design help you need.

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#### 12.3 Trademarks

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#### **12.4 Electrostatic Discharge Caution**



This integrated circuit can be damaged by ESD. Texas Instruments recommends that all integrated circuits be handled with appropriate precautions. Failure to observe proper handling and installation procedures can cause damage.

ESD damage can range from subtle performance degradation to complete device failure. Precision integrated circuits may be more susceptible to damage because very small parametric changes could cause the device not to meet its published specifications.

#### 12.5 Glossary

TI Glossary This glossary lists and explains terms, acronyms, and definitions.

## 13 Mechanical, Packaging, and Orderable Information

The following pages include mechanical, packaging, and orderable information. This information is the most current data available for the designated devices. This data is subject to change without notice and revision of this document. For browser-based versions of this data sheet, refer to the left-hand navigation.



## PACKAGING INFORMATION

Orderable Device	Status (1)	Package Type	Package Drawing	Pins	Package Qty	Eco Plan (2)	Lead finish/ Ball material (6)	MSL Peak Temp (3)	Op Temp (°C)	Device Marking (4/5)	Samples
TUSB217ARWBRQ1	ACTIVE	X2QFN	RWB	12	3000	RoHS & Green	NIPDAU	Level-2-260C-1 YEAR	-40 to 105	7A	Samples
TUSB217ARWBTQ1	ACTIVE	X2QFN	RWB	12	250	RoHS & Green	NIPDAU	Level-2-260C-1 YEAR	-40 to 105	7A	Samples

<sup>(1)</sup> The marketing status values are defined as follows:

ACTIVE: Product device recommended for new designs.

LIFEBUY: TI has announced that the device will be discontinued, and a lifetime-buy period is in effect.

NRND: Not recommended for new designs. Device is in production to support existing customers, but TI does not recommend using this part in a new design.

**PREVIEW:** Device has been announced but is not in production. Samples may or may not be available.

**OBSOLETE:** TI has discontinued the production of the device.

<sup>(2)</sup> RoHS: TI defines "RoHS" to mean semiconductor products that are compliant with the current EU RoHS requirements for all 10 RoHS substances, including the requirement that RoHS substance do not exceed 0.1% by weight in homogeneous materials. Where designed to be soldered at high temperatures, "RoHS" products are suitable for use in specified lead-free processes. TI may reference these types of products as "Pb-Free".

**RoHS Exempt:** TI defines "RoHS Exempt" to mean products that contain lead but are compliant with EU RoHS pursuant to a specific EU RoHS exemption.

Green: TI defines "Green" to mean the content of Chlorine (CI) and Bromine (Br) based flame retardants meet JS709B low halogen requirements of <=1000ppm threshold. Antimony trioxide based flame retardants must also meet the <=1000ppm threshold requirement.

<sup>(3)</sup> MSL, Peak Temp. - The Moisture Sensitivity Level rating according to the JEDEC industry standard classifications, and peak solder temperature.

<sup>(4)</sup> There may be additional marking, which relates to the logo, the lot trace code information, or the environmental category on the device.

<sup>(5)</sup> Multiple Device Markings will be inside parentheses. Only one Device Marking contained in parentheses and separated by a "~" will appear on a device. If a line is indented then it is a continuation of the previous line and the two combined represent the entire Device Marking for that device.

<sup>(6)</sup> Lead finish/Ball material - Orderable Devices may have multiple material finish options. Finish options are separated by a vertical ruled line. Lead finish/Ball material values may wrap to two lines if the finish value exceeds the maximum column width.

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# PACKAGE OPTION ADDENDUM

15-Sep-2021

#### OTHER QUALIFIED VERSIONS OF TUSB217A-Q1 :

• Catalog : TUSB217A

NOTE: Qualified Version Definitions:

Catalog - TI's standard catalog product



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STRUMENTS

## TAPE AND REEL INFORMATION





#### QUADRANT ASSIGNMENTS FOR PIN 1 ORIENTATION IN TAPE



*All dimensions are nominal												
Device	•	Package Drawing		SPQ	Reel Diameter (mm)	Reel Width W1 (mm)	A0 (mm)	B0 (mm)	K0 (mm)	P1 (mm)	W (mm)	Pin1 Quadrant
TUSB217ARWBRQ1	X2QFN	RWB	12	3000	180.0	9.5	1.8	1.8	0.45	4.0	8.0	Q1
TUSB217ARWBTQ1	X2QFN	RWB	12	250	180.0	9.5	1.8	1.8	0.45	4.0	8.0	Q1



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# PACKAGE MATERIALS INFORMATION

7-Oct-2023



\*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Length (mm)	Width (mm)	Height (mm)
TUSB217ARWBRQ1	X2QFN	RWB	12	3000	189.0	185.0	36.0
TUSB217ARWBTQ1	X2QFN	RWB	12	250	189.0	185.0	36.0

# **RWB0012A**



# **PACKAGE OUTLINE**

# X2QFN - 0.4 mm max height

PLASTIC QUAD FLATPACK - NO LEAD



NOTES:

- 1. All linear dimensions are in millimeters. Any dimensions in parenthesis are for reference only. Dimensioning and tolerancing per ASME Y14.5M.2. This drawing is subject to change without notice.



# **RWB0012A**

# **EXAMPLE BOARD LAYOUT**

# X2QFN - 0.4 mm max height

PLASTIC QUAD FLATPACK - NO LEAD



NOTES: (continued)

3. For more information, see Texas Instruments literature number SLUA271 (www.ti.com/lit/slua271).



# **RWB0012A**

# **EXAMPLE STENCIL DESIGN**

# X2QFN - 0.4 mm max height

PLASTIC QUAD FLATPACK - NO LEAD



NOTES: (continued)

4. Laser cutting apertures with trapezoidal walls and rounded corners may offer better paste release. IPC-7525 may have alternate design recommendations.



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